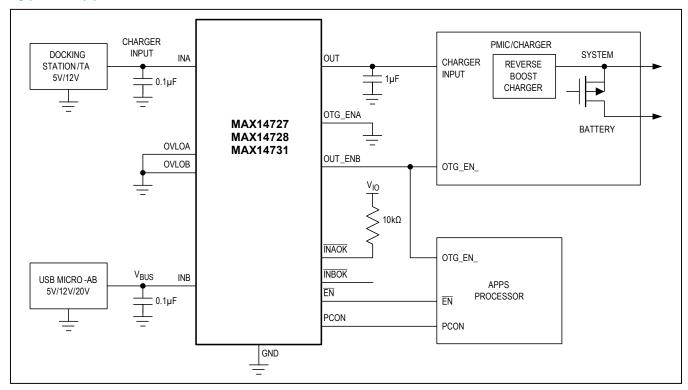
Typical Application Circuit



Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

Absolute Maximum Ratings

(All voltages are referenced to GND.)		Continuous Power Dissipation ($T_A = +70^{\circ}C$)	
INA, INB (Note 1)	0.3V to +30V	WLP (derate 20.25mW/°C above +70°C)	1620mW
OUT	0.3V to +16V	Operating Temperature Range	40°C to +85°C
INA, OUT	16V to +30V	Junction Temperature	+150°C
INB, OUT	16V to +30V	Storage Temperature Range6	5°C to +150°C
OTG_ENA, OTG_ENB,		Soldering Temperature (reflow)	+260°C
PCON, INAOK, INBOK, EN	0.3V to +6V		
OVLOA, OVLOB	0.3V to +16V		
Continuous-Current into INA, INB, OUT (No	ote 2)		
WLP	±5A		
Otreses becaused the self-tend and a #Absolute Mexicon Both		and the decise. These are stored actions only and founding the continue	- £ 41 4 41

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

- **Note 1:** Survive-to-burst pulse up to 100V, with 2Ω series resistance and hot plugs recur. Above the input clamp voltage, input must be a surge in nature with limited energy.
- Note 2: Limited by the PCB thermal design.

Package Thermal Characteristics (Note 3)

WLP

Junction-to-Ambient Thermal Resistance (θ_{JA})49.38°C/W

Note 3: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

Electrical Characteristics

 $(V_{INA,INB}$ = 3V to 28V, V_{OUT} = 3V to 14V, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at $V_{INA,INB}$ = 5V, T_A = +25°C) (Notes 4, 5)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Startup-Voltage Threshold	V _{IN_ST}			2.65	3.2	V
Input Sustaining Voltage	V _{IN_SU}	I _{OUT} = 0A		1.95	2.35	V
Input Valid-Voltage Threshold	V _{IN_VAL}		3.9	4.1	4.3	V
Input Clamp Voltage	V _{IN_CLAMP}	I _{INA,INB} = 10mA, T _A = +25°C		34		V
Input Supply Current	I _{INA,INB}	V _{OVLOA,OVLOB} = 0V, V _{INA,INB} = 5V, INAOK, INBOK is unconnected, I _{OUT} = 0A (Note 6)		230	360	μА
Output Startup-Voltage Threshold	V _{OUTST}			2.65	3.2	V
Output Sustaining Voltage	V _{OUTSU}	I _{INA,INB} = 0A		1.95	2.35	V
Output Supply Current	I _{OUT}	V _{OVLOA,OVLOB} = 0V, V _{OUT} = 5V, I _{INA,INB} = 0A, V _{OTG_ENA,OTG_ENB} = 1.8V		275	420	μA
Output Shutdown Current	I _{SHDN}	V _{OVLOA,OVLOB} = 3V, V _{OUT} = 5V, V _{INA,INB} = 0V, V _{OTG_ENA,OTG_ENB} = 0V		6.5	14	μА
Input Leakage Voltage		V _{OUT} = 14V, INA, INB unconnected, V _{OTG_ENA,OTG_ENB} = 0V		0.001	0.1	V
Input Discharge Current		V _{INA,INB} = V _{OUT} = 5V, INA, INB discharge current after an OTG_ENA, OTG_ENB transition from high to low		130	200	mA

