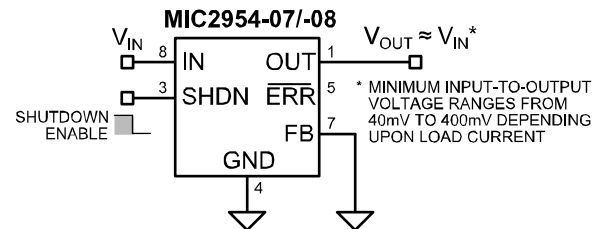
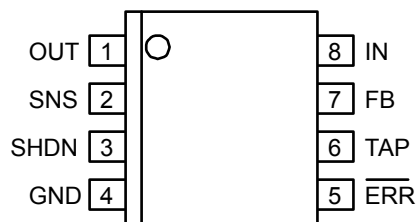


### 5V or 3V Selectable Regulator with Shutdown

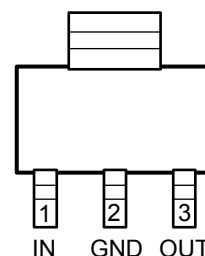


### Wide Input-Voltage-Range Current Limiter

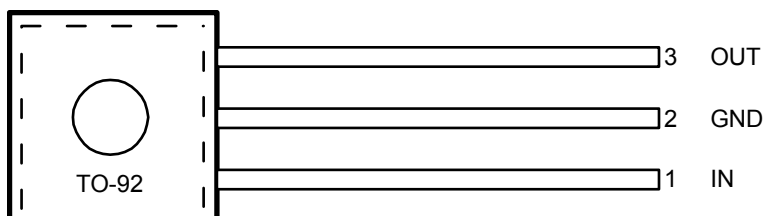
## Pin Configuration



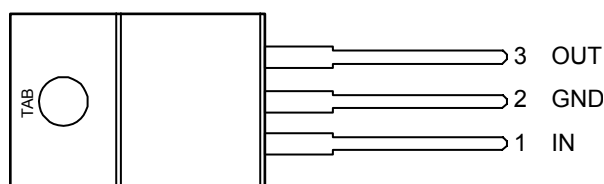
8-Pin SOIC (M)



3-Pin SOT-223 (S)



3-Pin TO-92 (Z)



3-Pin TO-220 (T)

## Pin Description

Pin No. SOIC-8	Pin No. SOT-223	Pin No. TO-92	Pin No. TO-220	Pin Name	Pin Name
8	1	1	1	IN	Supply Input.
4	2, TAB	2	2	GND	Ground.
1	3	3	3	OUT	Regulator Output.
2				SNS	Sense (Input): Output-sense-voltage end of internal resistive divider. Connect to OUT ( $V_{OUT} = 5V$ ) for fixed 5V operation; also see TAP. Not used in adjustable configuration.
3				SHDN	Shutdown (Input): Active-low input enables regulator. (Low = enable; high = shutdown.)
5				/ERR	Error Flag (Output): Open collector (active-low) output. Active state indicates an output ( $V_{OUT}$ ) undervoltage condition. (Low = error, floating = normal.)
6				TAP	Divider Tap (Output): Resistive voltage divider tap. With 5V applied to SNS, VTAP is approximately 1.23V. Connect to FB for 5V operation. Not used in adjustable configuration.
7				FB	Feedback (Input): Error amplifier input. Compared to internal 1.23V reference. Connect to external voltage divider for adjustable operation or internal voltage divider (TAP) for 5V operation (see SNS, TAP).

**Absolute Maximum Ratings<sup>(1)</sup>**

Supply Voltage ( $V_{IN}$ )	–20V to +60V
Feedback Voltage ( $V_{FB}$ ) <sup>(14, 15)</sup>	–1.5V to +26V
Shutdown Input Voltage ( $V_{SHDN}$ )	–0.3V to +30V
Error Output Voltage ( $V_{ERR}$ )	–0.3V to +30V
Power Dissipation ( $P_D$ ) <sup>(4)</sup>	Internally Limited
Lead Temperature (soldering, 5 sec.)	260°C
Storage Temperature ( $T_s$ )	–65°C to +150°C
ESD Rating <sup>(3)</sup>	

**Operating Ratings<sup>(2)</sup>**

Supply Voltage ( $V_{IN}$ )	+2.0V to +30V
Junction Temperature ( $T_J$ )	–40°C to +125°C
Package Thermal Resistance ( $\theta_{JC}, \theta_{JA}$ ) <sup>(5)</sup>	

**Electrical Characteristics**

MIC2954-07/08:  $V_{FB} = V_{TAP}$ ;  $V_{SNS} = V_{OUT}$ ;  $V_{SHDN} \leq 0.6V$ . All versions:  $V_{IN} = 6V$ ;  $I_L = 1mA$ ;  $C_L = 2.2\mu F$ ;  $T_J = 25^\circ C$ , **bold** values indicate  $-40^\circ C \leq T_J \leq +125^\circ C$ ; **Note 8**; unless noted.

