

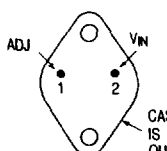
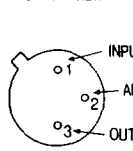
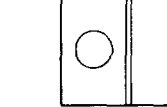
ABSOLUTE MAXIMUM RATINGS

Power Dissipation Internally Limited
Input to Output Voltage Differential 40V
Operating Junction Temperature Range
LT117A/LM117 -55°C to 150°C
LT317A/LM317 0°C to 125°C
Storage Temperature Range
LT117A/LM117 -65°C to 150°C
LT317A/LM317 -65°C to 150°C
Lead Temperature (Soldering, 10 sec.) 300°C

PRECONDITIONING:

100% THERMAL LIMIT BURN-IN

PACKAGE/ORDER INFORMATION

BOTTOM VIEW  K PACKAGE TO-3 METAL CAN (STEEL)	ORDER PART NO. LT117AK LT317AK LM117K LM317K	BOTTOM VIEW  H PACKAGE TO-39 METAL CAN	ORDER PART NO. LT117AH LT317AH LM117H LM317H
CASE IS OUTPUT  T PACKAGE TO-220 PLASTIC	ORDER PART NO. LT317AT LM317T		

ELECTRICAL CHARACTERISTICS (See Note 1) LT117A/LM117

SYMBOL	PARAMETER	CONDITIONS	LT117A			LM117			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{REF}	Reference Voltage	$I_{OUT} = 10\text{mA}$ $T_J = 25^{\circ}\text{C}$	1.238	1.250	1.262				V
		$3\text{V} \leq (V_{IN} - V_{OUT}) \leq 40\text{V}$ $10\text{mA} \leq I_{OUT} \leq I_{max}$, $P \leq P_{max}$	● 1.225	1.250	1.270	1.20	1.25	1.30	V
$\frac{\Delta V_{OUT}}{\Delta V_{IN}}$	Line Regulation	$3\text{V} \leq (V_{IN} - V_{OUT}) \leq 40\text{V}$, (See Note 2)		0.005	0.01		0.01	0.02	%/V
			● 0.01	0.02		0.02	0.05		%/V
$\frac{\Delta V_{OUT}}{\Delta I_{OUT}}$	Load Regulation	$10\text{mA} \leq I_{OUT} \leq I_{max}$, (See Note 2) $V_{OUT} \leq 5\text{V}$ $V_{OUT} \geq 5\text{V}$		5 0.1	15 0.3		5 0.1	15 0.3	mV %
		$V_{OUT} \leq 5\text{V}$ $V_{OUT} \geq 5\text{V}$	● 20 0.3	50 1		20 0.3	50 1		mV %
	Thermal Regulation	$T_A = 25^{\circ}\text{C}$, 20msec Pulse		0.002	0.02		0.03	0.07	%/W
	Ripple Rejection	$V_{OUT} = 10\text{V}$, $f = 120\text{Hz}$ $C_{ADJ} = 0$	● 65			65			dB
		$C_{ADJ} = 10\mu\text{F}$	● 66	80		66	80		dB
I_{ADJ}	Adjust Pin Current		● 50	100		50	100		μA
ΔI_{ADJ}	Adjust Pin Current Change	$10\text{mA} \leq I_{OUT} \leq I_{max}$ $2.5\text{V} \leq (V_{IN} - V_{OUT}) \leq 40\text{V}$	● 0.2	5		0.2	5		μA
I_{min}	Minimum Load Current	$(V_{IN} - V_{OUT}) = 40\text{V}$	● 3.5	5		3.5	5		mA
	Current Limit	$(V_{IN} - V_{OUT}) \leq 15\text{V}$ K Package H Package	● 1.5 0.5	2.2 0.8		1.5 0.5	2.2 0.8		A A
		$(V_{IN} - V_{OUT}) = 40\text{V}$, $T_J = 25^{\circ}\text{C}$ K Package H Package	0.3 0.15	0.5 0.2		0.3 0.15	0.4 0.2		A A
$\frac{\Delta V_{OUT}}{\Delta T_{emp}}$	Temperature Stability	$-55^{\circ}\text{C} \leq T_J \leq +150^{\circ}\text{C}$		1	2		1		%
$\frac{\Delta V_{OUT}}{\Delta T_{ime}}$	Long Term Stability	$T_A = 125^{\circ}\text{C}$		0.3	1		0.3	1	%
e_n	RMS Output Noise (% of V_{OUT})	$T_A = 25^{\circ}\text{C}$, $10\text{Hz} \leq f \leq 10\text{kHz}$		0.001			0.001		%
θ_{jc}	Thermal Resistance Junction to Case	H Package K Package		12 2.3	15 3		12 2.3	15 3	$^{\circ}\text{C/W}$ $^{\circ}\text{C/W}$

