

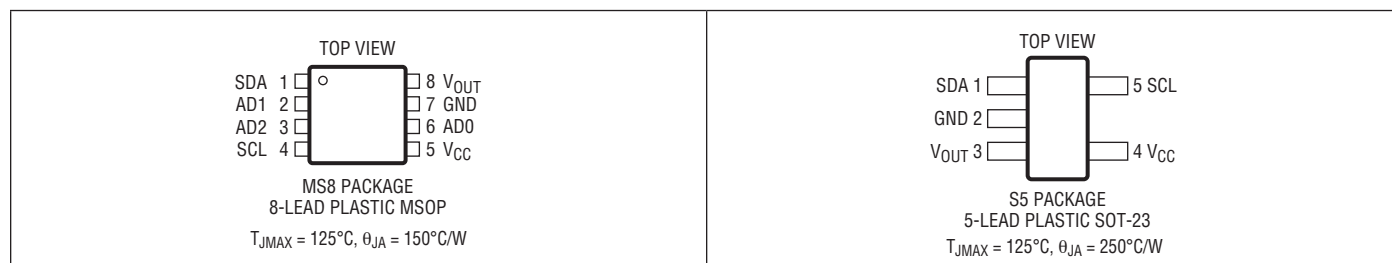
LTC1669

ABSOLUTE MAXIMUM RATINGS (Note 1)

V_{CC} to GND $-0.3V$ to $7.5V$
SDA, SCL $-0.3V$ to $7.5V$
AD0, AD1, AD2 (MSOP Only) $-0.3V$ to $(V_{CC} + 0.3V)$
 V_{OUT} $-0.3V$ to $(V_{CC} + 0.3V)$

Operating Temperature Range
LTC1669C $0^{\circ}C$ to $70^{\circ}C$
LTC1669I $-40^{\circ}C$ to $85^{\circ}C$
Storage Temperature Range $-65^{\circ}C$ to $150^{\circ}C$
Lead Temperature (Soldering, 10 sec) $300^{\circ}C$

PIN CONFIGURATION



ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LTC1669CMS8#PBF	LTC1669CMS8#TRPBF	LTAHV	8-Lead Plastic MSOP	$0^{\circ}C$ to $70^{\circ}C$
LTC1669IMS8#PBF	LTC1669IMS8#TRPBF	LTAHX	8-Lead Plastic MSOP	$-40^{\circ}C$ to $85^{\circ}C$
LTC1669-8CMS8#PBF	LTC1669-8CMS8#TRPBF	LTAHT	8-Lead Plastic MSOP	$0^{\circ}C$ to $70^{\circ}C$
LTC1669-8IMS8#PBF	LTC1669-8IMS8#TRPBF	LTAHU	8-Lead Plastic MSOP	$-40^{\circ}C$ to $85^{\circ}C$
TAPE AND REEL (MINI)	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LTC1669CS5#TRMPBF	LTC1669CS5#TRPBF	LTAHW	5-Lead Plastic TSOT-23	$0^{\circ}C$ to $70^{\circ}C$
LTC1669-1CS5#TRMPBF	LTC1669-1CS5#TRPBF	LTAHR	5-Lead Plastic TSOT-23	$0^{\circ}C$ to $70^{\circ}C$

TRM = 500 pieces.

Consult LTC Marketing for parts specified with wider operating temperature ranges.

Consult LTC Marketing for information on lead based finish parts.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreel/>

ELECTRICAL CHARACTERISTICS The ● denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_{CC} = 2.7\text{V}$ to 5.5V , V_{CC} set as reference, V_{OUT} unloaded, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
DAC							
	Resolution		●	10			Bits
	Monotonicity	(Note 2)	●	10			Bits
DNL	Differential Nonlinearity	Guaranteed Monotonic (Note 2)	●		± 0.2	± 0.75	LSB
INL	Integral Nonlinearity	(Note 2)	●		± 0.5	± 2.5	LSB
V_{OS}	Offset Error	Measured at Code 20	●		± 10	± 30	mV
V_{OSTC}	Offset Error Temperature Coefficient				± 15		$\mu\text{V}/^\circ\text{C}$
FSE	Full-Scale Error	Reference Set to V_{CC} Reference Set to Internal Bandgap	● ●		± 3 ± 3	± 15 ± 15	LSB LSB
V_{OUT}	DAC Output Span	Reference Set to V_{CC} Reference Set to Internal Bandgap			0 to V_{CC} 0 to 2.5		V V
V_{FSTC}	Full-Scale Voltage Temperature Coefficient	Reference Set to V_{CC} Reference Set to Internal Bandgap			± 30 ± 50		$\mu\text{V}/^\circ\text{C}$ $\mu\text{V}/^\circ\text{C}$
PSRR	Power Supply Rejection Ratio	Reference Set to Internal Bandgap, Code = 1023			± 0.4		LSB/V

Power Supply

V_{CC}	Positive Supply Voltage		●	2.7		5.5	V
I_{CC}	Supply Current	$V_{CC} = 3\text{V}$ (Note 3) $V_{CC} = 5\text{V}$ (Note 3)	● ●		60 75	100 125	μA μA
I_{SD}	Supply Current in Shutdown Mode	(Note 3)	●		12	24	μA