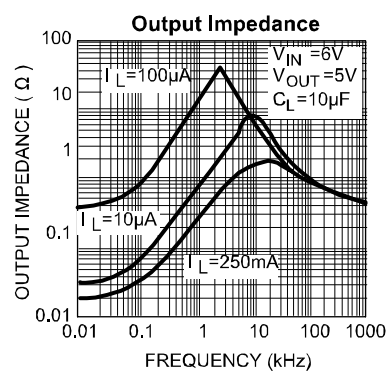
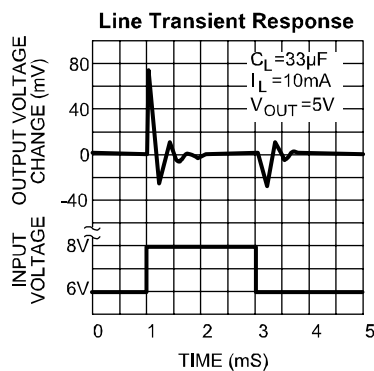
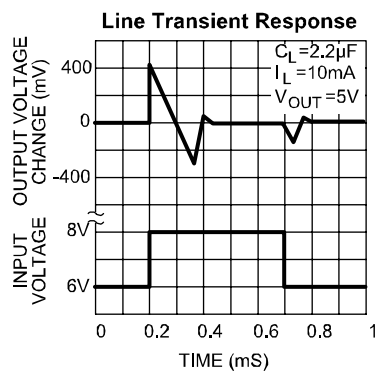
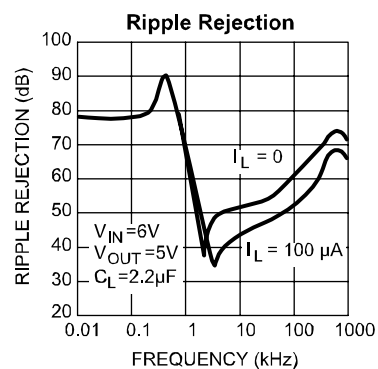
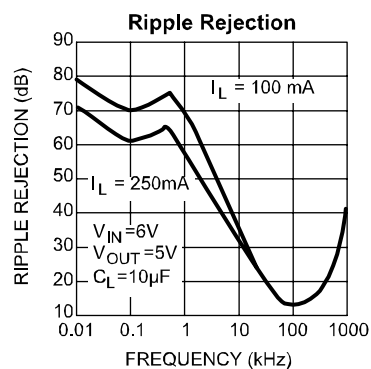
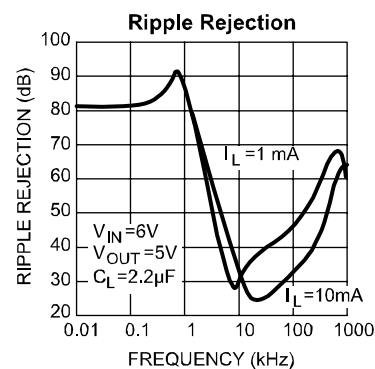
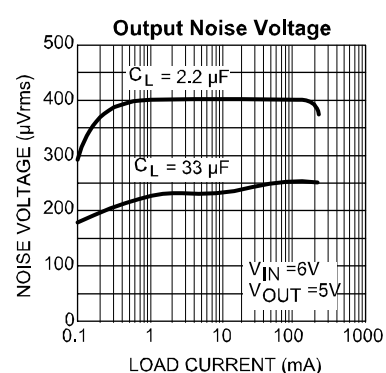
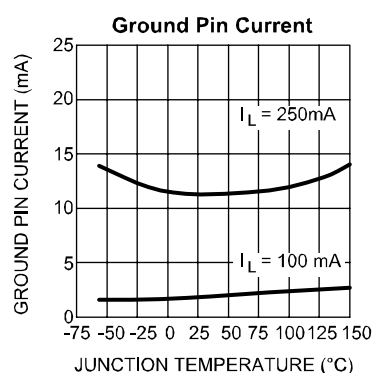
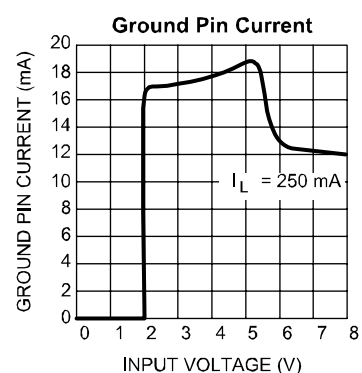
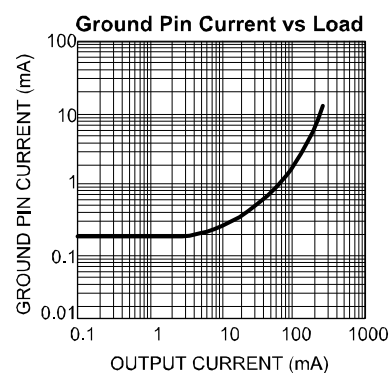
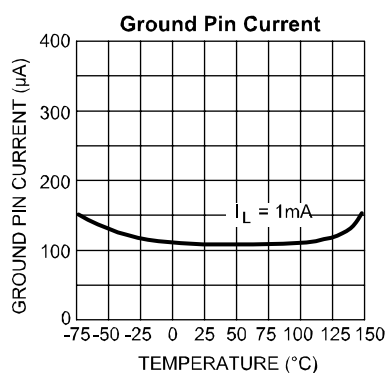
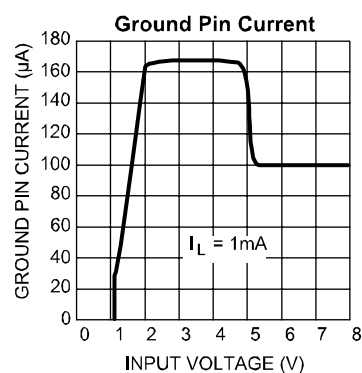
Symbol	Parameter	Condition	Min	Typ	Max	Units
$V_{OUT}$	Output Voltage	MIC2954-02/-07 ( $\pm 0.5\%$ )	4.975 <b>4.940</b>	5.000	5.025 <b>5.060</b>	V V
		MIC2954-02/-07 ( $\pm 0.5\%$ ), $1mA \leq I_L \leq 250mA$	<b>4.930</b>	<b>5.000</b>	<b>5.070</b>	V
		MIC2954-03/-08 ( $\pm 1\%$ )	4.950 <b>4.900</b>	5.000	5.050 <b>5.100</b>	V V
		MIC2954-03/-08 ( $\pm 1\%$ ), $1mA \leq I_L \leq 250mA$	<b>4.880</b>	<b>5.000</b>	<b>5.120</b>	V
$\Delta V_{OUT}/\Delta T$	Output Voltage Temperature Coefficient, <b>Note 6</b>	MIC2954-02/-07 ( $\pm 0.5\%$ )		<b>20</b>	<b>100</b>	ppm/°C
		MIC2954-03/-08 ( $\pm 1\%$ )		<b>20</b>	<b>150</b>	ppm/°C
$\Delta V_{OUT}/V_{OUT}$	Line Regulation, <b>Note 7</b>	MIC2954-02/-07 ( $\pm 0.5\%$ ), $V_{IN} = 6V$ to 26V		0.03	0.10 <b>0.20</b>	%/V %/V
		MIC2954-03/-08 ( $\pm 1\%$ ), $V_{IN} = 6V$ to 26V		0.03	0.20 <b>0.40</b>	%/V %/V
$\Delta V_{OUT}/V_{OUT}$	Load Regulation, <b>Note 8</b>	MIC2954-02/-07 ( $\pm 0.5\%$ ), $I_L = 1$ to 250mA		0.04	0.16 0.20	%/V %/V
		MIC2954-03/-08 ( $\pm 1\%$ ), $I_L = 1$ to 250mA		0.04	0.20 <b>0.30</b>	%/V %/V
$V_{IN} - V_{OUT}$	Dropout Voltage, <b>Note 9</b>	$I_L = 1mA$		60	100 <b>150</b>	mV mV
		$I_L = 50mA$		220	250 <b>420</b>	mV mV
		$I_L = 100mA$		250	300 <b>450</b>	mV mV
		$I_L = 250mA$		375	450 <b>600</b>	mV mV
$I_{GND}$	Ground Pin Current, <b>Note 10</b>	$I_L = 1mA$		140	200 <b>300</b>	$\mu A$ $\mu A$
		$I_L = 50mA$		0.5	1 <b>2</b>	mA mA
		$I_L = 100mA$		1.7	2.5 <b>3.5</b>	mA mA
		$I_L = 250mA$		5	9 <b>12</b>	mA mA
$I_{GND(DO)}$	Ground Pin Current at Dropout, <b>Note 10</b>	$V_{IN} = 4.5V$		180	<b>300</b>	$\mu A$