ELECTRICAL CHARACTERISTICS (See Note 1) LT317A/LM317

SYMBOL	PARAMETER	CONDITIONS	LT317A			LM317			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{REF}	Reference Voltage	$I_{OUT} = 10mA$ $T_j = 25^\circ C$	1.238	1.250	1.262				V
		$3V \leq (V_{IN} - V_O) \leq 40V$ $10mA \leq I_{OUT} \leq I_{MAX}$, $P \leq P_{MAX}$	● 1.225	1.250	1.270	1.20	1.25	1.30	V
$\frac{\Delta V_{OUT}}{\Delta V_{IN}}$	Line Regulation	$3V \leq (V_{IN} - V_{OUT}) \leq 40V$, (See Note 2)		0.005	0.01		0.01	0.04	%/V
			●	0.01	0.02		0.02	0.07	%/V
$\frac{\Delta V_{OUT}}{\Delta I_{OUT}}$	Load Regulation	$10mA \leq I_{OUT} \leq I_{MAX}$, (See Note 2) $V_O \leq 5V$ $V_O \geq 5V$		5 0.1	25 0.5		5 0.1	25 0.5	mV %
		$V_O \leq 5V$ $V_O \geq 5V$	● ●	20 0.3	50 1		20 0.3	70 1.5	mV %
	Thermal Regulation	$T_A = 25^\circ C$, 20msec Pulse	●	0.002	0.02		0.04	0.07	%/W
	Ripple Rejection	$V_O = 10V$, $f = 120Hz$ $C_{ADJ} = 0$		65			65		dB
		$C_{ADJ} = 10\mu F$		66	80		66	80	dB
I_{ADJ}	Adjust Pin Current			50	100		50	100	μA
ΔI_{ADJ}	Adjust Pin Current Change	$10mA \leq I_{OUT} \leq I_{MAX}$ $2.5V \leq (V_{IN} - V_{OUT}) \leq 40V$	●	0.2	5		0.2	5	μA
I_{min}	Minimum Load Current	$(V_{IN} - V_{OUT}) = 40V$	●	3.5	10		3.5	10	mA
	Current Limit	$(V_{IN} - V_{OUT}) \leq 15V$ K and T Package H Package	● ●	1.5 0.5	2.2 0.8		1.5 0.5	2.2 0.8	A A
		$(V_{IN} - V_{OUT}) = 40V$, $T_j = 25^\circ C$ K and T Package H Package		0.15 0.075	0.4 0.2		0.15 0.075	0.4 0.2	A A
$\frac{\Delta V_{OUT}}{\Delta Temp}$	Temperature Stability	$0^\circ C \leq T_j \leq 125^\circ C$		1	2		1		%
$\frac{\Delta V_{OUT}}{\Delta Time}$	Long Term Stability	$T_A = 125^\circ C$		0.3	1		0.3	1	%
e_n	RMS Output Noise (% of V_{OUT})	$T_A = 25^\circ C$, $10Hz \leq f \leq 10kHz$		0.001			0.001		%
θ_{jc}	Thermal Resistance Junction to Case	H Package K Package T Package		12 2.3 4	15 3 5		12 2.3 4	15 3	$^\circ C/W$ $^\circ C/W$ $^\circ C/W$

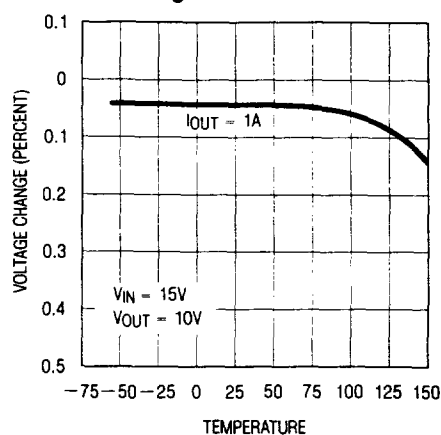
The ● denotes the specifications which apply over the full operating temperature range.

