

ADM690A/ADM692A/ADM802L/M/ADM805L/M—SPECIFICATIONS

($V_{CC} = 4.75\text{ V}$ to 5.5 V (ADM690A/ADM802L/ADM805L), $V_{CC} = 4.5\text{ V}$ to 5.5 V , (ADM692A/ADM802M/ADM805M), $V_{BATT} = +2.8\text{ V}$, $T_A = T_{MIN}$ to T_{MAX} unless otherwise noted)

Parameter	Min	Typ	Max	Units	Test Conditions/Comments
V _{CC} /V _{BATT} OPERATION ¹					
V _{CC} Operating Voltage Range	1.0		5.5	V	V _{CC} = 0 V, V _{BATT} = 2.8 V 5.5 V > V _{CC} > V _{BATT} + 0.2 V
Supply Current (Excludes I _{OUT})		70	100	μA	
Supply Current in Battery Backup Mode		0.05	1.0	μA	
Battery Standby Current (+ = Discharge, – = Charge)	–0.1		+0.02	μA	
V _{OUT} Output Voltage	V _{CC} – 0.02	V _{CC} – 0.01		V	I _{OUT} = 5 mA
	V _{CC} – 0.5	V _{CC} – 0.05		V	I _{OUT} = 50 mA
		V _{CC} – 0.02		V	I _{OUT} = 250 mA
V _{OUT} in Battery Backup Mode	V _{BATT} – 0.05	V _{BATT} – 0.002		V	I _{OUT} = 250 μA, V _{CC} < V _{BATT} – 0.2 V
Battery Switchover Threshold		20		mV	Power Up
		–20		mV	Power Down
Battery Switchover Hysteresis		40		mV	
RESET THRESHOLD					
Reset Voltage Threshold					T _A = 25°C, V _{CC} Falling T _A = 25°C, V _{CC} Falling
ADM690A, ADM802L, ADM805L	4.5	4.65	4.75	V	
ADM692A, ADM802M, ADM805M	4.25	4.4	4.5	V	
ADM802L	4.55		4.7	V	
ADM802M	4.30		4.45	V	
Reset Threshold Hysteresis		40		mV	I _{SOURCE} = 800 μA I _{SINK} = 3.2 mA I _{SINK} = 100 μA, V _{CC} = 1 V I _{SOURCE} = 4 μA, V _{CC} = 1.1 V ADM805L/M, I _{SOURCE} = 800 μA ADM805L/M, I _{SINK} = 3.2 mA
Reset Timeout Delay	140	200	280	ms	
RESET Output Voltage	V _{CC} – 1.5		0.4	V	
			0.3	V	
				V	
RESET Output Voltage	0.8			V	I _{SOURCE} = 800 μA I _{SINK} = 3.2 mA
	V _{CC} – 1.5		0.4	V	
WATCHDOG TIMER					
Watchdog Timeout Period	1.0	1.6	2.25	s	V _{IL} = 0.4, V _{IH} = 0.8 (V _{CC})
WDI Input Pulse Width	50			ns	
WDI Input Threshold					WDI = V _{CC} WDI = 0 V
Logic Low			0.8	V	
Logic High	3.5			V	
WDI Input Current	–10		10	μA	
				μA	
POWER FAIL DETECTOR					
PFI Input Threshold	1.20	1.25	1.30	V	ADM690A, ADM692A, ADM805L/M ADM802L/M
	1.225	1.25	1.275	V	
PFI Input Current	–25	0.01	+25	nA	I _{SOURCE} = 800 μA I _{SINK} = 3.2 mA
PFO Output Voltage	V _{CC} – 1.5			V	
			0.4	V	

NOTES

¹Either V_{CC} or V_{BATT} can be 0 V if the other $> +2.0\text{ V}$.

Specifications subject to change without notice.

ADM690A/ADM692A/ADM802L/M/ADM805L/M

ABSOLUTE MAXIMUM RATINGS*

(T_A = +25°C unless otherwise noted)

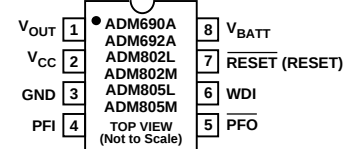
V _{CC}	−0.3 V to +6 V
V _{BATT}	−0.3 V to +6 V
All Other Inputs	−0.3 V to V _{CC} + 0.3 V
Input Current	
V _{CC}	200 mA
V _{BATT}	50 mA
GND	20 mA
Digital Output Current	20 mA
Power Dissipation, N-8 DIP	400 mW
θ _{JA} Thermal Impedance	120°C/W
Power Dissipation, SO-8 SOIC	500 mW
θ _{JA} Thermal Impedance	110°C/W
Operating Temperature Range	
Industrial (A Version)	−40°C to +85°C
Lead Temperature (Soldering, 10 sec)	+300°C
Vapor Phase (60 sec)	+215°C
Infrared (15 sec)	+220°C
Storage Temperature Range	−65°C to +150°C
ESD Rating	>4 kV

*Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those listed in the operational sections of this specification is not implied. Exposure to absolute maximum ratings for extended periods of time may affect device reliability.