Symbol	Parameter	Condition	Min	Typ	Max	Units
$I_{LIMIT}$	Current Limit, <b>Note 11</b>	$V_{OUT} = 0V$			750 <b>800</b>	mA mA
$\Delta V_{OUT}/\Delta P_D$	Thermal Regulation, <b>Note 12</b>			0.05	0.2	%/W
$e_n$	Output Noise Voltage (10Hz to 100kHz)	$I_L = 100mA, C_L = 2.2\mu F$		400		$\mu V_{RMS}$
		$I_L = 100mA, C_L = 33\mu F$		260		$\mu V_{RMS}$
	Reference Voltage	MIC2954-02/-07 ( $\pm 0.5\%$ )	1.220 <b>1.200</b>	1.235	1.250 <b>1.260</b>	V V
		MIC2954-03/-08 ( $\pm 1\%$ )	1.210 <b>1.200</b>	1.235	1.260 <b>1.270</b>	V V
	Reference Voltage	MIC2954-02/-07 ( $\pm 0.5\%$ ), <b>Note 13</b>	<b>1.190</b>		<b>1.270</b>	V
		MIC2954-03/-08 ( $\pm 1\%$ ), <b>Note 13</b>	<b>1.185</b>		<b>1.285</b>	V
	Feedback Pin Bias Current			20	40 <b>60</b>	nA nA
	Reference Voltage Temperature Coefficient, <b>Note 12</b>	MIC2954-02/-07 ( $\pm 0.5\%$ )		20		ppm/ $^{\circ}C$
		MIC2954-03/-08 ( $\pm 1\%$ )		50		ppm/ $^{\circ}C$
	Feedback Pin Bias Current Temperature Coefficient			0.1	40	nA/ $^{\circ}C$
<b>Error Comparator</b>						
	Output Leakage Current	$V_{OH} = 30V$		0.01	1.00 <b>2.00</b>	$\mu A$ $\mu A$
	Output Low Voltage	$V_{IN} = 4.5V, I_{OL} = 400\mu A$		150	250 <b>400</b>	mV mV
	Upper Threshold Voltage	<b>Note 14</b>		60	40 <b>25</b>	mV mV
	Lower Threshold Voltage	<b>Note 14</b>		75	95 <b>140</b>	mV mV
	Hysteresis	<b>Note 14</b>		15		mV mV
<b>Shutdown Input</b>						
	Input Logic Voltage	low (on)		1.3	<b>0.7</b>	V
		high (off)	2.0			V
	Shutdown Pin Input Current	$V_{SHDN} = 2.4V$		30	50 <b>100</b>	$\mu A$ $\mu A$
		$V_{SHDN} = 30V$		450	600 <b>750</b>	$\mu A$ $\mu A$
	Regular Output Current in Shutdown	<b>Note 15</b>		3	10 <b>20</b>	$\mu A$ $\mu A$