Note 1: Unless otherwise specified, these specifications apply for $V_{IN} - V_{OUT} = 5V$; and $I_{OUT} = 0.1A$ for the TO-39 and $I_{OUT} = 0.5A$ for the TO-3 and TO-220 packages. Although power dissipation is internally limited, these specifications are applicable for power dissipations of 2W for the TO-39, and 20W for the TO-3 and TO-220. I_{MAX} is 1.5A for the TO-3 and TO-220 packages and 0.5A for the TO-39.

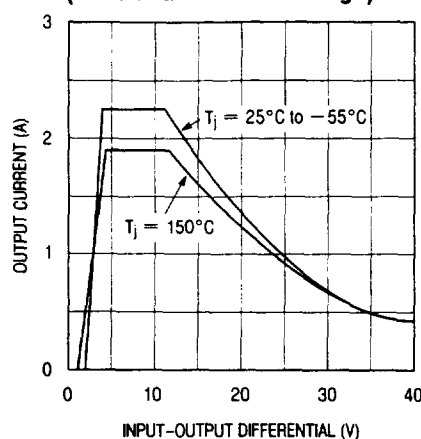
Note 2: Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specification for thermal regulation. Load regulation is measured on the output pin at a point $\frac{1}{8}$ " below the base of the K and H package and at the junction of the wide and narrow portion of the lead on the T package.

TYPICAL PERFORMANCE CHARACTERISTICS

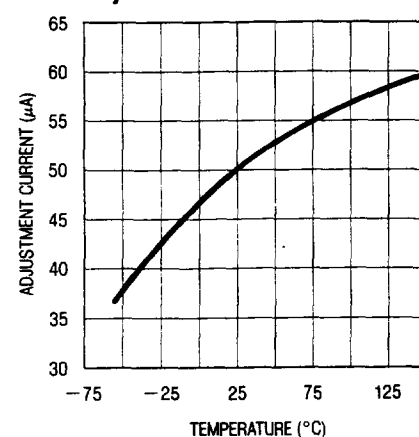
Load Regulation



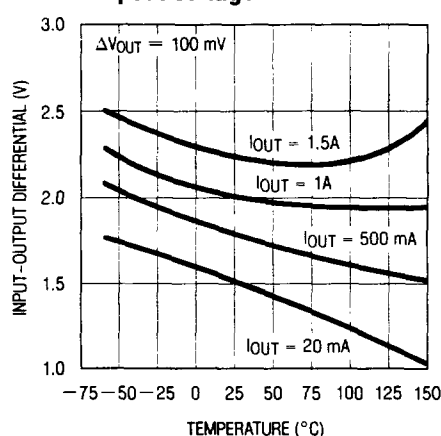
Current Limit (T0-3 and T0-220 Package)