ORDERING GUIDE

Model	Temperature Range	Package Option
ADM690AAN	−40°C to +85°C	N-8
ADM690AARN	−40°C to +85°C	SO-8
ADM690AARM	−40°C to +85°C	RM-8
ADM692AAN	−40°C to +85°C	N-8
ADM692AARN	−40°C to +85°C	SO-8
ADM802LAN	−40°C to +85°C	N-8
ADM802LARN	−40°C to +85°C	SO-8
ADM802MAN	−40°C to +85°C	N-8
ADM802MARN	−40°C to +85°C	SO-8
ADM805LAN	−40°C to +85°C	N-8
ADM805LARN	−40°C to +85°C	SO-8
ADM805MAN	−40°C to +85°C	N-8
ADM805MARN	−40°C to +85°C	SO-8

PIN CONFIGURATIONS



PIN FUNCTION DESCRIPTION

Mnemonic	Function
V _{CC}	Power Supply Input: +5 V Nominal.
V _{BATT}	Backup Battery Input. As V _{CC} falls below the reset threshold and below V _{BATT} by 20 mV, V _{BATT} will be switched to V _{OUT} . On power-up as V _{CC} rises to 20 mV above V _{BATT} , V _{OUT} will be switched back to V _{CC} .
V _{OUT}	Output Voltage. When V _{CC} is above the reset threshold, V _{OUT} is connected to V _{CC} through an on chip switch. When V _{CC} is below the reset threshold, the higher of V _{CC} or V _{BATT} is connected to V _{OUT} .
GND	0 V. Ground reference for all signals.
PFI	Power Fail Comparator Input. If PFI is less than 1.25 V, the power fail output $\overline{\text{PFO}}$ goes low. If unused, PFI should be connected to V _{CC} or GND.
$\overline{\text{PFO}}$	Power Fail Comparator Output. If PFI is less than 1.25 V, the power fail output $\overline{\text{PFO}}$ goes low.
$\overline{\text{RESET}}$	Logic Output. $\overline{\text{RESET}}$ goes low if 1. V _{CC} falls below the Reset Threshold 2. The watchdog timer is not serviced within its timeout period (1.6 seconds) The reset threshold is typically 4.65 V for the ADM690A/ADM802L/ADM805L and 4.4 V for the ADM692A/ADM802M/ADM805M. $\overline{\text{RESET}}$ remains low for 200 ms after V _{CC} returns above the threshold. $\overline{\text{RESET}}$ also goes low for 200 ms if the watchdog timer is enabled but not serviced within its timeout period.
RESET	Active high RESET output (ADM805L/M only). This is the inverse of $\overline{\text{RESET}}$. The asserted (high) level is V _{CC} or V _{BATT} whichever is higher.
WDI	Watchdog Input. WDI is a three level input. If WDI remains either high or low for longer than 1.6 s, (RESET) (RESET) is activated. The timer resets with each transition on the WDI line. The watchdog timer may be disabled if WDI is left floating or is connected to a high impedance three stated logic output.

