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REVISION HISTORY

6/20—Rev. D to Rev. E

This Rev E product data sheet removes the Flash Memory section, flash memory specifications, and all obsolete models that include 16M bit SPI flash memory.

These changes are reflected in the following sections:

Changes to Memory 1
Changes to Peripherals 1
Changes to Functional Block Diagram 1
Changes to Processor Comparison 3
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GENERAL DESCRIPTION

The ADSP-BF512/ADSP-BF514/ADSP-BF516/ADSP-BF518 processors are members of the Blackfin® family of products, incorporating the Analog Devices/Intel Micro Signal Architecture (MSA). Blackfin processors combine a dual-MAC state-of-the-art signal processing engine, the advantages of a clean, orthogonal RISC-like microprocessor instruction set, and single-instruction, multiple-data (SIMD) multimedia capabilities into a single instruction-set architecture.

The processors are completely code compatible with other Blackfin processors.

Table 1. Processor Comparison

Feature		ADSP-BF512	ADSP-BF514	ADSP-BF516	ADSP-BF518
IEEE-1588		–	–	–	1
Ethernet MAC		–	–	1	1
RSI		–	1	1	1
TWI		1	1	1	1
SPORTs		2	2	2	2
UARTs		2	2	2	2
SPIs		2	2	2	2
GP Timers		8	8	8	8
Watchdog Timers		1	1	1	1
RTC		1	1	1	1
PPI		1	1	1	1
Rotary Counter		1	1	1	1
3-Phase PWM Pairs		3	3	3	3
GPIOs		40	40	40	40
Memory (bytes)	L1 Instruction SRAM	32K			
	L1 Instruction SRAM/Cache	16K			
	L1 Data SRAM	32K			
	L1 Data SRAM/Cache	32K			
	L1 Scratchpad	4K			
	L3 Boot ROM	32K			
Maximum Speed Grade		400 MHz			
Package Options		176-Lead LQFP_EP (with Exposed Pad)			
		168-Ball CSP_BGA			

By integrating a rich set of industry-leading system peripherals and memory, Blackfin processors are the platform of choice for next-generation applications that require RISC-like programmability, multimedia support, and leading-edge signal processing in one integrated package.

PORTABLE LOW POWER ARCHITECTURE

Blackfin processors provide world-class power management and performance. They are produced with a low power and low voltage design methodology and feature on-chip dynamic power management, which is the ability to vary both the voltage and frequency of operation to significantly lower overall power consumption. This capability can result in a substantial reduction in power consumption, compared with just varying the frequency of operation. This allows longer battery life for portable appliances.

SYSTEM INTEGRATION

The ADSP-BF51x processors are highly integrated system-on-a-chip solutions for the next generation of embedded network connected applications. By combining industry-standard interfaces with a high performance signal processing core, cost-effective applications can be developed quickly, without the need for costly external components. The system peripherals include an IEEE-compliant 802.3 10/100 Ethernet MAC with IEEE-1588 support (ADSP-BF518 only), an RSI controller, a TWI controller, two UART ports, two SPI ports, two serial ports (SPORTs), nine general-purpose 32-bit timers (eight with PWM capability), 3-phase PWM for motor control, a real-time clock, a watchdog timer, and a parallel peripheral interface (PPI).

BLACKFIN PROCESSOR CORE

As shown in Figure 2, the Blackfin processor core contains two 16-bit multipliers, two 40-bit accumulators, two 40-bit ALUs, four video ALUs, and a 40-bit shifter. The computation units process 8-, 16-, or 32-bit data from the register file.

The compute register file contains eight 32-bit registers. When performing compute operations on 16-bit operand data, the register file operates as 16 independent 16-bit registers. All operands for compute operations come from the multiported register file and instruction constant fields.

Each MAC can perform a 16-bit by 16-bit multiply in each cycle, accumulating the results into the 40-bit accumulators. Signed and unsigned formats, rounding, and saturation are supported.

The ALUs perform a traditional set of arithmetic and logical operations on 16-bit or 32-bit data. In addition, many special instructions are included to accelerate various signal processing tasks. These include bit operations such as field extract and population count, modulo 2^{32} multiply, divide primitives, saturation and rounding, and sign/exponent detection. The set of video instructions include byte alignment and packing operations, 16-bit and 8-bit adds with clipping, 8-bit average operations, and 8-bit subtract/absolute value/accumulate (SAA) operations. The compare/select and vector search instructions are also provided.

For certain instructions, two 16-bit ALU operations can be performed simultaneously on register pairs (a 16-bit high half and 16-bit low half of a compute register). If the second ALU is used, quad 16-bit operations are possible.

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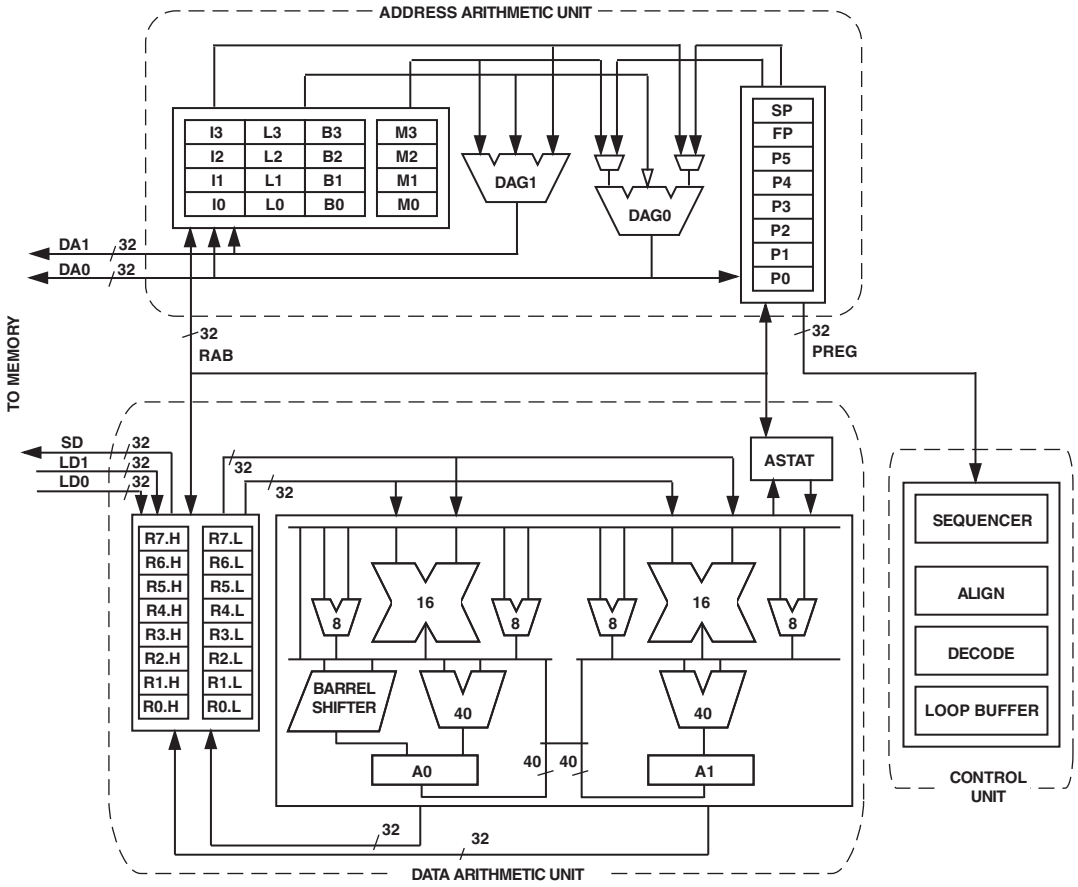

Figure 2. Blackfin Processor Core

The 40-bit shifter can perform shifts and rotates and is used to support normalization, field extract, and field deposit instructions.

The program sequencer controls the flow of instruction execution, including instruction alignment and decoding. For program flow control, the sequencer supports PC relative and indirect conditional jumps (with static branch prediction), and subroutine calls. Hardware is provided to support zero-overhead looping. The architecture is fully interlocked, meaning that the programmer need not manage the pipeline when executing instructions with data dependencies.

The address arithmetic unit provides two addresses for simultaneous dual fetches from memory. It contains a multiported register file consisting of four sets of 32-bit index, modify, length, and base registers (for circular buffering), and eight additional 32-bit pointer registers (for C-style indexed stack manipulation).

Blackfin processors support a modified Harvard architecture in combination with a hierarchical memory structure. Level 1 (L1) memories are those that typically operate at the full processor speed with little or no latency. At the L1 level, the instruction memory holds instructions only. The two data memories hold data, and a dedicated scratchpad data memory stores stack and local variable information.

In addition, multiple L1 memory blocks are provided, offering a configurable mix of SRAM and cache. The memory management unit (MMU) provides memory protection for individual tasks that may be operating on the core and can protect system registers from unintended access.

The architecture provides three modes of operation: user mode, supervisor mode, and emulation mode. User mode has restricted access to certain system resources, thus providing a protected software environment, while supervisor mode has unrestricted access to the system and core resources.

The Blackfin processor instruction set has been optimized so that 16-bit opcodes represent the most frequently used instructions, resulting in excellent compiled code density. Complex DSP instructions are encoded into 32-bit opcodes, representing fully featured multifunction instructions. Blackfin processors support a limited multi-issue capability, where a 32-bit instruction can be issued in parallel with two 16-bit instructions, allowing the programmer to use many of the core resources in a single instruction cycle.

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The Blackfin processor assembly language uses an algebraic syntax for ease of coding and readability. The architecture has been optimized for use in conjunction with the C/C++ compiler, resulting in fast and efficient software implementations.

MEMORY ARCHITECTURE

The ADSP-BF51x processors view memory as a single unified 4G byte address space, using 32-bit addresses. All resources, including internal memory, external memory, and I/O control registers, occupy separate sections of this common address space. The memory portions of this address space are arranged in a hierarchical structure to provide a good cost/performance balance of some very fast, low-latency on-chip memory as cache or SRAM, and larger, lower-cost and performance off-chip memory systems. The memory map for both internal and external memory space is shown in Figure 3.

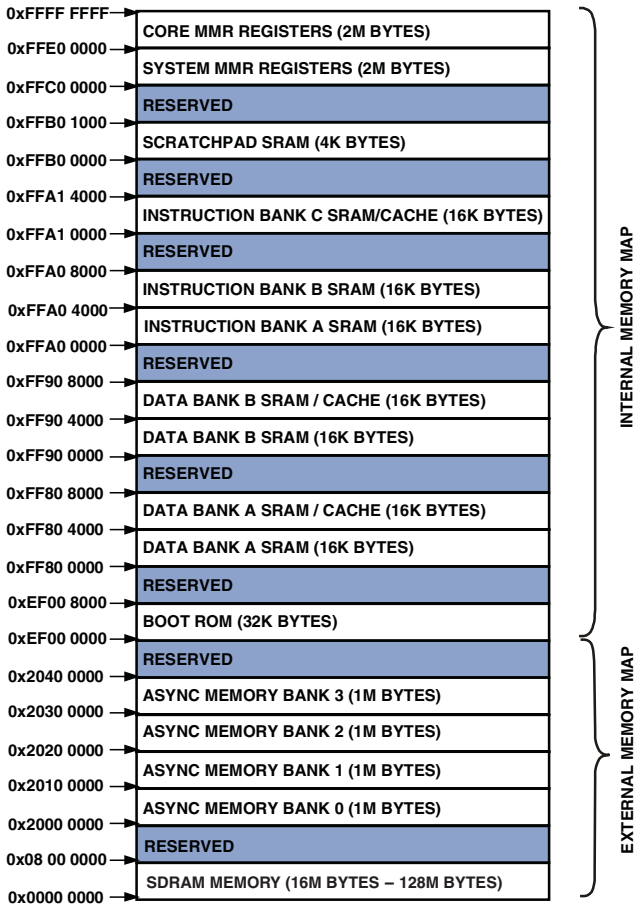

Figure 3. ADSP-BF51x Internal/External Memory Map

The on-chip L1 memory system is the highest-performance memory available to the Blackfin processor. The off-chip memory system, accessed through the external bus interface unit (EBIU), provides expansion with SDRAM, flash memory, and SRAM, optionally accessing up to 132M bytes of physical memory.

The memory DMA controller provides high bandwidth data-movement capability. It can perform block transfers of code or data between the internal memory and the external memory spaces.

Internal (On-Chip) Memory

The ADSP-BF51x processors have three blocks of on-chip memory that provide high bandwidth access to the core.

The first block is the L1 instruction memory, consisting of 48K bytes SRAM, of which 16K bytes can be configured as a four-way set-associative cache. This memory is accessed at full processor speed.

The second on-chip memory block is the L1 data memory, consisting of up to two banks of up to 32K bytes each. Each memory bank is configurable, offering both cache and SRAM functionality. This memory block is accessed at full processor speed.

The third memory block is a 4K byte scratchpad SRAM which runs at the same speed as the L1 memories, but is only accessible as data SRAM and cannot be configured as cache memory.

External (Off-Chip) Memory

External memory is accessed via the EBIU. This 16-bit interface provides a glueless connection to a bank of synchronous DRAM (SDRAM) as well as up to four banks of asynchronous memory devices including flash, EPROM, ROM, SRAM, and memory mapped I/O devices.

The SDRAM controller can be programmed to interface to up to 128M bytes of SDRAM. A separate row can be open for each SDRAM internal bank, and the SDRAM controller supports up to four internal SDRAM banks, improving overall performance.

The asynchronous memory controller can be programmed to control up to four banks of devices with very flexible timing parameters for a wide variety of devices. Each bank occupies a 1M byte segment regardless of the size of the devices used, so that these banks are only contiguous if each is fully populated with 1M byte of memory.

One-Time Programmable Memory

The processors have 64K bits of one-time programmable non-volatile memory that can be programmed by the developer only once. It includes the array and logic to support read access and programming. Additionally, its pages can be write protected.

The OTP memory allows both public and private data to be stored on-chip. In addition to storing public and private key data for applications requiring security, OTP allows developers to store completely user-definable data such as customer ID, product ID, and MAC address. Therefore, generic parts can be supplied which are then programmed and protected by the developer within this non-volatile memory.

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I/O Memory Space

The processors do not define a separate I/O space. All resources are mapped through the flat 32-bit address space. On-chip I/O devices have their control registers mapped into memory-mapped registers (MMRs) at addresses near the top of the 4G byte address space. These are separated into two smaller blocks, one which contains the control MMRs for all core functions, and the other which contains the registers needed for setup and control of the on-chip peripherals outside of the core. The MMRs are accessible only in supervisor mode and appear as reserved space to on-chip peripherals.

Booting from ROM

The processors contain a small on-chip boot kernel, which configures the appropriate peripheral for booting. If the processors are configured to boot from boot ROM memory space, the processor starts executing from the on-chip boot ROM. For more information, see Booting Modes.

EVENT HANDLING

The event controller handles all asynchronous and synchronous events to the processor. The processors provide event handling that supports both nesting and prioritization. Nesting allows multiple event service routines to be active simultaneously. Prioritization ensures that servicing of a higher priority event takes precedence over servicing of a lower priority event.

The controller provides support for five different types of events:

- Emulation—An emulation event causes the processor to enter emulation mode, allowing command and control of the processor through the JTAG interface.
- Reset—This event resets the processor.
- Nonmaskable Interrupt (NMI)—The NMI event can be generated by the software watchdog timer or by the NMI input signal to the processor. The NMI event is frequently used as a power-down indicator to initiate an orderly shutdown of the system.
- Exceptions—Events that occur synchronously to program flow; that is, the exception is taken before the instruction is allowed to complete. Conditions such as data alignment violations and undefined instructions cause exceptions.
- Interrupts—Events that occur asynchronously to program flow. They are caused by input signals, timers, and other peripherals, as well as by an explicit software instruction.

Each event type has an associated register to hold the return address and an associated return-from-event instruction. When an event is triggered, the state of the processor is saved on the supervisor stack.

The event controller consists of two stages, the core event controller (CEC) and the system interrupt controller (SIC). The core event controller works with the system interrupt controller to prioritize and control all system events. Conceptually, interrupts from the peripherals enter into the SIC, and are then routed directly into the general-purpose interrupts of the CEC.

Core Event Controller (CEC)

The CEC supports nine general-purpose interrupts (IVG15–7), in addition to the dedicated interrupt and exception events. Of these general-purpose interrupts, the two lowest priority interrupts (IVG15–14) are recommended to be reserved for software interrupt handlers, leaving seven prioritized interrupt inputs to support the peripherals of the processors. The inputs to the CEC, identifies their names in the event vector table (EVT), and lists their priorities are described in the *ADSP-BF51x Blackfin Processor Hardware Reference Manual* “System Interrupts” chapter.

System Interrupt Controller (SIC)

The system interrupt controller provides the mapping and routing of events from the many peripheral interrupt sources to the prioritized general-purpose interrupt inputs of the CEC. Although the processors provide a default mapping, the user can alter the mappings and priorities of interrupt events by writing the appropriate values into the interrupt assignment registers (SIC_IARx). See the *ADSP-BF51x Blackfin Processor Hardware Reference Manual* “System Interrupts” chapter for the inputs into the SIC and the default mappings into the CEC.

The SIC allows further control of event processing by providing three pairs of 32-bit interrupt control and status registers. Each register contains a bit corresponding to each of the peripheral interrupt events. For more information, see the *ADSP-BF51x Blackfin Processor Hardware Reference Manual* “System Interrupts” chapter.

DMA CONTROLLERS

The ADSP-BF51x processors have multiple independent DMA channels that support automated data transfers with minimal overhead for the processor core. DMA transfers can occur between the processor's internal memories and any of its DMA-capable peripherals. Additionally, DMA transfers can be accomplished between any of the DMA-capable peripherals and external devices connected to the external memory interfaces, including the SDRAM controller and the asynchronous memory controller. DMA-capable peripherals include the Ethernet MAC, RSI, SPORTs, SPIs, UARTs, and PPI. Each individual DMA-capable peripheral has at least one dedicated DMA channel.

The processors’ DMA controller supports both one-dimensional (1-D) and two-dimensional (2-D) DMA transfers. DMA transfer initialization can be implemented from registers or from sets of parameters called descriptor blocks.

The 2-D DMA capability supports arbitrary row and column sizes up to 64K elements by 64K elements, and arbitrary row and column step sizes up to ±32K elements. Furthermore, the column step size can be less than the row step size, allowing implementation of interleaved data streams. This feature is especially useful in video applications where data can be de-interleaved on the fly.

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Examples of DMA types supported by the DMA controller include:

- A single, linear buffer that stops upon completion
- A circular, auto-refreshing buffer that interrupts on each full or fractionally full buffer
- 1-D or 2-D DMA using a linked list of descriptors
- 2-D DMA using an array of descriptors, specifying only the base DMA address within a common page

In addition to the dedicated peripheral DMA channels, there are two memory DMA channels that transfer data between the various memories of the processor system. This enables transfers of blocks of data between any of the memories—including external SDRAM, ROM, SRAM, and flash memory—with minimal processor intervention. Memory DMA transfers can be controlled by a very flexible descriptor-based methodology or by a standard register-based autobuffer mechanism.

The processors also have an external DMA controller capability via dual external DMA request signals when used in conjunction with the external bus interface unit (EBIU). This functionality can be used when a high speed interface is required for external FIFOs and high bandwidth communications peripherals. It allows control of the number of data transfers for memory DMA. The number of transfers per edge is programmable. This feature can be programmed to allow memory DMA to have an increased priority on the external bus relative to the core.

PROCESSOR PERIPHERALS

The ADSP-BF51x processors contain a rich set of peripherals connected to the core via several high bandwidth buses, providing flexibility in system configuration as well as excellent overall system performance (see Figure 2). The processors contain dedicated network communication modules and high speed serial and parallel ports, an interrupt controller for flexible management of interrupts from the on-chip peripherals or external sources, and power management control functions to tailor the performance and power characteristics of the processor and system to many application scenarios.

All of the peripherals, except for the general-purpose I/O, rotary counter, TWI, three-phase PWM, real-time clock, and timers, are supported by a flexible DMA structure. There are also separate memory DMA channels dedicated to data transfers between the processor's various memory spaces, including external SDRAM and asynchronous memory. Multiple on-chip buses provide enough bandwidth to keep the processor core running along with activity on all of the on-chip and external peripherals.

Real-Time Clock

The real-time clock (RTC) provides a robust set of digital watch features, including current time, stopwatch, and alarm. The RTC is clocked by a 32.768 kHz crystal external to the processors. The RTC peripheral has a dedicated power supply so that it can remain powered up and clocked even when the rest of the processor is in a low power state. The RTC provides several programmable interrupt options, including interrupt per second, minute, hour, or day clock ticks, interrupt on programmable stopwatch countdown, or interrupt at a programmed alarm time.

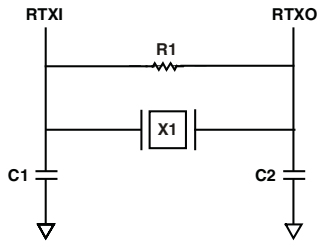
The 32.768 kHz input clock frequency is divided down to a 1 Hz signal by a prescaler. The counter function of the timer consists of four counters: a 60-second counter, a 60-minute counter, a 24-hour counter, and an 32,768-day counter.

When enabled, the alarm function generates an interrupt when the output of the timer matches the programmed value in the alarm control register. There are two alarms: The first alarm is for a time of day. The second alarm is for a day and time of that day.

The stopwatch function counts down from a programmed value, with one-second resolution. When the stopwatch is enabled and the counter underflows, an interrupt is generated.

Like the other peripherals, the RTC can wake up the processor from sleep mode upon generation of any RTC wakeup event. Additionally, an RTC wakeup event can wake up the processor from deep sleep mode or cause a transition from the hibernate state.

Connect RTC signals RTXI and RTXO with external components as shown in Figure 4.



SUGGESTED COMPONENTS:
X1 = ECLIPTEK EC38J (THROUGH-HOLE PACKAGE) OR
EPSON MC405 12 pF LOAD (SURFACE-MOUNT PACKAGE)
C1 = 22 pF
C2 = 22 pF
R1 = 10 MΩ
NOTE: C1 AND C2 ARE SPECIFIC TO CRYSTAL SPECIFIED FOR X1. CONTACT CRYSTAL MANUFACTURER FOR DETAILS. C1 AND C2 SPECIFICATIONS ASSUME BOARD TRACE CAPACITANCE OF 3 pF.

Figure 4. External Components for RTC

Watchdog Timer

The ADSP-BF51x processors include a 32-bit timer that can be used to implement a software watchdog function. A software watchdog can improve system availability by forcing the processor to a known state through generation of a hardware reset, nonmaskable interrupt (NMI), or general-purpose interrupt, if the timer expires before being reset by software. The programmer initializes the count value of the timer, enables the appropriate interrupt, then enables the timer. Thereafter, the software must reload the counter before it counts to zero from the programmed value. This protects the system from remaining in an unknown state where software, which would normally reset the timer, has stopped running due to an external noise condition or software error.

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If configured to generate a hardware reset, the watchdog timer resets both the core and the processor peripherals. After a reset, software can determine if the watchdog was the source of the hardware reset by interrogating a status bit in the watchdog timer control register.

The timer is clocked by the system clock (SCLK) at a maximum frequency of f_{SCLK}.

Timers

There are nine general-purpose programmable timer units in the ADSP-BF51x processors. Eight timers have an external signal that can be configured either as a pulse width modulator (PWM) or timer output, as an input to clock the timer, or as a mechanism for measuring pulse widths and periods of external events. These timers can be synchronized to an external clock input to the several other associated PF signals, an external clock input to the PPI_CLK input signal, or to the internal SCLK.