Electrical Characteristics (continued)

 $(V_{INA,INB}=3V\ to\ 28V,\ V_{OUT}=3V\ to\ 14V,\ T_A=-40^{\circ}C\ to\ +85^{\circ}C,\ unless\ otherwise\ noted.$ Typical values are at $V_{INA,INB}=5V,\ T_A=+25^{\circ}C)$ (Notes 4, 5)

PARAMETER	SYMBOL	COI	NDITIONS	MIN	TYP	MAX	UNITS
OVP (INA,INB TO OUT)						,	
On-Resistance (INA to OUT)	R _{ONA}	V _{INA} = 5V, I _{OUT} = 100mA, T _A = +25°C			68	115	mΩ
On-Resistance (INB to OUT)	R _{ONB}	V _{INB} = 5V, I _{OUT} = 100mA, T _A = +25°C			68	115	mΩ
		MAN/44707	V _{IN} rising	13.5	13.75	14.0	
		MAX14727	V _{IN} falling	13.0			
Internal Overvoltage-Trip Level	\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	MAV44700	V _{IN} rising	9.85	10	10.15	
	V _{IN} OVLO	MAX14728	V _{IN} falling	9.5			V
		MAN44724	V _{IN} rising	5.83	5.92	6.01	
		MAX14731	V _{IN} falling	5.6			
OVLOA, OVLOB Set Threshold	V _{OVLO_TH}			1.2		1.24	V
Adjustable OVLOA, OVLOB Threshold Range	V _{OVLO_EXT}			4		14	V
External OVLOA,OVLOB Select Threshold	V _{OVLO_SEL}			0.2	0.25	0.3	V
Active OUT Discharge Current	I _{OUT_DIS}	V _{OUT} = V _{INA,INB} = 5V, OUT discharge current during a transition between channel A and B when PCON = 0			130		mA
DIGITAL SIGNALS (INAOK, INB	OK, OTG_ENA	, OTG_ENB, PCO	I, EN)	•			
INAOK, INBOK Output Low Voltage	V _{OL}	V _{I/O} = 3.3V, I _{SINK} = 1mA				0.4	V
INAOK, INBOK Leakage Current		V _{I/O} = 3.3V, ĪNAOK, ĪNBOK deasserted				1	μA
OTG_ENA, OTG_ENB, PCON, EN Input Logic-High	V _{IH}			1.6			V
OTG_ENA, OTG_ENB, PCON, EN Input Logic-Low	V _{IL}					0.4	V
OTG_ENA, OTG_ENB, PCON, EN Input Leakage Current	I _{IN}	$0V \le V_{IN} \le V_{IL}$ or $V_{IH} \le V_{IN} \le 5.5V$		-1		+1	μA
TIMING CHARACTERISTICS	<u>'</u>						
INA, INB Debounce Time	t _{DEB}		harge pump on (10% of = 100Ω, C_{LOAD} = 10μF		17		ms
INA, INB Soft-Start Time	t _{SS}	V _{INA,INB} = 5V, INAOK, INBOK assert to V _{OUT} = 90% of V _{INA,INB} , C _{LOAD} = 1000μF			5		ms
INA, INB Break-Before-Make Time	t _{BBM}	V _{INB} = 5V, V _{INA} = (C _{LOAD} = 10µF	OV to 5V or 5V to 0V,		8		ms

Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

Electrical Characteristics (continued)

 $(V_{INA,INB}$ = 3V to 28V, V_{OUT} = 3V to 14V, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at $V_{INA,INB}$ = 5V, T_A = +25°C) (Notes 4, 5)