ADM690A/ADM692A/ADM802L/M/ADM805L/M

Typical Performance Curves

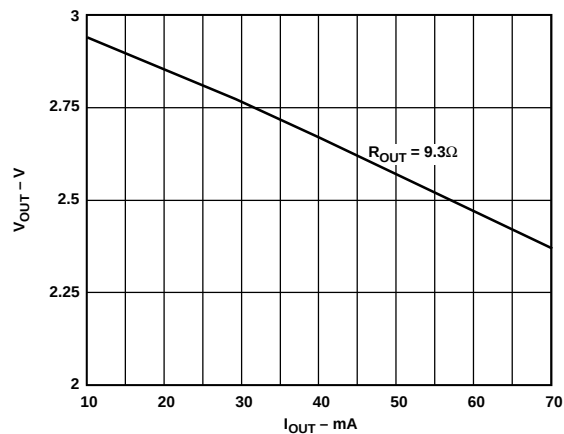


Figure 1. Output Voltage vs. Load Current in Battery Backup

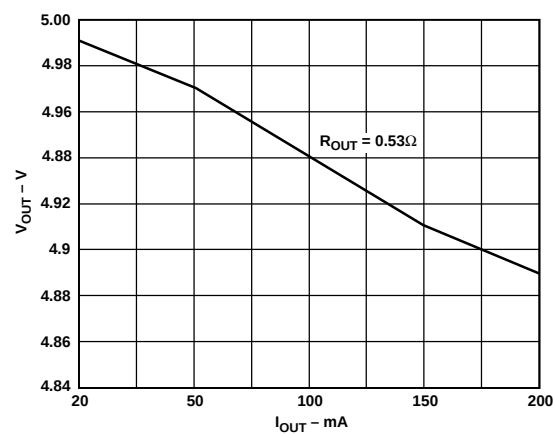


Figure 4. Output Voltage vs. Load Current in Normal Operation

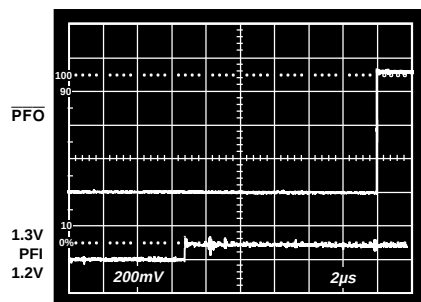


Figure 2. Power Fail Comparator Response Time L → H

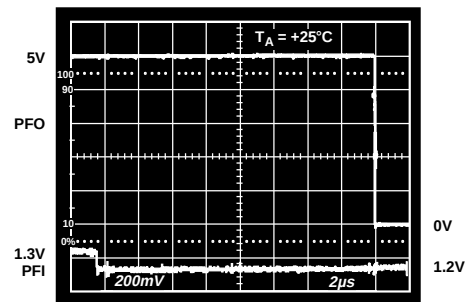


Figure 5. Power Fail Comparator Response Time H → L

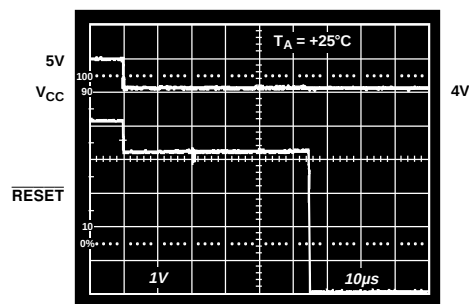


Figure 3. ADM690A $\overline{RESET}$ Response Time

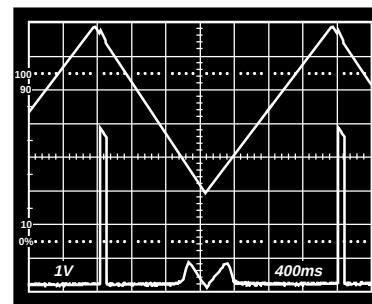


Figure 6. $\overline{RESET}$ Output Voltage vs. V_{CC}

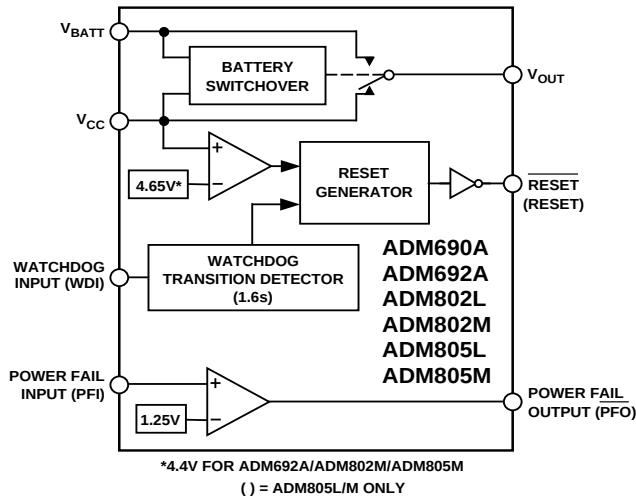


Figure 7. Functional Block Diagram

POWER FAIL $\overline{\text{RESET}}$, RESET

RESET is an active low output which provides a $\overline{\text{RESET}}$ signal to the microprocessor whenever V_{CC} is at an invalid level. When V_{CC} falls below the reset threshold, the $\overline{\text{RESET}}$ output is forced low. The nominal reset voltage threshold is 4.65 V (ADM690A/ADM802L/ADM805L or 4.4 V ADM692A/ADM802M/ADM805M).