The timer units can be used in conjunction with the two UARTs to measure the width of the pulses in the data stream to provide a software auto-baud detect function for the respective serial channels.

The timers can generate interrupts to the processor core providing periodic events for synchronization, either to the system clock or to a count of external signals.

In addition to the eight general-purpose programmable timers, a ninth timer is also provided. This extra timer is clocked by the internal processor clock and is typically used as a system tick clock for generation of operating system periodic interrupts.

3-Phase PWM

The processors integrate a flexible and programmable 3-phase PWM waveform generator that can be programmed to generate the required switching patterns to drive a 3-phase voltage source inverter for ac induction (ACIM) or permanent magnet synchronous (PMSM) motor control. In addition, the PWM block contains special functions that considerably simplify the generation of the required PWM switching patterns for control of the electronically commutated motor (ECM) or brushless dc motor (BDCM). Software can enable a special mode for switched reluctance motors (SRM).

Features of the 3-phase PWM generation unit are:

- 16-bit center-based PWM generation unit
- Programmable PWM pulse width
- Single/double update modes
- Programmable dead time and switching frequency
- Twos-complement implementation which permits smooth transition to full ON and full OFF states
- Possibility to synchronize the PWM generation to an external synchronization
- Special provisions for BDCM operation (crossover and output enable functions)
- Wide variety of special switched reluctance (SR) operating modes
- Output polarity and clock gating control
- Dedicated asynchronous PWM shutdown signal

General-Purpose (GP) Counter

A 32-bit GP counter is provided that can sense 2-bit quadrature or binary codes as typically emitted by industrial drives or manual thumb wheels. The counter can also operate in general-purpose up/down count modes. Then, count direction is either controlled by a level-sensitive input signal or by two edge detectors.

A third input can provide flexible zero marker support and can alternatively be used to input the push-button signal of thumb wheels. All three signals have a programmable debouncing circuit.

An internal signal forwarded to the GP timer unit enables one timer to measure the intervals between count events. Boundary registers enable auto-zero operation or simple system warning by interrupts when programmable count values are exceeded.

Serial Ports

The ADSP-BF51x processors incorporate two dual-channel synchronous serial ports (SPORT0 and SPORT1) for serial and multiprocessor communications. The SPORTs support the following features:

Serial port data can be automatically transferred to and from on-chip memory/external memory via dedicated DMA channels. Each of the serial ports can work in conjunction with another serial port to provide TDM support. In this configuration, one SPORT provides two transmit signals while the other SPORT provides the two receive signals. The frame sync and clock are shared.

Serial ports operate in five modes:

- Standard DSP serial mode
- Multichannel (TDM) mode
- I^2S mode
- Packed I^2S mode
- Left-justified mode

Serial Peripheral Interface (SPI) Ports

The processors have two SPI-compatible ports (SPI0 and SPI1) that enable the processor to communicate with multiple SPI-compatible devices.

The SPI interface uses three signals for transferring data: two data signals (master output-slave input–MOSI, and master input-slave output–MISO) and a clock signal (serial clock–SCK). An SPI chip select input signal ($\overline{\text{SPIxSS}}$) lets other SPI devices select the processor, and multiple SPI chip select output signals let the processor select other SPI devices. The SPI select signals are reconfigured general-purpose I/O signals. Using these signals, the SPI port provides a full-duplex, synchronous serial interface, which supports both master/slave modes and multimaster environments.

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The SPI port baud rate and clock phase/polarities are programmable, and it has an integrated DMA channel, configurable to support transmit or receive data streams. The DMA channel of the SPI can only service unidirectional accesses at any given time.

UART Ports

The processors provide two full-duplex universal asynchronous receiver/transmitter (UART) ports, which are fully compatible with PC-standard UARTs. Each UART port provides a simplified UART interface to other peripherals or hosts, supporting full-duplex, DMA-supported, asynchronous transfers of serial data. A UART port includes support for five to eight data bits, and none, even, or odd parity. Optionally, an additional address bit can be transferred to interrupt only addressed nodes in multi-drop bus (MDB) systems. A frame is terminates by one, one and a half, two or two and a half stop bits.

The UART ports support automatic hardware flow control through the Clear To Send (CTS) input and Request To Send (RTS) output with programmable assertion FIFO levels.

To help support the Local Interconnect Network (LIN) protocols, a special command causes the transmitter to queue a break command of programmable bit length into the transmit buffer. Similarly, the number of stop bits can be extended by a programmable inter-frame space.

The capabilities of the UARTs are further extended with support for the Infrared Data Association (IrDA®) serial infrared physical layer link specification (SIR) protocol.

2-Wire Interface (TWI)

The processors include a TWI module for providing a simple exchange method of control data between multiple devices. The TWI is compatible with the widely used $I^2C^®$ bus standard. The TWI module offers the capabilities of simultaneous master and slave operation, support for both 7-bit addressing and multimedia data arbitration. The TWI interface utilizes two signals for transferring clock (SCL) and data (SDA) and supports the protocol at speeds up to 400k bits/sec. The TWI interface signals are compatible with 5 V logic levels.

Additionally, the processor's TWI module is fully compatible with serial camera control bus (SCCB) functionality for easier control of various CMOS camera sensor devices.

Removable Storage Interface (RSI)

The RSI controller, available on the ADSP-BF514/ADSP-BF516/ADSP-BF518 processors, acts as the host interface for multi-media cards (MMC), secure digital memory cards (SD Card), secure digital input/output cards (SDIO), and CE-ATA hard disk drives. The following list describes the main features of the RSI controller.

- Support for a single MMC, SD memory, SDIO card or CE-ATA hard disk drive
- Support for 1-bit and 4-bit SD modes
- Support for 1-bit, 4-bit and 8-bit MMC modes
- Support for 4-bit and 8-bit CE-ATA hard disk drives
- A ten-signal external interface with clock, command, and up to eight data lines
- Card detection using one of the data signals
- Card interface clock generation from SCLK
- SDIO interrupt and read wait features
- CE-ATA command completion signal recognition and disable

10/100 Ethernet MAC

The ADSP-BF516 and ADSP-BF518 processors offer the capability to directly connect to a network by way of an embedded fast Ethernet media access controller (MAC) that supports both 10-BaseT (10M bits/sec) and 100-BaseT (100M bits/sec) operation. The 10/100 Ethernet MAC peripheral on the processor is fully compliant to the IEEE 802.3-2002 standard and it provides programmable features designed to minimize supervision, bus use, or message processing by the rest of the processor system.

Some standard features are:

- Support of MII and RMII protocols for external PHYs
- Full duplex and half duplex modes
- Data framing and encapsulation: generation and detection of preamble, length padding, and FCS
- Media access management (in half-duplex operation): collision and contention handling, including control of retransmission of collision frames and of back-off timing
- Flow control (in full-duplex operation): generation and detection of pause frames
- Station management: generation of MDC/MDIO frames for read-write access to PHY registers
- Operating range for active and sleep operating modes, see Table 39 and Table 40
- Internal loopback from transmit to receive

Some advanced features are:

- Buffered crystal output to external PHY for support of a single crystal system
- Automatic checksum computation of IP header and IP payload fields of Rx frames
- Independent 32-bit descriptor-driven receive and transmit DMA channels
- Frame status delivery to memory through DMA, including frame completion semaphores for efficient buffer queue management in software
- Tx DMA support for separate descriptors for MAC header and payload to eliminate buffer copy operations
- Convenient frame alignment modes support even 32-bit alignment of encapsulated receive or transmit IP packet data in memory after the 14-byte MAC header

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- Programmable Ethernet event interrupt supports any combination of:
  - Selected receive or transmit frame status conditions
  - PHY interrupt condition
  - Wakeup frame detected
  - Selected MAC management counter(s) at half-full
  - DMA descriptor error
- 47 MAC management statistics counters with selectable clear-on-read behavior and programmable interrupts on half maximum value
- Programmable receive address filters, including a 64-bin address hash table for multicast and/or unicast frames, and programmable filter modes for broadcast, multicast, unicast, control, and damaged frames
- Advanced power management supporting unattended transfer of receive and transmit frames and status to/from external memory via DMA during low power sleep mode
- System wakeup from sleep operating mode upon magic packet or any of four user-definable wakeup frame filters
- Support for 802.3Q tagged VLAN frames
- Programmable MDC clock rate and preamble suppression
- In RMII operation, seven unused signals may be configured as GPIO signals for other purposes

IEEE 1588 Support

The IEEE 1588 standard is a precision clock synchronization protocol for networked measurement and control systems. The ADSP-BF518 processor includes hardware support for IEEE 1588 with an integrated precision time protocol synchronization engine (PTP_TSYNC). This engine provides hardware assisted time stamping to improve the accuracy of clock synchronization between PTP nodes. The main features of the PTP_SYNC engine are:

- Support for both IEEE 1588-2002 and IEEE 1588-2008 protocol standards
- Hardware assisted time stamping capable of up to 12.5 ns resolution
- Lock adjustment
- Programmable PTM message support
- Dedicated interrupts
- Programmable alarm
- Multiple input clock sources (SCLK, MII clock, external clock)
- Programmable pulse per second (PPS) output
- Auxiliary snapshot to time stamp external events

Ports

Because of the rich set of peripherals, the processors group the many peripheral signals to four ports—port F, port G, port H, and port J. Most of the associated pins/balls are shared by multiple signals. The ports function as multiplexer controls.

General-Purpose I/O (GPIO)

The ADSP-BF51x processors have 40 bidirectional, general-purpose I/O (GPIO) signals allocated across three separate GPIO modules—PORTFIO, PORTGIO, and PORTHIO, associated with Port F, Port G, and Port H, respectively. Each GPIO-capable signal shares functionality with other peripherals via a multiplexing scheme; however, the GPIO functionality is the default state of the device upon power-up. Neither GPIO output nor input drivers are active by default. Each general-purpose port signal can be individually controlled by manipulation of the port control, status, and interrupt registers.

Parallel Peripheral Interface (PPI)

The ADSP-BF51x processors provide a parallel peripheral interface (PPI) that can connect directly to parallel analog-to-digital and digital-to-analog converters, ITU-R-601/656 video encoders and decoders, and other general-purpose peripherals. The PPI consists of a dedicated input clock signal, up to three frame synchronization signals, and up to 16 data signals.

In ITU-R-656 modes, the PPI receives and parses a data stream of 8-bit or 10-bit data elements. On-chip decode of embedded preamble control and synchronization information is supported.

Three distinct ITU-R-656 modes are supported:

- Active video only mode—The PPI does not read in any data between the End of Active Video (EAV) and Start of Active Video (SAV) preamble symbols, or any data present during the vertical blanking intervals. In this mode, the control byte sequences are not stored to memory; they are filtered by the PPI.
- Vertical blanking only mode—The PPI only transfers vertical blanking interval (VBI) data, as well as horizontal blanking information and control byte sequences on VBI lines.
- Entire field mode—The entire incoming bitstream is read in through the PPI. This includes active video, control preamble sequences, and ancillary data that may be embedded in horizontal and vertical blanking intervals.

Though not explicitly supported, ITU-R-656 output functionality can be achieved by setting up the entire frame structure (including active video, blanking, and control information) in memory and streaming the data out the PPI in a frame sync-less mode. The processor's 2-D DMA features facilitate this transfer by allowing the static frame buffer (blanking and control codes) to be placed in memory once, and simply updating the active video information on a per-frame basis.

The general-purpose modes of the PPI are intended to suit a wide variety of data capture and transmission applications. The modes are divided into four main categories, each allowing up to 16 bits of data transfer per PPI_CLK cycle:

- Data receive with internally generated frame syncs
- Data receive with externally generated frame syncs
- Data transmit with internally generated frame syncs
- Data transmit with externally generated frame syncs

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These modes support ADC/DAC connections, as well as video communication with hardware signaling. Many of the modes support more than one level of frame synchronization. If desired, a programmable delay can be inserted between assertion of a frame sync and reception/transmission of data.

Code Security with Lockbox Secure Technology

A security system consisting of a blend of hardware and software provides customers with a flexible and rich set of code security features with Lockbox® secure technology. Key features include:

- OTP memory
- Unique chip ID
- Code authentication
- Secure mode of operation

The security scheme is based upon the concept of authentication of digital signatures using standards-based algorithms and provides a secure processing environment in which to execute code and protect assets.

LOCKBOX SECURE TECHNOLOGY DISCLAIMER

Analog Devices does not guarantee that the Code Security with Lockbox Secure Technology described herein provides absolute security. ACCORDINGLY, ANALOG DEVICES HEREBY DISCLAIMS ANY AND ALL EXPRESS AND IMPLIED WARRANTIES THAT THE SECURITY FEATURES CANNOT BE BREACHED, COMPROMISED, OR OTHERWISE CIRCUMVENTED AND IN NO EVENT SHALL ANALOG DEVICES BE LIABLE FOR ANY LOSS, DAMAGE, DESTRUCTION, OR RELEASE OF DATA, INFORMATION, PHYSICAL PROPERTY, OR INTELLECTUAL PROPERTY.

DYNAMIC POWER MANAGEMENT

The ADSP-BF51x processors provide four operating modes, each with a different performance/power profile. In addition, dynamic power management provides the control functions to dynamically alter the processor core supply voltage, further reducing power dissipation. When configured for a 0 V core supply voltage, the processor enters the hibernate state. Control of clocking to each of the processor peripherals also reduces power consumption. See Table 2 for a summary of the power settings for each mode.

Table 2. Power Settings

Mode/State	PLL	PLL Bypassed	Core Clock (CCLK)	System Clock (SCLK)	Core Power
Full On	Enabled	No	Enabled	Enabled	On
Active	Enabled/ Disabled	Yes	Enabled	Enabled	On
Sleep	Enabled	—	Disabled	Enabled	On
Deep Sleep	Disabled	—	Disabled	Disabled	On
Hibernate	Disabled	—	Disabled	Disabled	Off

Full-On Operating Mode—Maximum Performance

In the full-on mode, the PLL is enabled and is not bypassed, providing capability for maximum operational frequency. This is the power-up default execution state in which maximum performance can be achieved. The processor core and all enabled peripherals run at full speed.

Active Operating Mode—Moderate Power Savings

In the active mode, the PLL is enabled but bypassed. Because the PLL is bypassed, the processor's core clock (CCLK) and system clock (SCLK) run at the input clock (CLKIN) frequency. In this mode, the CLKIN to CCLK multiplier ratio can be changed, although the changes are not realized until the full-on mode is entered. DMA access is available to appropriately configured L1 memories.

In the active mode, it is possible to disable the PLL through the PLL control register (PLL_CTL). If disabled, the PLL must be re-enabled before transitioning to the full-on or sleep modes.

Sleep Operating Mode—High Dynamic Power Savings

The sleep mode reduces dynamic power dissipation by disabling the clock to the processor core (CCLK). The PLL and system clock (SCLK), however, continue to operate in this mode. Typically an external event or RTC activity wakes up the processor. When in the sleep mode, asserting wakeup causes the processor to sense the value of the BYPASS bit in the PLL control register (PLL_CTL). If BYPASS is disabled, the processor transitions to the full on mode. If BYPASS is enabled, the processor transitions to the active mode.

System DMA access to L1 memory is not supported in sleep mode.

Deep Sleep Operating Mode—Maximum Dynamic Power Savings

The deep sleep mode maximizes dynamic power savings by disabling the clocks to the processor core (CCLK) and to all synchronous peripherals (SCLK). Asynchronous peripherals, such as the RTC, may still be running but cannot access internal resources or external memory. This powered-down mode can only be exited by assertion of the reset interrupt ($\overline{\text{RESET}}$) or by an asynchronous interrupt generated by the RTC. When in deep sleep mode, an RTC asynchronous interrupt causes the processor to transition to the Active mode. Assertion of $\overline{\text{RESET}}$ while in deep sleep mode causes the processor to transition to the full on mode.

Hibernate State—Maximum Static Power Savings

The hibernate state maximizes static power savings by disabling the voltage and clocks to the processor core (CCLK) and system blocks (SCLK). Any critical information stored internally (for example memory contents, register contents) must be written to a non-volatile storage device prior to removing power if the processor state is to be preserved. Writing b#00 to the FREQ bits in the VR_CTL register also causes the EXT_WAKE signal to transition low, which can be used to signal an external voltage regulator to shut down.

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Since V_{DDEXT} is still supplied in this mode, all of the external signals three-state, unless otherwise specified. This allows other devices that may be connected to the processor to still have power applied without drawing unwanted current.

The Ethernet module can signal an external regulator to wake up using the EXT_WAKE signal. If PF15 does not connect as a $\overline{\text{PHYINT}}$ signal to an external PHY device, it can be pulled low by any other device to wake the processor up. The processor can also be woken up by a real-time clock wakeup event or by asserting the $\overline{\text{RESET}}$ pin. All hibernate wakeup events initiate the hardware reset sequence. Individual sources are enabled by the VR_CTL register. The EXT_WAKE signal is provided to indicate the occurrence of wakeup events.

With the exception of the VR_CTL and the RTC registers, all internal registers and memories lose their content in the hibernate state. State variables may be held in external SRAM or SDRAM. The SCKELOW bit in the VR_CTL register controls whether or not SDRAM operates in self-refresh mode, which allows it to retain its content while the processor is in hibernation and through the subsequent reset sequence.

Power Savings

As shown in Table 3, the processors support up to six different power domains, which maximizes flexibility while maintaining compliance with industry standards and conventions. By isolating the internal logic of the processor into its own power domain, separate from the RTC and other I/O, the processor can take advantage of dynamic power management without affecting the RTC or other I/O devices. There are no sequencing requirements for the various power domains, but all domains must be powered according to the appropriate Specifications table for processor Operating Conditions; even if the feature/peripheral is not used.

Table 3. Power Domains

Power Domain	V_{DD} Range
All internal logic, except RTC, Memory, OTP	V_{DDINT}
RTC internal logic and crystal I/O	V_{DDRTC}
Memory logic	V_{DDMEM}
OTP logic	V_{DDOTP}
All other I/O	V_{DDEXT}

The dynamic power management feature of the processor allows both the processor's input voltage (V_{DDINT}) and clock frequency (f_{CCLK}) to be dynamically controlled.

The power dissipated by a processor is largely a function of its clock frequency and the square of the operating voltage. For example, reducing the clock frequency by 25% results in a 25% reduction in dynamic power dissipation, while reducing the voltage by 25% reduces dynamic power dissipation by more than 40%.

Further, these power savings are additive, in that if the clock frequency and supply voltage are both reduced, the power savings can be dramatic, as shown in the following equations.

$$\text{Power Savings Factor} = \frac{f_{CCLKRED}}{f_{CCLKNOM}} \times \left(\frac{V_{DDINTRED}}{V_{DDINTNOM}}\right)^2 \times \left(\frac{T_{RED}}{T_{NOM}}\right)$$

$$\%\ \text{Power Savings} = (1 - \text{Power Savings Factor}) \times 100\%$$

where the variables in the equations are:

$f_{CCLKNOM}$ is the nominal core clock frequency

$f_{CCLKRED}$ is the reduced core clock frequency

$V_{DDINTNOM}$ is the nominal internal supply voltage

$V_{DDINTRED}$ is the reduced internal supply voltage

T_{NOM} is the duration running at $f_{CCLKNOM}$

T_{RED} is the duration running at $f_{CCLKRED}$

VOLTAGE REGULATION INTERFACE

The ADSP-BF51x processors require an external voltage regulator to power the V_{DDINT} domain. To reduce standby power consumption in the hibernate state, the external voltage regulator can be signaled through EXT_WAKE to remove power from the processor core. The EXT_WAKE signal is high-true for power-up and may be connected directly to the low-true shut down input of many common regulators.

The Power Good ($\overline{\text{PG}}$) input signal allows the processor to start only after the internal voltage has reached a chosen level. In this way, the startup time of the external regulator is detected after hibernation. For a complete description of the $\overline{\text{PG}}$ functionality, refer to the *ADSP-BF51x Blackfin Processor Hardware Reference*.

CLOCK SIGNALS

The ADSP-BF51x processors can be clocked by an external crystal, a sine wave input, or a buffered, shaped clock derived from an external clock oscillator.

If an external clock is used, it should be a TTL compatible signal and must not be halted, changed, or operated below the specified frequency during normal operation. This signal is connected to the processor CLKIN signal. When an external clock is used, the XTAL pin/ball must be left unconnected.

Alternatively, because the processor includes an on-chip oscillator circuit, an external crystal may be used. For fundamental frequency operation, use the circuit shown in Figure 5. A parallel-resonant, fundamental frequency, microprocessor-grade crystal is connected across the CLKIN and XTAL pins/balls. The on-chip resistance between the CLKIN pin/ball and the XTAL pin/ball is in the 500 kΩ range. Further parallel resistors are typically not recommended. The two capacitors and the series resistor shown in Figure 5 fine tune phase and amplitude of the sine frequency.

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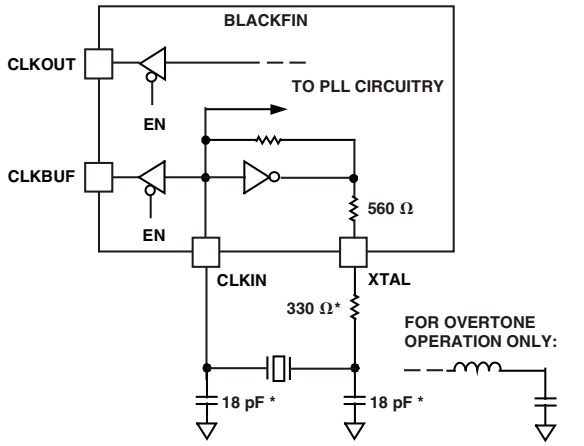
The capacitor and resistor values shown in Figure 5 are typical values only. The capacitor values are dependent upon the crystal manufacturers' load capacitance recommendations and the PCB physical layout. The resistor value depends on the drive level specified by the crystal manufacturer. The user should verify the customized values based on careful investigations on multiple devices over temperature range.



NOTE: VALUES MARKED WITH * MUST BE CUSTOMIZED, DEPENDING ON THE CRYSTAL AND LAYOUT. PLEASE ANALYZE CAREFULLY. FOR FREQUENCIES ABOVE 33 MHz, THE SUGGESTED CAPACITOR VALUE OF 18 pF SHOULD BE TREATED AS A MAXIMUM, AND THE SUGGESTED RESISTOR VALUE SHOULD BE REDUCED TO 0 Ω.

Figure 5. External Crystal Connections

A third-overtone crystal can be used for frequencies above 25 MHz. The circuit is then modified to ensure crystal operation only at the third overtone, by adding a tuned inductor circuit as shown in Figure 5. A design procedure for third-overtone operation is discussed in detail in application note *(EE-168) Using Third Overtone Crystals with the ADSP-218x DSP* on the Analog Devices website (www.analog.com)—use site search on "EE-168."

The CLKBUF signal is an output signal, which is a buffered version of the input clock. This signal is particularly useful in Ethernet applications to limit the number of required clock sources in the system. In this type of application, a single 25 MHz or 50 MHz crystal may be applied directly to the processor. The 25 MHz or 50 MHz output of CLKBUF can then be connected to an external Ethernet MII or RMII PHY device.

The Blackfin core runs at a different clock rate than the on-chip peripherals. As shown in Figure 6, the core clock (CCLK) and system peripheral clock (SCLK) are derived from the input clock (CLKIN) signal. An on-chip PLL is capable of multiplying the CLKIN signal by a programmable 5× to 64× multiplication factor (bounded by specified minimum and maximum VCO frequencies). The default multiplier is 6×, but it can be modified by a software instruction sequence.