Op Amp DC Performance

	Short-Circuit Current (Sourcing)	V_{OUT} Shorted to GND, Input Code = 1023	●		25	100	mA
	Short-Circuit Current (Sinking)	V_{OUT} Shorted to V_{CC} , Input Code = 0	●		30	120	mA
	Output Impedance to GND	Input Code = 0, $V_{CC} = 5\text{V}$ Input Code = 0, $V_{CC} = 5\text{V}$ In Shutdown Mode			65 150 500		Ω Ω k Ω
	Output Impedance to V_{CC}	Input Code = 1023, $V_{CC} = 5\text{V}$ Input Code = 1023, $V_{CC} = 5\text{V}$			80 120		Ω Ω

AC Performance

	Voltage Output Slew Rate	Rising (Notes 4, 5) Falling (Notes 4, 5)			0.75 0.25		V/ μs V/ μs
	Voltage Output Settling Time	To $\pm 0.5\text{LSB}$ (Notes 4, 5)			30		μs
	Digital Feedthrough				0.75		nV•s
	Digital-to-Analog Glitch Impulse	1LSB Change Around Major Carry			70		nV•s

Digital Inputs SCL, SDAs

V_{IH}	High Level Input Voltage		●	2.1			V
V_{IL}	Low Level Input Voltage		●			1.5	V
V_{LTH}	Logic Threshold Voltage				1.8		V
I_{LEAK}	Digital Input Leakage	$V_{CC} = 5.5\text{V}$ and 0V, $V_{IN} = \text{GND to } V_{CC}$	●			± 1	μA
C_{IN}	Digital Input Capacitance	(Note 7)	●			10	pF

Digital Output SDA

V_{OL}	Digital Output Low Voltage	$I_{PULLUP} = 3\text{mA}$	●			0.4	V
----------	----------------------------	---------------------------	---	--	--	-----	---

ELECTRICAL CHARACTERISTICS The ● denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_{CC} = 2.7\text{V}$ to 5.5V , V_{CC} set as reference, V_{OUT} unloaded, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Address Inputs AD0, AD1, AD2 (MSOP Only)						
I_{UP}	Address Pin Pull-Up Current	$V_{IN} = 0\text{V}$	●	0.5	1.5	μA
V_{IH}	High Level Input Voltage		●	$V_{CC} - 0.3$		V
V_{IL}	Low Level Input Voltage		●		0.8	V

TIMING CHARACTERISTICS The ● denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_{CC} = 2.7\text{V}$ to 5.5V , V_{CC} set as reference, V_{OUT} unloaded, unless otherwise noted.

SYMBOL	PARAMETER		MIN	TYP	MAX	UNITS
Timing Characteristics (Notes 6, 7)						
f _{SCL}	Clock Operating Frequency	●	100			kHz
t _{BUF}	Bus Free Time Between Stop and Start Condition	●	4.7			μs
t _{HD, STA}	Hold Time After (Repeated) Start Condition	●	4			μs
t _{SU, STA}	Repeated Start Condition Setup Time	●	4.7			μs
t _{SU, STO}	Stop Condition Setup Time	●	4			μs
t _{HD, DAT (IN)}	Data Hold Time (Input)	●	0			ns
t _{HD, DAT (OUT)}	Data Hold Time (Output)	●	225	500	3450	ns
t _{SU, DAT}	Data Setup Time	●	250			ns
t _{LOW}	Clock Low Period	●	4.7			μs
t _{HIGH}	Clock High Period	●	4			μs
t _f	Clock, Data Fall Time	●	20	300		ns
t _r	Clock, Data Rise Time	●	20	1000		ns

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: Nonlinearity and monotonicity are defined from code 20 to code 1003 (full scale). See Applications Information.

Note 3: Digital inputs at 0V or V_{CC} .

Note 4: Load is $10\text{k}\Omega$ in parallel with 100pF .

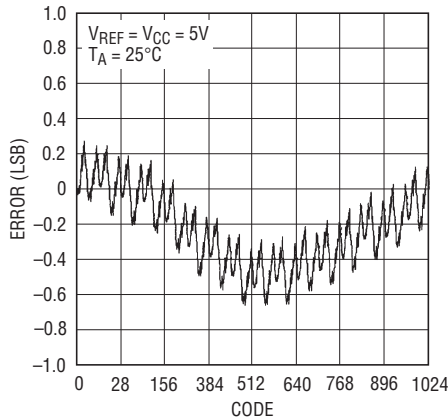
Note 5: $V_{CC} = V_{REF} = 5\text{V}$. DAC switched between $0.1V_{FS}$ and $0.9V_{FS}$, i.e., codes $k = 102$ and $k = 922$.

Note 6: All values are referenced to V_{IH} and V_{IL} levels.

Note 7: Guaranteed by design and not subject to test.

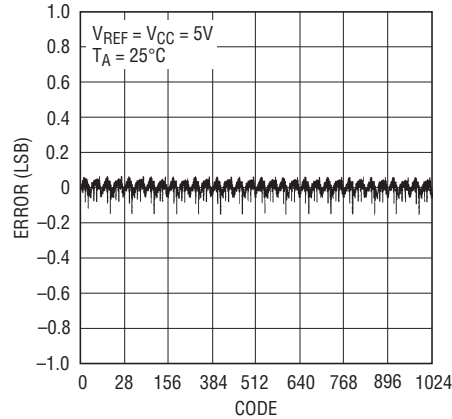
TYPICAL PERFORMANCE CHARACTERISTICS

Integral Nonlinearity (INL)