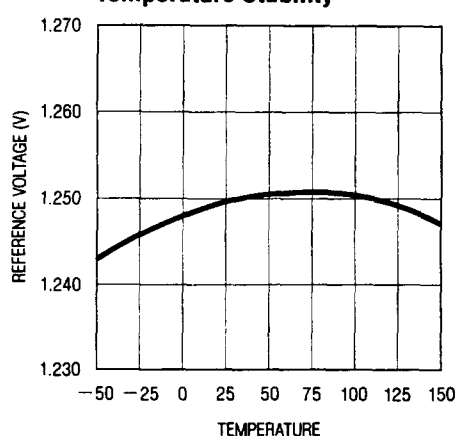
Adjustment Current



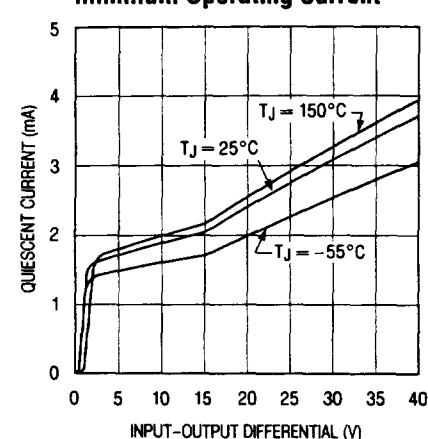
Dropout Voltage



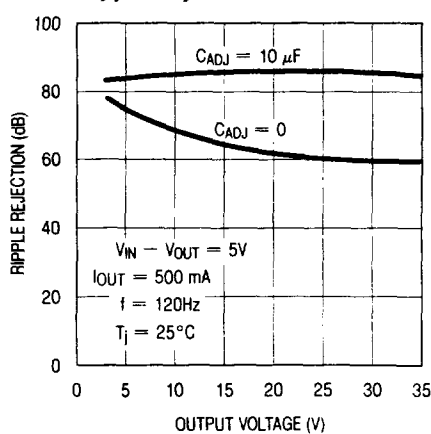
Temperature Stability



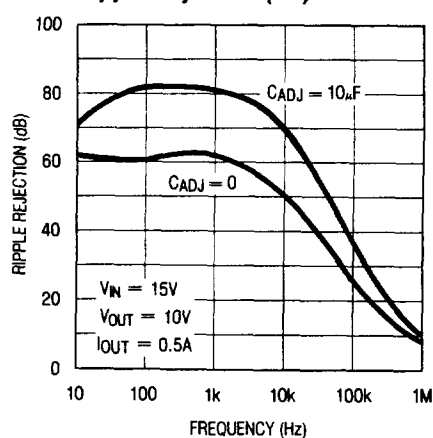
Minimum Operating Current



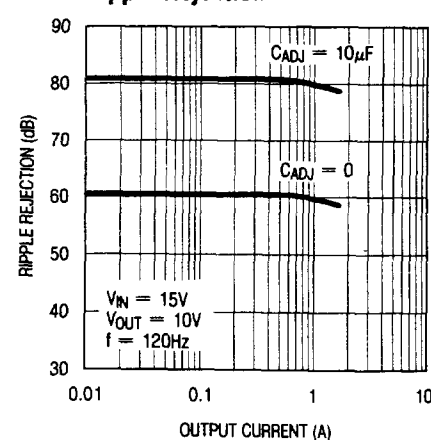
Ripple Rejection



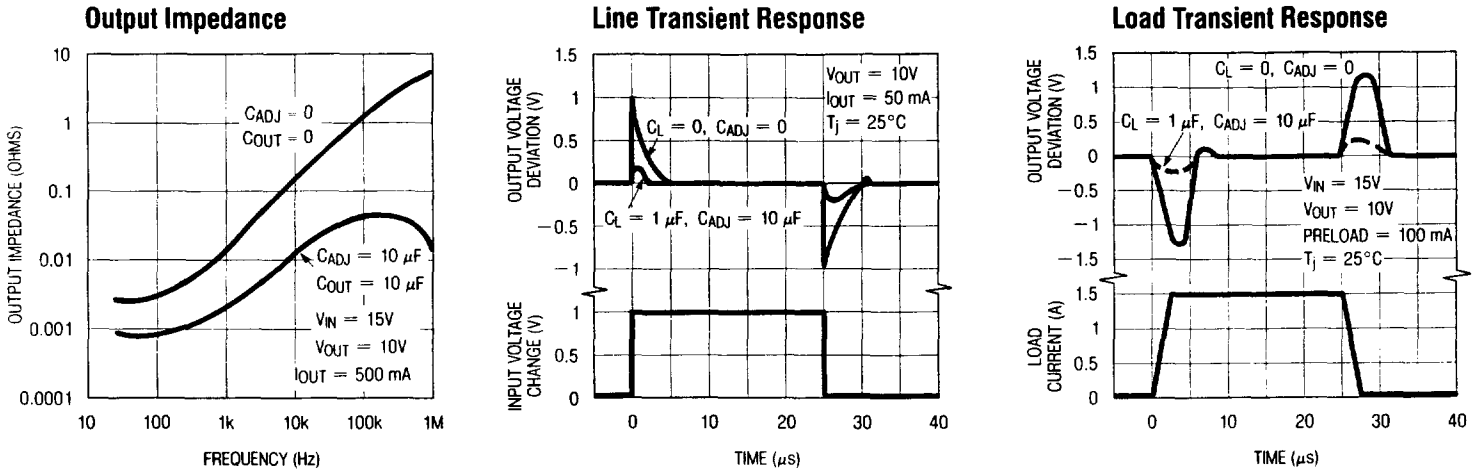
Ripple Rejection (dB)



Ripple Rejection



TYPICAL PERFORMANCE CHARACTERISTICS



APPLICATIONS INFORMATION

General: The LT117A develops a 1.25V reference voltage between the output and the adjustable terminal (see Figure 1). By placing a resistor, R1, between these two terminals, a constant current is caused to flow through R1 and down through R2 to set the overall output voltage. Normally this current is the specified minimum load current of 5mA or 10mA.