On-the-fly frequency changes can be done simply by writing to the PLL_DIV register. The maximum allowed CCLK and SCLK rates depend on the applied voltages V_{DDINT}, V_{DDEXT}, and V_{DDMEM}, and the VCO is always permitted to run up to the frequency specified by the part's speed grade. The CLKOUT signal reflects the SCLK frequency to the off-chip world. It belongs to the SDRAM interface, but it functions as a reference signal in other timing specifications as well. While active by default, it can be disabled using the EBIU_SDGCTL and EBIU_AMGCTL registers.

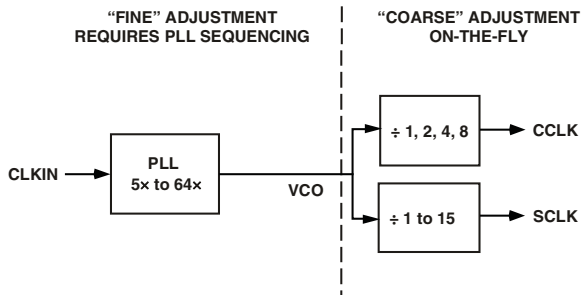


Figure 6. Frequency Modification Methods

All on-chip peripherals are clocked by the system clock (SCLK). The system clock frequency is programmable by means of the SSEL3–0 bits of the PLL_DIV register. The values programmed into the SSEL fields define a divide ratio between the PLL output (VCO) and the system clock. SCLK divider values are 1 through 15. Table 4 illustrates typical system clock ratios.

Table 4. Example System Clock Ratios

Signal Name SSEL3–0	Divider Ratio VCO/SCLK	Example Frequency Ratios (MHz)	
		VCO	SCLK
0010	2:1	100	50
0110	6:1	300	50
1010	10:1	400	40

Note that the divisor ratio must be chosen to limit the system clock frequency to its maximum of f_{SCLK}. The SSEL value can be changed dynamically without any PLL lock latencies by writing the appropriate values to the PLL divisor register (PLL_DIV).

The core clock (CCLK) frequency can also be dynamically changed by means of the CSEL1–0 bits of the PLL_DIV register. Supported CCLK divider ratios are 1, 2, 4, and 8, as shown in Table 5. This programmable core clock capability is useful for fast core frequency modifications.

Table 5. Core Clock Ratios

Signal Name CSEL1–0	Divider Ratio VCO/CCLK	Example Frequency Ratios (MHz)	
		VCO	CCLK
00	1:1	300	300
01	2:1	300	150
10	4:1	400	100
11	8:1	200	25

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The maximum CCLK frequency not only depends on the part's speed grade (see Page 63), it also depends on the applied V_{DDINT} voltage. See Table 9 for details. The maximal system clock rate (SCLK) depends on the chip package and the applied V_{DDINT}, V_{DDEXT}, and V_{DDMEM} voltages (see Table 11).

BOOTING MODES

The processor has several mechanisms (listed in Table 6) for automatically loading internal and external memory after a reset. The boot mode is defined by three BMODE input bits dedicated to this purpose. There are two categories of boot modes. In master boot modes the processor actively loads data from parallel or serial memories. In slave boot modes the processor receives data from external host devices.

The boot modes listed in Table 6 provide a number of mechanisms for automatically loading the processor's internal and external memories after a reset. By default, all boot modes use the slowest meaningful configuration settings. Default settings can be altered via the initialization code feature at boot time or by proper OTP programming at pre-boot time. The BMODE bits of the reset configuration register, sampled during power-on resets and software-initiated resets, implement the modes shown in Table 6.

Table 6. Booting Modes

BMODE2–0	Description
000	Idle - No boot
001	Boot from 8- or 16-bit external flash memory
010	Reserved
011	Boot from external SPI memory (EEPROM or flash)
100	Boot from SPI0 host
101	Boot from OTP memory
110	Boot from SDRAM
111	Boot from UART0 Host

- Idle/no boot mode (BMODE = 0x0)—In this mode, the processor goes into idle. The idle boot mode helps recover from illegal operating modes, such as when the user has mis configured the OTP memory.
- Boot from 8-bit or 16-bit external flash memory (BMODE = 0x1)—In this mode, the boot kernel loads the first block header from address 0x2000 0000 and—depending on instructions containing in the header—the boot kernel performs 8-bit or 16-bit boot or starts program execution at the address provided by the header. By default, all configuration settings are set for the slowest device possible (3-cycle hold time, 15-cycle R/W access times, 4-cycle setup).

  The ARDY is not enabled by default, but it can be enabled by OTP programming. Similarly, all interface behavior and timings can be customized by OTP programming. This includes activation of burst-mode or page-mode operation. In this mode, all signals belonging to the asynchronous interface are enabled at the port muxing level.
- Boot from external SPI EEPROM or flash (BMODE = 0x3)—8-bit, 16-bit, 24-bit or 32-bit addressable devices are supported. The processor uses the PG15 GPIO signal (at $\overline{\text{SPI0SEL2}}$) to select a single SPI EEPROM/flash device connected to the SPI0 interface; then submits a read command and successive address bytes (0x00) until a valid 8-, 16-, 24-, or 32-bit addressable device is detected. Pull-up resistors are required on the SSEL and MISO signals. By default, a value of 0x85 is written to the SPI0_BAUD register.
- Boot from SPI0 host device (BMODE = 0x4)—The processor operates in SPI slave mode and is configured to receive the bytes of the LDR file from an SPI host (master) agent. In the host, the HWAIT signal must be interrogated by the host before every transmitted byte. A pull-up resistor is required on the $\overline{\text{SPI0SS}}$ input. A pull-down on the serial clock may improve signal quality and booting robustness.
- Boot from OTP memory (BMODE = 0x5)—This provides a stand-alone booting method. The boot stream is loaded from on-chip OTP memory. By default the boot stream is expected to start from OTP page 0x40 on and can occupy all public OTP memory up to page 0xDF. This is 2560 bytes. Since the start page is programmable the maximum size of the boot stream can be extended to 3072 bytes.
- Boot from SDRAM (BMODE = 0x6)—This is a warm boot scenario, where the boot kernel starts booting from address 0x0000 0010. The SDRAM is expected to contain a valid boot stream and the SDRAM controller must be configured by the OTP settings.
- Boot from UART0 host (BMODE = 0x7)—Using an autobaud handshake sequence, a boot-stream formatted program is downloaded by the host. The host selects a bit rate within the UART clocking capabilities.

  When performing the autobaud, the UART expects a "@" (0x40) character (eight bits data, one start bit, one stop bit, no parity bit) on the RX0 signal to determine the bit rate. The UART then replies with an acknowledgment composed of 4 bytes (0xBF—the value of UART0_DLL and 0x00—the value of UART0_DLH). The host can then download the boot stream. To hold off the host the Blackfin processor signals the host with the boot host wait (HWAIT) signal. Therefore, the host must monitor HWAIT before every transmitted byte.

For each of the boot modes, a 16-byte header is first read from an external memory device. The header specifies the number of bytes to be transferred and the memory destination address. Multiple memory blocks may be loaded by any boot sequence. Once all blocks are loaded, program execution commences from the address stored in the EVT1 register.

Prior to booting, the pre-boot routine interrogates the OTP memory. Individual boot modes can be customized or even disabled based on OTP programming. External hardware, especially booting hosts may watch the HWAIT signal to determine when the pre-boot has finished and the boot kernel starts

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the boot process. By programming OTP memory, the user can instruct the preboot routine to also customize the PLL, the SDRAM Controller, and the Asynchronous Interface.

The boot kernel differentiates between a regular hardware reset and a wakeup-from-hibernate event to speed up booting in the later case. Bits 6-4 in the system reset configuration (SYSCR) register can be used to bypass pre-boot routine and/or boot kernel in case of a software reset. They can also be used to simulate a wakeup-from-hibernate boot in the software reset case.

The boot process can be further customized by "initialization code." This is a piece of code that is loaded and executed prior to the regular application boot. Typically, this is used to configure the SDRAM controller or to speed up booting by managing PLL, clock frequencies, wait states, or serial bit rates.

The boot ROM also features C-callable function entries that can be called by the user application at run time. This enables second-stage boot or boot management schemes to be implemented with ease.

INSTRUCTION SET DESCRIPTION

The Blackfin processor family assembly language instruction set employs an algebraic syntax designed for ease of coding and readability. The instructions have been specifically tuned to provide a flexible, densely encoded instruction set that compiles to a very small final memory size. The instruction set also provides fully featured multifunction instructions that allow the programmer to use many of the processor core resources in a single instruction. Coupled with many features more often seen on microcontrollers, this instruction set is very efficient when compiling C and C++ source code. In addition, the architecture supports both user (algorithm/application code) and supervisor (O/S kernel, device drivers, debuggers, ISRs) modes of operation, allowing multiple levels of access to core processor resources.

The assembly language, which takes advantage of the processor's unique architecture, offers the following advantages:

- Seamlessly integrated DSP/MCU features are optimized for both 8-bit and 16-bit operations.
- A multi-issue load/store modified-harvard architecture, which supports two 16-bit MACs or four 8-bit ALUs plus two load/store plus two pointer updates per cycle.
- All registers, I/O, and memory are mapped into a unified 4G byte memory space, providing a simplified programming model.
- Microcontroller features, such as arbitrary bit and bit-field manipulation, insertion, and extraction; integer operations on 8-, 16-, and 32-bit data-types; and separate user and supervisor stack pointers.
- Code density enhancements, which include intermixing of 16-bit and 32-bit instructions (no mode switching, no code segregation). Frequently used instructions are encoded in 16 bits.

DEVELOPMENT TOOLS

Analog Devices supports its processors with a complete line of software and hardware development tools, including integrated development environments (which include CrossCore® Embedded Studio and/or VisualDSP++®), evaluation products, emulators, and a wide variety of software add-ins.

Integrated Development Environments (IDEs)

For C/C++ software writing and editing, code generation, and debug support, Analog Devices offers two IDEs.

The newest IDE, CrossCore Embedded Studio, is based on the Eclipse™ framework. Supporting most Analog Devices processor families, it is the IDE of choice for future processors, including multicore devices. CrossCore Embedded Studio seamlessly integrates available software add-ins to support real time operating systems, file systems, TCP/IP stacks, USB stacks, algorithmic software modules, and evaluation hardware board support packages. For more information visit www.analog.com/cces.

The other Analog Devices IDE, VisualDSP++, supports processor families introduced prior to the release of CrossCore Embedded Studio. This IDE includes the Analog Devices VDK real time operating system and an open source TCP/IP stack. For more information, visit www.analog.com/visualdsp. Note that VisualDSP++ will not support future Analog Devices processors.

EZ-KIT Lite Evaluation Board

For processor evaluation, Analog Devices provides wide range of EZ-KIT Lite® evaluation boards. Including the processor and key peripherals, the evaluation board also supports on-chip emulation capabilities and other evaluation and development features. Also available are various EZ-Extenders®, which are daughter cards delivering additional specialized functionality, including audio and video processing. For more information visit www.analog.com and search on "ezkit" or "ezextender".

EZ-KIT Lite Evaluation Kits

For a cost-effective way to learn more about developing with Analog Devices processors, Analog Devices offer a range of EZ-KIT Lite evaluation kits. Each evaluation kit includes an EZ-KIT Lite evaluation board, directions for downloading an evaluation version of the available IDE(s), a USB cable, and a power supply. The USB controller on the EZ-KIT Lite board connects to the USB port of the user's PC, enabling the chosen IDE evaluation suite to emulate the on-board processor in-circuit. This permits the customer to download, execute, and debug programs for the EZ-KIT Lite system. It also supports in-circuit programming of the on-board flash device to store user-specific boot code, enabling standalone operation. With the full version of CrossCore Embedded Studio or VisualDSP++ installed (sold separately), engineers can develop software for supported EZ-KITs or any custom system utilizing supported Analog Devices processors.

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Software Add-Ins for CrossCore Embedded Studio

Analog Devices offers software add-ins which seamlessly integrate with CrossCore Embedded Studio to extend its capabilities and reduce development time. Add-ins include board support packages for evaluation hardware, various middleware packages, and algorithmic modules. Documentation, help, configuration dialogs, and coding examples present in these add-ins are viewable through the CrossCore Embedded Studio IDE once the add-in is installed.

Board Support Packages for Evaluation Hardware

Software support for the EZ-KIT Lite evaluation boards and EZ-Extender daughter cards is provided by software add-ins called Board Support Packages (BSPs). The BSPs contain the required drivers, pertinent release notes, and select example code for the given evaluation hardware. A download link for a specific BSP is located on the web page for the associated EZ-KIT or EZ-Extender product. The link is found in the *Product Download* area of the product web page.

Middleware Packages

Analog Devices separately offers middleware add-ins such as real time operating systems, file systems, USB stacks, and TCP/IP stacks. For more information see the following web pages:

- www.analog.com/ucos3
- www.analog.com/ucfs
- www.analog.com/ucusbd
- www.analog.com/lwip

Algorithmic Modules

To speed development, Analog Devices offers add-ins that perform popular audio and video processing algorithms. These are available for use with both CrossCore Embedded Studio and VisualDSP++. For more information visit www.analog.com and search on "Blackfin software modules" or "SHARC software modules".

Designing an Emulator-Compatible DSP Board (Target)

For embedded system test and debug, Analog Devices provides a family of emulators. On each JTAG DSP, Analog Devices supplies an IEEE 1149.1 JTAG Test Access Port (TAP). In-circuit emulation is facilitated by use of this JTAG interface. The emulator accesses the processor's internal features via the processor's TAP, allowing the developer to load code, set breakpoints, and view variables, memory, and registers. The processor must be halted to send data and commands, but once an operation is completed by the emulator, the DSP system is set to run at full speed with no impact on system timing. The emulators require the target board to include a header that supports connection of the JTAG port of the DSP to the emulator.

For details on target board design issues including mechanical layout, single processor connections, signal buffering, signal termination, and emulator pod logic, see the *EE-68: Analog Devices JTAG Emulation Technical Reference* on the Analog Devices website (www.analog.com)—use site search on "EE-68." This document is updated regularly to keep pace with improvements to emulator support.

ADDITIONAL INFORMATION

The following publications that describe ADSP-BF512/ADSP-BF514/ADSP-BF516/ADSP-BF518 processors (and related processors) can be accessed electronically on our website:

- *Getting Started With Blackfin Processors*
- *ADSP-BF51x Blackfin Processor Hardware Reference*
- *Blackfin Processor Programming Reference*
- *ADSP-BF512/BF514/BF516/BF518 Blackfin Processor Anomaly List*

RELATED SIGNAL CHAINS

A *signal chain* is a series of signal-conditioning electronic components that receive input (data acquired from sampling either real-time phenomena or from stored data) in tandem, with the output of one portion of the chain supplying input to the next. Signal chains are often used in signal processing applications to gather and process data or to apply system controls based on analysis of real-time phenomena.

Analog Devices eases signal processing system development by providing signal processing components that are designed to work together well. A tool for viewing relationships between specific applications and related components is available on the www.analog.com website.

The Application Signal Chains page in the Circuits from the Lab™ site (www.analog.com/circuits) provides:

- Graphical circuit block diagram presentation of signal chains for a variety of circuit types and applications
- Drill down links for components in each chain to selection guides and application information
- Reference designs applying best practice design techniques

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SIGNAL DESCRIPTIONS

The processors' signal definitions are listed in Table 7. In order to maintain maximum function and reduce package size and signal count, some signals have dual, multiplexed functions. In cases where signal function is reconfigurable, the default state is shown in plain text, while the alternate function is shown in italics.

All pins are three-stated during and immediately after reset, with the exception of the external memory interface, asynchronous and synchronous memory control, and the buffered XTAL output pin (CLKBUF). On the external memory interface, the control and address lines are driven high, with the exception of CLKOUT, which toggles at the system clock rate. During hibernate all outputs are three-stated unless otherwise noted in Table 7.

All I/O signals have their input buffers disabled with the exception of the signals noted in the data sheet that need pull-ups or pull downs if unused.

The SDA (serial data) and SCL (serial clock) pins/balls are open drain and therefore require a pullup resistor. Consult version 2.1 of the I^2C specification for the proper resistor value.

It is strongly advised to use the available IBIS models to ensure that a given board design meets overshoot/undershoot and signal integrity requirements. If no IBIS simulation is performed, it is strongly recommended to add series resistor terminations for all Driver Types A, C and D. The termination resistors should be placed near the processor to reduce transients and improve signal integrity. The resistance value, typically 33 Ω or 47 Ω, should be chosen to match the average board trace impedance. Additionally, adding a parallel termination to CLKOUT may prove useful in further enhancing signal integrity. Be sure to verify overshoot/undershoot and signal integrity specifications on actual hardware.

Table 7. Signal Descriptions

Signal Name	Type	Function	Driver Type[1]
EBIU			
ADDR19–1	O	Address Bus	A
DATA15–0	I/O	Data Bus	A
$\overline{\text{ABE1–0}}$/*SDQM1–0*	O	Byte Enable or Data Mask	A
$\overline{\text{AMS1–0}}$	O	Asynchronous Memory Bank Selects (Require pull-ups if hibernate is used)	A
$\overline{\text{ARE}}$	O	Asynchronous Memory Read Enable	A
$\overline{\text{AWE}}$	O	Asynchronous Memory Write Enable	A
$\overline{\text{SRAS}}$	O	SDRAM Row Address Strobe	A
$\overline{\text{SCAS}}$	O	SDRAM Column Address Strobe	A
$\overline{\text{SWE}}$	O	SDRAM Write Enable	A
SCKE	O	SDRAM Clock Enable (Requires a pull-down if hibernate with SDRAM self-refresh is used)	A
CLKOUT	O	SDRAM Clock Output	B
SA10	O	SDRAM A10 Signal	A
$\overline{\text{SMS}}$	O	SDRAM Bank Select	A
Port F: GPIO and Multiplexed Peripherals			
PF0/*ETxD2/PPI D0/*$\overline{\textit{SPI1SEL2}}$*/TACLK6*	I/O	GPIO/*Ethernet MII Transmit D2/PPI Data 0/SPI1 Slave Select 2/Timer6 Alternate Clock*	C
PF1/*ERxD2/PPI D1/PWM AH/TACLK7*	I/O	GPIO/*Ethernet MII Receive D2/PPI Data 1/PWM AH Output/Timer7 Alternate Clock*	C
PF2/*ETxD3/PPI D2/PWM AL*	I/O	GPIO/*Ethernet Transmit D3/PPI Data 2/PWM AL Output*	C
PF3/*ERxD3/PPI D3/PWM BH/TACLK0*	I/O	GPIO/*Ethernet MII Data Receive D3/PPI Data 3/PWM BH Output/Timer0 Alternate Clock*	C
PF4/*ERxCLK/PPI D4/PWM BL/TACLK1*	I/O	GPIO/*Ethernet MII Receive Clock/PPI Data 4/PWM BL Out/Timer1 Alternate CLK*	C
PF5/*ERxDV/PPI D5/PWM CH/TACI0*	I/O	GPIO/*Ethernet MII Receive Data Valid/PPI Data 5/PWM CH Out /Timer0 Alternate Capture Input*	C
PF6/*COL/PPI D6/PWM CL/TACI1*	I/O	GPIO/*Ethernet MII Collision/PPI Data 6/PWM CL Out/Timer1 Alternate Capture Input*	C
PF7/$\overline{\textit{SPI0SEL1}}$*/PPI D7/PWMSYNC*	I/O	GPIO/*SPI0 Slave Select 1/PPI Data 7/PWM Sync*	C

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Table 7. Signal Descriptions (Continued)

Signal Name	Type	Function	Driver Type[1]
PF8/*MDC*/PPI D8/*SPI1SEL4*	I/O	GPIO/*Ethernet Management Channel Clock/PPI Data 8/SPI1 Slave Select 4*	C
PF9/*MDIO*/*PPI D9/TMR2*	I/O	GPIO/*Ethernet Management Channel Serial Data/PPI Data 9/Timer 2*	C
PF10/*ETxD0*/*PPI D10/TMR3*	I/O	GPIO/*Ethernet MII or RMII Transmit D0/PPI Data 10/Timer 3*	C
PF11/*ERxD0*/*PPI D11*/*PWM AH*/*TACI3*	I/O	GPIO/*Ethernet MII Receive D0/PPI Data 11/PWM AH output /Timer3 Alternate Capture Input*	C
PF12/*ETxD1*/*PPI D12*/*PWM AL*	I/O	GPIO/*Ethernet MII Transmit D1/PPI Data 12/PWM AL Output*	C
PF13/*ERxD1*/*PPI D13*/*PWM BH*	I/O	GPIO/*Ethernet MII or RMII Receive D1/PPI Data 13/PWM BH Output*	C
PF14/*ETxEN*/*PPI D14*/*PWM BL*	I/O	GPIO/*Ethernet MII Transmit Enable/PPI Data 14/PWM BL Out*	C
PF15[2]/*RMII PHYINT*/*PPI D15*/*PWM_SYNCA*	I/O	GPIO/*Ethernet MII PHY Interrupt/PPI Data 15/Alternate PWM Sync*	C
Port G: GPIO and Multiplexed Peripherals			
PG0/*MIICRS*/*RMIICRS*/*HWAIT*[3]/*SPI1SEL3*	I/O	GPIO/*Ethernet MII or RMII Carrier Sense or RMII Data Valid/HWAIT/SPI1 Slave Select3*	C
PG1/*ERxER*/*DMAR1*/*PWM CH*	I/O	GPIO/*Ethernet MII or RMII Receive Error/DMA Req 1/PWM CH Out*	C
PG2/*MIITxCLK*/*RMIIREF_CLK*/*DMAR0*/*PWM CL*	I/O	GPIO/*Ethernet MII or RMII Reference Clock/DMA Req 0/PWM CL Out*	C
PG3/*DR0PRI*/*RSI_DATA0*/*SPI0SEL5*/*TACLK3*	I/O	GPIO/*SPORT0 Primary Rx Data/RSI Data 0/SPI0 Slave Select 5/Timer3 Alternate CLK*	C
PG4/*RSCLK0*/*RSI_DATA1*/*TMR5*/*TACI5*	I/O	GPIO/*SPORT0 Rx Clock/RSI Data 1/Timer 5/Timer5 Alternate Capture Input*	D
PG5/*RFS0*/*RSI_DATA2*/*PPICLK*/*TMRCLK*	I/O	GPIO/*SPORT0 Rx Frame Sync/RSI Data 2/PPI Clock/External Timer Reference*	C
PG6/*TFS0*/*RSI_DATA3*/*TMR0*/*PPIFS1*	I/O	GPIO/*SPORT0 Tx Frame Sync/RSI Data 3/Timer0/PPI Frame Sync1*	C
PG7/*DT0PRI*/*RSI_CMD*/*TMR1*/*PPIFS2*	I/O	GPIO/*SPORT0 Tx Primary Data/RSI Command/Timer 1/PPI Frame Sync2*	C
PG8/*TSCLK0*/*RSI_CLK*/*TMR6*/*TACI6*	I/O	GPIO/*SPORT0 Tx Clock/RSI Clock/Timer 6/Timer6 Alternate Capture Input*	D
PG9/*DT0SEC*/*UART0TX*/*TMR4*	I/O	GPIO/*SPORT0 Secondary Tx Data/UART0 Transmit/Timer 4*	C
PG10/*DR0SEC*/*UART0RX*/*TACI4*	I/O	GPIO/*SPORT0 Secondary Rx Data/UART0 Receive/Timer4 Alternate Capture Input*	C
PG11/*SPI0SS*/*AMS2*/*SPI1SEL5*/*TACLK2*	I/O	GPIO/*SPI0 Slave Device Select/Asynchronous Memory Bank Select 2/SPI1 Slave Select 5/Timer2 Alternate CLK*	C
PG12/*SPI0SCK*/*PPICLK*/*TMRCLK*/*PTP_PPS*	I/O	GPIO/*SPI0 Clock/PPI Clock/External Timer Reference/PTP Pulse Per Second Out*	D
PG13/*SPI0MISO*[4]/*TMR0*/*PPIFS1*/ *PTP_CLKOUT*	I/O	GPIO/*SPI0 Master In Slave Out/Timer0/PPI Frame Sync1/PTP Clock Out*	C
PG14/*SPI0MOSI*/*TMR1*/*PPIFS2*/*PWM TRIP* /*PTP_AUXIN*	I/O	GPIO/*SPI0 Master Out Slave In/Timer 1/PPI Frame Sync2/PWM Trip/PTP Auxiliary Snapshot Trigger Input*	C
PG15/*SPI0SEL2*/*PPIFS3*/*AMS3*	I/O	GPIO/*SPI0 Slave Select 2/PPI Frame Sync3/Asynchronous Memory Bank Select 3*	C
Port H: GPIO and Multiplexed Peripherals			
PH0/*DR1PRI*/*SPI1SS*/*RSI_DATA4*	I/O	GPIO/*SPORT1 Primary Rx Data/SPI1 Device Select/RSI Data 4*	C
PH1/*RFS1*/*SPI1MISO*/*RSI_DATA5*	I/O	GPIO/*SPORT1 Rx Frame Sync/SPI1 Master In Slave Out/RSI Data 5*	C
PH2/*RSCLK1*/*SPI1SCK*/*RSI DATA6*	I/O	GPIO/*SPORT1 Rx Clock/SPI1 Clock/RSI Data 6*	D
PH3/*DT1PRI*/*SPI1MOSI*/*RSI DATA7*	I/O	GPIO/*SPORT1 Primary Tx Data/SPI1 Master Out Slave In/RSI Data 7*	C
PH4/*TFS1*/*AOE*/*SPI0SEL3*/*CUD*	I/O	GPIO/*SPORT1 Tx Frame Sync/Asynchronous Memory Output Enable/SPI0 Slave Select 3/Counter Up Direction*	C
PH5/*TSCLK1*/*ARDY*/*PTP_EXT_CLKIN*/*CDG*	I/O	GPIO/*SPORT1 Tx Clock/Asynchronous Memory Hardware Ready Control/ External Clock for PTP TSYNC/Counter Down Gate*	D
PH6/*DT1SEC*/*UART1TX*/*SPI1SEL1*/*CZM*	I/O	GPIO/*SPORT1 Secondary Tx Data/UART1 Transmit/SPI1 Slave Select 1 /Counter Zero Marker*	C
PH7/*DR1SEC*/*UART1RX*/*TMR7*/*TACI2*	I/O	GPIO/*SPORT1 Secondary Rx Data/UART1 Receive/Timer 7/Timer2 Alternate Clock Input*	C