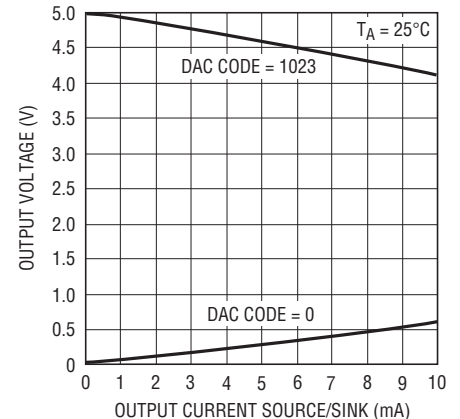


1669 G01

Differential Nonlinearity (DNL)

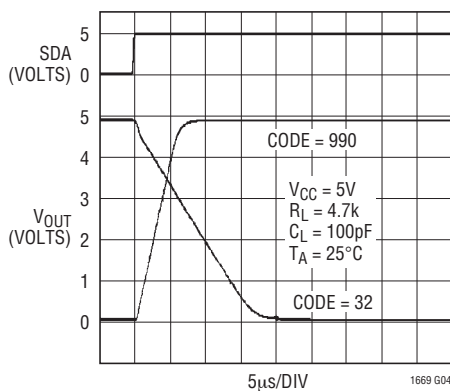


1669 G02

Source and Sink Current Capability with $V_{CC} = 5V$ 

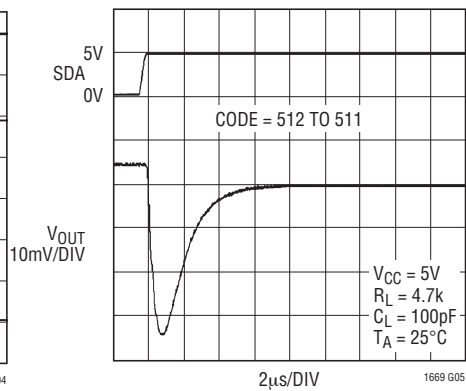
1669 G03

Large-Signal Step Response



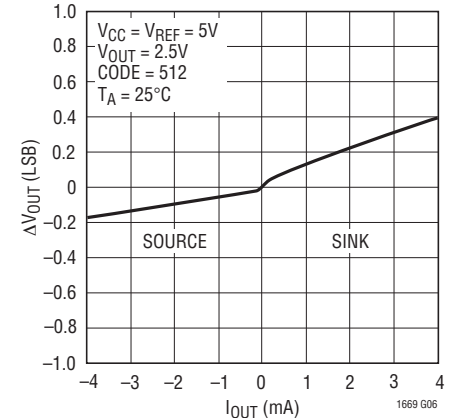
1669 G04

Midscale Glitch



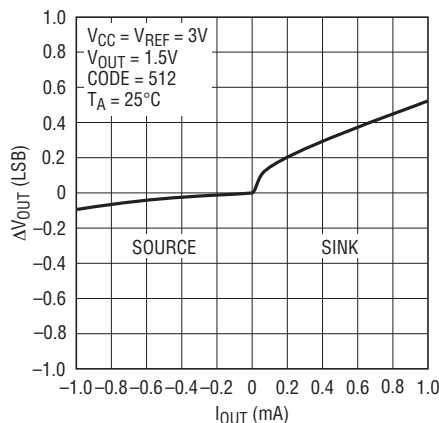
1669 G05

Load Regulation vs Output Current



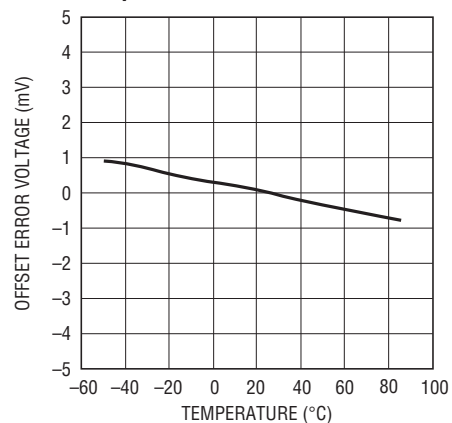
1669 G06

Load Regulation vs Output Current



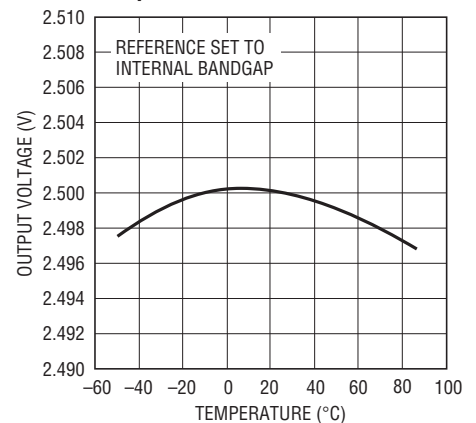
1669 G07

Offset Error Voltage vs Temperature



1669 G08

Full-Scale Output Voltage vs Temperature



1669 G09

PIN FUNCTIONS

SDA (Pin 1, Pin 1 on SOT-23): Serial Data Bidirectional Pin. Data is shifted into the SDA pin and acknowledged by the SDA pin. High impedance pin while data is shifted in. Open-drain N-channel output during acknowledgment. Requires a pull-up resistor or current source to V_{CC} .

AD1 (Pin 2): Slave Address Select Bit 1. Tie this pin to either V_{CC} or GND to modify the corresponding bit of the LTC1669's slave address.