PARAMETER	SYMBOL	CONDITIONS MIN TYP MAX		UNITS		
OUT Discharge Pulse Duration	tout_dis		7.2	8	8.8	ms
INA, INB OVP Turn-Off Response Time	t _{OFF}	From $V_{INA,INB} > V_{OVLOA,OVLOB}$ to $V_{OUT} = 80\%$ of $V_{INA,INB}$, $R_{LOAD} = 100\Omega$		100		ns
OTG Turn-On Time	t _{OTG_ON}	Time from OTG_ENA, OTG_ENB high to V _{INA,INB} = 80% of V _{OUT} , V _{OUT} = 5V, C _{INA,INB} = 10µF	30		μs	
INA, INB Discharge Pulse Duration		V _{INA,INB} = V _{OUT} = 5V, current pulse duration after an OTG_ENA, OTG_ENB transition from high to low	16		ms	
THERMAL PROTECTION						
Thermal Shutdown	T _{SHDN}			150		°C
Thermal Hysteresis	T _{HYST}			20		°C
ESD PROTECTION						
Human Body Model		INA, INB		±15		kV
IEC 61000-4-2 Contact Discharge		INA, INB	±8 k		kV	
IEC 61000-4-2 Air Gap Discharge		INA, INB ±15			kV	

- Note 4: All devices are 100% production tested at $T_A = +25$ °C. Limits over the operating temperature range are guaranteed by design and not production tested.
- Note 5: The input is considered not present when it is lower than a rough threshold of TP = 1.5V (typ). When the B path is active and the INA > Tp is connected, the device goes to the A path (since it has the priority), but the INA has to be higher than the startup voltage to turn on the switch.
- Note 6: When both inputs are connected, supply current is drawn from the highest one.

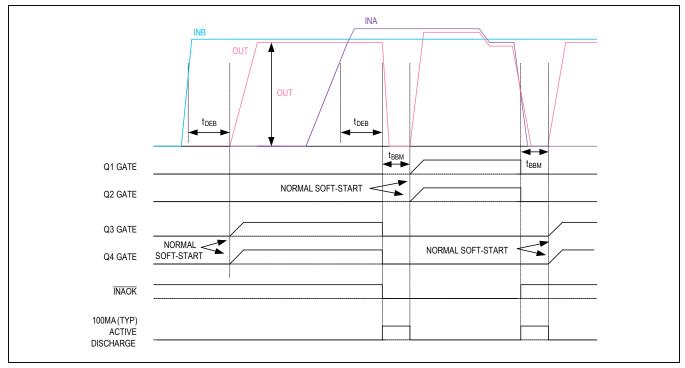


Figure 1. Automatic Switch Operation, PCON = 0

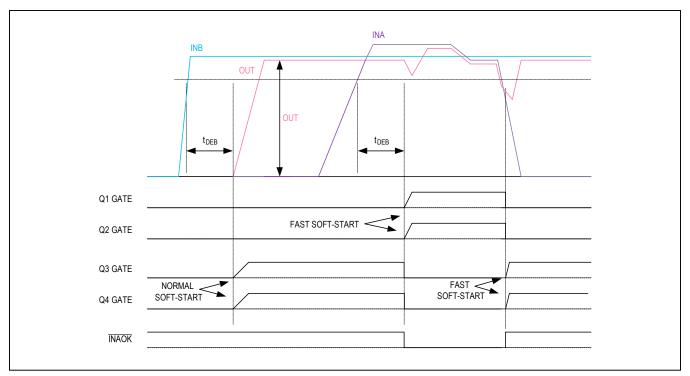
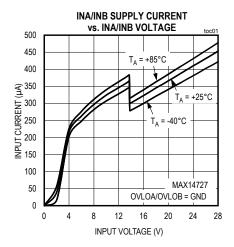
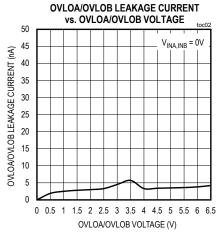


