#### **ABSOLUTE MAXIMUM RATINGS**

V <sub>CC</sub> to GND	0.3V to +6V
LNAIN Input Power ( $50\Omega$ source)	+10dBm
LO Input Power ( $50\Omega$ source)	+10dBm
MIXIN Input Power (50 $\Omega$ source)	+10dBm
IFOUT+, IFOUT- to GND	0.3V to +6V
LNAOUT to GND	0.3V to +6V
GAIN, SHDN to GND	0.3V to $(V_{CC} + 0.3V)$

Continuous Power Dissipation (T <sub>A</sub> = +70°C)	
16-Pin QSOP (derate 8.3mW/°C above +70	0°C)667mW
Junction Temperature	+150°C
Operating Temperature Range	40°C to +85°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +2.7V \text{ to } +5.5V, V \overline{SHDN} = +2V, V_{GAIN} = +2V, LNAIN = LNAOUT = MIXIN = LO = unconnected, IFOUT+ = IFOUT- = V_{CC}, T_{A} = -40^{\circ}C$  to +85°C. Typical values are at  $T_{A} = +25^{\circ}C$  and  $V_{CC} = +3V$ , unless otherwise noted.)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS	
Supply Voltage Range		2.7		5.5	V	
Operating Supply Current	GAIN = Vcc (Note 1)		8.5	14.1	mA	
	GAIN = GND (Note 1)		3.8	6.4		
Shutdown Supply Current	SHDN = GND (Note 1)		0.1	1.0	μA	
Logic Input Voltage High	GAIN, SHDN	2.0			V	
Logic Input Voltage Low	GAIN, SHDN			0.5	V	
Logic Input Current	VSHDN = VGAIN = 0 to 5.5V (Note 1)		±0.01	±1	μΑ	

#### **AC ELECTRICAL CHARACTERISTICS**

(MAX2685 EV kit,  $V_{CC} = V_{\overline{SHDN}} = +3V$ ,  $f_{LNAIN} = f_{MIXIN} = 880MHz$ ,  $f_{LO} = 960MHz$ ,  $P_{LNAIN} = -30dBm$ ,  $P_{LO} = -8dBm$ ,  $P_{MIXIN} = -25dBm$ , differential IFOUT operation,  $Z_0 = 50\Omega$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS				
LOW-NOISE AMPLIFIER (LNA)									
RF Frequency Range (Note 2)		800		1000	MHz				
LNIA Coin	GAIN = V <sub>CC</sub> (Note 1)	13	15	16.2	-ID				
LNA Gain	GAIN = GND (Note 1)	-14	-12	-10.3	dB				
LNA Gain Variation over Temperature	GAIN = VCC, TA = TMIN to TMAX (Note 3)		0.9	1.6	dB				
LNA Noise Figure	GAIN = V <sub>CC</sub>	1.4			d۵				
LNA Noise Figure	GAIN = GND		12.2		dB				
LNA Input ID2	GAIN = V <sub>CC</sub> (Note 4)		-4.1		dBm				
LNA Input IP3	GAIN = GND (Note 5) +		+16.2		dbiii				
LNA Input 1dB Compression	GAIN = VCC		-18.4		dBm				
LNAOUT Port Return Loss	GAIN = V <sub>CC</sub> -18.6  GAIN = GND -11.3			- dB					
LIVACOT FOIL RETUITI LOSS									
DOWNCONVERTER MIXER	,	-							
RF Frequency Range (Note 2)		800		1000	MHz				
Mixer Conversion Gain	GAIN = Vcc (Note 1)	4.7	6.1	7.0	dB				
IVIIAEI COTIVEISIOTI GAITI	GAIN = GND (Note 1)	2.5	4.6	6.0					