On power-up $\overline{\text{RESET}}$ will remain low for 200 ms after V_{CC} rises above the reset threshold. This allows time for the power supply and microprocessor to stabilize. On power-down, the $\overline{\text{RESET}}$ output remains low with V_{CC} as low as 1 V. This ensures that the microprocessor is held in a stable shutdown condition.

The guaranteed minimum and maximum thresholds are as follows:
ADM690A/ADM802L/ADM805L: 4.5 V and 4.75 V

ADM692A: 4.25 V and 4.5 V.

ADM802L: 4.55 V and 4.7 V

ADM802M: 4.3 V and 4.45 V

The ADM805L and ADM805M contain an active high reset output. This is the complement of $\overline{\text{RESET}}$ and is intended for processors requiring an active high RESET signal.

The guaranteed minimum and maximum thresholds for the ADM805 are:

ADM805L: 4.5 V and 4.75 V

ADM805M: 4.25 V and 4.5 V.

Watchdog Timer $\overline{\text{RESET}}$, RESET

The watchdog timer circuit monitors the activity of the microprocessor in order to check that it is not stalled in an indefinite loop. An output line on the processor is used to toggle the Watchdog Input (WDI) line. If this line is not toggled within 1.6 seconds, a RESET pulse is generated. The watchdog timeout period restarts with each transition on the WDI pin. To ensure that the watchdog timer does not time out, either a high-to-low or low-to-high transition on the WDI pin must occur at or less than the minimum timeout period. If WDI remains permanently either high or low, reset pulses will be issued after each timeout period (1.6 s). The watchdog monitor can be deactivated by floating the Watchdog Input (WDI) or by connecting it to midsupply.

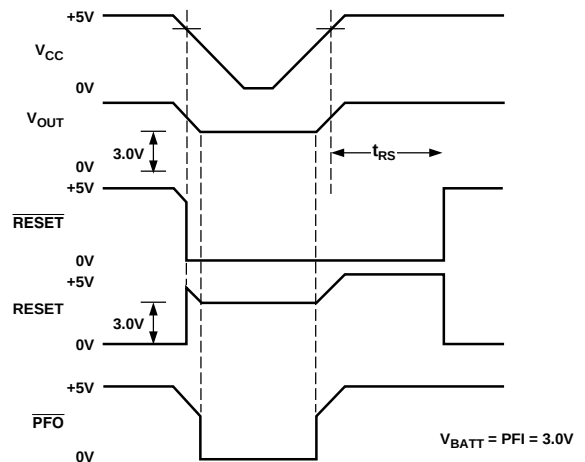


Figure 8. Timing Diagram

BATTERY SWITCHOVER SECTION

During normal operation with V_{CC} higher than the reset threshold, V_{CC} is internally switched to V_{OUT} via an internal PMOS transistor switch. This switch has a typical on-resistance of less than 1 Ω and can supply up to 100 mA at the V_{OUT} terminal. Once V_{CC} falls below the reset threshold, the higher of V_{CC} or V_{BATT} is switched to V_{OUT} . This means that V_{BATT} connects to V_{OUT} only when V_{CC} is below the reset threshold and V_{BATT} is greater than V_{CC} .

V_{OUT} is normally used to drive a RAM memory bank which may require instantaneous currents of greater than 100 mA. If this is the case, then a bypass capacitor should be connected to V_{OUT} . The capacitor will provide the peak current transients to the RAM. A capacitance value of 0.1 μF or greater may be used.

A 9 Ω MOSFET switch connects the V_{BATT} input to V_{OUT} during battery backup. This MOSFET has very low input-to-output differential (dropout voltage) at the low current levels required for battery backup of CMOS RAM or other low power CMOS circuitry. The supply current in battery backup is typically 0.05 μA .

Typically 3 V batteries are used as the backup supply. High value capacitors, either standard electrolytic or the farad size double layer capacitors, can also be used for short-term memory back up. A small charging current of typically 10 nA (0.1 μA max) flows out of the V_{BATT} terminal. This current is useful for maintaining rechargeable batteries in a fully charged condition. This extends the life of the back up battery by compensating for its self discharge current. Also note that this current poses no problem when lithium batteries are used for back up since the maximum charging current (0.1 μA) is safe for even the smallest lithium cells.

If the battery-switchover section is not used, V_{BATT} should be connected to GND and V_{OUT} should be connected to V_{CC} .