AD2 (Pin 3): Slave Address Select Bit 2. Tie this pin to either V_{CC} or GND to modify the corresponding bit of the LTC1669's slave address.

SCL (Pin 4, Pin 5 on SOT-23): Serial Clock Input Pin. Data is shifted into the SDA pin at the rising edges of the clock. This high impedance pin requires a pull-up resistor or current source to V_{CC} .

V_{CC} (Pin 5, Pin 4 on SOT-23): Power Supply. $2.7V \leq V_{CC} \leq 5.5V$. Also used as the reference voltage input when the part is programmed to use V_{CC} as the reference.

AD0 (Pin 6): Slave Address Select Bit 0. Tie this pin to either V_{CC} or GND to modify the corresponding bit of the LTC1669's slave address.

GND (Pin 7, Pin 2 on SOT-23): System Ground.

V_{OUT} (Pin 8, Pin 3 on SOT-23): Voltage Output. Buffered rail-to-rail DAC output.

DEFINITIONS

Differential Nonlinearity (DNL): The difference between the measured change and the ideal 1LSB change for any two adjacent codes. The DNL error between any two codes is calculated as follows:

$$\text{DNL} = (\Delta V_{\text{OUT}} - \text{LSB}) / \text{LSB}$$

Where ΔV_{OUT} is the measured voltage difference between two adjacent codes.

Digital Feedthrough: The glitch that appears at the analog output caused by AC coupling from the digital inputs when they change state. The area of the glitch is specified in (nV)(sec).

Full-Scale Error (FSE): The deviation of the actual full-scale voltage from ideal. FSE includes the effects of offset and gain errors (see Applications Information).

Integral Nonlinearity (INL): The deviation from a straight line passing through the endpoints of the DAC transfer curve (Endpoint INL). Because the output cannot go below zero, the linearity is measured between full scale and the lowest code that guarantees the output will be

greater than zero. The INL error at a given input code is calculated as follows:

$$\text{INL} = [V_{\text{OUT}} - V_{\text{OS}} - (V_{\text{FS}} - V_{\text{OS}})(\text{code}/1023)] / \text{LSB}$$

Where V_{OUT} is the output voltage of the DAC measured at the given input code.

Least Significant Bit (LSB): The ideal voltage difference between two successive codes.

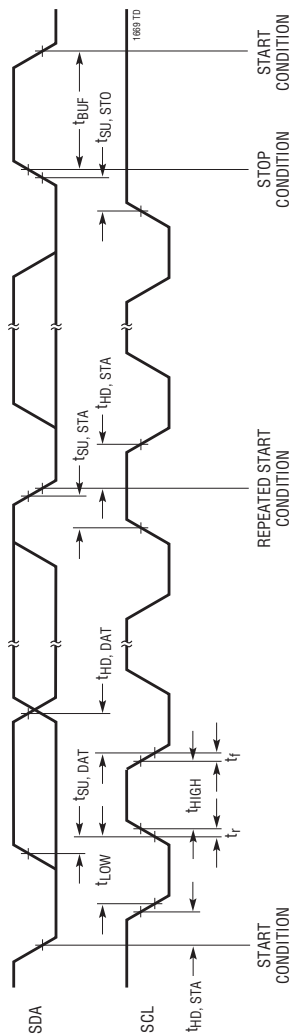
$$\text{LSB} = V_{\text{REF}} / 1024$$

Resolution (n): Defines the number of DAC output states (2^n) that divide the full-scale range. Resolution does not imply linearity.

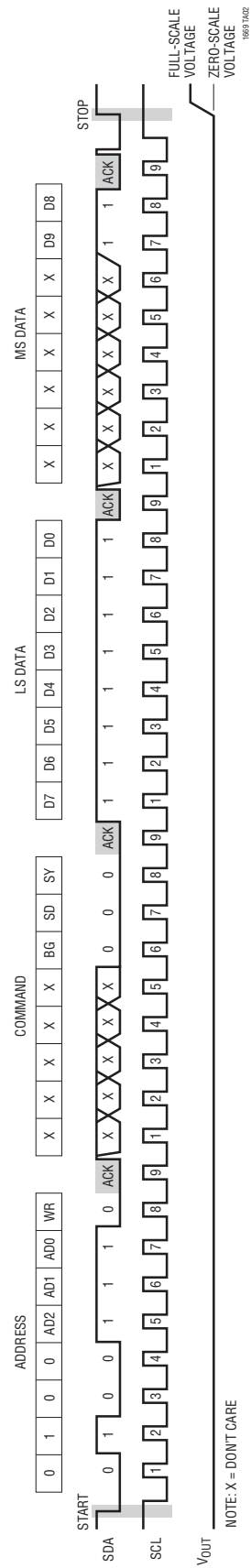
Voltage Offset Error (V_{OS}): Nominally, the voltage at the output when the DAC is loaded with all zeros. A single supply DAC can have a true negative offset, but the output cannot go below zero (see Applications Information).

For this reason, single supply DAC offset is measured at the lowest code that guarantees the output will be greater than zero.

TIMING DIAGRAM

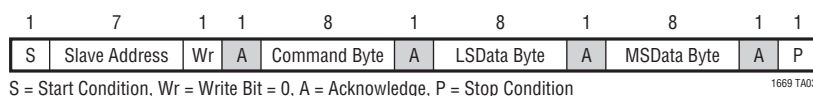


Typical LTC1669 Input Waveform—Programming DAC Output for Full Scale (AD2 to AD0 Set High)



APPLICATIONS INFORMATION

Write Word Protocol Used by the LTC1669



Serial Digital Interface

The LTC1669 communicates with a host (master) using the standard 2-wire interface. The Timing Diagram shows the timing relationship of the signals on the bus. The two bus lines, SDA and SCL, must be high when the bus is not in use. External pull-up resistors or current sources, such as the LTC1694 SMBus/I²C Accelerator, are required on these lines.