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Table 7. Signal Descriptions (Continued)

Signal Name	Type	Function	Driver Type[1]
Port J			
PJ0:SCL	I/O 5V	TWI Serial Clock (This signal is an open-drain output and requires a pull-up resistor. Consult version 2.1 of the I^2C specification for the proper resistor value.)	E
PJ1:SDA	I/O 5V	TWI Serial Data (This signal is an open-drain output and requires a pull-up resistor. Consult version 2.1 of the I^2C specification for the proper resistor value.)	E
Real Time Clock			
RTXI	I	RTC Crystal Input (This ball should be pulled low when not used.)	
RTXO	O	RTC Crystal Output (Does not three-state during hibernate)	
JTAG Port			
TCK	I	JTAG Clock	
TDO	O	JTAG Serial Data Out	C
TDI	I	JTAG Serial Data In	
TMS	I	JTAG Mode Select	
$\overline{\text{TRST}}$	I	JTAG Reset (This signal should be pulled low if the JTAG port is not used.)	
$\overline{\text{EMU}}$	O	Emulation Output	C
Clock			
CLKIN	I	Clock/Crystal Input	
XTAL	O	Crystal Output (If CLKBUF is enabled, does not three-state during hibernate)	
CLKBUF	O	Buffered XTAL Output (If enabled, does not three-state during hibernate)	C
Mode Controls			
$\overline{\text{RESET}}$	I	Reset	
$\overline{\text{NMI}}$	I	Non-maskable Interrupt (This signal should be pulled high when not used.)	
BMODE2-0	I	Boot Mode Strap 2-0	
Voltage Regulation Interface			
$\overline{\text{PG}}$	I	Power Good (This signal should be pulled low when not used.)	
EXT_WAKE	O	Wake up Indication (Does not three-state during hibernate)	C
Power Supplies		**ALL SUPPLIES MUST BE POWERED** See Operating Conditions.	
V_{DDEXT}	P	I/O Power Supply	
V_{DDINT}	P	Internal Power Supply	
V_{DDRTC}	P	Real Time Clock Power Supply	
V_{DDMEM}	P	MEM Power Supply	
V_{PPOTP}	P	OTP Programming Voltage	
V_{DDOTP}	P	OTP Power Supply	
GND	G	Ground for All Supplies	

[1] See Output Drive Currents for more information about each driver type.
[2] When driven low, the PF15 signal can be used to wake up the processor from the hibernate state, either in normal GPIO mode or in Ethernet mode as $\overline{\text{PHYINT}}$. If the pin/ball is used for wake up, enable the feature with the PHYWE bit in the VR_CTL register, and pull-up the signal with a resistor.
[3] Boot host wait is a GPIO signal toggled by the boot kernel. The mandatory external pull-up/pull-down resistor defines the signal polarity.
[4] A pull-up resistor is required for the boot from external SPI EEPROM or flash (BMODE = 0x3).

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SPECIFICATIONS

Note that component specifications are subject to change without notice.

OPERATING CONDITIONS

Parameter		Conditions	Min	Nominal	Max	Unit
V_{DDINT}	Internal Supply Voltage	Industrial Models	1.14		1.47	V
	Internal Supply Voltage	Commercial Models	1.10		1.47	V
	Internal Supply Voltage	Automotive Models	1.33		1.47	V
V_{DDEXT}[1, 2]	External Supply Voltage	1.8 V I/O, Nonautomotive Models	1.7	1.8	1.9	V
	External Supply Voltage	2.5 V I/O, Nonautomotive Models	2.25	2.5	2.75	V
	External Supply Voltage	3.3 V I/O, All Models	3.0	3.3	3.6	V
V_{DDMEM}[3]	MEM Supply Voltage	1.8 V I/O, Nonautomotive Models	1.7	1.8	1.9	V
	MEM Supply Voltage	2.5 V I/O, Nonautomotive Models	2.25	2.5	2.75	V
	MEM Supply Voltage	3.3 V I/O, All Models	3.0	3.3	3.6	V
V_{DDRTC}[4]	RTC Power Supply Voltage		2.25		3.6	V
V_{DDOTP}[1]	OTP Supply Voltage		2.25	2.5	2.75	V
V_{PPOTP}	OTP Programming Voltage					
	For Reads[1]		2.25	2.5	2.75	V
	For Writes[5]		6.9	7.0	7.1	V
V_{IH}	High Level Input Voltage[6, 7]	V_{DDEXT}/V_{DDMEM} = 1.90 V	1.2			V
	High Level Input Voltage[6, 8]	V_{DDEXT}/V_{DDMEM} = 2.75 V	1.7			V
	High Level Input Voltage[6, 8]	V_{DDEXT}/V_{DDMEM} = 3.6 V	2			V
V_{IHTWI}	High Level Input Voltage	V_{DDEXT} = 1.90 V/2.75 V/3.6 V	0.7 × V_{BUSTWI}		V_{BUSTWI}[9]	V
V_{IL}	Low Level Input Voltage[6, 7]	V_{DDEXT}/V_{DDMEM} = 1.7 V			0.6	V
	Low Level Input Voltage[6, 8]	V_{DDEXT}/V_{DDMEM} = 2.25 V			0.7	V
	Low Level Input Voltage[6, 8]	V_{DDEXT}/V_{DDMEM} = 3.0 V			0.8	V
V_{ILTWI}	Low Level Input Voltage	V_{DDEXT} = Minimum			0.3 × V_{BUSTWI}[10]	V
	Junction Temperature	168-Ball CSP_BGA @ $T_{AMBIENT}$ = 0°C to +70°C	0		+95	°C
	Junction Temperature	168-Ball CSP_BGA @ $T_{AMBIENT}$ = –40°C to +85°C	–40		+105	°C
	Junction Temperature	176-Lead LQFP_EP @ $T_{AMBIENT}$ = 0°C to +70°C	0		+95	°C
	Junction Temperature	176-Lead LQFP_EP @ $T_{AMBIENT}$ = –40°C to +85°C	–40		+105	°C

[1] Must remain powered (even if the associated function is not used).

[2] V_{DDEXT} is the supply to the GPIO.

[3] Pins/balls that use V_{DDMEM} are DATA15–0, ADDR19–1, $\overline{ABE1–0}$, $\overline{ARE}$, $\overline{AWE}$, $\overline{AMS1–0}$, SA10, $\overline{SWE}$, $\overline{SCAS}$, CLKOUT, $\overline{SRAS}$, $\overline{SMS}$, SCKE. These pins/balls are not tolerant to voltages higher than V_{DDMEM}. When using any of the asynchronous memory signals $\overline{AMS3–2}$, ARDY, or $\overline{AOE}$ V_{DDMEM} and V_{DDEXT} must be shorted externally.

[4] If not used, power with V_{DDEXT}.

[5] The V_{PPOTP} voltage for writes must only be applied when programming OTP memory. There is a finite amount of cumulative time that this voltage may be applied (dependent on voltage and junction temperature) over the lifetime of the part.

[6] Parameter value applies to all input and bidirectional pins/balls except SDA and SCL.

[7] Bidirectional balls (PF15–0, PG15–0, PH15–0) and input balls (RTXI, TCK, TDI, TMS, $\overline{TRST}$, CLKIN, $\overline{RESET}$, $\overline{NMI}$, and BMODE3–0) of the ADSP-BF51x processors are 2.5 V tolerant (always accept up to 2.7 V maximum V_{IH}). Voltage compliance (on outputs, V_{OH}) is limited by the V_{DDEXT} supply voltage.

[8] Bidirectional pins/balls (PF15–0, PG15–0, PH7–0) and input pins/balls (RTXI, TCK, TDI, TMS, $\overline{TRST}$, CLKIN, $\overline{RESET}$, $\overline{NMI}$, and BMODE2–0) of the ADSP-BF51x are 3.3 V tolerant (always accept up to 3.6 V maximum V_{IH}). Voltage compliance (on outputs, V_{OH}) is limited by the V_{DDEXT} supply voltage.

[9] The V_{IHTWI} min and max value vary with the selection in the TWI_DT field of the NONGPIO_DRIVE register. See V_{BUSTWI} min and max values in Table 8.

[10] SDA and SCL are pulled up to V_{BUSTWI}. See Table 8.

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Table 8 shows settings for TWI_DT in the NONGPIO_DRIVE register. Set this register prior to using the TWI port.

Table 8. TWI_DT Field Selections and V_{DDEXT}/V_{BUSTWI}

TWI_DT	V_{DDEXT} Nominal	V_{BUSTWI} Minimum	V_{BUSTWI} Nominal	V_{BUSTWI} Maximum	Unit
000 (default)	3.3	2.97	3.3	3.63	V
001	1.8	1.7	1.8	1.98	V
010	2.5	2.97	3.3	3.63	V
011	1.8	2.97	3.3	3.63	V
100	3.3	4.5	5	5.5	V
101	1.8	2.25	2.5	2.75	V
110	2.5	2.25	2.5	2.75	V
111 (reserved)	—	—	—	—	—

Clock Related Operating Conditions

Table 9 describes the timing requirements for the processor clocks. Take care in selecting MSEL, SSEL, and CSEL ratios so as not to exceed the maximum core clock and system clock. Table 10 describes phase-locked loop operating conditions.

Table 9. Core Clock (CCLK) Requirements

Parameter		Nominal Voltage Setting	Maximum	Unit
f_{CCLK}	Core Clock Frequency (V_{DDINT} = 1.33 V Minimum, All Models)	1.400 V	400	MHz
	Core Clock Frequency (V_{DDINT} = 1.23 V Minimum, Industrial/Commercial Models)	1.300 V	300	MHz
	Core Clock Frequency (V_{DDINT} = 1.14 V Minimum, Industrial Models Only)	1.200 V	200	MHz
	Core Clock Frequency (V_{DDINT} = 1.10 V Minimum, Commercial Models Only)	1.150 V	200	MHz

Table 10. Phase-Locked Loop Operating Conditions

Parameter		Min	Max	Unit
f_{VCO}	Voltage Controlled Oscillator (VCO) Frequency (Commercial/Industrial Models)	72	Instruction Rate[1]	MHz
	Voltage Controlled Oscillator (VCO) Frequency (Automotive Models)	84	Instruction Rate[1]	MHz

[1] For more information, see Ordering Guide.

Table 11. SCLK Conditions

Parameter[1]		V_{DDEXT}/V_{DDMEM} 1.8 V Nominal Max	V_{DDEXT}/V_{DDMEM} 2.5 V or 3.3 V Nominal Max	Unit
f_{SCLK}	CLKOUT/SCLK Frequency (V_{DDINT} ≥ 1.230 V Minimum)	80	100	MHz
f_{SCLK}	CLKOUT/SCLK Frequency (V_{DDINT} < 1.230 V)	80	80	MHz

[1] f_{SCLK} must be less than or equal to f_{CCLK} and is subject to additional restrictions for SDRAM interface operation. See Table 24.

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ELECTRICAL CHARACTERISTICS

Parameter		Conditions	Min	Typical	Max	Unit
V_{OH}	High Level Output Voltage	V_{DDEXT}/V_{DDMEM} = 1.7 V, I_{OH} = –0.5 mA	1.35			V
	High Level Output Voltage	V_{DDEXT}/V_{DDMEM} = 2.25 V, I_{OH} = –0.5 mA	2			V
	High Level Output Voltage	V_{DDEXT}/V_{DDMEM} = 3.0 V, I_{OH} = –0.5 mA	2.4			V
V_{OL}	Low Level Output Voltage	V_{DDEXT}/V_{DDMEM} = 1.7/2.25/3.0 V, I_{OL} = 2.0 mA			0.4	V
I_{IH}[1]	High Level Input Current	V_{DDEXT}/V_{DDMEM} = 3.6 V, V_{IN} = 3.6 V			10	µA
I_{IL}[1]	Low Level Input Current	V_{DDEXT}/V_{DDMEM} = 3.6 V, V_{IN} = 0 V			10	µA
I_{IHP}[2]	High Level Input Current JTAG	V_{DDEXT} = 3.6 V, V_{IN} = 3.6 V			75	µA
I_{OZH}[3]	Three-State Leakage Current	V_{DDEXT}/V_{DDMEM} = 3.6 V, V_{IN} = 3.6 V			10	µA
I_{OZHTWI}[4]	Three-State Leakage Current	V_{DDEXT} = 3.0 V, V_{IN} = 5.5 V			10	µA
I_{OZL}[3]	Three-State Leakage Current	V_{DDEXT}/V_{DDMEM} = 3.6 V, V_{IN} = 0 V			10	µA
C_{IN}[5, 6]	Input Capacitance	f_{IN} = 1 MHz, $T_{AMBIENT}$ = 25°C, V_{IN} = 2.5 V		5	8	pF
C_{INTWI}[4, 6]	Input Capacitance	f_{IN} = 1 MHz, $T_{AMBIENT}$ = 25°C, V_{IN} = 2.5 V			15	pF
$I_{DDDEEPSLEEP}$[7]	V_{DDINT} Current in Deep Sleep Mode	V_{DDINT} = 1.3 V, f_{CCLK} = 0 MHz, f_{SCLK} = 0 MHz, T_J = 25°C, ASF = 0.00		2.1		mA
$I_{DDSLEEP}$	V_{DDINT} Current in Sleep Mode	V_{DDINT} = 1.3 V, f_{SCLK} = 25 MHz, T_J = 25°C		5.5		mA
$I_{DD-IDLE}$	V_{DDINT} Current in Idle	V_{DDINT} = 1.3 V, f_{CCLK} = 50 MHz, f_{SCLK} = 25 MHz, T_J = 25°C, ASF = 0.41		12		mA
I_{DD-TYP}	V_{DDINT} Current	V_{DDINT} = 1.3 V, f_{CCLK} = 300 MHz, f_{SCLK} = 25 MHz, T_J = 25°C, ASF = 1.00		77		mA
I_{DD-TYP}	V_{DDINT} Current	V_{DDINT} = 1.4 V, f_{CCLK} = 400 MHz, f_{SCLK} = 25 MHz, T_J = 25°C, ASF = 1.00		108		mA
$I_{DDHIBERNATE}$[8]	Hibernate State Current	V_{DDEXT}=V_{DDMEM}=V_{DDRTC} = 3.3 V V_{DDOTP}=V_{PPOTP}=2.5 V, T_J = 25°C, CLKIN = 0 MHz		40		µA
I_{DDRTC}	V_{DDRTC} Current	V_{DDRTC} = 3.3 V, T_J = 25°C		20		µA
$I_{DDSLEEP}$[8, 9]	V_{DDINT} Current in Sleep Mode	f_{CCLK} = 0 MHz, f_{SCLK} > 0 MHz			Table 13 + (0.20 × V_{DDINT} × f_{SCLK})	mA[10]
$I_{DDDEEPSLEEP}$[8, 10]	V_{DDINT} Current in Deep Sleep Mode	f_{CCLK} = 0 MHz, f_{SCLK} = 0 MHz			Table 13	mA
I_{DDINT}[10, 11]	V_{DDINT} Current	f_{CCLK} > 0 MHz, f_{SCLK} ≥ 0 MHz			Table 13 + (Table 14 × ASF) + (0.20 × V_{DDINT} × f_{SCLK})	mA

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Parameter		Conditions	Min	Typical	Max	Unit
I_{DDOTP}	V_{DDOTP} Current	V_{DDOTP} = 2.5 V, T_J = 25°C, OTP Memory Read		2		mA
I_{DDOTP}	V_{DDOTP} Current	V_{DDOTP} = 2.5 V, T_J = 25°C, OTP Memory Write		2		mA
I_{PPOTP}	V_{PPOTP} Current	V_{PPOTP} = 2.5 V, T_J = 25°C, OTP Memory Read		100		µA
I_{PPOTP}	V_{PPOTP} Current	V_{PPOTP} = Table 17 V, T_J = 25°C, OTP Memory Write		3		mA

[1] Applies to input balls.
[2] Applies to JTAG input balls (TCK, TDI, TMS, $\overline{TRST}$).
[3] Applies to three-statable balls.
[4] Applies to bidirectional balls SCL and SDA.
[5] Applies to all signal balls, except SCL and SDA.
[6] Guaranteed, but not tested.
[7] See the *ADSP-BF51x Blackfin Processor Hardware Reference Manual* for definition of sleep, deep sleep, and hibernate operating modes.
[8] Includes current on V_{DDEXT}, V_{DDMEM}, V_{DDOTP}, and V_{PPOTP} supplies. Clock inputs are tied high or low.
[9] Guaranteed maximum specifications.
[10] Unit for V_{DDINT} is V (Volts). Unit for f_{SCLK} is MHz.
[11] See Table 12 for the list of I_{DDINT} power vectors covered.

Total Power Dissipation

Total power dissipation has two components:

1. Static, including leakage current
2. Dynamic, due to transistor switching characteristics

Many operating conditions can also affect power dissipation, including temperature, voltage, operating frequency, and processor activity. Electrical Characteristics shows the current dissipation for internal circuitry (V_{DDINT}). $I_{DDDEEPSLEEP}$ specifies static power dissipation as a function of voltage (V_{DDINT}) and temperature (see Table 13), and I_{DDINT} specifies the total power specification for the listed test conditions, including the dynamic component as a function of voltage (V_{DDINT}) and frequency (Table 14).

There are two parts to the dynamic component. The first part is due to transistor switching in the core clock (CCLK) domain. This part is subject to an Activity Scaling Factor (ASF) which represents application code running on the processor core and L1 memories (Table 12).

The ASF is combined with the CCLK Frequency and V_{DDINT} dependent data in Table 14 to calculate this part. The second part is due to transistor switching in the system clock (SCLK) domain, which is included in the I_{DDINT} specification equation.

Table 12. Activity Scaling Factors (ASF)[1]

I_{DDINT} Power Vector	Activity Scaling Factor (ASF)
$I_{DD-PEAK}$	1.29
$I_{DD-HIGH}$	1.25
I_{DD-TYP}	1.00
I_{DD-APP}	0.85
I_{DD-NOP}	0.70
$I_{DD-IDLE}$	0.41

[1] See *Estimating Power for ADSP-BF534/BF536/BF537 Blackfin Processors (EE-297)*. The power vector information also applies to the ADSP-BF51x processors.

Table 13. Static Current—$I_{DD-DEEPSLEEP}$ (mA)

	Voltage (V_{DDINT})[1]								
T_J (°C)[1]	1.10 V	1.15 V	1.20 V	1.25 V	1.30 V	1.35 V	1.40 V	1.45 V	1.50 V
–40	0.9	1.0	1.0	1.1	1.1	1.2	1.3	1.7	1.9
–20	1.0	1.1	1.2	1.3	1.4	1.6	1.7	1.9	2.0
0	1.2	1.3	1.4	1.6	1.8	2.0	2.2	2.3	2.5
25	1.8	1.9	2.1	2.3	2.5	2.8	3.1	3.3	3.7
40	2.4	2.6	2.8	3.0	3.3	3.7	4.0	4.4	4.9
55	3.3	3.5	3.8	4.3	4.6	5.0	5.5	6.1	6.7
70	4.6	5.0	5.4	6.0	6.4	7.0	7.7	8.4	9.2

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Table 13. Static Current—$I_{DD\text{-}DEEPSLEEP}$ (mA) (Continued)

T_J (°C)[1]	Voltage (V_{DDINT})[1]								
	1.10 V	1.15 V	1.20 V	1.25 V	1.30 V	1.35 V	1.40 V	1.45 V	1.50 V
85	6.5	7.1	7.7	8.3	9.1	9.9	10.8	11.8	12.8
100	9.2	10.0	10.8	11.7	12.7	13.7	15.0	16.1	17.5
105	10.3	11.1	12.1	13.1	14.2	15.3	16.6	18.0	19.4

[1] Valid frequency and voltage ranges are model-specific. See Operating Conditions.

Table 14. Dynamic Current in CCLK Domain (mA, with ASF = 1.0)[1]

f_{CCLK} (MHz)[2]	Voltage (V_{DDINT})[2]								
	1.10 V	1.15 V	1.20 V	1.25 V	1.30 V	1.35 V	1.40 V	1.45 V	1.50 V
400	N/A	N/A	N/A	N/A	N/A	N/A	102.1	106.5	111.0
350	N/A	N/A	N/A	N/A	N/A	86.2	90.1	94.0	98.0
300	N/A	N/A	N/A	N/A	71.4	74.7	78.1	81.5	85.0
250	N/A	N/A	N/A	57.5	60.4	63.2	66.1	69.0	71.9
200	N/A	42.5	44.7	47.0	49.4	51.7	54.1	56.5	58.9
150	31.1	32.9	34.7	36.5	38.4	40.2	42.1	44.0	45.9
100	22.0	23.4	24.7	26.0	27.4	28.7	30.1	31.5	33.0

[1] The values are not guaranteed as standalone maximum specifications. They must be combined with static current per the equations of Electrical Characteristics.
[2] Valid frequency and voltage ranges are model-specific. See Operating Conditions.