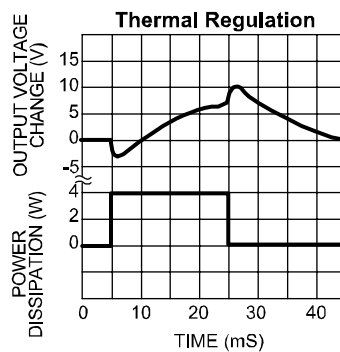
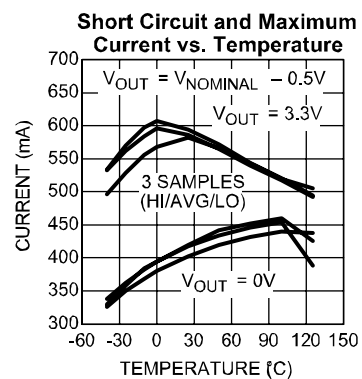
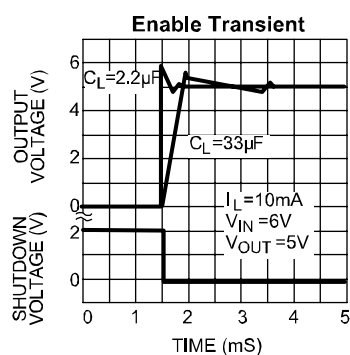
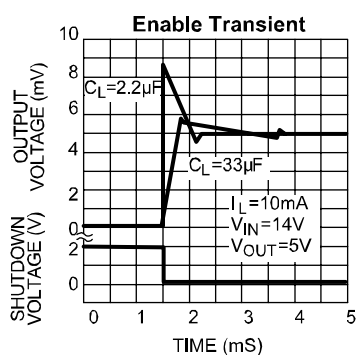
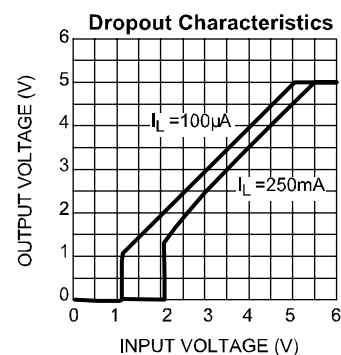
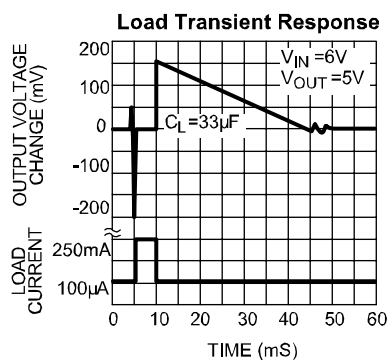
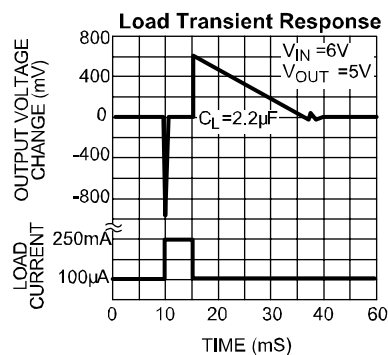
**Notes:**

1. Exceeding the absolute maximum rating may damage the device.
2. The device is not guaranteed to function outside its operating rating.
3. Devices are ESD sensitive. Handling precautions recommended.
4.  $P_{D(max)} = (T_{J(max)} - T_A) \div \theta_{JC}$ . Exceeding  $T_{J(max)}$  will cause thermal shutdown.
5. Thermal resistance ( $\theta_{JC}$ ) of the TO-220 package is 2.5°C/W, and 15°C/W for the SOT-223. Thermal resistance ( $\theta_{JC}$ ) of the TO-92 package is 180°C/W with 0.4" leads and 160°C/W with 0.25" leads. Thermal resistance ( $\theta_{JA}$ ) of the SOP-8 is 160°C/W mounted on a printed circuit board (See "Application Information: Thermal Calculation").
6. Output voltage temperature coefficient is defined as the worst case voltage change divide by the total temperature range.
7. Line regulation for the MIC2954 is tested at 125°C for  $I_L = 1\text{mA}$ . For  $I_L = 100\mu\text{A}$  and  $T_J = 125^\circ\text{C}$ , line regulation is guaranteed by design to 0.2%. See "Typical Characteristics" for line regulation versus temperature and load current.
8. Regulation is measured at constant junction temperature using low duty cycle pulse testing. Changes in output voltage due to heating effects are covered by the thermal regulation specification.
9. Dropout Voltage is defined as the input to output differential at which the output voltage drops 100 mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2 V (2.3V over temperature) must be taken into account.
10. Ground pin current is the regulator quiescent current. The total current drawn from the source is the sum of the load current plus the ground pin current.
11. The MIC2954 features fold-back current limiting. The short circuit ( $V_{OUT} = 0\text{V}$ ) current limit is less than the maximum current with normal output voltage.
12. Thermal regulation is defined as the change in output voltage at a time  $t$  after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 200mA load pulse at  $V_{IN} = 20\text{V}$  (a 4W pulse) for  $t = 10\text{ms}$ .
13.  $V_{REF} \leq V_{OUT} \leq (V_{IN} - 1\text{V})$ ,  $2.3\text{V} \leq V_{IN} \leq 30\text{V}$ ,  $100\mu\text{A} < I_L \leq 250\text{mA}$ ,  $T_J \leq T_{J(max)}$ .
14. Comparator thresholds are expressed in terms of a voltage differential at the FB pin below the nominal reference voltage measured at 6Vinput. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain  $= V_{OUT}/V_{REF} = (R1 + R2)/R2$ . For example, at a programmed output voltage of 5V, the error output is guaranteed to go low when the output drops by  $95\text{mV} \times 5\text{V}/1.235\text{V} = 384\text{mV}$ . Thresholds remain constant as a percent of  $V_{OUT}$  as  $V_{OUT}$  is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.
15.  $V_{SHDN} \geq 2\text{V}$ ,  $V_{IN} \leq 30\text{V}$ ,  $V_{OUT} = 0$ , with the FB pin connected to TAP.
16. When used in dual supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.
17. Maximum positive supply voltage of 60V must be of limited duration (<10ms) and duty cycle (<1%). The maximum continuous supply voltage is 30V.