The LTC1669 is a receive-only (slave) device. The master can communicate with the LTC1669 using the Quick Command, Send Byte or Write Word protocols as explained later.

The START and STOP Conditions

When the bus is not in use, both SCL and SDA must be high. A bus master signals the beginning of a communication to a slave device by transmitting a START condition. A START condition is generated by transitioning SDA from high to low while SCL is high.

When the master has finished communicating with the slave, it issues a STOP condition. A STOP condition is generated by transitioning SDA from low to high while SCL is high. The bus is then free for communication with another I²C device.

Acknowledge

The Acknowledge signal is used for handshaking between the master and the slave. An Acknowledge (active LOW) generated by the slave lets the master know that the latest byte of information was received. The Acknowledge related clock pulse is generated by the master. The master releases the SDA line (HIGH) during the Acknowledge clock pulse. The slave-receiver must pull down the SDA line during the Acknowledge clock pulse so that it remains a stable LOW during the HIGH period of this clock pulse.

Write Word Protocol

The master initiates communication with the LTC1669 with a START condition and a 7-bit address followed by the Write Bit (Wr) = 0. The LTC1669 acknowledges and the master delivers the command byte. The LTC1669 acknowledges and latches the command byte into the command byte input register. The master then delivers the least significant data byte. Again the LTC1669 acknowledges and the data is latched into the least significant data byte input register. The master then delivers the most significant data byte. The LTC1669 acknowledges once more and latches the data into the most significant data byte input register. Lastly, the master terminates the communication with a STOP condition. On the reception of the STOP condition, the LTC1669 transfers the input register information to output registers and the DAC output is updated.

Slave Address (MSOP Package Only)

The LTC1669 can respond to one of eight 7-bit addresses. The first 4 bits (MSBs) have been factory programmed to 0100. The first 4 bits of the LTC1669-8 have been factory programmed to 0011. The three address bits, AD2, AD1 and AD0 are programmed by the user and determine the LSBs of the slave address, as shown in the table below:

			LTC1669	LTC-1669-8
AD2	AD1	AD0	0100 xxx	0011 xxx
L	L	L	0100 000	0011 000
L	L	H	0100 001	0011 001
L	H	L	0100 010	0011 010
L	H	H	0100 011	0011 011
H	L	L	0100 100	0011 100
H	L	H	0100 101	0011 101
H	H	L	0100 110	0011 110
H	H	H	0100 111	0011 111

APPLICATIONS INFORMATION

Slave Address (SOT-23 Package)

The slave address for the SOT-23 package has been factory programmed to be “0100 000” (LTC1669) and “0100 001” (LTC1669-1). If another address is required, please consult the factory.

Command Byte

7	6	5	4	3	2	1	0
X	X	X	X	X	BG	SD	SY
SY	1 0	Allows update on Acknowledge of SYNC Address only Update on Stop condition only (Power-On Default)					
SD	1 0	Puts the device in power-down mode Puts the device in standard operating mode (Power-On Default)					
BG	1 0	Selects the internal bandgap reference Selects the supply as the reference (Power-On Default)					
X	X	Don't Care					

The stop condition normally initiates the update of the DAC's output latches. Simultaneous update of more than one DAC or other devices on the bus can be achieved by reissuing new start bit, address, command and data bytes before issuing a final stop condition (which will update all the devices). An alternate way to achieve simultaneous LTC1669 updates is to override the stop condition update by setting the “SY” bit of the command byte. Setting this bit sets the device to update the DAC output latches only at the reception of a SYNC address quick command. The actual update occurs on the rising edge of SCL during the Acknowledge. In this way, all devices can update on the reception of the SYNC address quick command instead of the STOP condition.

A Shutdown (SD) bit = HIGH will put the device in a low power state but retain all data latch information. Shutdown will occur at the reception of a STOP condition. This way shutdown could be synchronized to other devices. The output impedance of the DAC will go to a high impedance state ($\approx 500k\Omega$ to GND).

The Bandgap (BG) bit when set to “0” selects the DAC supply voltage as its voltage reference. The full-scale output of the DAC with this setting is equal to the supply voltage. When the BG bit is set to “1,” the internal bandgap reference ($\approx 1.25V$) is selected as the DAC's reference. The full-scale output voltage for this setting is 2.5V.

Data Bytes

Least Significant Data Byte

7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0

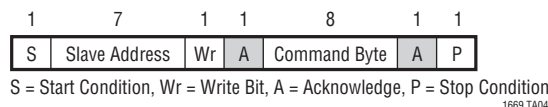
Most Significant Data Byte

7	6	5	4	3	2	1	0
X	X	X	X	X	X	D9	D8

X = Don't care

Send Byte Protocol

The Send Byte protocol used on the LTC1669 is actually a subset of the Write Word protocol described previously. The Send Byte protocol can only be used to send the command byte information to the LTC1669.



The Send Byte protocol is also used whenever the Write Word protocol is interrupted for any reason. Reception of a START or STOP condition after the Acknowledge of the command byte, but before the Acknowledge of the last data byte, will cause both data bytes to be ignored and the command byte to be accepted.