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ABSOLUTE MAXIMUM RATINGS

Stresses greater than those listed in Table 15 may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions greater than those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 15. Absolute Maximum Ratings

Parameter	Rating
Internal Supply Voltage (V_{DDINT})	–0.3 V to +1.50 V
External (I/O) Supply Voltage (V_{DDEXT}/V_{DDMEM})	–0.3 V to +3.8 V
Input Voltage[1, 2]	–0.5 V to +3.6 V
Input Voltage[1, 3]	–0.5 V to +5.5 V
Output Voltage Swing	–0.5 V to V_{DDEXT}/V_{DDMEM} + 0.5 V
I_{OH}/I_{OL} Current per Pin Group[4]	80 mA (max)
Storage Temperature Range	–65°C to +150°C
Junction Temperature While biased	+110°C

[1] Applies to 100% transient duty cycle. For other duty cycles see Table 16.
[2] Applies only when V_{DDEXT} is within specifications. When V_{DDEXT} is outside specifications, the range is V_{DDEXT} ± 0.2.
[3] Applies to signals SCL, SDA.
[4] For more information, see the information preceding Table 18 and Table 19.

Table 16. Maximum Duty Cycle for Input Transient Voltage[1]

V_{IN} Min (V)[2]	V_{IN} Max (V)[2]	Maximum Duty Cycle[3]
–0.50	+3.80	100%
–0.70	+4.00	40%
–0.80	+4.10	25%
–0.90	+4.20	15%
–1.00	+4.30	10%

[1] Applies to all signal pins/balls with the exception of CLKIN, XTAL.
[2] The individual values cannot be combined for analysis of a single instance of overshoot or undershoot. The worst case observed value must fall within one of the voltages specified and the total duration of the overshoot or undershoot (exceeding the 100% case) must be less than or equal to the corresponding duty cycle.
[3] Duty cycle refers to the percentage of time the signal exceeds the value for the 100% case. It is equivalent to the measured duration of a single instance of overshoot or undershoot as a percentage of the period of occurrence.

When programming OTP memory on the ADSP-BF51x processor, the V_{PPOTP} pin/ball must be set to the write value specified in the Operating Conditions. There is a finite amount of cumulative time that the write voltage may be applied (dependent on voltage and junction temperature) to V_{PPOTP} over the lifetime of the part. Therefore, maximum OTP memory programming time for the processor is shown in Table 17.

Table 17. Maximum OTP Memory Programming Time

VPPOTP Voltage (V)	Temperature		
	25°C	85°C	110°C
6.9	6000 sec	100 sec	25 sec
7.0	2400 sec	44 sec	12 sec
7.1	1000 sec	18 sec	4.5 sec

Table 18 and Table 19 specify the maximum total source/sink (I_{OH}/I_{OL}) current for a group of pins. Permanent damage can occur if this value is exceeded. To understand this specification, if pins PF9, PF8, PF7, PF6, and PF5 from Group 1 in Table 19 table were sourcing or sinking 2 mA each, the total current for those pins would be 10 mA. This would allow up to 70 mA total that could be sourced or sunk by the remaining pins in the group without damaging the device. Note that the V_{OH} and V_{OL} specifications have separate per-pin maximum current requirements as shown in the Electrical Characteristics table.

Table 18. Total Current Pin Groups–V_{DDMEM} Groups

Group	Pins in Group
1	DATA15, DATA14, DATA13, DATA12, DATA11, DATA10
2	DATA9, DATA8, DATA7, DATA6, DATA5, DATA4
3	DATA3, DATA2, DATA1, DATA0, ADDR19, ADDR18
4	ADDR17, ADDR16, ADDR15, ADDR14, ADDR13
5	ADDR12, ADDR11, ADDR10, ADDR9, ADDR8, ADDR7
6	ADDR6, ADDR5, ADDR4, ADDR3, ADDR2, ADDR1
7	$\overline{ABE1}$, $\overline{ABE0}$, SA10, $\overline{SWE}$, $\overline{SCAS}$, $\overline{SRAS}$
8	$\overline{SMS}$, SCKE, $\overline{AMS1}$, $\overline{ARE}$, $\overline{AWE}$, $\overline{AMS0}$, CLKOUT

Table 19. Total Current Pin Groups–V_{DDEXT} Groups

Group	Pins in Group
1	PF9, PF8, PF7, PF6, PF5, PF4, PF3, PF2
2	PF1, PF0, PG15, PG14, PG13, PG12, PG11, PG10
3	PG9, PG8, PG7, PG6, PG5, PG4, PG3, PG2, BMODE0, BMODE1, BMODE2
4	PG1, PG0, TDO, $\overline{EMU}$, TDI, TCK, $\overline{TRST}$, TMS
5	$\overline{RESET}$, $\overline{NMI}$, CLKBUF
6	PH7, PH6, PH5, PH4, PH3, PH2, PH1, PH0
7	PF15, PF14, PF13, PF12, PF11, SDA, SCL, PF10

ESD SENSITIVITY

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

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TIMING SPECIFICATIONS

Clock and Reset Timing

Table 20 and Figure 7 describe clock and reset operations. Per the CCLK and SCLK timing specifications in Table 9, Table 10, and Table 11, combinations of CLKIN and clock multipliers must not select core/peripheral clocks in excess of the processor's speed grade.

Table 20. Clock and Reset Timing

Parameter		Min	Max	Unit
Timing Requirements				
f_{CKIN}	CLKIN Frequency (Commercial/Industrial Models[1, 2, 3, 4]	12	50	MHz
f_{CKIN}	CLKIN Frequency (Automotive Models)[1, 2, 3, 4]	14	50	MHz
t_{CKINL}	CLKIN Low Pulse[1]	10		ns
t_{CKINH}	CLKIN High Pulse[1]	10		ns
t_{WRST}	$\overline{\text{RESET}}$ Asserted Pulse Width Low[5]	11 × t_{CKIN}		ns
Switching Characteristic				
$t_{BUFDLAY}$	CLKIN to CLKBUF Delay		11	ns

[1] Applies to PLL bypass mode and PLL nonbypass mode.
[2] Combinations of the CLKIN frequency and the PLL clock multiplier must not exceed the allowed f_{VCO}, f_{CCLK}, and f_{SCLK} settings discussed in Table 9 through Table 11.
[3] The t_{CKIN} period (see Figure 7) equals $1/f_{CKIN}$.
[4] If the DF bit in the PLL_CTL register is set, the minimum f_{CKIN} specification is 24 MHz for commercial/industrial models and 28 MHz for automotive models.
[5] Applies after power-up sequence is complete. See Table 21 and Figure 8 for power-up reset timing.

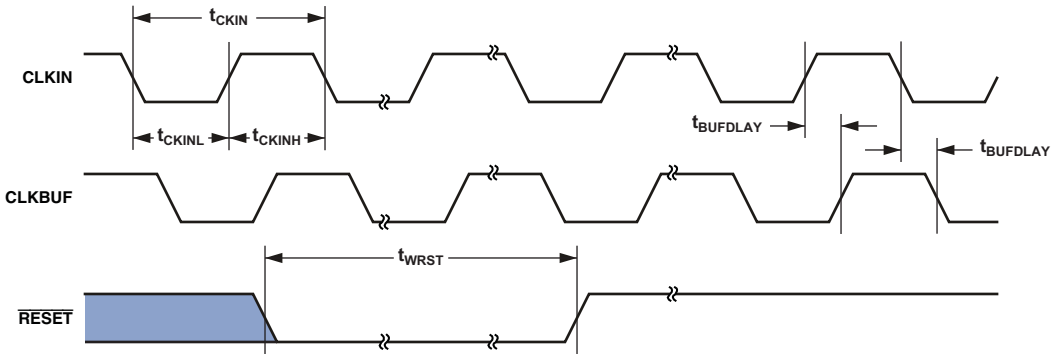

Figure 7. Clock and Reset Timing

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Table 21. Power-Up Reset Timing

Parameter		Min	Max	Unit
Timing Requirements				
$t_{RST_IN_PWR}$	$\overline{\text{RESET}}$ Deasserted after the V_{DDINT}, V_{DDEXT}, V_{DDRTC}, V_{DDMEM}, V_{DDOTP}, and CLKIN Pins are Stable and Within Specification	$3500 \times t_{CKIN}$		ns

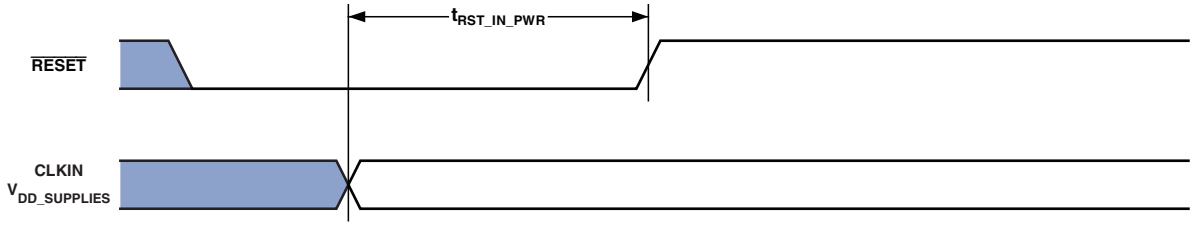

Figure 8. Power-Up Reset Timing

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Asynchronous Memory Read Cycle Timing

Table 22. Asynchronous Memory Read Cycle Timing

Parameter		V_{DDMEM} 1.8 V Nominal Min	V_{DDMEM} 1.8 V Nominal Max	V_{DDMEM} 2.5 V/3.3 V Nominal Min	V_{DDMEM} 2.5 V/3.3 V Nominal Max	Unit
Timing Requirements						
t_{SDAT}	DATA15–0 Setup Before CLKOUT	2.1		2.1		ns
t_{HDAT}	DATA15–0 Hold After CLKOUT	1.2		0.8		ns
t_{SARDY}	ARDY Setup Before CLKOUT	4		4		ns
t_{HARDY}	ARDY Hold After CLKOUT	0.2		0.2		ns
Switching Characteristics						
t_{DO}	Output Delay After CLKOUT[1]		6		6	ns
t_{HO}	Output Hold After CLKOUT [1]	0.8		0.8		ns

[1] Output pins/balls include $\overline{AMS3-0}$, $\overline{ABE1-0}$, ADDR19–1, $\overline{AOE}$, $\overline{ARE}$.

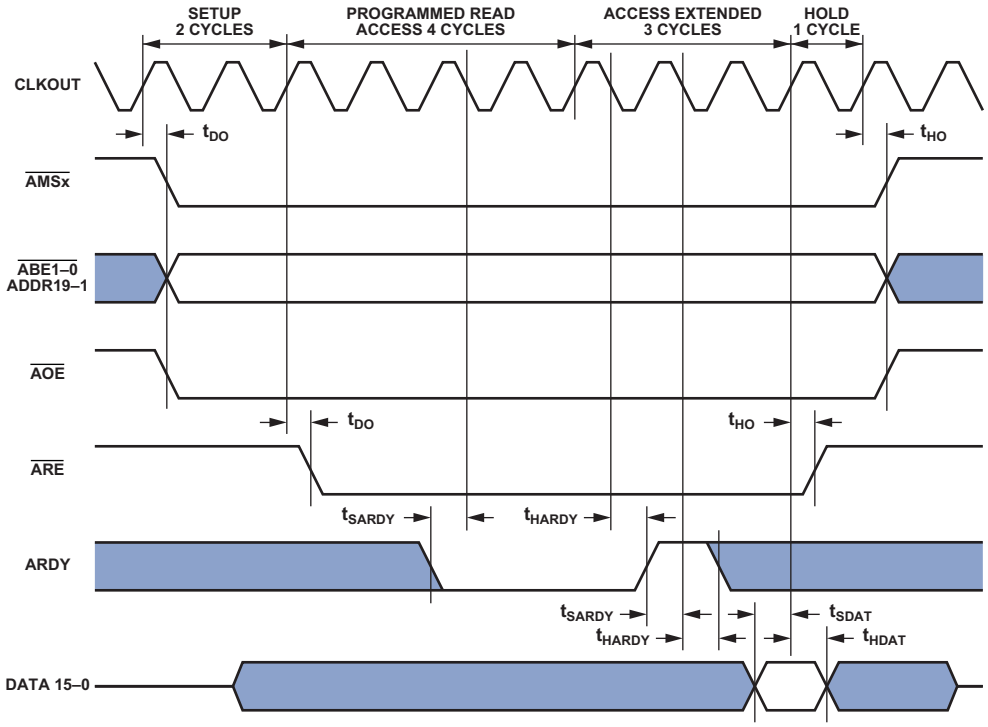

Figure 9. Asynchronous Memory Read Cycle Timing

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Asynchronous Memory Write Cycle Timing

Table 23. Asynchronous Memory Write Cycle Timing

Parameter		Min	Max	Unit
Timing Requirements				
t_{SARDY}	ARDY Setup Before CLKOUT	4		ns
t_{HARDY}	ARDY Hold After CLKOUT	0.2		ns
Switching Characteristics				
t_{DDAT}	DATA15–0 Disable After CLKOUT		6	ns
t_{ENDAT}	DATA15–0 Enable After CLKOUT	0		ns
t_{DO}	Output Delay After CLKOUT[1]		6	ns
t_{HO}	Output Hold After CLKOUT [1]	0.8		ns

[1] Output pins/balls include $\overline{AMS3-0}$, $\overline{ABE1-0}$, ADDR19–1, DATA15–0, $\overline{AOE}$, $\overline{AWE}$.

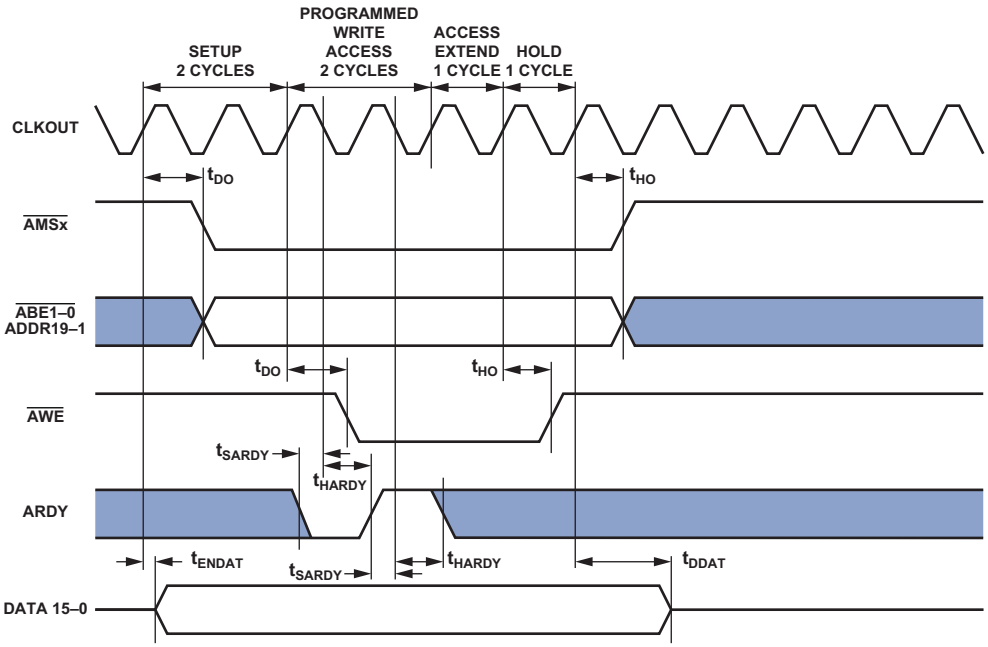

Figure 10. Asynchronous Memory Write Cycle Timing

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SDRAM Interface Timing

Table 24. SDRAM Interface Timing

Parameter		V_{DDMEM} 1.8 V Nominal		V_{DDMEM} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirements						
t_{SSDAT}	Data Setup Before CLKOUT	1.5		1.5		ns
t_{HSDAT}	Data Hold After CLKOUT	1.3		0.8		ns
Switching Characteristics						
t_{SCLK}	CLKOUT Period[1]	12.5		10		ns
t_{SCLKH}	CLKOUT Width High	5		4		ns
t_{SCLKL}	CLKOUT Width Low	5		4		ns
t_{DCAD}	Command, Address, Data Delay After CLKOUT[2]		5		4	ns
t_{HCAD}	Command, Address, Data Hold After CLKOUT[2]	1		1		ns
t_{DSDAT}	Data Disable After CLKOUT		5.5		5	ns
t_{ENSDAT}	Data Enable After CLKOUT	0		0		ns

[1] The t_{SCLK} value is the inverse of the f_{SCLK} specification discussed in Table 11. Package type and reduced supply voltages affect the best-case value listed here.
[2] Command pins/balls include: $\overline{SRAS}$, $\overline{SCAS}$, $\overline{SWE}$, SDQM, $\overline{SMS}$, SA10, SCKE.

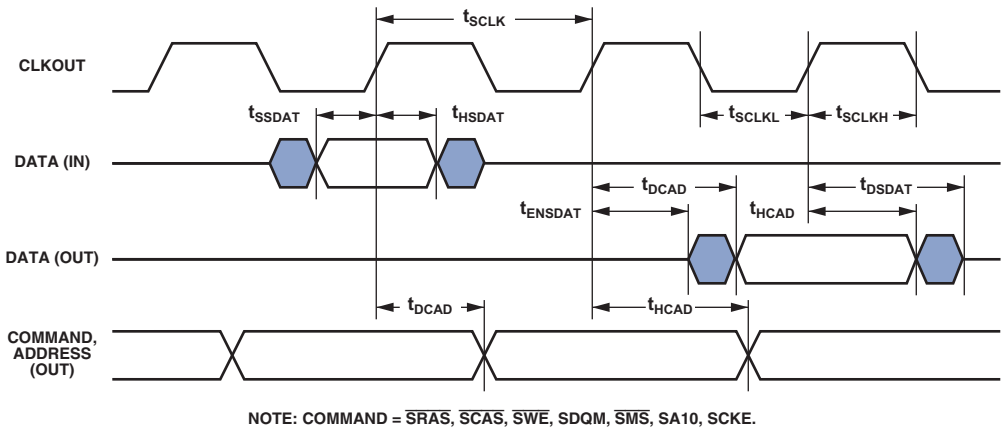

Figure 11. SDRAM Interface Timing

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External DMA Request Timing

Table 25 and Figure 12 describe the External DMA Request operations.

Table 25. External DMA Request Timing[1]

Parameter		V_{DDMEM}/V_{DDEXT} 1.8 V Nominal		V_{DDMEM}/V_{DDEXT} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirements						
t_{DS}	DMARx Asserted to CLKOUT High Setup	9		7.2		ns
t_{DH}	CLKOUT High to DMARx Deasserted Hold Time	0		0		ns
$t_{DMARACT}$	DMARx Active Pulse Width	$t_{SCLK} + 1$		$t_{SCLK} + 1$		ns
$t_{DMARINACT}$	DMARx Inactive Pulse Width	$1.75 \times t_{SCLK}$		$1.75 \times t_{SCLK}$		ns

[1] Because the external DMA control pins are part of the V_{DDEXT} power domain and the CLKOUT signal is part of the V_{DDMEM} power domain, systems in which V_{DDEXT} and V_{DDMEM} are NOT equal may require level shifting logic for correct operation.

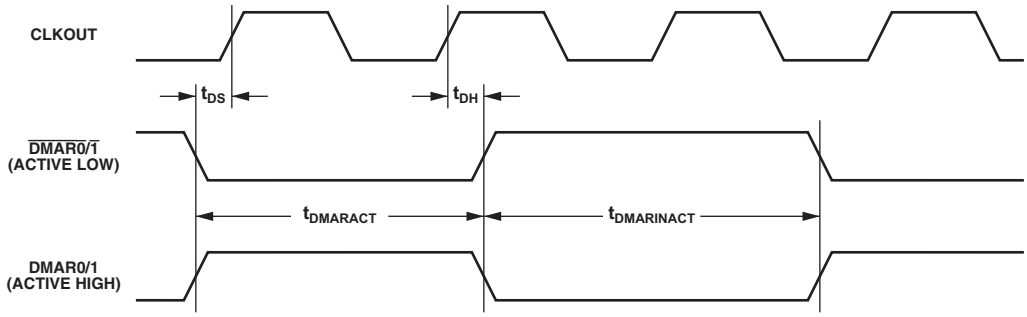

Figure 12. External DMA Request Timing

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Parallel Peripheral Interface Timing

Table 26 and Figure 13 through Figure 17 and describe parallel peripheral interface operations.

Table 26. Parallel Peripheral Interface Timing

Parameter		V_{DDEXT} 1.8 V Nominal		V_{DDEXT} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirements						
t_{PCLKW}	PPI_CLK Width	t_{SCLK} – 1.5		t_{SCLK} – 1.5		ns
t_{PCLK}	PPI_CLK Period	2 × t_{SCLK} – 1.5		2 × t_{SCLK} – 1.5		ns
Timing Requirements - GP Input and Frame Capture Modes						
t_{PSUD}	External Frame Sync Startup Delay[1]	4 × t_{PCLK}		4 × t_{PCLK}		ns
t_{SFSPE}	External Frame Sync Setup Before PPI_CLK (Nonsampling Edge for Rx, Sampling Edge for Tx)	6.7		6.7		ns
t_{HFSPE}	External Frame Sync Hold After PPI_CLK	1.75		1.75		ns
t_{SDRPE}	Receive Data Setup Before PPI_CLK	4.1		3.5		ns
t_{HDRPE}	Receive Data Hold After PPI_CLK	2		1.6		ns
Switching Characteristics - GP Output and Frame Capture Modes						
t_{DFSPE}	Internal Frame Sync Delay After PPI_CLK		8		8	ns
t_{HOFSPE}	Internal Frame Sync Hold After PPI_CLK	1.7		1.7		ns
t_{DDTPE}	Transmit Data Delay After PPI_CLK		8.2		8	ns
t_{HDTPE}	Transmit Data Hold After PPI_CLK	2.3		1.9		ns

[1] The PPI port is fully enabled 4 PPI clock cycles after the PAB write to the PPI port enable bit. Only after the PPI port is fully enabled are external frame syncs and data words guaranteed to be received correctly by the PPI peripheral.