### AC ELECTRICAL CHARACTERISTICS (continued)

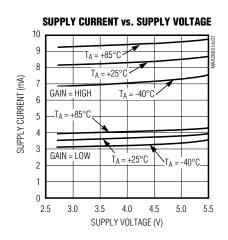
(MAX2685 EV kit,  $V_{CC} = V_{\overline{SHDN}} = +3V$ ,  $f_{LNAIN} = f_{MIXIN} = 880MHz$ ,  $f_{LO} = 960MHz$ ,  $P_{LNAIN} = -30dBm$ ,  $P_{LO} = -8dBm$ ,  $P_{MIXIN} = -25dBm$ , differential IFOUT operation,  $Z_0 = 50\Omega$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

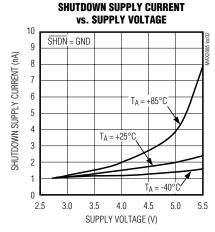
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS	
Mixer Conversion Gain Variation over Temperature	GAIN = V <sub>CC</sub> , T <sub>A</sub> = T <sub>MIN</sub> to T <sub>MAX</sub> (Note 3)		1	2	dB	
Miyor Noigo Eiguro (SSP)	GAIN = V <sub>CC</sub>		13		dВ	
Mixer Noise Figure (SSB)	GAIN = GND		12.1		dB	
Mixer Input IP3 (Note 6)	GAIN = V <sub>CC</sub>		7	dBm		
	GAIN = GND		-1.5		— ubiii	
LO Port Return Loss			11		dB	
LO-to-LNAIN Isolation	SHDN = V <sub>CC</sub> or GND		53		dB	
LO-to-MIXIN Isolation			31		dB	
LNAOUT-to-MIXIN Isolation			28		dB	
OVERALL SYSTEM					•	
Receiver Enable Time	(Note 7)		0.8		μs	

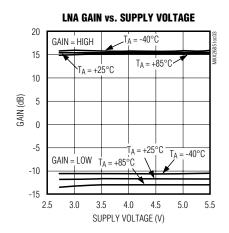
- **Note 1:** Performance at temperatures greater than or equal to +25°C are guaranteed by production test; performance at temperatures less than +25°C are guaranteed by design and characterization.
- **Note 2:** This is the recommended operating frequency range.
- Note 3: Maximum and minimum limits are guaranteed by design and device characterization and are not production tested.
- Note 4: Two tones at 880MHz and 880.1MHz, -30dBm per tone.
- Note 5: Two tones at 880MHz and 880.1MHz, -10dBm per tone.
- Note 6: Two tones at 880MHz and 880.1MHz, -25dBm per tone.
- Note 7: Time from SHDN = high, until the cascaded receive gain is within 1dB of its final value. Measured with 47pF blocking capacitors on LNAIN and LNAOUT. Matching network removed from IFOUT output.

### Typical Operating Characteristics

 $(\text{MAX}2685 \text{ EV kit, V}_{\text{CC}} = \text{V}_{\overline{\text{SHDN}}} = +3\text{V, } f_{\text{LNAIN}} = f_{\text{MIXIN}} = 880\text{MHz, } f_{\text{LO}} = 960\text{MHz, } P_{\text{LNAIN}} = -30\text{dBm, } P_{\text{LO}} = -8\text{dBm, } P_{\text{MIXIN}} = -25\text{dBm, } differential \text{ IFOUT operation, } Z_0 = 50\Omega, T_A = +25^{\circ}\text{C, unless otherwise noted.})$ 

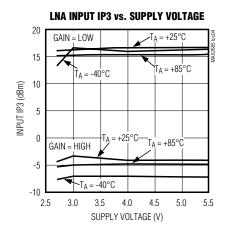


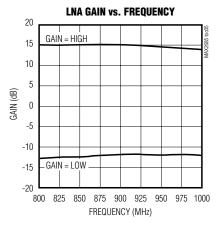


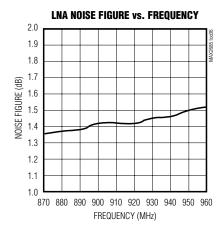


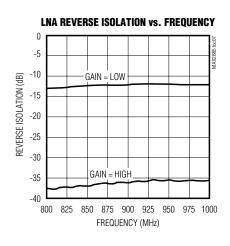
### Typical Operating Characteristics (continued)