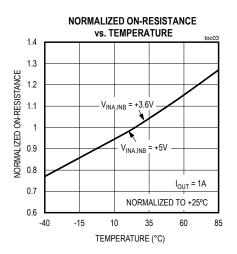
Figure 2. Automatic Switch Operation, PCON = 1

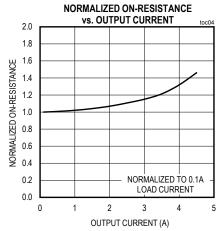
Typical Operating Characteristics

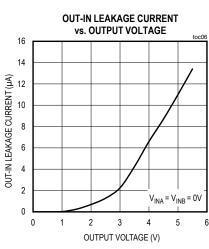
 $(V_{INA,\ INB}$ = +5V, OVLO = GND, C_{IN} = 0.1 μ F, C_{OUT} = 1 μ F, T_A = +25°C, unless otherwise noted.)

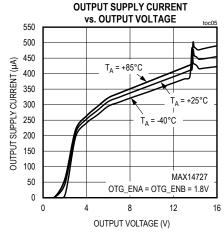


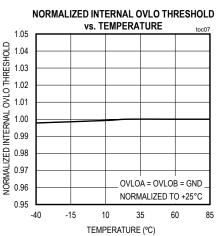






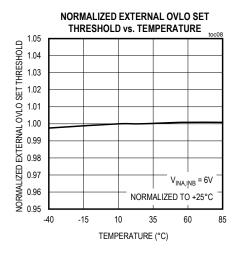


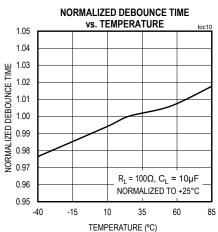


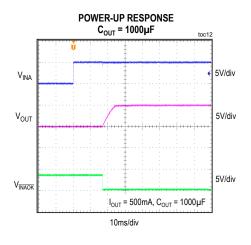


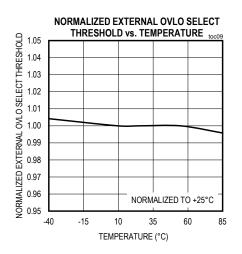
Typical Operating Characteristics (continued)

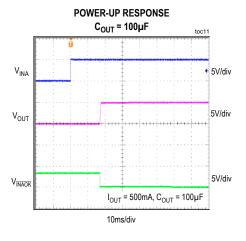
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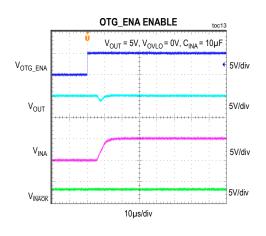






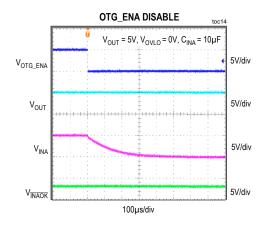


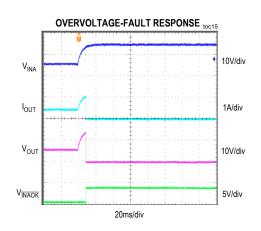


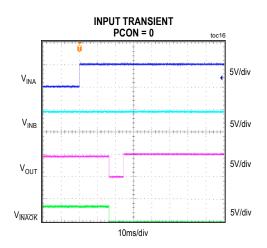


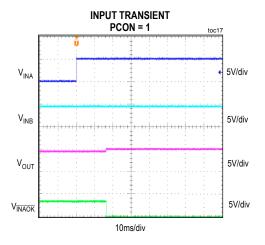
Typical Operating Characteristics (continued)

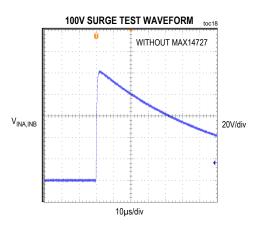
 $(V_{INA,\ INB}$ = +5V, OVLO = GND, C_{IN} = 0.1 μ F, C_{OUT} = 1 μ F, T_A = +25°C, unless otherwise noted.)