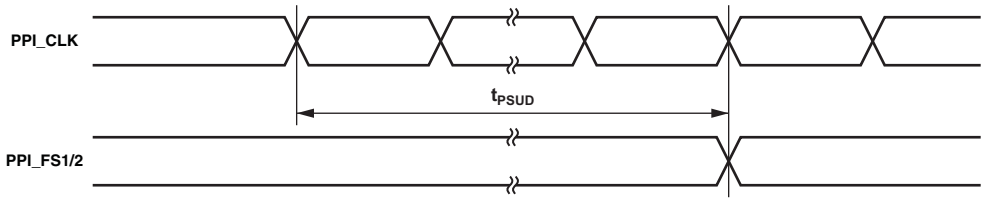

Figure 13. PPI with External Frame Sync Timing

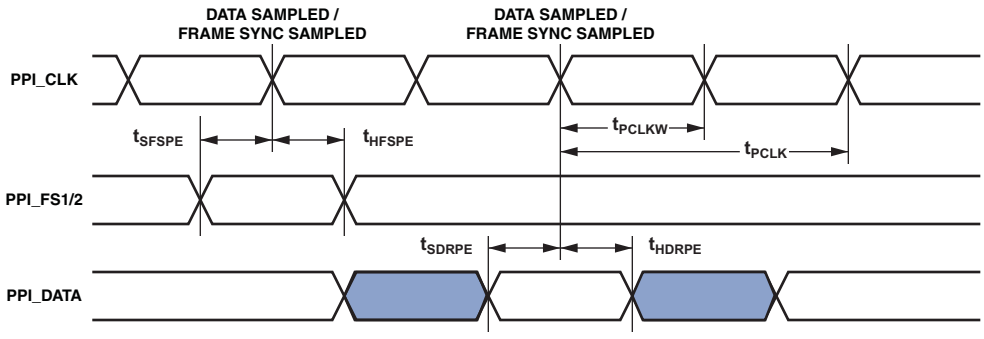

Figure 14. PPI GP Rx Mode with External Frame Sync Timing

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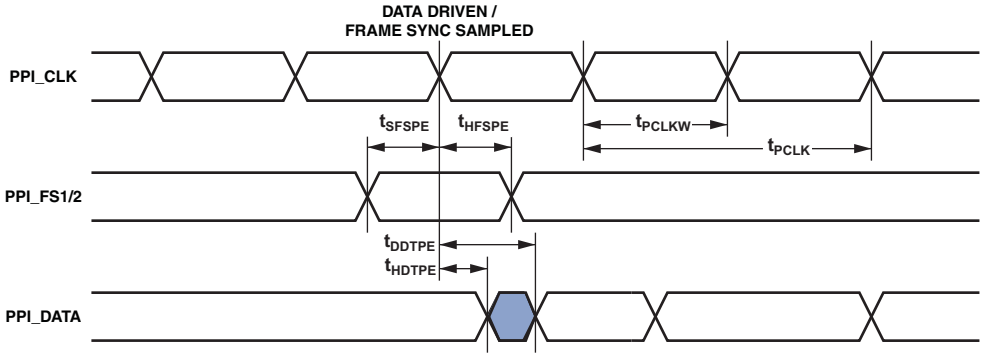

Figure 15. PPI GP Tx Mode with External Frame Sync Timing

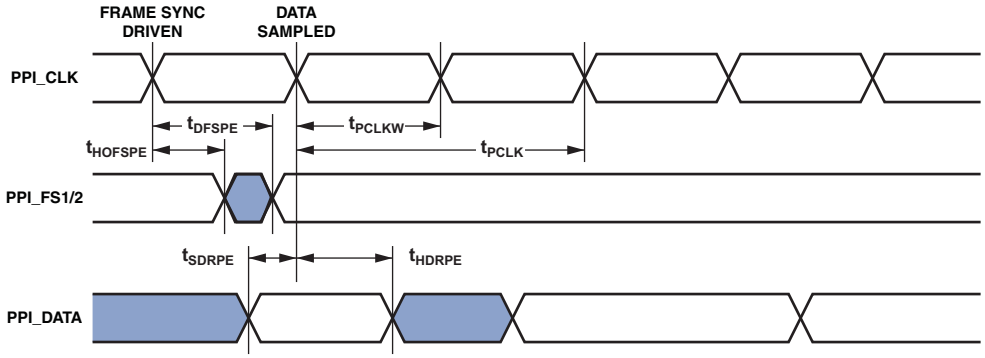

Figure 16. PPI GP Rx Mode with Internal Frame Sync Timing

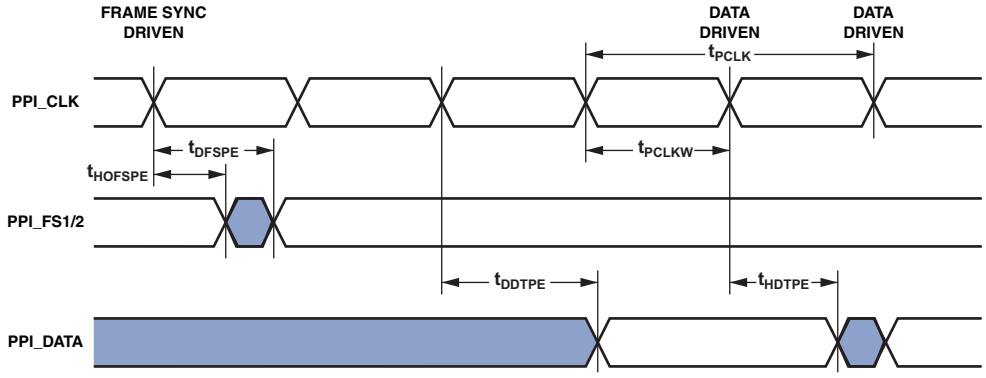

Figure 17. PPI GP Tx Mode with Internal Frame Sync Timing

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RSI Controller Timing

Table 27 and Figure 18 describe RSI controller timing. Table 28 and Figure 19 describe RSI controller (high speed) timing.

Table 27. RSI Controller Timing

Parameter		Min	Max	Unit
Timing Requirements				
t_{ISU}	Input Setup Time	5.6		ns
t_{IH}	Input Hold Time	2		ns
Switching Characteristics				
f_{PP}[1]	Clock Frequency Data Transfer Mode	0	25	MHz
f_{OD}	Clock Frequency Identification Mode	100[2]	400	kHz
t_{WL}	Clock Low Time	10		ns
t_{WH}	Clock High Time	10		ns
t_{TLH}	Clock Rise Time		10	ns
t_{THL}	Clock Fall Time		10	ns
t_{ODLY}	Output Delay Time During Data Transfer Mode		14	ns
t_{ODLY}	Output Delay Time During Identification Mode		50	ns

[1] $t_{PP} = 1/f_{PP}$
[2] Specification can be 0 kHz, which means to stop the clock. The given minimum frequency range is for cases where a continuous clock is required.

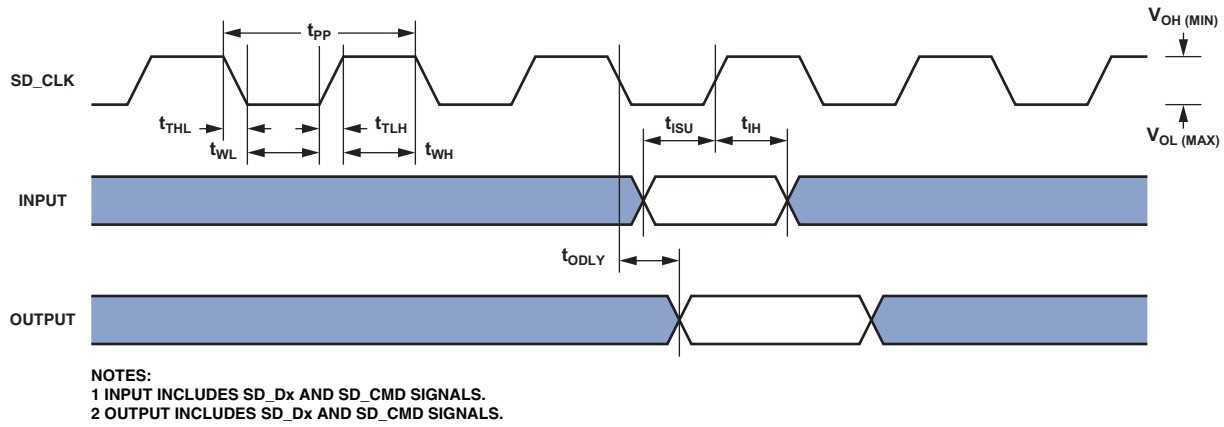

Figure 18. RSI Controller Timing

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Table 28. RSI Controller Timing (High Speed Mode)

Parameter		Min	Max	Unit
Timing Requirements				
t_{ISU}	Input Setup Time	5.6		ns
t_{IH}	Input Hold Time	2		ns
Switching Characteristics				
f_{PP}[1]	Clock Frequency Data Transfer Mode	0	50	MHz
t_{WL}	Clock Low Time	7		ns
t_{WH}	Clock High Time	7		ns
t_{TLH}	Clock Rise Time		3	ns
t_{THL}	Clock Fall Time		3	ns
t_{ODLY}	Output Delay Time During Data Transfer Mode		4	ns
t_{OH}	Output Hold Time	2.75		ns

[1] $t_{PP} = 1/f_{PP}$

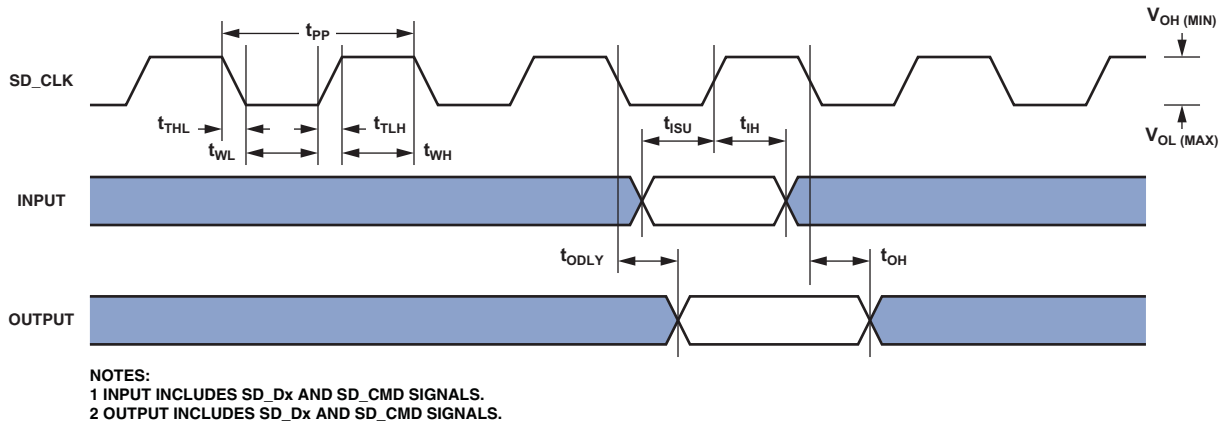

Figure 19. RSI Controller Timing (High Speed Mode)

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Serial Ports

Table 29 through Table 32 and Figure 20 through Figure 23 describe serial port operations.

Table 29. Serial Ports—External Clock

Parameter		V_{DDEXT} 1.8 V Nominal		V_{DDEXT} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirements						
t_{SFSE}[1]	TFSx/RFSx Setup Before TSCLKx/RSCLKx	3		3		ns
t_{HFSE}[1]	TFSx/RFSx Hold After TSCLKx/RSCLKx	3		3		ns
t_{SDRE}[1]	Receive Data Setup Before RSCLKx	3		3		ns
t_{HDRE}[1]	Receive Data Hold After RSCLKx	3.5		3		ns
t_{SCLKEW}	TSCLKx/RSCLKx Width	7		4.5		ns
t_{SCLKE}	TSCLKx/RSCLKx Period	$2 \times t_{SCLK}$		$2 \times t_{SCLK}$		ns
t_{SUDTE}[2]	Start-Up Delay From SPORT Enable To First External TFSx	$4 \times t_{SCLKE}$		$4 \times t_{SCLKE}$		ns
t_{SUDRE}[2]	Start-Up Delay From SPORT Enable To First External RFSx	$4 \times t_{SCLKE}$		$4 \times t_{SCLKE}$		ns
Switching Characteristics						
t_{DFSE}[3]	TFSx/RFSx Delay After TSCLKx/RSCLKx (Internally Generated TFSx/RFSx)		10		10	ns
t_{HOFSE}[3]	TFSx/RFSx Hold After TSCLKx/RSCLKx (Internally Generated TFSx/RFSx)	0		0		ns
t_{DDTE}[3]	Transmit Data Delay After TSCLKx		10		10	ns
t_{HDTE}[3]	Transmit Data Hold After TSCLKx	0		0		ns

[1] Referenced to sample edge.
[2] Verified in design but untested.
[3] Referenced to drive edge.

Table 30. Serial Ports—Internal Clock

Parameter		V_{DDEXT} 1.8 V Nominal		V_{DDEXT} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirements						
t_{SFSI}[1]	TFSx/RFSx Setup Before TSCLKx/RSCLKx	11		9.6		ns
t_{HFSI}[1]	TFSx/RFSx Hold After TSCLKx/RSCLKx	−1.5		−1.5		ns
t_{SDRI}[1]	Receive Data Setup Before RSCLKx	11		9.6		ns
t_{HDRI}[1]	Receive Data Hold After RSCLKx	−1.5		−1.5		ns
Switching Characteristics						
t_{DFSI}[2]	TFSx/RFSx Delay After TSCLKx/RSCLKx (Internally Generated TFSx/RFSx)		3		3	ns
t_{HOFSI}[2]	TFSx/RFSx Hold After TSCLKx/RSCLKx (Internally Generated TFSx/RFSx)	−2		−1		ns
t_{DDTI}[2]	Transmit Data Delay After TSCLKx		3		3	ns
t_{HDTI}[2]	Transmit Data Hold After TSCLKx	−1.8		−1.5		ns
t_{SCLKIW}	TSCLKx/RSCLKx Width	10		8		ns

[1] Referenced to sample edge.
[2] Referenced to drive edge.

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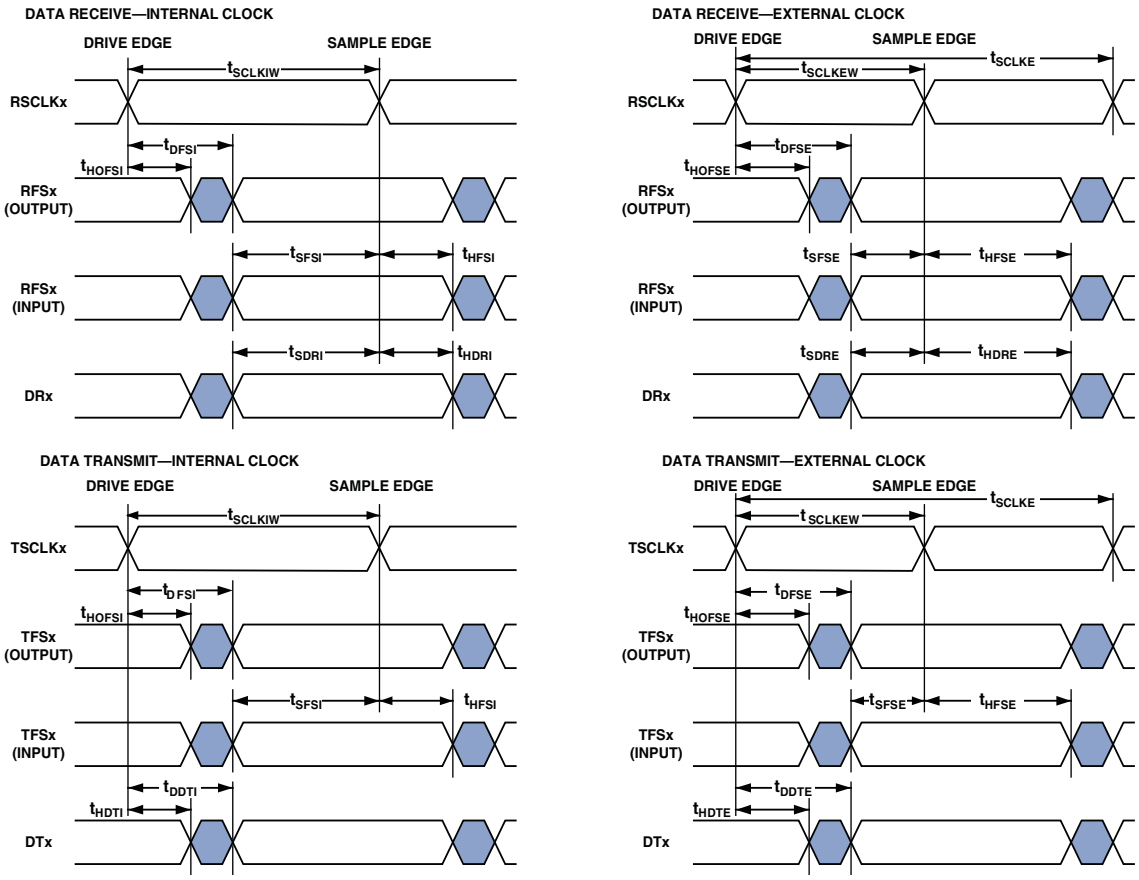

Figure 20. Serial Ports

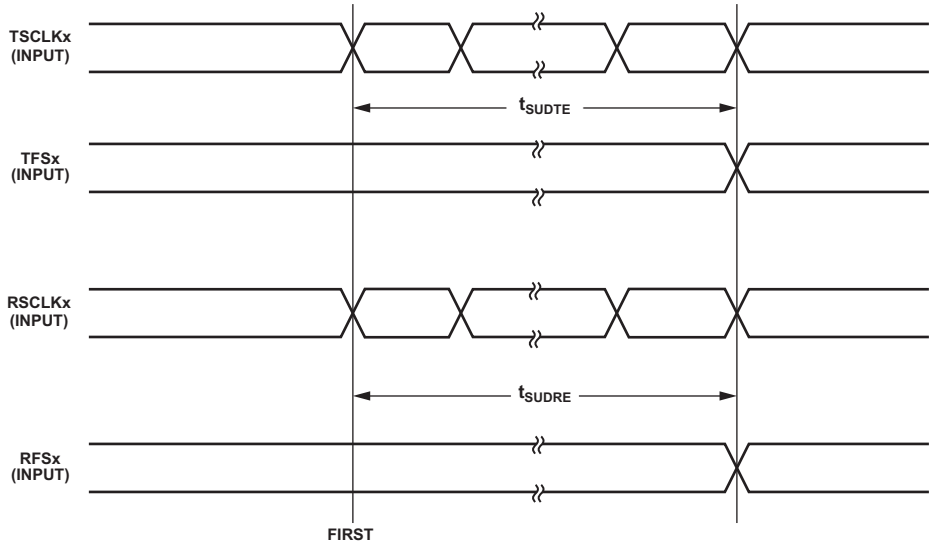

Figure 21. Serial Port Start Up with External Clock and Frame Sync

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Table 31. Serial Ports—Enable and Three-State[1]

Parameter		Min	Max	Unit
Switching Characteristics				
t_{DTENE}	Data Enable Delay from External TSCLKx	0		ns
t_{DDTTE}	Data Disable Delay from External TSCLKx		t_{SCLK} + 1	ns
t_{DTENI}	Data Enable Delay from Internal TSCLKx	–2.0		ns
t_{DDTTI}	Data Disable Delay from Internal TSCLKx		t_{SCLK} + 1	ns

[1] Referenced to drive edge.

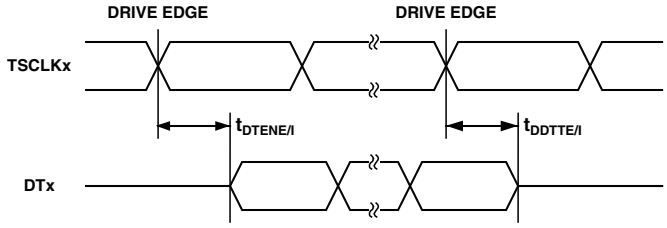

Figure 22. Enable and Three-State

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Table 32. External Late Frame Sync

Parameter		V_{DDEXT} 1.8 V Nominal Min	V_{DDEXT} 1.8 V Nominal Max	V_{DDEXT} 2.5 V/3.3 V Nominal Min	V_{DDEXT} 2.5 V/3.3 V Nominal Max	Unit
Switching Characteristics						
$t_{DDTLFSE}$[1, 2]	Data Delay from Late External TFSx or External RFSx with MCE = 1, MFD = 0		12		10	ns
$t_{DTENLFSE}$[1, 2]	Data Enable from Late FS or MCE = 1, MFD = 0	0		0		ns

[1] MCE = 1, TFSx enable and TFSx valid follow $t_{DDTENFS}$ and $t_{DDTLFSE}$.
[2] If external RFSx/TFSx setup to RSCLKx/TSCLKx > $t_{SCLKE}/2$ then $t_{DDTTE/I}$ and $t_{DTENE/I}$ apply, otherwise $t_{DDTLFSE}$ and $t_{DTENLFS}$ apply.

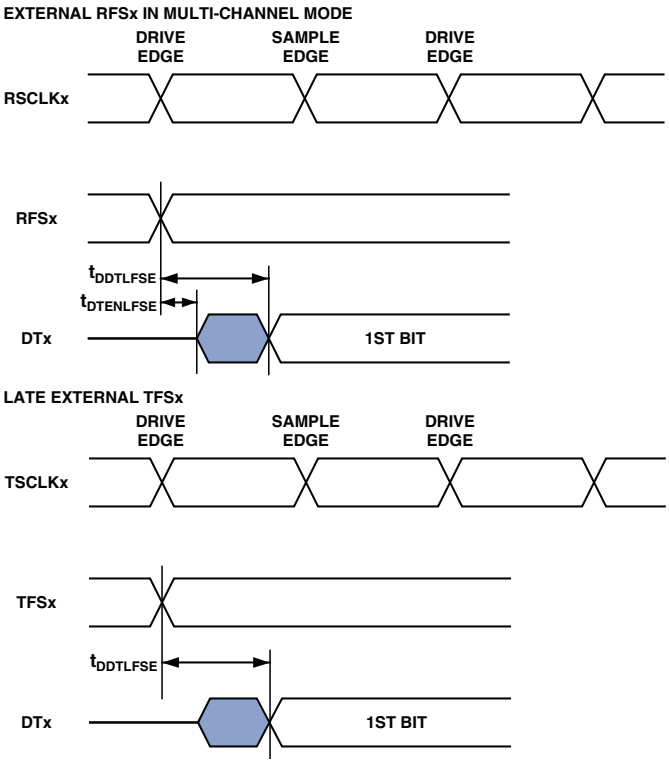

Figure 23. External Late Frame Sync

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Serial Peripheral Interface (SPI) Port—Master Timing

Table 33 and Figure 24 describe SPI port master operations.