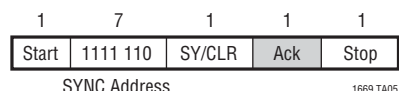
Reception of a START or STOP condition before the Acknowledge of the command byte will cause the interrupted command byte to be ignored.

APPLICATIONS INFORMATION

SYNC Address/Quick Command

In addition to the slave address, the LTC1669 has an address that can be shared by other devices so that they may be updated synchronously. The address is called the SYNC address and uses the quick command protocol.

The SYNC Address is 1111 110



SY/CLR	1	Update output latches on rising edge of SCL during Acknowledge of SYNC Address
	0	Clear all internal latches on rising edge of SCL during Acknowledge of SYNC Address

The SY/CLR bit set high only has meaning when the “SY” bit of the command byte was previously set HIGH. On the otherhand, the SY/CLR bit set LOW will always clear the part, independent of the state of the “SY” bit in the command byte.

Voltage Output

The output amplifier contained in the LTC1669 can source or sink up to 5mA. The output stage swings to within a few millivolts of either supply rail when unloaded and has an equivalent output resistance of 85Ω when driving a load to the rails. The output amplifier is stable driving capacitive loads up to 1000pF.

A small resistor placed in series with the output can be used to achieve stability for any load capacitance greater than 1000pF. For example, a 0.1μF load can be driven by the LTC1669 if a 110Ω series resistance is used. The phase margin of the resulting circuit is 45° and increases monotonically from this point if larger values of resistance, capacitance or both are substituted for the values given.

Rail-to-Rail Output Considerations

As in any rail-to-rail device, the output is limited to voltages within the supply range.

If the DAC offset is negative, the output for the lowest codes limits at 0V as shown in Figure 1b.

Similarly, limiting can occur near full scale when V_{CC} is used as the reference. If $V_{REF} = V_{CC}$ and the DAC full-scale error (FSE) is positive, the output for the highest codes limits at V_{CC} as shown in Figure 1c. No full-scale limiting can occur if the internal reference is used.

Offset and linearity are defined and tested over the region of the DAC transfer function where no output limiting can occur.

Internal Reference

In applications where a predictable output is required that is independent of supply voltage, the LTC1669 has a user-selectable internal reference. Selecting the internal reference will set the full-scale output voltage to 2.5V. This can be useful in applications where the supply voltage is poorly regulated.

Using the LT[®]1460 Micropower Series Reference as a Power Supply for the LTC1669

In applications where the advantages of using the internal reference are required but the full-scale range needs to be greater than 2.5V, an external series reference can be used. The LT1460 is ideal for use as a power supply for the LTC1669 and can provide 3V, 3.3V and 5V full-scale output voltage ranges. The LT1460 provides accuracy, noise immunity and extended supply range to the LTC1669 when the LTC1669 is operated ratiometric to V_{CC} . Since both parts are available in SOT-23 packages, the PC board space for this application is extremely small. See Figure 2.