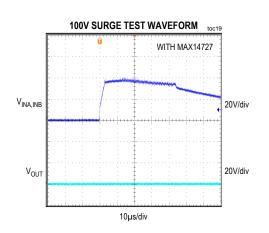






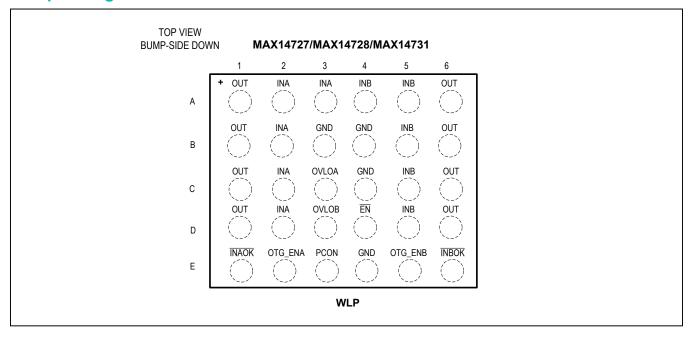






Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

Bump Configurations



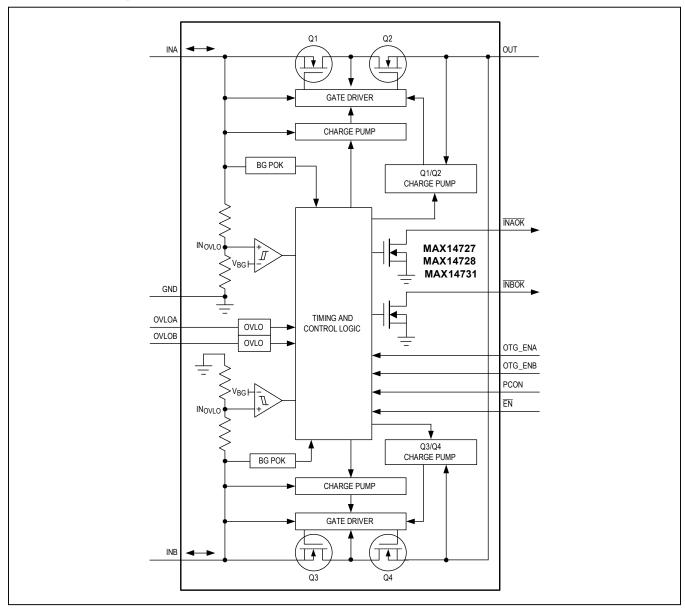
Bump Description

BUMP	NAME	FUNCTION
A1, A6, B1, B6, C1, C6, D1, D6	OUT	Overvoltage-Protection Output. Bypass OUT with a 1µF ceramic capacitor. Externally connect all OUT together.
A2, A3, B2, C2, D2	INA	Overvoltage-Protection Input A. Bypass INA with a 0.1µF ceramic capacitor as close as possible to the device. Externally connect all INA together.
A4, A5, B5, C5, D5	INB	Overvoltage-Protection Input B. Bypass INB with a 0.1µF ceramic capacitor as close as possible to the device as possible. Externally connect all INB together.
B3, B4, C4, E4	GND	Ground
C3	OVLOA	Overvoltage-Lockout Input A. Connect OVLOA to GND to use internal OVLO threshold for channel A. Connect OVLOA to a resistor divider for different voltage threshold for channel A.
D3	OVLOB	Overvoltage-Lockout Input B. Connect OVLOB to GND to use internal OVLO threshold for channel B. Connect OVLOB to a resistor divider for different voltage threshold for channel B.
D4	ĒN	Active-Low Enable for Both OVP. When $\overline{\text{EN}}$ is high, both channels are off. When $\overline{\text{EN}}$ is low, both channels are active. $\overline{\text{EN}}$ does not affect $\overline{\text{INAOK}}$ or $\overline{\text{INBOK}}$.
E1	ĪNAOK	Open-Drain Flag Output. $\overline{\text{INAOK}}$ is driven low after INA voltage is stable between minimum V_{IN} and V_{OVLO} when OTG_ENA = 0. Connect a pullup resistor from $\overline{\text{INAOK}}$ to the logic I/O voltage of the host system.
E2	OTG_ENA	Enable Input for Channel A OTG Supply Operation. When OTG_ENA is low, channel A OTG operation is disabled. When OTG_ENA is high, channel A OTG operation is enabled, and OUT is connected to INA.
E3	PCON	Path Overlap Control. When PCON is low, there is a break-before-make time, 8ms (typ). When PCON is high, there is no break-before-make time. See the <i>Automatic Switch Operation</i> section.