Table 33. Serial Peripheral Interface (SPI) Port—Master Timing

Parameter		V_{DDEXT} 1.8 V Nominal		V_{DDEXT} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirements						
t_{SSPIDM}	Data Input Valid to SCK Edge (Data Input Setup)	11.6		9.6		ns
t_{HSPIDM}	SCK Sampling Edge to Data Input Invalid	–1.5		–1.5		ns
Switching Characteristics						
t_{SDSCIM}	$\overline{SPISEL}$x low to First SCK Edge	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SPICHM}	Serial Clock High Period	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SPICLM}	Serial Clock Low Period	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SPICLK}	Serial Clock Period	$4 \times t_{SCLK}$		$4 \times t_{SCLK}$		ns
t_{HDSM}	Last SCK Edge to $\overline{SPISEL}$x High	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SPITDM}	Sequential Transfer Delay	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
$t_{DDSPIDM}$	SCK Edge to Data Out Valid (Data Out Delay)		6		6	ns
$t_{HDSPIDM}$	SCK Edge to Data Out Invalid (Data Out Hold)	–1		–1		ns

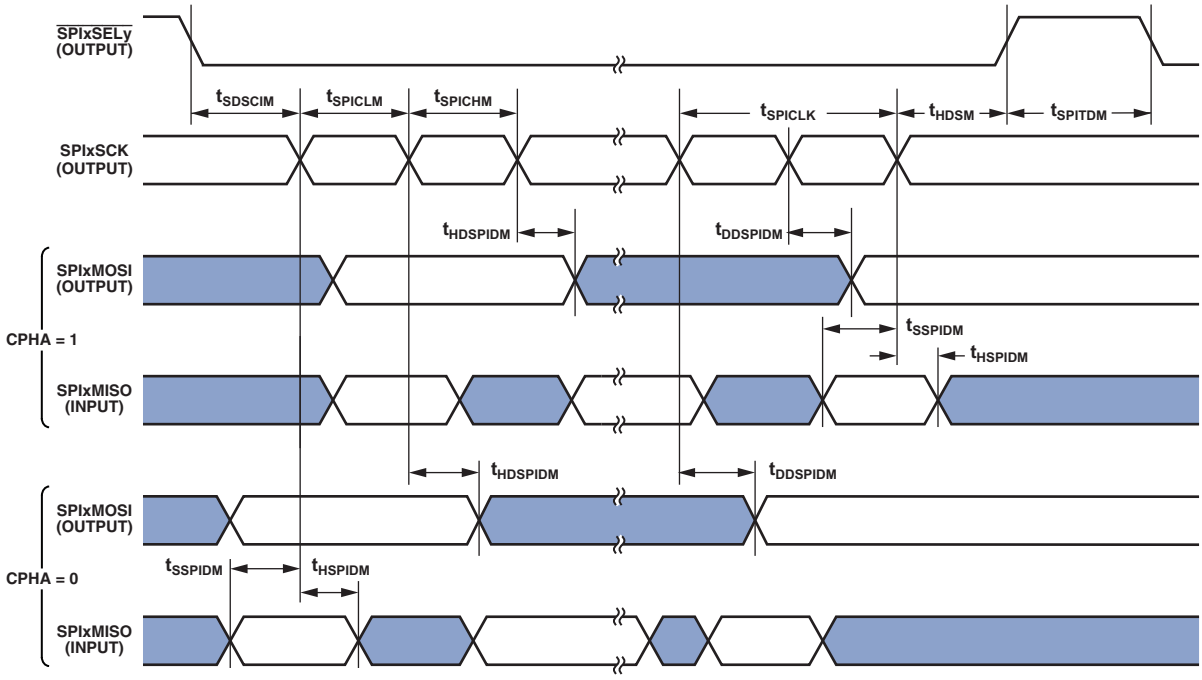

Figure 24. Serial Peripheral Interface (SPI) Port—Master Timing

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Serial Peripheral Interface (SPI) Port—Slave Timing

Table 34 and Figure 25 describe SPI port slave operations.

Table 34. Serial Peripheral Interface (SPI) Port—Slave Timing

Parameter		V_{DDEXT} 1.8 V Nominal		V_{DDEXT} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirements						
t_{SPICHS}	Serial Clock High Period	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SPICLS}	Serial Clock Low Period	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SPICLK}	Serial Clock Period	$4 \times t_{SCLK} - 1.5$		$4 \times t_{SCLK} - 1.5$		ns
t_{HDS}	Last SCK Edge to $\overline{SPISS}$ Not Asserted	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SPITDS}	Sequential Transfer Delay	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SDSCI}	$\overline{SPISS}$ Assertion to First SCK Edge	$2 \times t_{SCLK} - 1.5$		$2 \times t_{SCLK} - 1.5$		ns
t_{SSPID}	Data Input Valid to SCK Edge (Data Input Setup)	1.6		1.6		ns
t_{HSPID}	SCK Sampling Edge to Data Input Invalid	2		1.6		ns
Switching Characteristics						
t_{DSOE}	$\overline{SPISS}$ Assertion to Data Out Active	0	12	0	10.3	ns
t_{DSDHI}	$\overline{SPISS}$ Deassertion to Data High Impedance	0	11	0	9	ns
t_{DDSPID}	SCK Edge to Data Out Valid (Data Out Delay)		10		10	ns
t_{HDSPID}	SCK Edge to Data Out Invalid (Data Out Hold)	0		0		ns

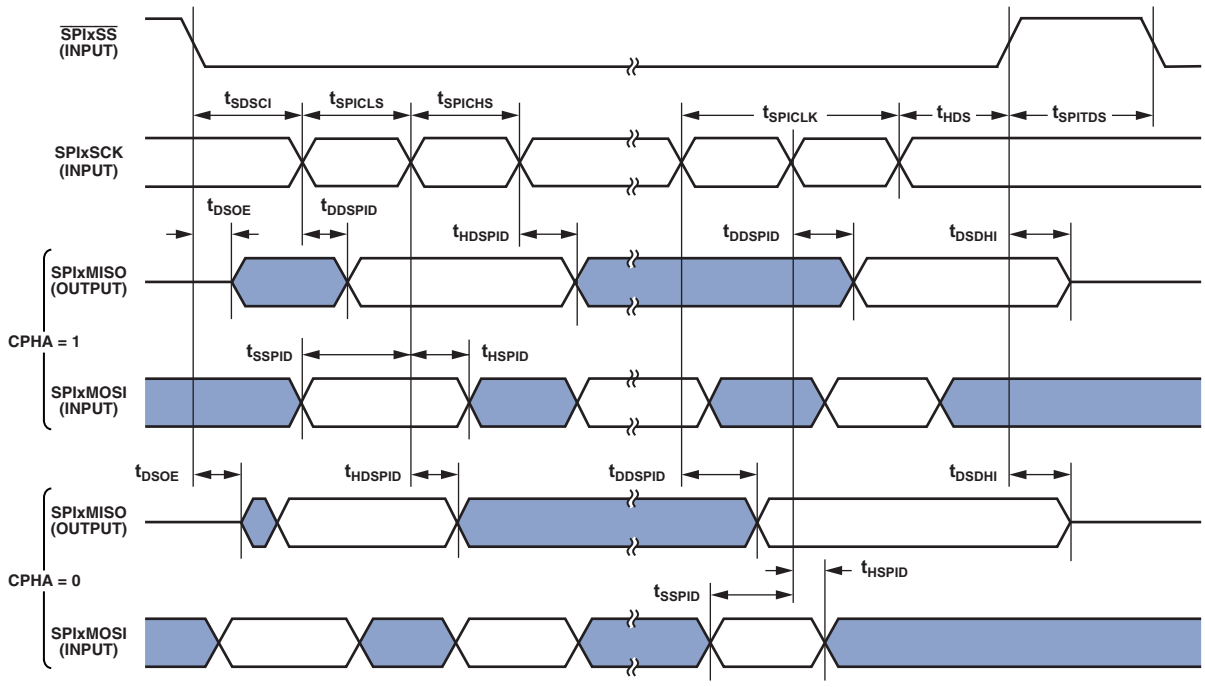

Figure 25. Serial Peripheral Interface (SPI) Port—Slave Timing

Universal Asynchronous Receiver-Transmitter (UART) Ports—Receive and Transmit Timing

The UART ports receive and transmit operations are described in the *ADSP-BF51x Hardware Reference Manual.*

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General-Purpose Port Timing

Table 35 and Figure 26 describe general-purpose port operations.

Table 35. General-Purpose Port Timing

Parameter		V_{DDEXT} 1.8 V Nominal		V_{DDEXT} 2.5 V/3.3 V Nominal		Unit
		Min	Max	Min	Max	
Timing Requirement						
t_{WFI}	General-Purpose Port Signal Input Pulse Width	t_{SCLK} + 1		t_{SCLK} + 1		ns
Switching Characteristic						
t_{GPOD}	General-Purpose Port Signal Output Delay from CLKOUT Low	0	11	0	8.5	ns

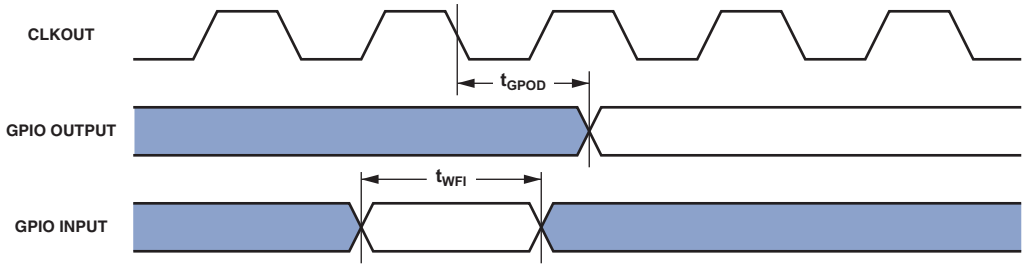

Figure 26. General-Purpose Port Timing

Timer Clock Timing

Table 36 and Figure 27 describe timer clock timing.

Table 36. Timer Clock Timing

Parameter		Min	Max	Unit
Switching Characteristic				
t_{TODP}	Timer Output Update Delay After PPICLK High		12	ns

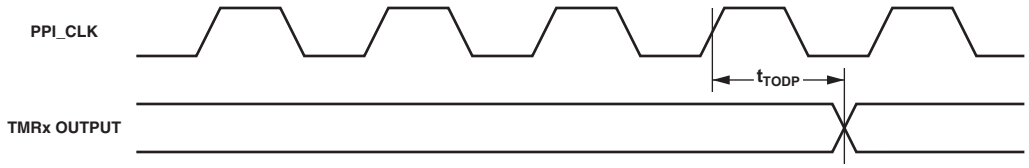

Figure 27. Timer Clock Timing

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Timer Cycle Timing

Table 37 and Figure 28 describe timer expired operations. The input signal is asynchronous in "width capture mode" and "external clock mode" and has an absolute maximum input frequency of (f_{SCLK}/2) MHz.

Table 37. Timer Cycle Timing

		V_{DDEXT} 1.8 V Nominal		V_{DDEXT} 2.5 V/3.3 V Nominal		
Parameter		**Min**	**Max**	**Min**	**Max**	**Unit**
Timing Characteristics						
t_{WL}[1]	Timer Pulse Width Input Low (Measured In SCLK Cycles)	t_{SCLK}		t_{SCLK}		ns
t_{WH}[1]	Timer Pulse Width Input High (Measured In SCLK Cycles)	t_{SCLK}		t_{SCLK}		ns
t_{TIS}[2]	Timer Input Setup Time Before CLKOUT Low	10		7		ns
t_{TIH}[2]	Timer Input Hold Time After CLKOUT Low	–2		–2		ns
Switching Characteristics						
t_{HTO}	Timer Pulse Width Output (Measured In SCLK Cycles)	t_{SCLK} – 1.5	$(2^{32}–1)t_{SCLK}$	t_{SCLK} – 1	$(2^{32}–1)t_{SCLK}$	ns
t_{TOD}	Timer Output Update Delay After CLKOUT High		6		6	ns

[1] The minimum pulse widths apply for TMRx signals in width capture and external clock modes. They also apply to the PF15 or PPI_CLK signals in PWM output mode.
[2] Either a valid setup and hold time or a valid pulse width is sufficient. There is no need to resynchronize programmable flag inputs.

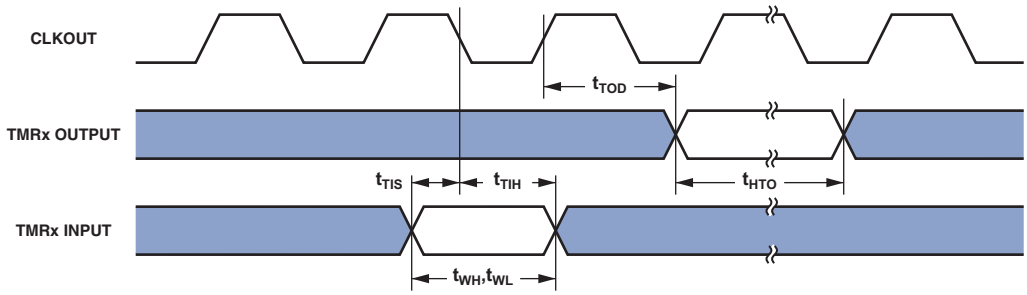

Figure 28. Timer Cycle Timing

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Up/Down Counter/Rotary Encoder Timing

Table 38. Up/Down Counter/Rotary Encoder Timing

Parameter		V_{DDEXT} 1.8 V Nominal Min	V_{DDEXT} 1.8 V Nominal Max	V_{DDEXT} 2.5 V/3.3 V Nominal Min	V_{DDEXT} 2.5 V/3.3 V Nominal Max	Unit
Timing Requirements						
t_{WCOUNT}	Up/Down Counter/Rotary Encoder Input Pulse Width	t_{SCLK} + 1		t_{SCLK} + 1		ns
t_{CIS}	Counter Input Setup Time Before CLKOUT Low[1]	9		7		ns
t_{CIH}	Counter Input Hold Time After CLKOUT Low[1]	0		0		ns

[1] Either a valid setup and hold time or a valid pulse width is sufficient. There is no need to resynchronize counter inputs.

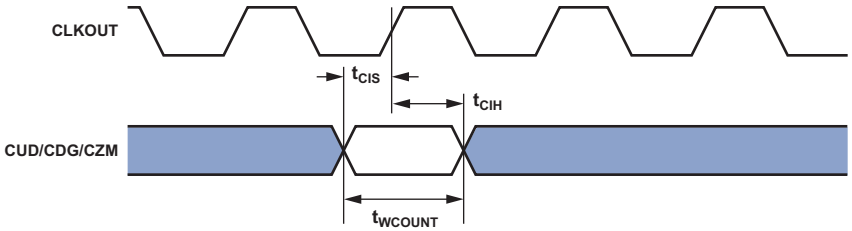

Figure 29. Up/Down Counter/Rotary Encoder Timing

10/100 Ethernet MAC Controller Timing

Table 39 through Table 44 and Figure 30 through Figure 35 describe the 10/100 Ethernet MAC Controller operations.

Table 39. 10/100 Ethernet MAC Controller Timing: MII Receive Signal

Parameter[1]		V_{DDEXT} 1.8 V Nominal Min	V_{DDEXT} 1.8 V Nominal Max	V_{DDEXT} 2.5 V/3.3 V Nominal Min	V_{DDEXT} 2.5 V/3.3 V Nominal Max	Unit
Timing Requirements						
$t_{ERXCLKF}$	ERxCLK Frequency (f_{SCLK} = SCLK Frequency)	None	25 + 1%	None	25 + 1%	MHz
$t_{ERXCLKW}$	ERxCLK Width (t_{ERxCLK} = ERxCLK Period)	t_{ERxCLK} × 40%	t_{ERxCLK} × 60%	t_{ERxCLK} × 35%	t_{ERxCLK} × 65%	ns
$t_{ERXCLKIS}$	Rx Input Valid to ERxCLK Rising Edge (Data In Setup)	7.5		7.5		ns
$t_{ERXCLKIH}$	ERxCLK Rising Edge to Rx Input Invalid (Data In Hold)	7.5		7.5		ns

[1] MII inputs synchronous to ERxCLK are ERxD3–0, ERxDV, and ERxER.

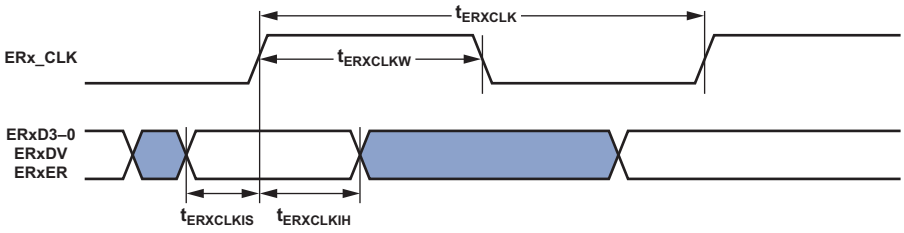

Figure 30. 10/100 Ethernet MAC Controller Timing: MII Receive Signal

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Table 40. 10/100 Ethernet MAC Controller Timing: MII Transmit Signal

Parameter[1]		V_{DDEXT} 1.8 V Nominal Min	V_{DDEXT} 1.8 V Nominal Max	V_{DDEXT} 2.5 V/3.3 V Nominal Min	V_{DDEXT} 2.5 V/3.3 V Nominal Max	Unit
Switching Characteristics						
t_{ETF}	ETxCLK Frequency (f_{SCLK} = SCLK Frequency)	None	25 + 1%	None	25 + 1%	MHz
$t_{ETXCLKW}$	ETxCLK Width (t_{ETxCLK} = ETxCLK Period)	$t_{ETxCLK} \times 40\%$	$t_{ETxCLK} \times 60\%$	$t_{ETxCLK} \times 35\%$	$t_{ETxCLK} \times 65\%$	ns
$t_{ETXCLKOV}$	ETxCLK Rising Edge to Tx Output Valid (Data Out Valid)		20		20	ns
$t_{ETXCLKOH}$	ETxCLK Rising Edge to Tx Output Invalid (Data Out Hold)	0		0		ns

[1] MII outputs synchronous to ETxCLK are ETxD3–0.

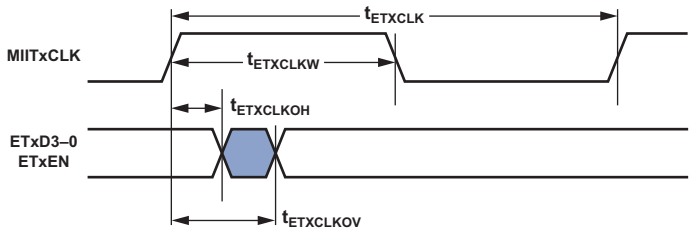

Figure 31. 10/100 Ethernet MAC Controller Timing: MII Transmit Signal

Table 41. 10/100 Ethernet MAC Controller Timing: RMII Receive Signal

Parameter[1]		V_{DDEXT} 1.8 V Nominal Min	V_{DDEXT} 1.8 V Nominal Max	V_{DDEXT} 2.5 V/3.3 V Nominal Min	V_{DDEXT} 2.5 V/3.3 V Nominal Max	Unit
Timing Requirements						
$t_{EREFCLKF}$	REF_CLK Frequency (f_{SCLK} = SCLK Frequency)	None	50 + 1%	None	50 + 1%	MHz
$t_{EREFCLKW}$	EREF_CLK Width ($t_{EREFCLK}$ = EREFCLK Period)	$t_{EREFCLK} \times 40\%$	$t_{EREFCLK} \times 60\%$	$t_{EREFCLK} \times 35\%$	$t_{EREFCLK} \times 65\%$	ns
$t_{EREFCLKIS}$	Rx Input Valid to RMII REF_CLK Rising Edge (Data In Setup)	4		4		ns
$t_{EREFCLKIH}$	RMII REF_CLK Rising Edge to Rx Input Invalid (Data In Hold)	2		2		ns

[1] RMII inputs synchronous to RMII REF_CLK are ERxD1–0, RMII CRS_DV, and ERxER.

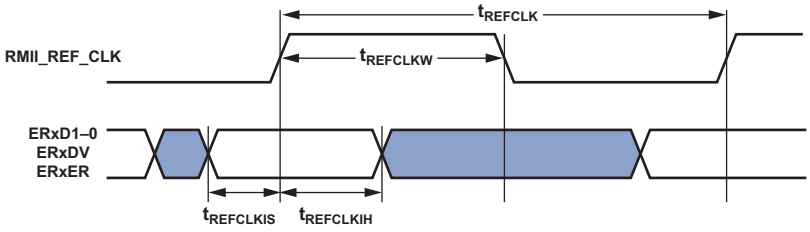

Figure 32. 10/100 Ethernet MAC Controller Timing: RMII Receive Signal

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Table 42. 10/100 Ethernet MAC Controller Timing: RMII Transmit Signal

Parameter[1]		Min	Max	Unit
Switching Characteristics				
$t_{EREFCLKOV}$	RMII REF_CLK Rising Edge to Tx Output Valid (Data Out Valid)		8.1	ns
$t_{EREFCLKOH}$	RMII REF_CLK Rising Edge to Tx Output Invalid (Data Out Hold)	2		ns

[1] RMII outputs synchronous to RMII REF_CLK are ETxD1–0.

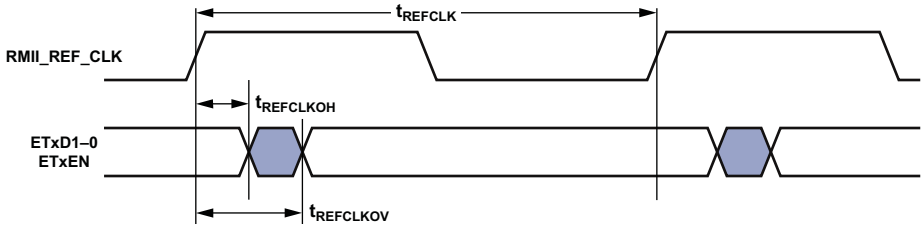

Figure 33. 10/100 Ethernet MAC Controller Timing: RMII Transmit Signal

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Table 43. 10/100 Ethernet MAC Controller Timing: MII/RMII Asynchronous Signal

Parameter		Min	Max	Unit
Timing Requirements				
t_{ECOLH}	COL Pulse Width High[1]	$t_{ETxCLK} \times 1.5$		ns
		$t_{ERxCLK} \times 1.5$		ns
t_{ECOLL}	COL Pulse Width Low[1]	$t_{ETxCLK} \times 1.5$		ns
		$t_{ERxCLK} \times 1.5$		ns
t_{ECRSH}	CRS Pulse Width High[2]	$t_{ETxCLK} \times 1.5$		ns
t_{ECRSL}	CRS Pulse Width Low[2]	$t_{ETxCLK} \times 1.5$		ns

[1] MII/RMII asynchronous signals are COL, CRS. These signals are applicable in both MII and RMII modes. The asynchronous COL input is synchronized separately to both the ETxCLK and the ERxCLK, and must have a minimum pulse width high or low at least 1.5 times the period of the slower of the two clocks.
[2] The asynchronous CRS input is synchronized to the ETxCLK, and must have a minimum pulse width high or low at least 1.5 times the period of ETxCLK.

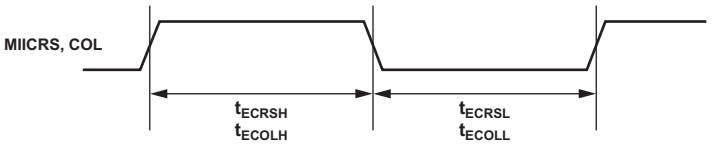

Figure 34. 10/100 Ethernet MAC Controller Timing: Asynchronous Signal

Table 44. 10/100 Ethernet MAC Controller Timing: MII Station Management

Parameter[1]		Min	Max	Unit
Timing Requirements				
t_{MDIOS}	MDIO Input Valid to MDC Rising Edge (Setup)	11.5		ns
t_{MDCIH}	MDC Rising Edge to MDIO Input Invalid (Hold)	0		ns
Switching Characteristics				
t_{MDCOV}	MDC Falling Edge to MDIO Output Valid		25	ns
t_{MDCOH}	MDC Falling Edge to MDIO Output Invalid (Hold)	–1.25		ns

[1] MDC/MDIO is a 2-wire serial bidirectional port for controlling one or more external PHYs. MDC is an output clock whose minimum period is programmable as a multiple of the system clock SCLK. MDIO is a bidirectional data line.

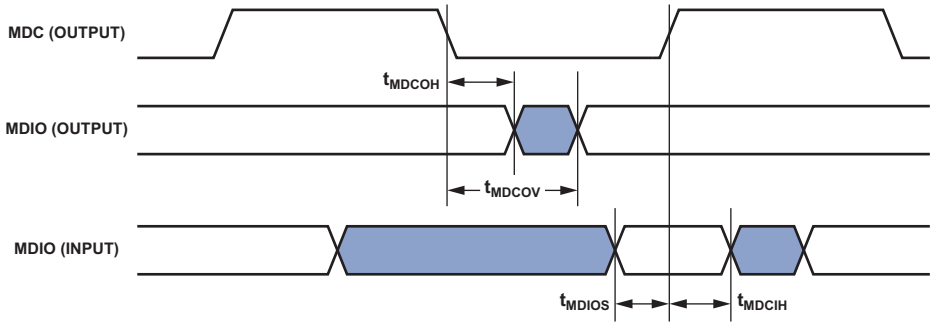

Figure 35. 10/100 Ethernet MAC Controller Timing: MII Station Management

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JTAG Test And Emulation Port Timing

Table 45 and Figure 36 describe JTAG port operations.