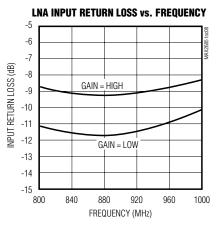
 $(\text{MAX2685 EV kit, V}_{\text{CC}} = \text{V}_{\overline{\text{SHDN}}} = +3\text{V, f}_{\text{LNAIN}} = \text{f}_{\text{MIXIN}} = 880\text{MHz, f}_{\text{LO}} = 960\text{MHz, P}_{\text{LNAIN}} = -30\text{dBm, P}_{\text{LO}} = -8\text{dBm, P}_{\text{MIXIN}} = -25\text{dBm, differential IFOUT operation, Z}_{\text{O}} = 50\Omega, \text{TA} = +25^{\circ}\text{C, unless otherwise noted.})$ 

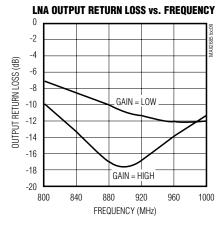


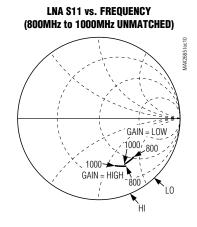


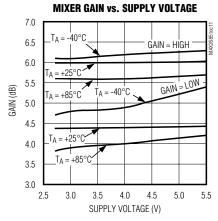


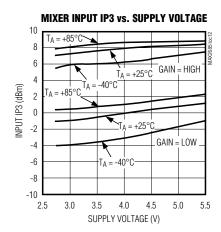








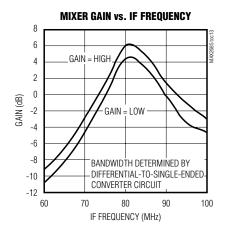


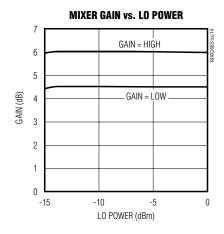


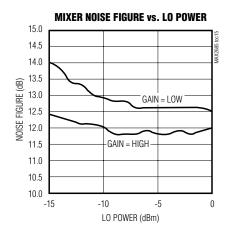
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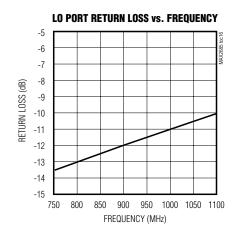
### **Typical Operating Characteristics (continued)**

 $(\text{MAX2685 EV kit, V}_{\text{CC}} = \text{V}_{\overline{\text{SHDN}}} = +3\text{V}, \\ f_{\text{LNAIN}} = f_{\text{MIXIN}} = 880\text{MHz}, \\ f_{\text{LO}} = 960\text{MHz}, \\ P_{\text{LNAIN}} = -30\text{dBm}, \\ P_{\text{LO}} = -8\text{dBm}, \\ P_{\text{MIXIN}} = -25\text{dBm}, \\ P_{\text{MIXIN}} = -25\text{dBm}, \\ P_{\text{LNAIN}} = -30\text{dBm}, \\ P_{\text{LO}} = -8\text{dBm}, \\ P_{\text{MIXIN}} = -25\text{dBm}, \\ P_{\text{LNAIN}} = -30\text{dBm}, \\ P_{\text{LNAIN}} = -30$ 