ADM690A/ADM692A/ADM802L/M/ADM805L/M

Table I. Input and Output Status in Battery Backup Mode

Signal	Status
V _{OUT}	V _{OUT} is connected to V _{BATT} via an internal PMOS switch.
RESET	Logic low.
RESET	Logic high (ADM805L, ADM805M). The open circuit output voltage is equal to V _{OUT} .
PFI	The power fail comparator is disabled
PFO	Logic low.
WDI	The watchdog timer is disabled

Power Fail Comparator

The power fail comparator is an independent comparator that may be used to monitor the input power supply. The comparator's inverting input is internally connected to a 1.25 V reference voltage. The noninverting input is available at the PFI input. This input may be used to monitor the input power supply via a resistive divider network. When the voltage on the PFI input drops below 1.25 V, the comparator output (PFO) goes low indicating a power failure. For early warning of power failure the comparator may be used to monitor the preregulator input simply by choosing an appropriate resistive divider network. The PFO output can be used to interrupt the processor so that a shutdown procedure is implemented before the power is lost.

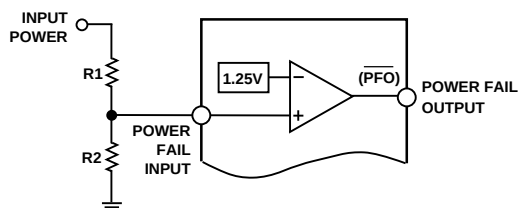
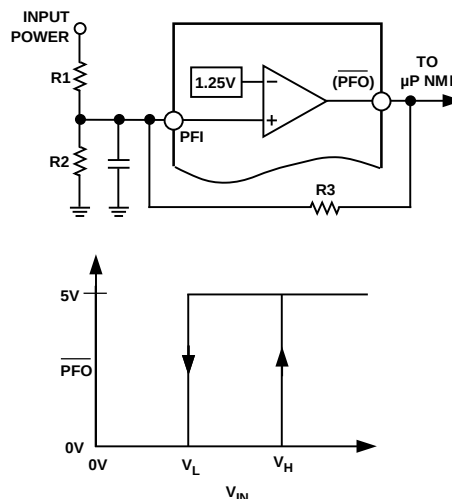


Figure 9. Power Fail Comparator

Adding Hysteresis to the Power Fail Comparator

For increased noise immunity, hysteresis may be added to the power fail comparator. Since the comparator circuit is non-inverting, hysteresis can be added simply by connecting a resistor between the PFO output and the PFI input as shown in Figure 10. When PFO is low, resistor R3 sinks current from the summing junction at the PFI pin. When PFO is high, resistor R3 sources current into the PFI summing junction. This results in differing trip levels for the comparator. Further noise immunity may be achieved by connecting a capacitor between PFI and GND.



$$V_H = 1.25 \left[1 + \left(\frac{R2 + R3}{R2 \times R3} \right) R1 \right]$$

$$V_L = 1.25 + R1 \left(\frac{1.25}{R2} - \frac{V_{CC} - 1.25}{R3} \right)$$

$$V_{MID} = 1.25 \left(\frac{R1 + R2}{R2} \right)$$

Figure 10. Adding Hysteresis to the Power Fail Comparator

TYPICAL APPLICATIONS

Figure 11 shows a typical power monitoring, battery backup application. V_{OUT} powers the CMOS RAM. Under normal operating conditions with V_{CC} present, V_{OUT} is internally connected to V_{CC}. If a power failure occurs, V_{CC} will decay and V_{OUT} will be switched to V_{BATT} thereby maintaining power for the CMOS RAM. A RESET pulse is also generated when V_{CC} falls below the reset threshold.

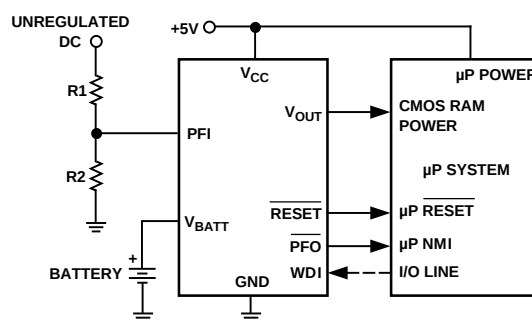


Figure 11. Typical Application Circuit

The watchdog timer input (WDI) monitors an I/O line from the μP system. This line must be toggled once every 1.6 seconds to verify correct software execution. Failure to toggle the line indicates that the μP system is not correctly executing its program and may be tied up in an endless loop. If this happens, a reset pulse is generated to initialize the processor.