Bump Description (continued)

BUMP	NAME	FUNCTION
E5	OTG_ENB	Enable Input for Channel B OTG Supply Operation. When OTG_ENB is low, channel B OTG operation is disabled. When OTG_ENB is high, channel B OTG operation is enabled, and OUT is connected to INB.
E6	ĪNBOK	Open-Drain Flag Output. $\overline{\text{INBOK}}$ is driven low after INB voltage is stable between minimum V_{IN} and V_{OVLO} when OTG_ENB = 0. Connect a pullup resistor from $\overline{\text{INBOK}}$ to the logic I/O voltage of the host system.

Functional Diagram



Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

Detailed Description

The MAX14727/MAX14728/MAX14731 provide overvoltage protection for applications with two input sources. The devices feature low on-resistance and protect low-voltage systems against voltage faults up to 28V_{DC}. An internal clamp also protects the devices from surges up to 100V. If the input voltage exceeds the overvoltage threshold, the output is disconnected from the input to prevent damage to the protected components. The 15ms debounce time prevents false turn-on of the internal FETs during startup.

Automatic Input Power Path Control

Automatic power path select has two modes of operation based on PCON. At least one of the inputs has to cross the startup voltage (2.75V, typ) at power-up for the circuitry to work. INA always has priority and INB only takes over INA when either INB is the only present input or when INB is valid (> 4.1V, typ) and INA is invalid (< 4.1V, typ) (Figure 1 and Figure 2).

When PCON = 0 (Figure 1):

When there is an input source change, the device has a break-before-make time (8ms, typ) with active discharge of 100mA (typ) during the off period. Additionally, during the off period, the battery-charger mode can reset so the charger can draw the proper amount of current.

INAOK and INBOK assert when the appropriate INA, INB is above the sustaining voltage (1.9V, typ) and below the OVLOA, OVLOB threshold.

When PCON = 1 (Figure 2):

When INA is detected valid while INB is valid, channel B is turned off and channel A is turned on quickly, with a higher soft-start current. Therefore, the transition from INB charging to INA charging is without shutoff. When INA becomes invalid while INB is still valid, channel B is turned on immediately, without debounce time, by using a higher soft-start current again.

INAOK and INBOK assert when the respective INA,INB is above valid voltage (4.1V, typ) and below the OVLOA, OVLOB threshold.

Table 1. Functionality Table

OVLOA OVLOB	OTG_ENA	OTG_ENB	EN	INAOK	ĪNBOK	CHANNEL A	CHANNEL B
0	0	0	0	INA status (Note 7)	INB status (Note 7)	(Note 8)	(Note 8)
0	0	0	1	INA status (Note 9)	INB status (Note 9)	Open	Open
0	0	1	0	INA status (Note 7)	High	Based on INA status	Close
0	0	1	1	INA status (Note 9)	High	Open	Open
0	1	0	0	High	INB status (Note 7)	Close	Based on INB status
0	1	0	1	High	INB status (Note 9)	Open	Open
0	1	1	0	High	High	Close	Close
0	1	1	1	High	High	Open	Open
1	Х	Х	Х	High when OVLOA high	High when OVLOB high	Open when OVLOA high	Open when OVLOB high

X = Don't care.

Note 7: PCON = 0, INAOK, INBOK asserts low when respective INA,INB is greater than sustaining voltage (1.9V, typ). PCON = 1, INAOK, and INBOK assert low when respective INA,INB is greater than valid voltage (4.1V, typ).