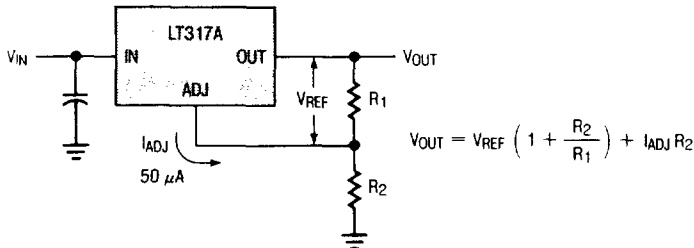


Figure 1

Because I_{ADJ} is very small and constant when compared with the current through R1, it represents a small error and can usually be ignored.

It is easily seen from the above equation, that even if the resistors were of exact value, the accuracy of the output is limited by the accuracy of V_{REF} . Earlier adjustable regulators had a reference tolerance of $\pm 4\%$. This tolerance is dangerously close to the $\pm 5\%$ supply tolerance required in many logic and analog systems. Further, many 1% resistors can drift $0.01\%/^\circ C$ adding another 1% to the output voltage tolerance.

For example, using 2% resistors and $\pm 4\%$ tolerance for V_{REF} , calculations will show that the expected range of a 5V regulator design would be $4.66V \leq V_{OUT} \leq 5.36V$ or approximately $\pm 7\%$. If the same example were used for a 15V regulator, the expected tolerance would be $\pm 8\%$. With these results most applications require some method of trimming, usually a trim pot. This solution is both expensive and not conducive to volume production.

One of the enhancements of Linear Technology's adjustable regulators over existing devices is tightened initial tolerance. This allows relatively inexpensive 1% or 2% film resistors to be used for R1 and R2 while setting output voltage within an acceptable tolerance range.

With a guaranteed 1% reference, a 5V power supply design, using $\pm 2\%$ resistors, would have a worst case manufacturing tolerance of $\pm 4\%$. If 1% resistors were used, the tolerance would drop to $\pm 2.5\%$. A plot of the worst case output voltage tolerance as a function of resistor tolerance is shown on the front page.

LT117A/LT317A LM117/LM317

For convenience, a table of standard 1% resistor values is shown below.

Table of ½% and 1% Standard Resistance Values

1.00	1.47	2.15	3.16	4.64	6.81
1.02	1.50	2.21	3.24	4.75	6.98
1.05	1.54	2.26	3.32	4.87	7.15
1.07	1.58	2.32	3.40	4.99	7.32
1.10	1.62	2.37	3.48	5.11	7.50
1.13	1.65	2.43	3.57	5.23	7.68
1.15	1.69	2.49	3.65	5.36	7.87
1.18	1.74	2.55	3.74	5.49	8.06
1.21	1.78	2.61	3.83	5.62	8.25
1.24	1.82	2.67	3.92	5.76	8.45
1.27	1.87	2.74	4.02	5.90	8.66
1.30	1.91	2.80	4.12	6.04	8.87
1.33	1.96	2.87	4.22	6.19	9.09
1.37	2.00	2.94	4.32	6.34	9.31
1.40	2.05	3.01	4.42	6.49	9.53
1.43	2.10	3.09	4.53	6.65	9.76

Standard Resistance Values are obtained from the Decade Table by multiplying by multiples of 10. As an example, 1.21 can represent 1.21Ω, 12.1Ω, 121Ω, 1.21KΩ etc.

Bypass Capacitors: Input bypassing using a 1μF tantalum or 25μF electrolytic is recommended when the input filter capacitors are more than 5 inches from the device. Improved ripple rejection (80 dB) can be accomplished by adding a 10μF capacitor from the adjust pin to ground. Increasing the size of the capacitor to 20μF will help ripple rejection at low output voltage since the reactance of this capacitor should be small compared to the voltage setting resistor, R2. For improved AC transient response and to prevent the possibility of oscillation due to unknown reactive load, a 1μF capacitor is also recommended at the output. Because of their low impedance at high frequencies, the best type of capacitor to use is solid tantalum.