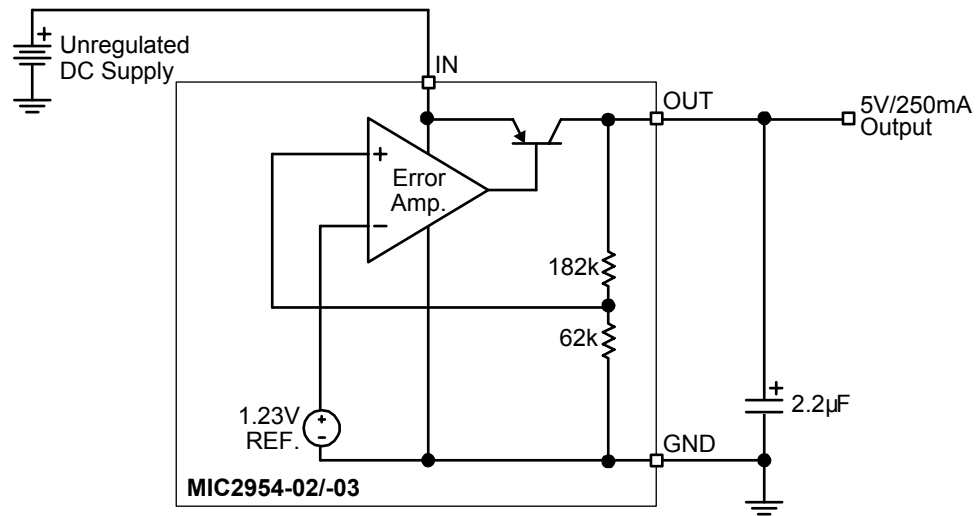
## Typical Characteristics



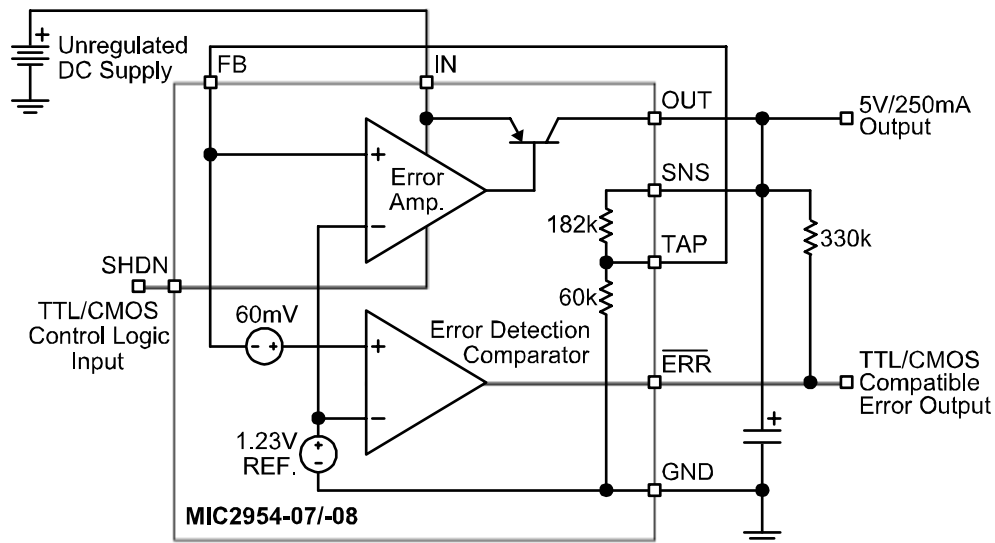
## Typical Characteristics



## Functional Diagrams



**MIC2954-02 and MIC2954-03**



**MIC2954-07 and MIC2954-08**



## Application Information

### External Capacitors

A 2.2 $\mu$ F (or greater) capacitor is required between the MIC2954 output and ground to prevent oscillations due to instability. Most types of tantalum or aluminum electrolytics will be adequate; film types will work, but are costly and therefore not recommended. Many aluminum electrolytics have electrolytes that freeze at about  $-30^{\circ}\text{C}$ , so solid tantalums are recommended for operation below  $-25^{\circ}\text{C}$ . The important parameters of the capacitor are an effective series resistance of about 5 $\Omega$  or less and a resonant frequency above 500kHz. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.5 $\mu$ F for current below 10mA or 0.15 $\mu$ F for currents below 1mA. Adjusting the MIC2954-07/-08 to voltages below 5V runs the error amplifier at lower gains so that more output capacitance is needed. For the worst-case situation of a 250mA load at 1.23V output (output shorted to feedback) a 5 $\mu$ F (or greater) capacitor should be used.

The MIC2954 will remain in regulation with a minimum load of 1mA. When setting the output voltage of the MIC2954-07/-08 version with external resistors, the current through these resistors may be included as a portion of the minimum load.

A 0.1 $\mu$ F capacitor should be placed from the MIC2954 input to ground if there is more than 10 inches of wire between the input and the ac filter capacitor or if a battery is used as the input.

### Error Detection Comparator Output (MIC2954-07/-08)

A logic-low output will be produced by the comparator whenever the MIC2954-07/-08 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 60mV divided by the 1.235V reference voltage. (Refer to the block diagram on page 1). This trip level remains "5% below normal" regardless of the programmed output voltage of the MIC2954-07/-08. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage, current limiting, or thermal limiting.