Note 8: INA always has priority. INB takes over INA when INB is the only present input or when INB is valid (above 4.1V, typ) and INA is invalid (below 4.1V, typ).

Note 9: When EN is high, both channels are open and the minimum input functionality voltage is startup voltage (2.75V, typ).

Soft-Start

To minimize inrush current, the devices feature a soft-start capability to slowly turn on the internal FETs. Soft-start begins when INAOK and INBOK are asserted, and ends after 15ms (typ).

Overvoltage Lockout (OVLOA, OVLOB)

Connect OVLOA, OVLOB to ground to use the internal OVLO comparator with the internally set OVLO value. When INA, INB go above the overvoltage-lockout threshold ($V_{\mbox{\footnotesize{IN}}}$ OVLO), OUT is disconnected from INA, INB and INAOK and INBOK is deasserted. When INA, INB drops below V_{IN OVLO}, the debounce time starts counting. After the debounce time, OUT follows INA, INB again and INAOK and INBOK is asserted.

The OVLOA, OVLOB input can be used as an additional enable input (Table 2).

External OVLO Adjustment Functionality

When an external resistor-divider is connected to OVLOA. OVLOB, and V_{OVLOA}, V_{OVLOB} exceeds the OVLOA, OVLOB select voltage (VOVLO SEL), the internal OVLOA, OVLOB comparator read INA,INB by the external resistor-divider.

R1 = $1M\Omega$ is a good starting value for minimum current consumption. Since V_{IN OVLO}, V_{OVLO TH}, and R1 are known, R2 can be calculated from the following formula:

$$V_{IN_OVLO} = V_{OVLO_TH} \times \left[1 + \frac{R1}{R2}\right]$$

This external resistor-divider is completely independent from the internal resistor-divider.

Table 2. Truth Table

SWITCH OPEN/CLOSE		OVLOA, OVLOB		
SWITCH OF	SWITCH OPEN/CLOSE		High	
EN	Low	Close	Open	
	High	Open	Open	

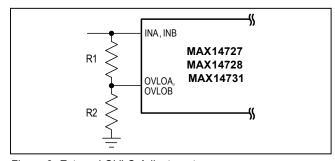


Figure 3. External OVLO Adjustment

Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

Reverse-Bias Blocking

The devices feature reverse-bias blocking. When INA, INB voltage is below the input startup voltage and OTG ENA and OTG_ENB is low, the switch between INA, INB and OUT is open and the two back-to-back diodes of the two series switches block reverse bias. Therefore, when the voltage is applied at the output, current does not travel back to the input.

OTG Enable

The devices feature reverse-turn-on capability. OTG_ENA and OTG ENB can be used to turn on the switch for OUT to feed back to INA, INB when voltage applied at OUT is above the startup voltage (2.75V, typ). When OTG ENA, OTG ENB is high, INAOK and INBOK is deasserted. During the OTG operation, if INA, INB go above OVLOA, OVLOB, the OVP switch turns off. It is recommended that power be supplied to OUT prior to OTG operation. Power is also removed from OUT prior to disabling OTG operation. When either OTG ENA or OTG ENB is on, the internal power path control is disabled.

Thermal-Shutdown Protection

The devices feature thermal-shutdown protection to protect the device from overheating. The internal FETs turn off when the junction temperature exceeds +150°C (typ). The device returns to normal operation once the temperature drops by approximately 20°C (typ).

Applications Information

Input Bypass Capacitor

For most applications, bypass INA and INB to GND with a 0.1µF ceramic capacitor as close as possible to the device. If the power source has significant inductance due to long lead length, take care to prevent overshoots due to the LC tank circuit and provide protection by clamping the overshoot.

Output Capacitor

The slow turn-on time provides a soft-start function that allows the device to charge an output capacitor up to 1000µF without turning off due to an overcurrent condition.