Protection Diodes: The LT117A/317A do not require a protection diode from the adjustment terminal to the output (see Figure 2). Improved internal circuitry

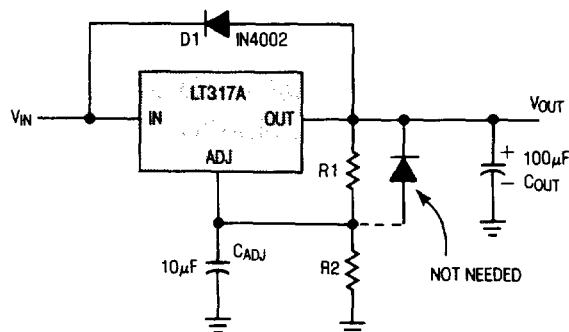


Figure 2

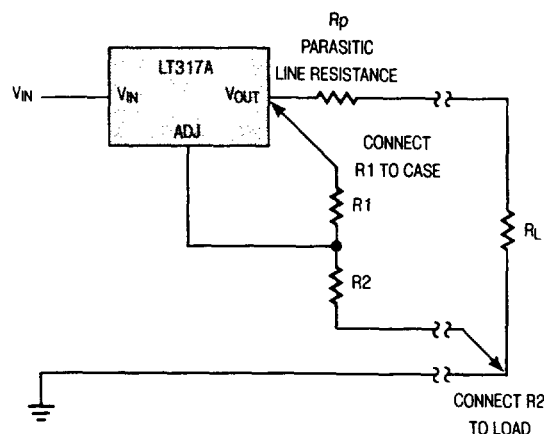
eliminates the need for this diode when the adjustment pin is bypassed with a capacitor to improve ripple rejection.

If a very large output capacitor is used, such as a 100μF shown in Figure 2, the regulator could be damaged or destroyed if the input is accidentally shorted to ground or crowbarred. This is due to the output capacitor discharging into the output terminal of the regulator. To prevent damage a diode D1 is recommended to safely discharge the capacitor.

Load Regulation: Because the LT117A is a three-terminal device, it is not possible to provide true remote load sensing. Load regulation will be limited by the resistance of the wire connecting the regulator to the load. For the data sheet specification, regulation is measured at the bottom of the package. Negative side sensing is a true Kelvin connection, with the bottom of the output divider returned to the negative side of the load. Although it may not be immediately obvious, best load regulation is obtained when the top of the divider is connected *directly* to the case *not to the load*. This is illustrated in Figure 3. If R1 were connected to the load, the effective resistance between the regulator and the load would be

$$R_p \times \left(\frac{R_2 + R_1}{R_1} \right), R_p = \text{Parasitic Line Resistance.}$$

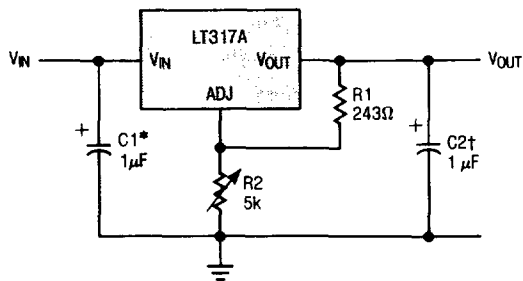
Connected as shown, R_p is not multiplied by the divider ratio. R_p is about 0.004Ω per foot using 16 gauge wire. This translates to 4mV/ft at 1A load current, so it is important to keep the positive lead between regulator and load as short as possible.