Figure 1 is a timing diagram depicting the /ERR signal and the regulated output voltage as the MIC2954-07/-08 input is ramped up and down. The /ERR signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which  $V_{\text{OUT}} = 4.75$ ). Since the MIC2954-07/-08's dropout voltage is load-dependent (see curve in "Typical Characteristics"), the input voltage trip point (about 5V) will vary with the load current. The output voltage trip point (approximately 4.75V) does not vary with load.

The error comparator has an open-collector output which requires an external pull-up resistor. Depending on system requirements, this resistor may be returned to the 5V output or some other supply voltage. In determining a value for this resistor, note that while the output is rated to sink 400 $\mu$ A, this sink current adds to battery drain in a low battery condition. Suggested values range from 100k to 1M $\Omega$ . The resistor is not required if this output is unused.

### Programming the Output Voltage (MIC2954-07/-08)

The MIC2954-07/-08 may be pin-strapped for 5V using its internal voltage divider by tying pin 1 (OUT) to pin 2 (SNS) and pin 7 (FB) to pin 6 (TAP). Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. An external pair of resistors is required, as shown in Figure 3.

The complete equation for the output voltage is:

$$V_{\text{OUT}} = V_{\text{REF}} \left( 1 + \frac{R1}{R2} \right) + I_{\text{FB}} R1$$

where:

$V_{\text{REF}}$  = nominal 1.235V reference voltage

$I_{\text{FB}}$  = nominal FB pin bias current ( $-20\text{nA}$ )

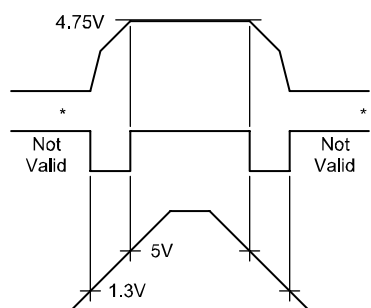
The minimum recommended load current of 1 $\mu$ A forces an upper limit of 1.2M $\Omega$  on the value of R2, if the regulator must work with no load (a condition often found in CMOS in standby), IFB will produce a 2% typical error in  $V_{\text{OUT}}$  which may be eliminated at room temperature by trimming R1. For better accuracy, choosing R2 = 100k reduces this error to 0.17% while increasing the resistor program current to 12 $\mu$ A. Since the MIC2954-07/-08 typically draws 60 $\mu$ A at no load with pin 2 (SNS) open-circuited, this is a negligible addition.

### Reducing Output Noise

In reference applications it may be advantageous to reduce the ac noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is relatively inefficient, as increasing the capacitor from 1 $\mu$ F to 220 $\mu$ F only decreases the noise from 430 $\mu$ V to 160 $\mu$ V<sub>RMS</sub> for a 100kHz bandwidth at 5V output. Noise can be reduced fourfold by a bypass capacitor across R1, since it reduces the high frequency gain from 4 to unity.

Pick:

$$C_{\text{BYPASS}} \cong \frac{1}{2\pi R1 \times 200\text{Hz}}$$



\* See Application Information

**Figure 1. /Error Output Timing**

or about  $0.01\mu\text{F}$ . When doing this, the output capacitor must be increased to  $3.3\mu\text{F}$  to maintain stability. These changes reduce the output noise from  $430\mu\text{V}$  to  $100\mu\text{V}_{\text{rms}}$  for a  $100\text{kHz}$  bandwidth at  $5\text{V}$  output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

### Automotive Applications

The MIC2954 is ideally suited for automotive applications for a variety of reasons. It will operate over a wide range of input voltages with very low dropout voltages ( $40\text{mV}$  at light loads), and very low quiescent currents ( $75\mu\text{A}$  typical). These features are necessary for use in battery powered systems, such as automobiles. It is a “bulletproof” device with the ability to survive both reverse battery (negative transients up to  $20\text{V}$  below ground), and load dump (positive transients up to  $60\text{V}$ ) conditions. A wide operating temperature range with low temperature coefficients is yet another reason to use these versatile regulators in automotive designs.

## Thermal Calculations

### Layout Considerations

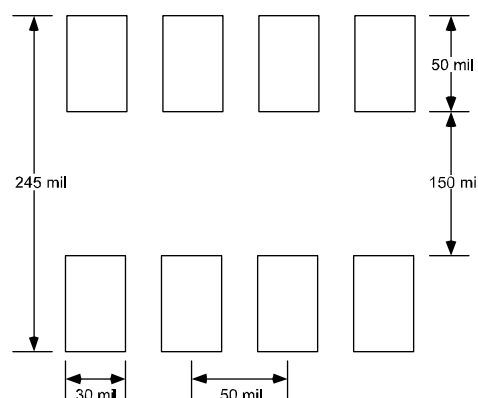
The MIC2954-07BM/-08BM (8-pin surface-mount package) has the following thermal characteristics when mounted on a single layer copper-clad printed circuit board.

PC Board Dielectric Material	$\theta_{\text{JA}}$
FR4	$160^{\circ}\text{C/W}$
Ceramic	$120^{\circ}\text{C/W}$

Multilayer boards having a ground plane, wide traces near the pads, and large supply bus lines provide better thermal conductivity.

Our calculations will use the “worst case” value of  $160^{\circ}\text{C/W}$ , which assumes no ground plane, minimum trace widths, and a FR4 material board.