### **Pin Description**

PIN	NAME	FUNCTION
1, 3, 8, 11, 12, 14	GND	Ground. Connect to ground plane with a low-inductance connection.
2	LNAIN	RF Input to LNA and LNA Bypass Switch. Requires an external matching network and a series DC-blocking capacitor.
4	GAIN	Gain Control Logic-Level Input. Drive high to enable the LNA, open the LNA bypass switch, and increase the receiver's gain. Drive low to disable the LNA, close the LNA bypass switch, and reduce the receiver's gain.
5, 15	Vcc	Supply Voltage. Bypass V <sub>CC</sub> to GND at each pin with a 47pF capacitor as close to the pin as possible.
6	SHDN	Shutdown Control Logic-Level Input. Drive high or connect to V <sub>CC</sub> for normal operation. Drive low to place the device in low-power shutdown mode.
7	LO	Local-Oscillator Input to Downconverter Mixer. Requires a series DC-blocking capacitor and an impedance-setting resistor (typically 75 $\Omega$ to ground).
9	IFOUT-	Inverting Side to Downconverter Mixer's Differential Open-Collector IF Output. Requires a pull-up inductor to VCC for proper biasing, as well as a matching network to ensure optimum output power.
10	IFOUT+	Noninverting Side of Downconverter Mixer's Differential Open-Collector IF Output. Requires a pull-up inductor to V <sub>CC</sub> for proper biasing, as well as a matching network to ensure optimum output power.
13	MIXIN	RF Input to Downconverter Mixer. Requires an external matching network and series DC-blocking capacitor.
16	LNAOUT	LNA Output. Internally matched to $50\Omega$ . LNAOUT has an internal blocking capacitor.

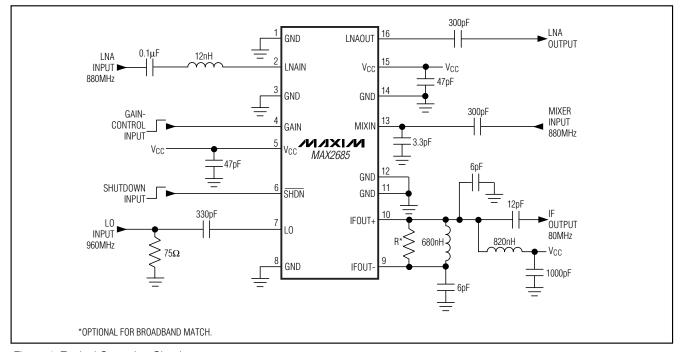


Figure 1. Typical Operating Circuit

### **Detailed Description**

The MAX2685 consists of five major components: a low-noise amplifier (LNA), an LNA bypass switch, a downconverter mixer, a local-oscillator (LO) buffer, and a power-management block.

#### Low-Noise Amplifier (LNA)

The LNA is a wideband, single-ended cascode amplifier that operates over a wide range of frequencies. The input of the LNA (LNAIN) requires an appropriate matching network and a DC-blocking capacitor. The typical operating circuit shown in Figure 1 is optimized for frequencies around 880MHz, requiring only a 0.1µF capacitor in series with a 12nH inductor. See Table 1 for the LNA "S" parameters for matching to other frequencies.

The output of the LNA (LNAOUT) is internally biased to VCC. It is internally matched to  $50\Omega$  and incorporates an internal DC-blocking capacitor.

### LNA Bypass Switch and Gain Control

When a large input signal is present, enable the LNA bypass function to increase linearity and reduce supply current. Set GAIN low to enable the LNA bypass function.

#### **Receive Mixer**

The downconverter mixer is a wideband, single-balanced design with a low noise figure and high linearity. The RF signal at the MIXIN port is mixed with the signal at the LO port, and is downconverted to an IF frequency at the differential IF port.

#### **RF** Input

The MIXIN input requires a simple external matching network and a series DC-blocking capacitor. See Figure 1 for a matching network example, optimized for 880MHz operation. Table 2 lists mixer "S" parameters for matching to other frequencies.

Table 1. LNA Typical S-Parameters ( $V_{CC} = +3V$ ,  $T_A = +25^{\circ}C$ )