Connections for Best Load Regulation
Figure 3

TYPICAL APPLICATIONS

1.2V–25V Adjustable Regulator

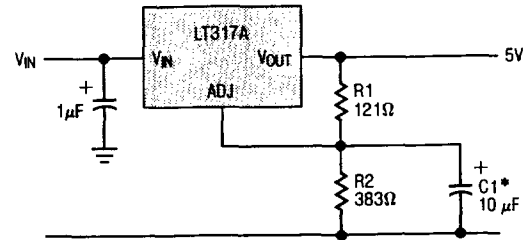


† Optional — improves transient response

* Needed if device is far from filter capacitors

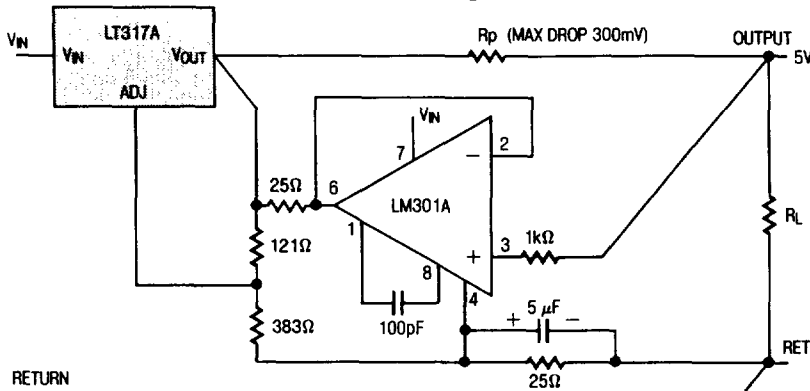
$$V_{OUT} = 1.25V \left(1 + \frac{R_2}{R_1} \right)$$

Improving Ripple Rejection

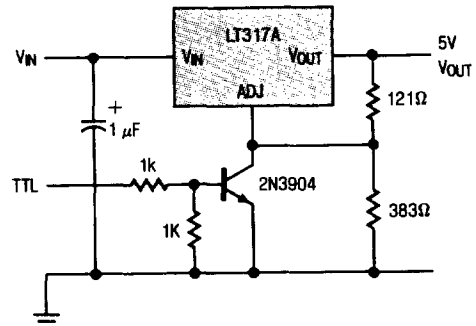


* C1 IMPROVES RIPPLE REJECTION
XC SHOULD BE SMALL
COMPARED TO R2

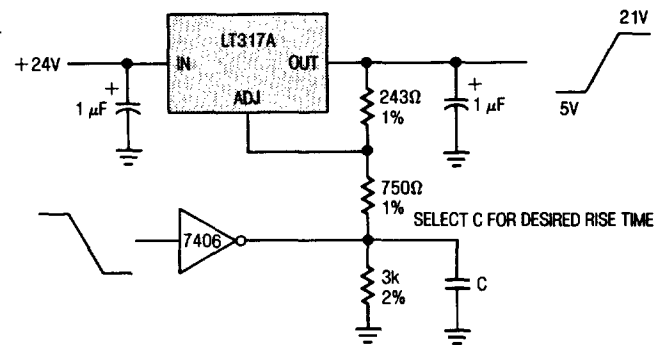
Remote Sensing



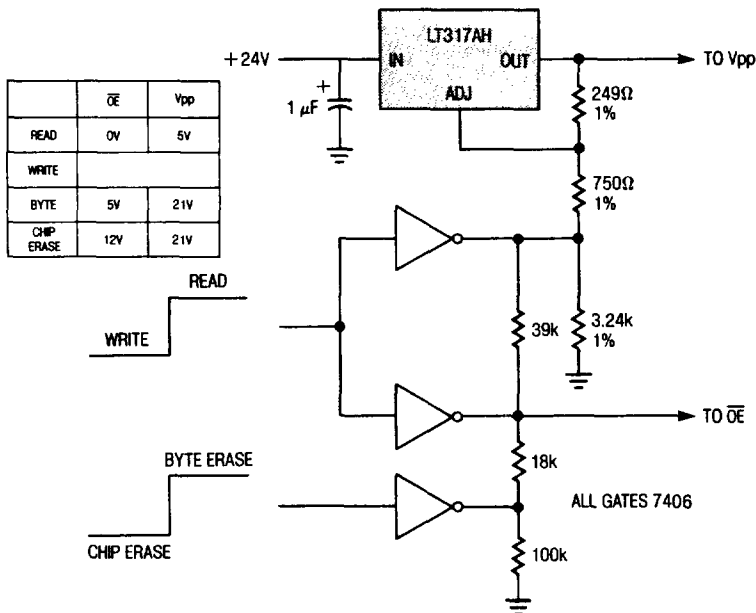
5V Regulator with Shut Down



21V Programming Supply for UV PROM/EEROM



2816 EEPROM Supply Programmer for Read/Write Control



Temperature Compensated Lead Acid Battery Charger