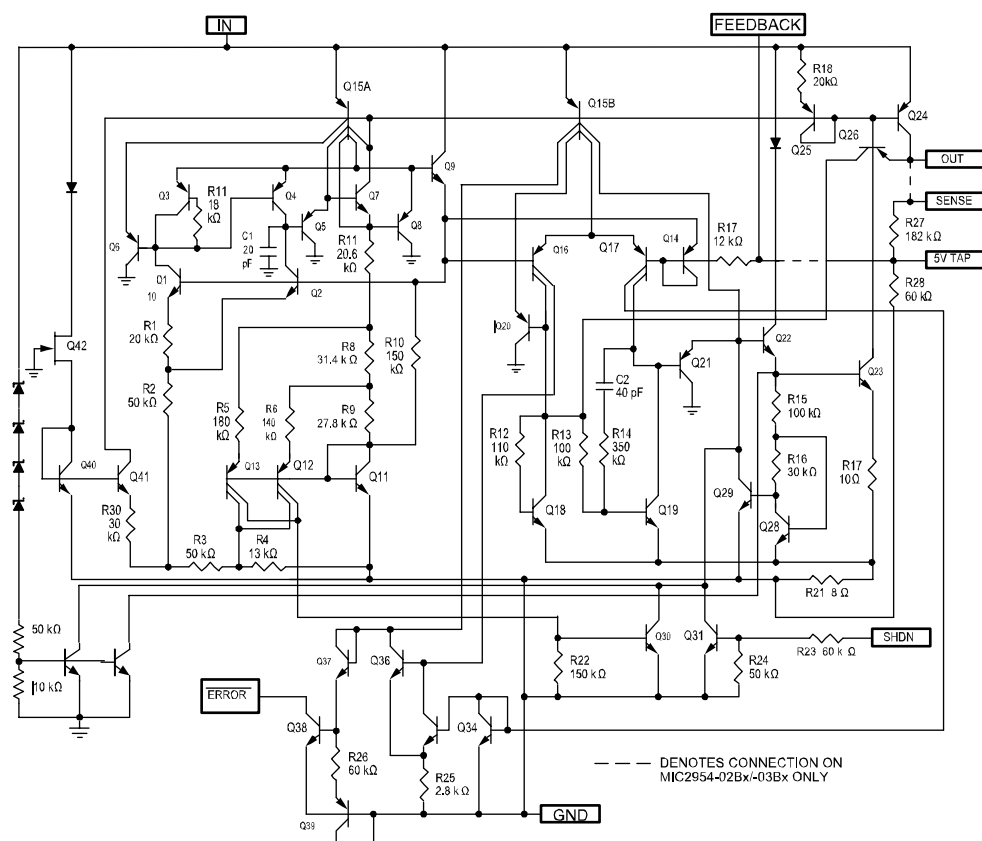
### Pad Layout (minimum recommended geometry)



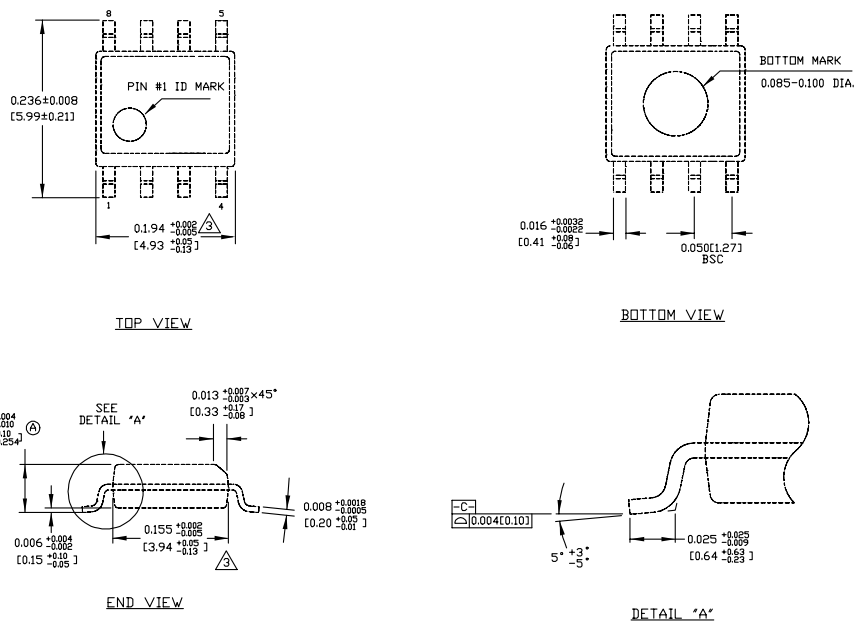
### Nominal Power Dissipation and Die Temperature

The MIC2954-07BM/-08BM at a  $55^{\circ}\text{C}$  ambient temperature will operate reliably at up to  $440\text{mW}$  power dissipation when mounted in the “worst case” manner described above. This power level is equivalent to a die temperature of  $125^{\circ}\text{C}$ , the recommended maximum temperature for nonmilitary grade silicon integrated circuits.