APPLICATIONS INFORMATION

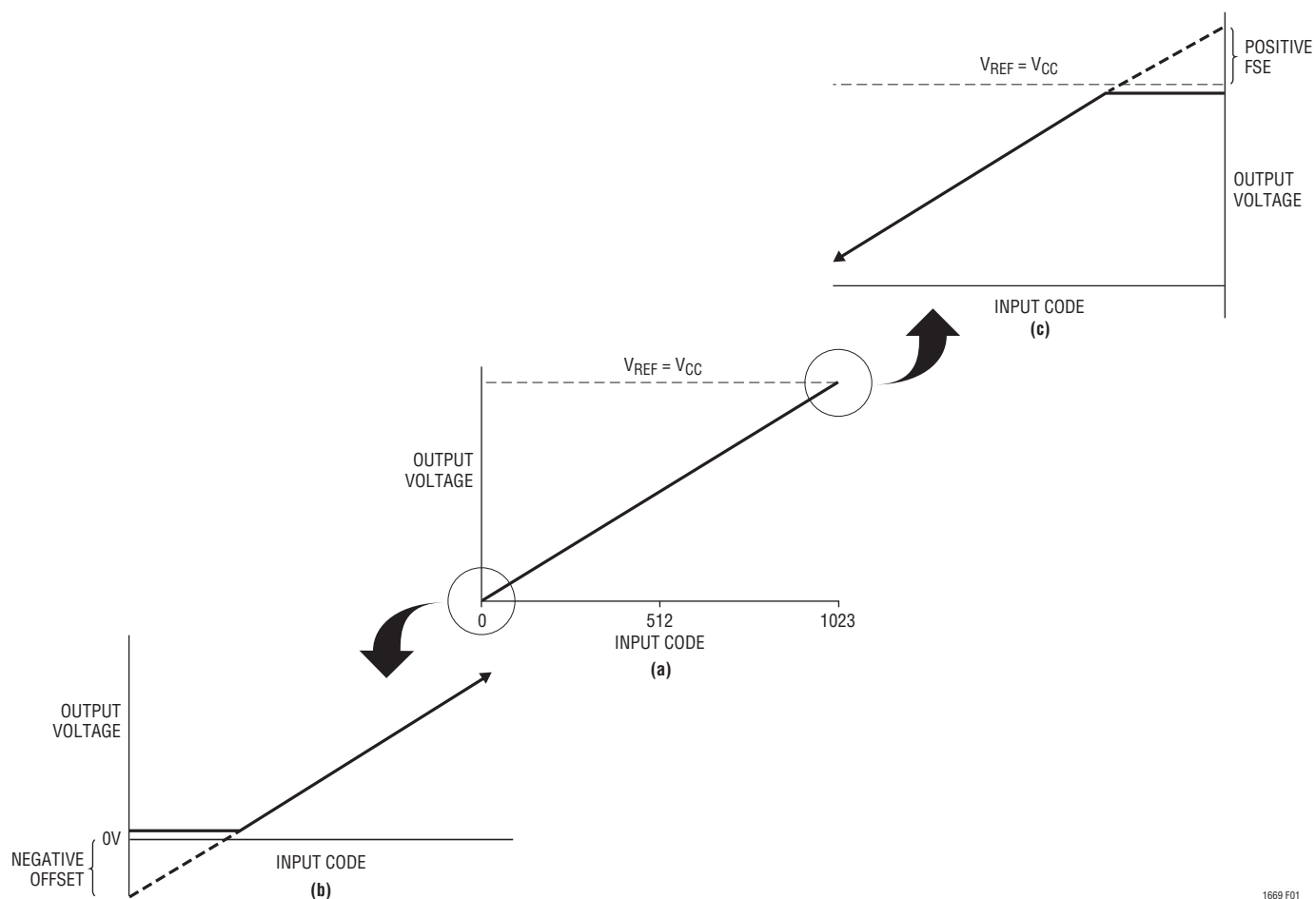


Figure 1. Effects of Rail-to-Rail Operation On a DAC Transfer Curve. (a) Overall Transfer Function (b) Effect of Negative Offset for Codes Near Zero Scale (c) Effect of Positive Full-Scale Error for Input Codes Near Full Scale When $V_{REF} = V_{CC}$

1669 F01

APPLICATIONS INFORMATION

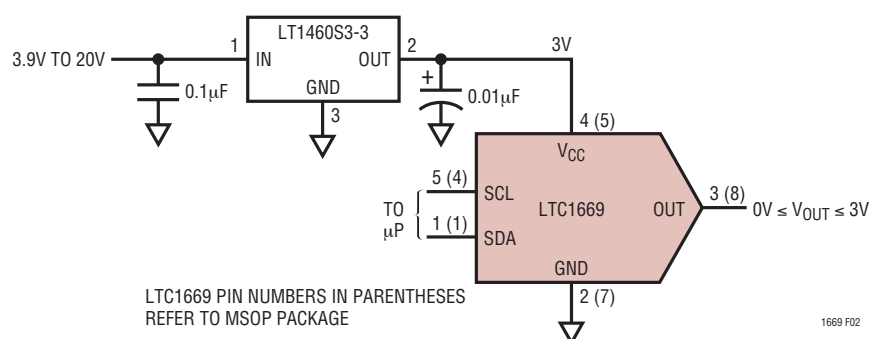
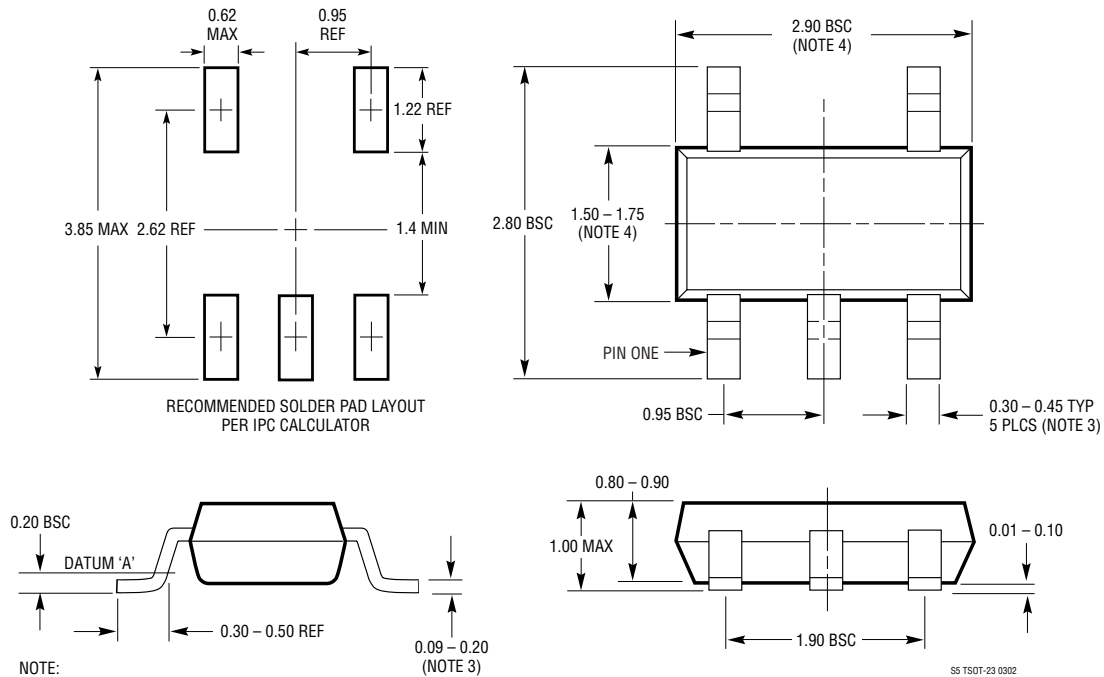