If the watchdog timer is not needed, the WDI input should be left floating.

The Power Fail Input, PFI, monitors the input power supply via a resistive divider network. The voltage on the PFI input is compared with a precision 1.25 V internal reference. If the input voltage drops below 1.25 V, a power fail output (PFO) signal is generated. This warns of an impending power failure and may be used to interrupt the processor so that the system may be shut down in an orderly fashion. The resistors in the sensing network are ratioed to give the desired power fail threshold voltage V_T .

$$V_T = (1.25 R1/R2) + 1.25 V$$

$$R1/R2 = (V_T/1.25) - 1$$

Alternate Watchdog Input Drive Circuits

The watchdog feature can be enabled and disabled under program control by driving WDI with a 3-state buffer. When three-stated, the WDI input will float thereby disabling the watchdog timer.

High Capacity Backup Capacitors

High capacity (0.1 μ F or greater) capacitors may be used as a backup power source. A typical application is shown in Figure 12.

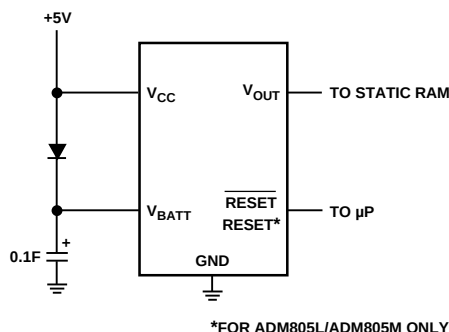


Figure 12. High Capacity Capacitor

Operation Without a Backup Supply

If a backup power source is not used, V_{BATT} should be connected to GND and V_{OUT} should be connected to V_{CC} .

Replacing the Backup Battery

The backup battery may be replaced without any danger of spurious resetting when V_{CC} is present. Since V_{CC} is above the reset threshold, a reset will not occur even if V_{BATT} is floating while a replacement battery is being inserted. This differs from older generation products where leakage currents flowing out V_{BATT} could cause spurious resetting during battery replacement.

μPs With Bidirectional RESET

In order to prevent contention for microprocessors with a bidirectional reset line, a current limiting resistor should be inserted between the ADM69xA/ADM80xx RESET output pin and the μ P reset pin. This will limit the current to a safe level if there are conflicting output reset levels. A suitable resistor value is 4.7 k Ω . If the reset output is required for other uses, then it should be buffered as shown in Figure 13.

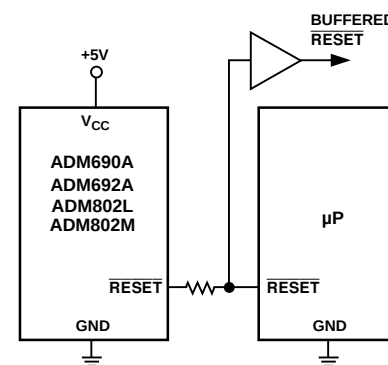


Figure 13. Bidirectional Reset

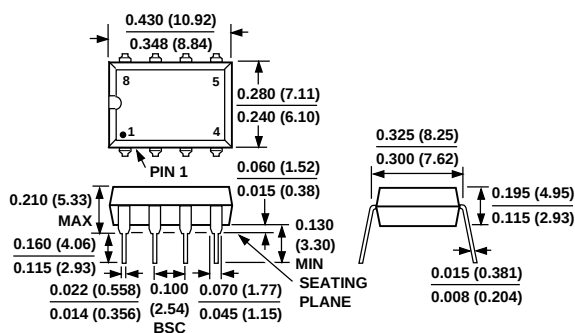
ADM690A/ADM692A/ADM802L/M/ADM805L/M

OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

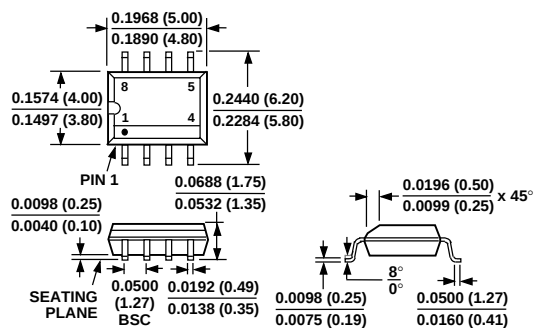
8-Pin Plastic DIP

(N-8)



8-Lead SOIC

(SO-8)



8-Lead MicroSOIC

(RM-8)