FREQUENCY (MHz)	S11  MAG	S11 PHASE (degrees)	S21  MAG	S21 PHASE (degrees)	S12  MAG	S12 PHASE (degrees)	S22  MAG	S22 PHASE (degrees)
High-Gain Mode	(GAIN = VCC	)						
800	0.761	-64.5	4.98	177.9	0.018	-163.7	0.376	-107.3
840	0.753	-68.6	5.06	167.2	0.022	-167.1	0.264	-107.0
880	0.747	-73.2	5.07	156.6	0.026	-171.3	0.172	-94.6
920	0.733	-78.0	4.91	146.6	0.030	-175.7	0.149	-62.9
960	0.719	-82.8	4.68	137.7	0.035	178.0	0.200	-42.4
1000	0.693	-87.5	4.40	130.3	0.039	171.0	0.263	-38.8
Low-Gain Mode	(GAIN = GND)				1	1		
800	0.625	-45.6	0.188	73.0	0.191	71.9	0.483	-91.3
840	0.621	-48.1	0.195	65.5	0.198	64.2	0.423	-91.3
880	0.619	-50.9	0.199	58.1	0.201	56.7	0.370	-89.9
920	0.611	-53.3	0.200	51.6	0.202	50.3	0.337	-86.1
960	0.608	-55.5	0.200	46.1	0.201	44.7	0.322	-80.9
1000	0.607	-57.5	0.200	41.2	0.200	40.0	0.317	-76.7

Table 2. Mixer Typical S-Parameters ( $V_{CC} = +3V$ ,  $T_A = +25$ °C)

RF FREQUENCY (MHz)	S11 MAG	S11  PHASE (degrees)	IF FREQUENCY (MHz)	S22  MAG (IFOUT+ Port Only)	S22 PHASE (IFOUT+ Port Only) (degrees)					
High-Gain Mode (GAIN = VCC)										
800	0.355	152.7	10	0.996	-0.4					
840	0.352	153.7	40	0.994	-1.8					
880	0.351	154.5	80	0.993	-3.2					
920	0.349	155.8	110	0.989	-4.2					
960	0.352	156.2	170	0.988	-6.2					
1000	0.353	156.9	240	0.983	-8.0					
Low-Gain Mode (GAIN = GI	ND)				L					
800	0.275	142.8	10	0.996	-0.5					
840	0.268	144.1	40	0.995	-1.8					
880	0.262	145.5	80	0.993	-3.2					
920	0.255	147.7	110	0.989	-4.2					
960	0.254	149.0	170	0.987	-6.2					
1000	0.245	156.9	240	0.982	-7.9					

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#### **Local-Oscillator Input**

The LO port is the high-impedance input of the local-oscillator buffer. It requires a series DC-blocking capacitor and a shunt resistor to ground to set the input impedance. See the *Typical Operating Characteristics* for a graph of LO Port Return Loss vs. Frequency.

#### **IF Output Port**

The mixer's downconverted output appears on the differential IFOUT+ and IFOUT- pins. The differential output can be converted to a single-ended output, as shown in the MAX2685 evaluation kit (EV kit). Refer to the *Detailed Description* in the MAX2685 EV kit data sheet.

#### Shutdown

Drive SHDN low to disable all device functions and place the MAX2685 in low-power shutdown mode. Drive SHDN high or connect it to VCC to enable all device functions.

### **Applications Information**

#### **Layout Considerations**

A properly designed PC board is an essential part of any RF/microwave circuit. Note the IC's high-frequency inputs and outputs, and be sure to decouple the DC supply and control pins.

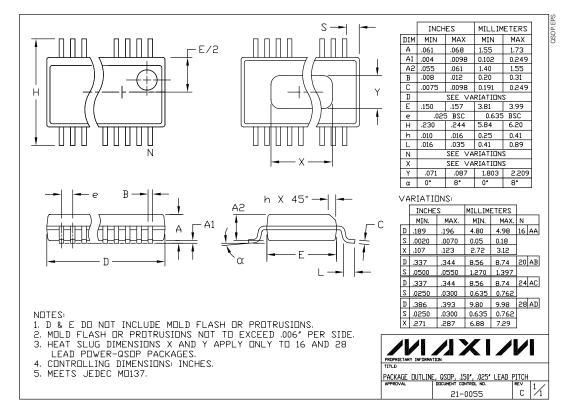
For power-supply traces and connections, a star topology works well. Each VCC node in the circuit has its own path to the central VCC node and a decoupling capacitor that provides a low impedance at the RF frequency of interest. The central VCC also has a large decoupling capacitor. This provides good isolation between the different sections of the MAX2685.

**Chip Information** 

TRANSISTOR COUNT: 295



### **Package Information**



Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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