Figure 2. LT1460 As Power Supply for the LTC1669

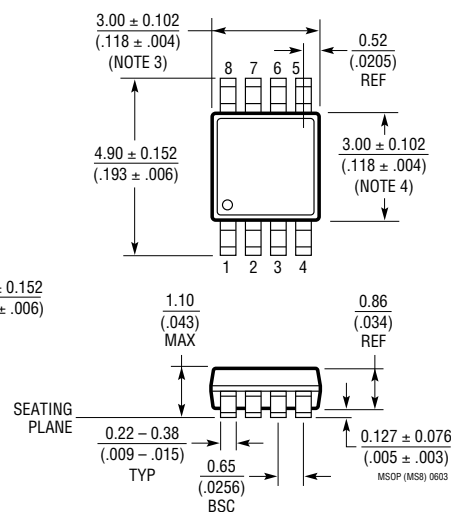
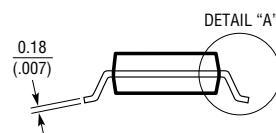
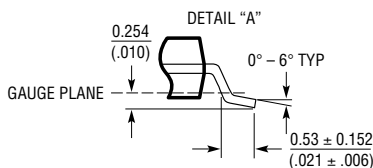
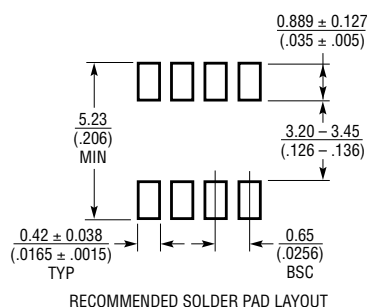
PACKAGE DESCRIPTION

S5 Package
5-Lead Plastic TSOT-23
 (Reference LTC DWG # 05-08-1635)



PACKAGE DESCRIPTION

MS8 Package 8-Lead Plastic MSOP (Reference LTC DWG # 05-08-1660)

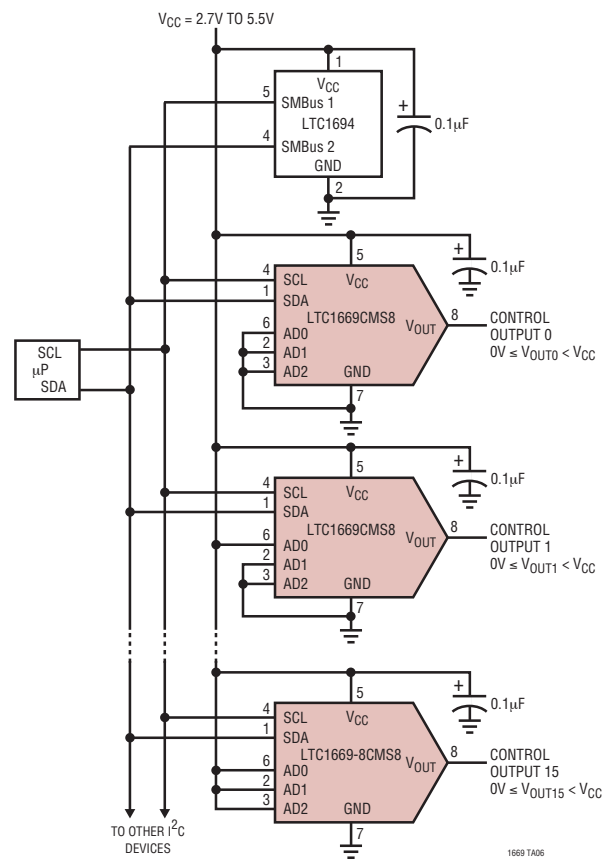


NOTE:

1. DIMENSIONS IN MILLIMETER/(INCH)
2. DRAWING NOT TO SCALE
3. DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.152mm (.006") PER SIDE
4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (.006") PER SIDE
5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (.004") MAX

TYPICAL APPLICATION

Program Up to 16 Control Outputs Per BUS and Place Them Where They Are Needed



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LTC1694	SMBus/I ² C Accelerator	Dual SMBus Accelerator with Active AC and DC Pull-Up Current Sources
LTC1694-1	SMBus/I ² C Accelerator	Dual SMBus Accelerator with Active AC Pull-Up Current Only
DACs		
LTC1659	Single Rail-to-Rail 12-Bit V _{OUT} DAC in 8-Lead MSOP Package. V _{CC} = 2.7V to 5.5V	Low Power Multiplying V _{OUT} DAC. Output Swings from GND to REF. REF Input Can Be Tied to V _{CC} . 3-Wire Interface.
LTC1660/LTC1664	Octal/Quad 10-Bit V _{OUT} DACs in 16-Pin Narrow SSOP	V _{CC} = 2.7V to 5.5V Micropower Rail-to-Rail Output. 3-Wire Interface.
LTC1661	Dual 10-Bit V _{OUT} in 8-Lead MSOP Package	V _{CC} = 2.7V to 5.5V Micropower Rail-to-Rail Output. 3-Wire Interface.
LTC1663	10-Bit V _{OUT} in SOT-23, SMBUS Interface	Pin Compatible with LTC1669
ADCs		
LTC1285/LTC1288	8-Pin SO, 3V Micropower ADCs	1- or 2-Channel, Autoshtutdown
LTC1286/LTC1298	8-Pin SO, 5V Micropower ADCs	1- or 2-Channel, Autoshtutdown