## Schematic Diagram

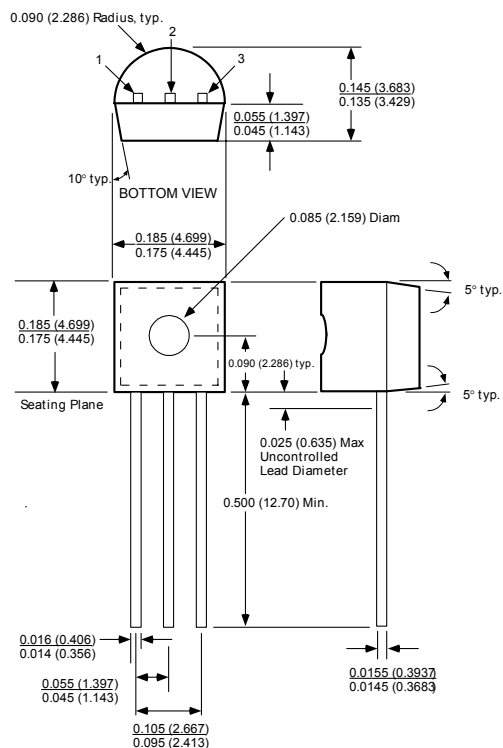


## Package Information

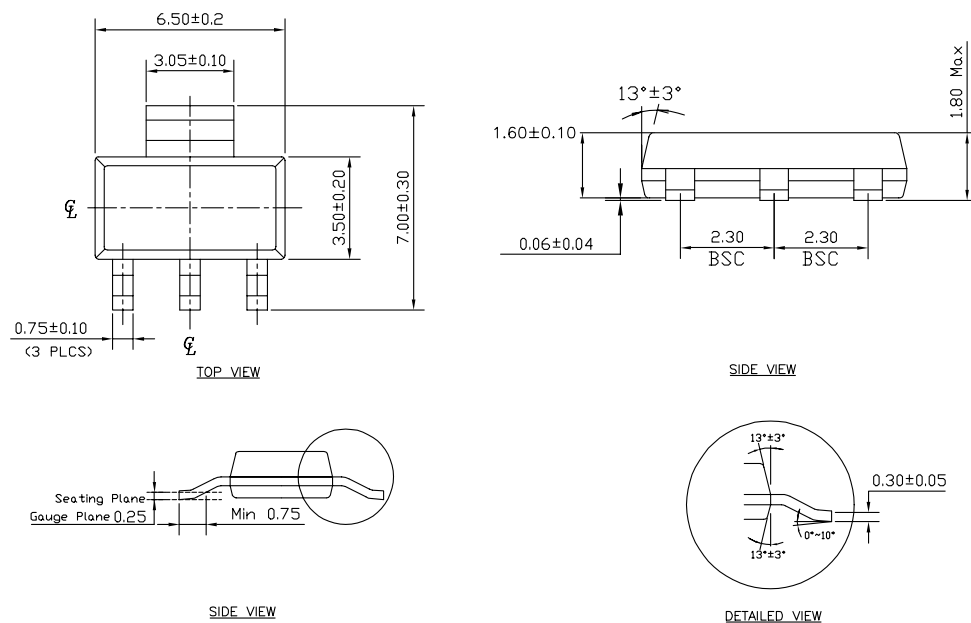


- NOTES:**
1. DIMENSIONS ARE IN INCHES[MM].
  2. CONTROLLING DIMENSION: INCHES.
  3. DIMENSION DOES NOT INCLUDE MOLD FLASH OR PROTRUSIONS, EITHER OF WHICH SHALL NOT EXCEED 0.010[0.25] PER SIDE.

### 8-Pin SOIC (M)



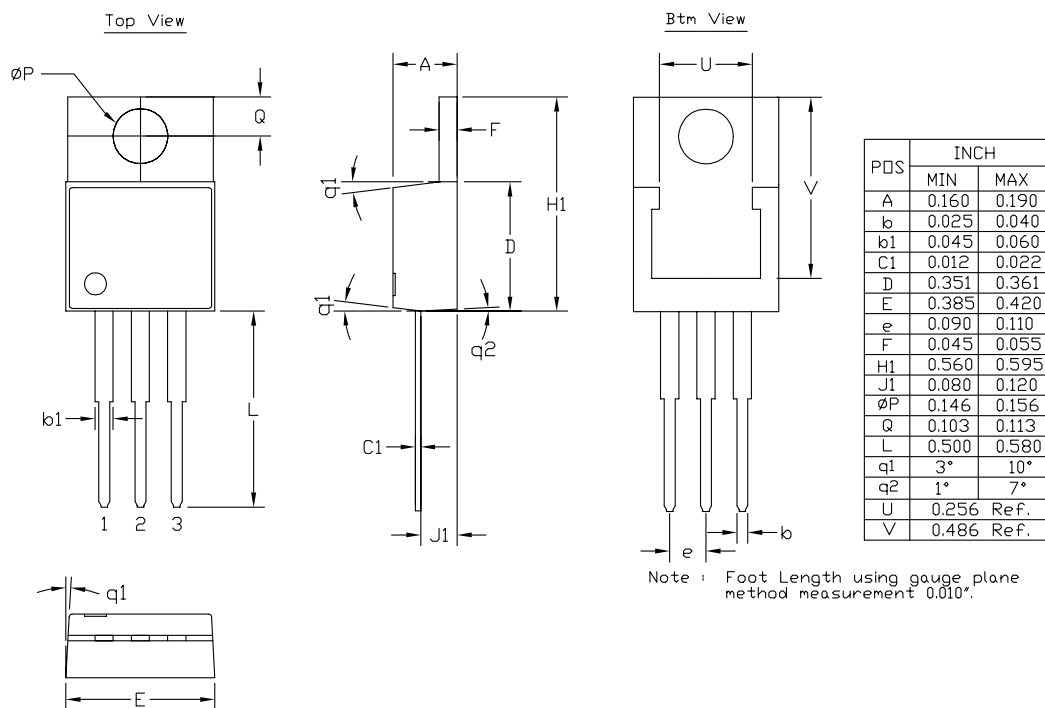
### 3-Pin TO-92 (Z)



## NOTE:

1. Dimensions and tolerances are as per ANSI Y14.5M, 1982.
2. Controlling dimension: Millimeters.
3. Dimensions are exclusive of mold flash and gate burr.
4. All specification comply to Jedec spec T0261 Issue C.

## 3-Pin SOT-223 (S)



## 3-Pin TO-220 (T)

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