When PCON = 0, the output is actively discharged during the channel B to channel A switchover event. The inductance of the input cable with a small input capacitance (0.1µF) would cause an input voltage drop. This might trigger a switch back to channel B and output latchoff. Therefore, the output capacitor should be limited so that the input voltage stays above a valid voltage during the output recharge event. Usually, the output capacitor value of 33µF or less is recommended for 1m charger cable.

Extended ESD Protection

ESD-protection structures are incorporated on all pins to protect against electrostatic discharges up to ≤ 2kV (Human Body Model) encountered during handling and assembly. INA and INB are further protected against ESD up to ±15kV (Human Body Model), ±15kV (Air-Gap Discharge method described in IEC 61000-4-2) and ±8kV (Contact discharge method described in IEC 61000-4-2) without damage.

The ESD structures withstand high ESD both in normal operation and when the device is powered down. After an ESD event, the devices continue to function without latchup.

ESD Test Conditions

ESD performance depends on a variety of conditions. Contact Maxim for a reliability report that documents test setup, test methodology, and test results.

Human Body Model ESD Protection

Figure 4 shows the HBM and Figure 5 shows the current waveform it generates when discharged into a

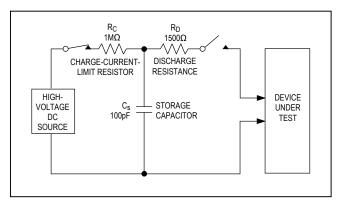


Figure 4. Human Body ESD Test Model

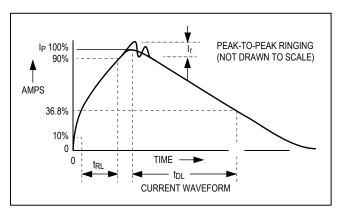


Figure 5. Human Body Current Waveform

Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

low-impedance state. This model consists of a 100pF capacitor charged to the ESD voltage of interest, which is then discharged into the device through a $1.5k\Omega$ resistor.

IEC 61000-4-2

The IEC 61000-4-2 standard covers ESD testing and performance of finished equipment. However, it does not specifically refer to integrated circuits. The major difference between tests done using the Human Body Model and IEC 61000-4-2 is higher peak current in IEC 61000-4-2 because series resistance is lower in the IEC 61000-4-2 model. Hence, the ESD withstand voltage measured to IEC 61000-4-2 is generally lower than that measured using the Human Body Model. Figure 6 shows the IEC 61000-4-2 model. Figure 7 shows the current waveform for the IEC 61000-4-2 ESD contact discharge test.

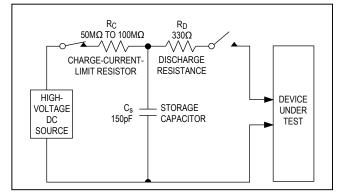


Figure 6. IEC 61000-4-2 ESD Test Model

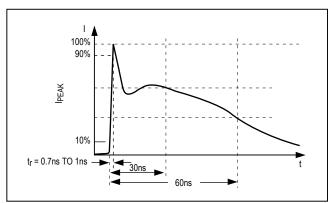


Figure 7. IEC 61000-4-2 ESD Generator Current Waveform

Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

Ordering Information/Selector Guide

PART	OVLO (V)	PIN- PACKAGE	TEMP
MAX14727EWV+	13.75	30 WLP	-40°C TO +85°C
MAX14728EWV+	10	30 WLP	-40°C TO +85°C
MAX14731EWV+	5.92	30 WLP	-40°C TO +85°C

⁺Denotes a lead(Pb)-free package/RoHS-compliant package.

Chip Information

PROCESS: BICMOS

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
30 WLP	W302F2+1	<u>21-0762</u>	Refer to Application Note 1891

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Dual-Input, Bidirectional Overvoltage Protector with Automatic Path Control

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	3/15	Initial release	_
1	6/15	Removed future product references from MAX14728 and MAX14731	4, 15

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim Integrated's website at www.maximintegrated.com.

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