Table 45. JTAG Port Timing

Parameter		Min	Max	Unit
Timing Requirements				
t_{TCK}	TCK Period	20		ns
t_{STAP}	TDI, TMS Setup Before TCK High	4		ns
t_{HTAP}	TDI, TMS Hold After TCK High	4		ns
t_{SSYS}[1]	System Inputs Setup Before TCK High	4		ns
t_{HSYS}[1]	System Inputs Hold After TCK High	5		ns
t_{TRSTW}	$\overline{\text{TRST}}$ Pulse Width[2] (measured in TCK cycles)	4		TCK
Switching Characteristics				
t_{DTDO}	TDO Delay from TCK Low		10	ns
t_{DSYS}[3]	System Outputs Delay After TCK Low	0	13	ns

[1] System Inputs = DATA15–0, SCL, SDA, TFS0, TSCLK0, RSCLK0, RFS0, DR0PRI, DR0SEC, PF15–0, PG15–0, PH7–0, MDIO, TD1, TMS, $\overline{\text{RESET}}$, $\overline{\text{NMI}}$, BMODE2–0.
[2] 50 MHz Maximum.
[3] System Outputs = DATA15–0, ADDR19–1, $\overline{\text{ABE1–0}}$, $\overline{\text{ARE}}$, $\overline{\text{AWE}}$, $\overline{\text{AMS1–0}}$, $\overline{\text{SRAS}}$, $\overline{\text{SCAS}}$, $\overline{\text{SWE}}$, SCKE, CLKOUT, SA10, $\overline{\text{SMS}}$, SCL, SDA, TSCLK0, TFS0, RFS0, RSCLK0, DT0PRI, DT0SEC, PF15–0, PG15–0, PH7–0, MDC, MDIO.

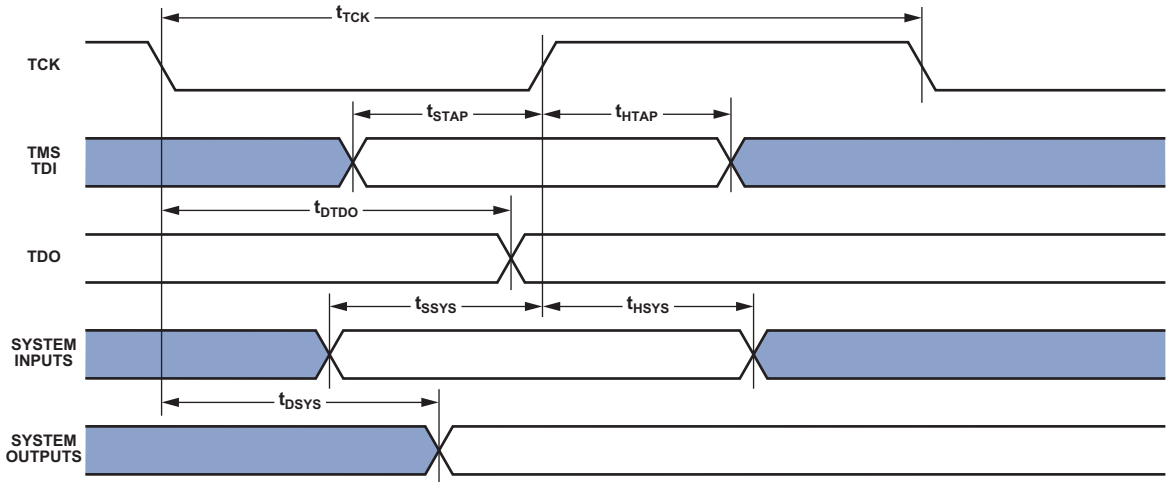

Figure 36. JTAG Port Timing

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OUTPUT DRIVE CURRENTS

Figure 37 through Figure 51 show typical current-voltage characteristics for the output drivers of the ADSP-BF51x processors. The curves represent the current drive capability of the output drivers. See Table 7 for information about which driver type corresponds to a particular ball.

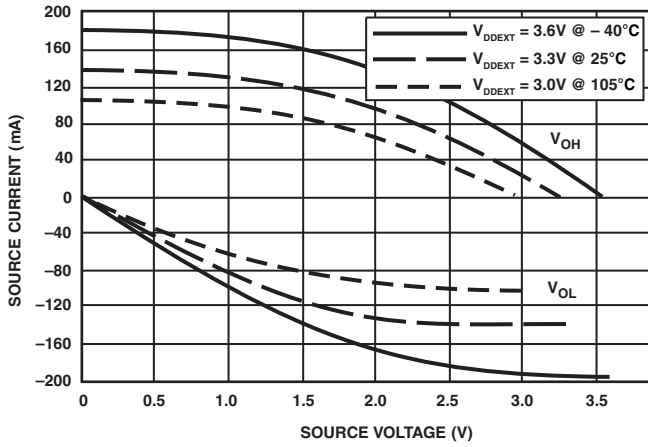

Figure 37. Driver Type A Current (3.3V V_{DDEXT}/V_{DDMEM})

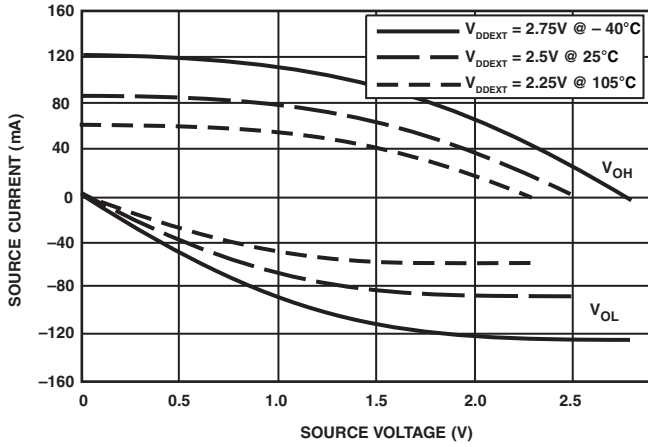

Figure 38. Driver Type A Current (2.5V V_{DDEXT}/V_{DDMEM})

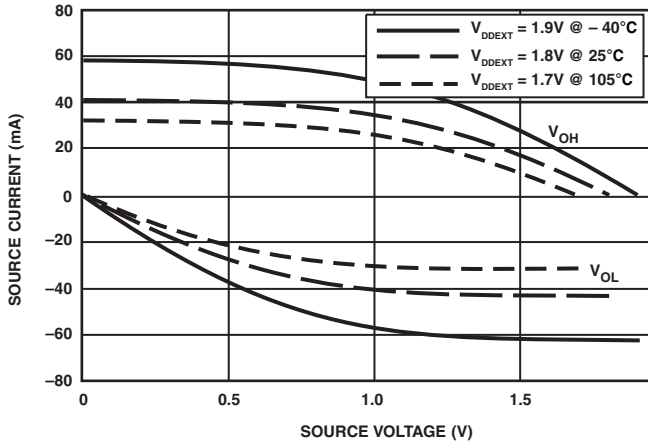

Figure 39. Driver Type A Current (1.8V V_{DDEXT}/V_{DDMEM})

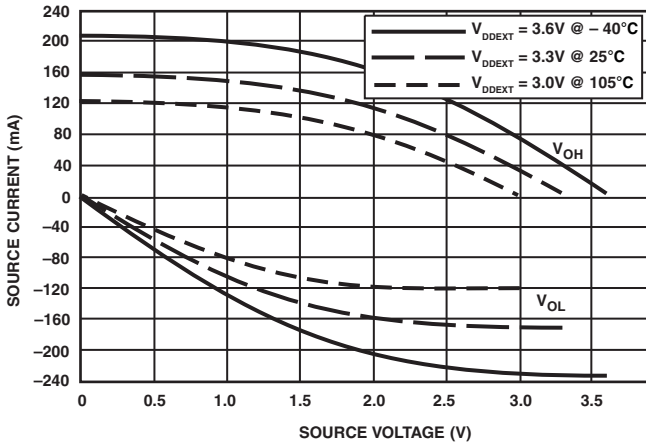

Figure 40. Driver Type B Current (3.3V V_{DDEXT}/V_{DDMEM})

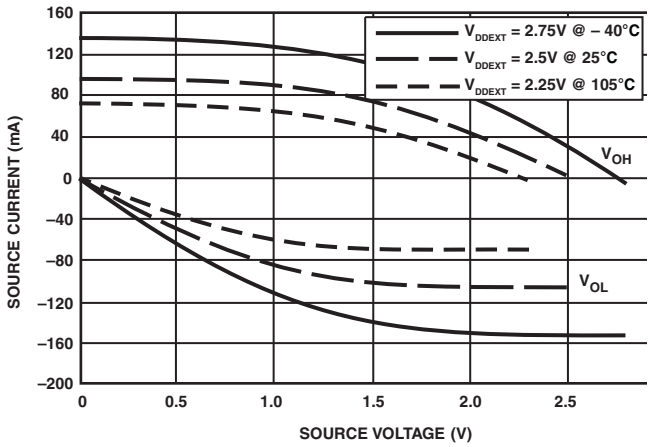

Figure 41. Driver Type B Current (2.5V V_{DDEXT}/V_{DDMEM})

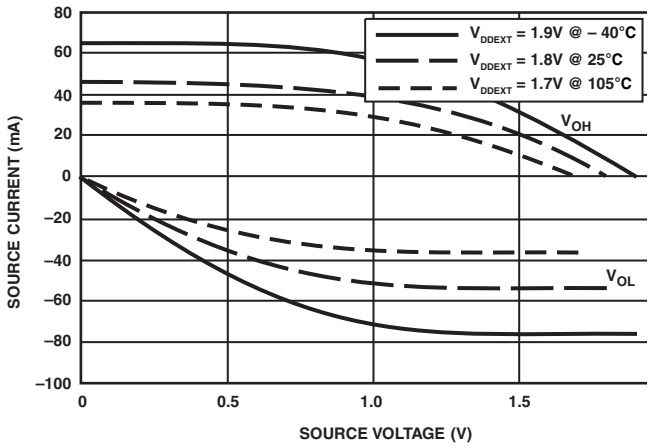

Figure 42. Driver Type B Current (1.8V V_{DDEXT}/V_{DDMEM})

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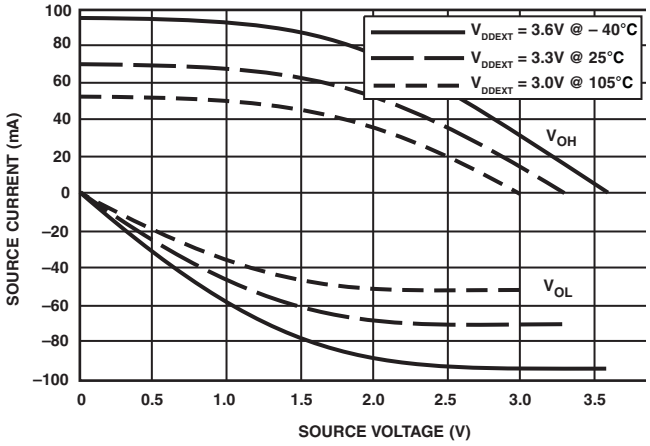

Figure 43. Driver Type C Current (3.3V V_{DDEXT}/V_{DDMEM})

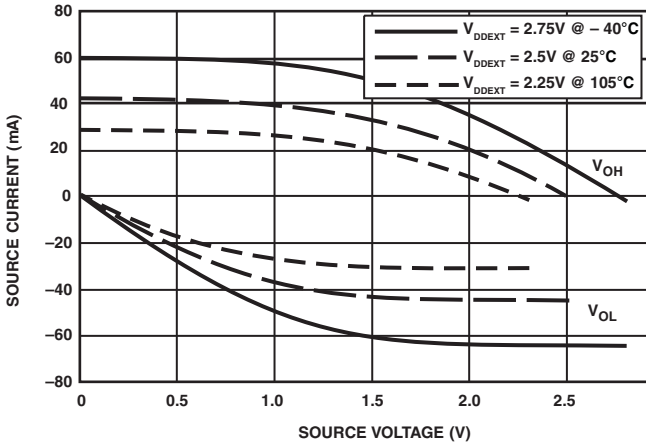

Figure 44. Drive Type C Current (2.5V V_{DDEXT}/V_{DDMEM})

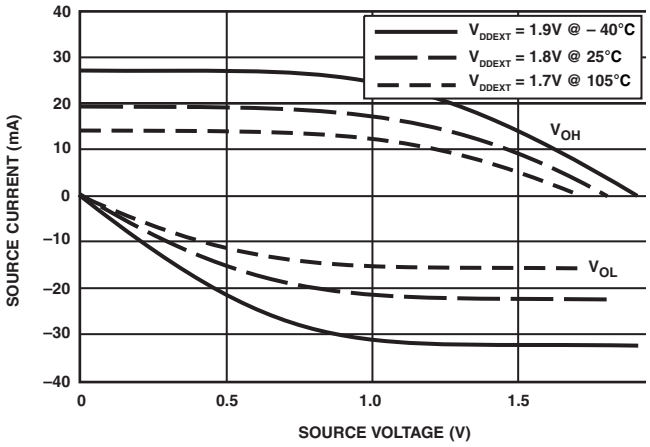

Figure 45. Driver Type C Current (1.8V V_{DDEXT}/V_{DDMEM})

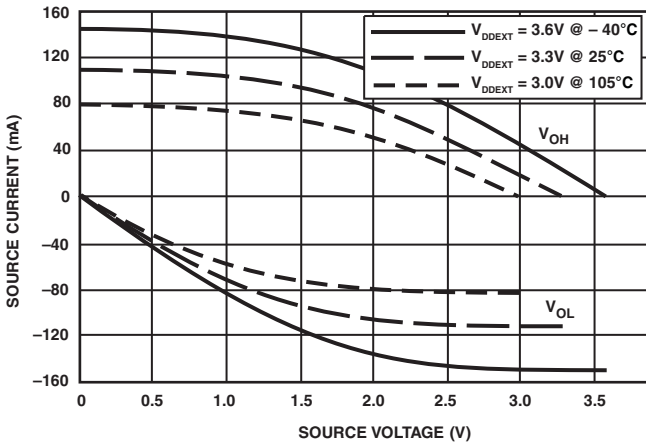

Figure 46. Driver Type D Current (3.3V V_{DDEXT}/V_{DDMEM})

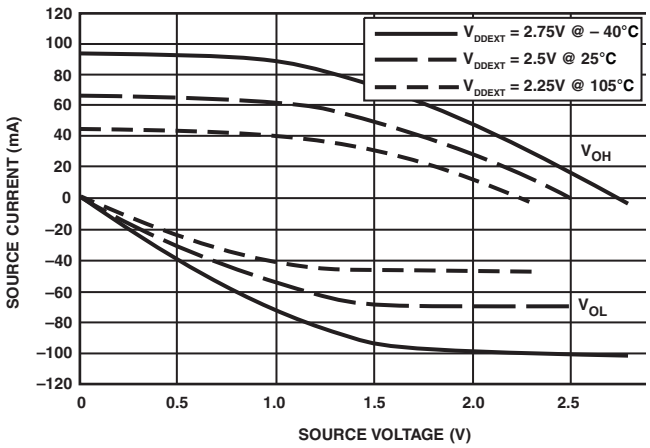

Figure 47. Driver Type D Current (2.5V V_{DDEXT}/V_{DDMEM})

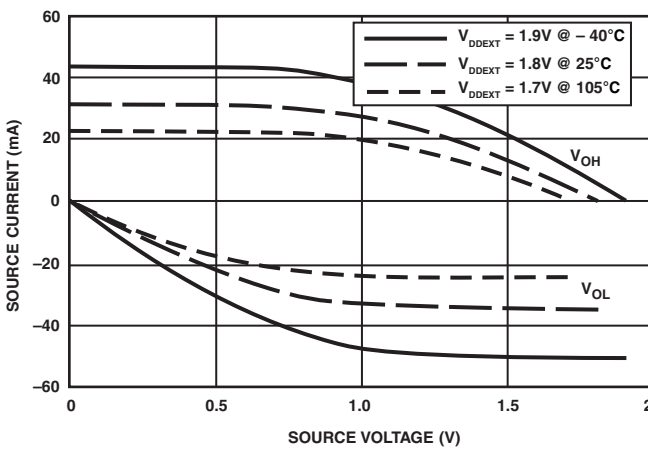

Figure 48. Driver Type D Current (1.8V V_{DDEXT}/V_{DDMEM})

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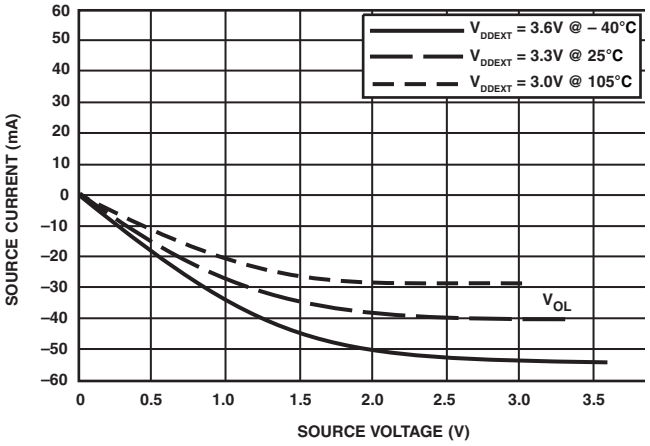

Figure 49. Driver Type E Current (3.3V V_{DDEXT}/V_{DDMEM})

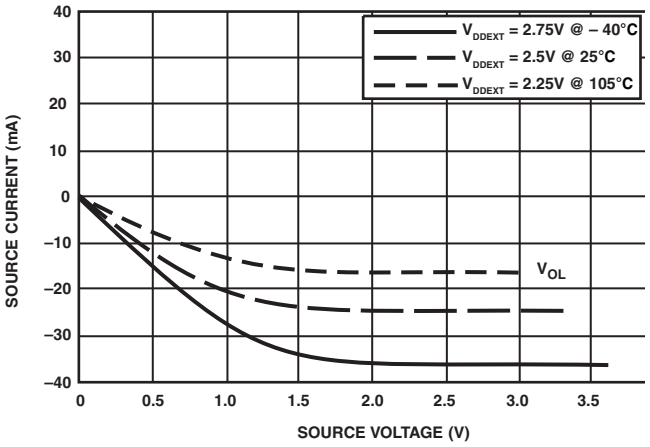

Figure 50. Driver Type E Current (2.5V V_{DDEXT}/V_{DDMEM})

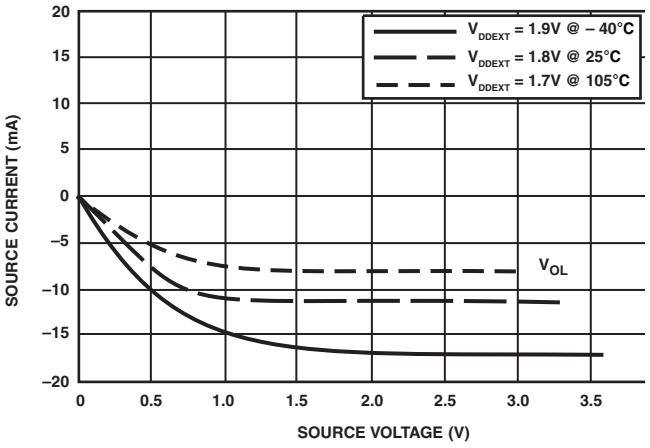

Figure 51. Driver Type E Current (1.8V V_{DDEXT}/V_{DDMEM})

TEST CONDITIONS

All timing parameters appearing in this data sheet were measured under the conditions described in this section. Figure 52 shows the measurement point for ac measurements (except output enable/disable). The measurement point V_{MEAS} is V_{DDEXT}/2 or V_{DDMEM}/2 for V_{DDEXT}/V_{DDMEM} (nominal) = 1.8 V/2.5 V/3.3 V.

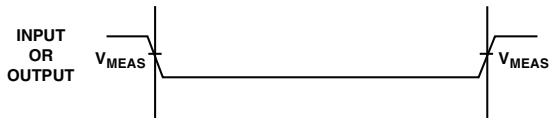

Figure 52. Voltage Reference Levels for AC Measurements (Except Output Enable/Disable)

Output Enable Time Measurement

Output signals are considered to be enabled when they have made a transition from a high impedance state to the point when they start driving.

The output enable time t_{ENA} is the interval from the point when a reference signal reaches a high or low voltage level to the point when the output starts driving as shown on the right side of Figure 53.

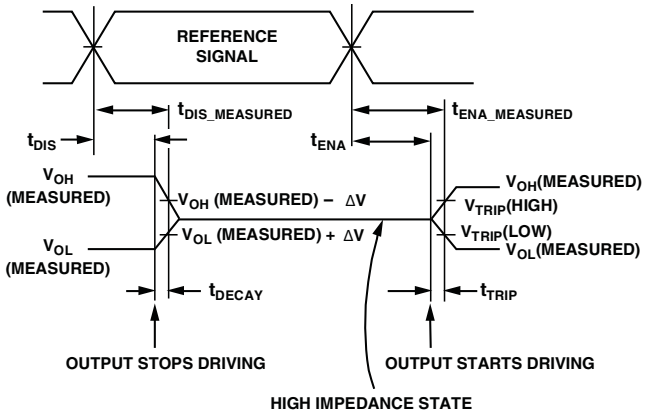

Figure 53. Output Enable/Disable

The time $t_{ENA_MEASURED}$ is the interval from when the reference signal switches to when the output voltage reaches V_{TRIP}(high) or V_{TRIP}(low). For V_{DDEXT} (nominal) = 1.8 V, V_{TRIP} (high) is 0.95 V, and V_{TRIP} (low) is 0.85 V. For V_{DDEXT} (nominal) = 2.5 V, V_{TRIP} (high) is 1.3 V and V_{TRIP} (low) is 1.2 V. For V_{DDEXT} (nominal) = 3.3 V, V_{TRIP} (high) is 1.7 V, and V_{TRIP} (low) is 1.6 V. Time t_{TRIP} is the interval from when the output starts driving to when the output reaches the V_{TRIP}(high) or V_{TRIP}(low) trip voltage.

Time t_{ENA} is calculated as shown in the equation:

$$t_{ENA} = t_{ENA_MEASURED} - t_{TRIP}$$

If multiple signals (such as the data bus) are enabled, the measurement value is that of the first signal to start driving.

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Output Disable Time Measurement

Output signals are considered to be disabled when they stop driving, go into a high impedance state, and start to decay from their output high or low voltage. The output disable time t_{DIS} is the difference between $t_{DIS_MEASURED}$ and t_{DECAY} as shown on the left side of Figure 53.

$$t_{DIS} = t_{DIS_MEASURED} - t_{DECAY}$$

The time for the voltage on the bus to decay by ΔV is dependent on the capacitive load C_L and the load current I_L. This decay time can be approximated by the equation:

$$t_{DECAY} = (C_L \Delta V)/I_L$$

The time t_{DECAY} is calculated with test loads C_L and I_L and with ΔV equal to 0.25 V for V_{DDEXT}/V_{DDMEM} (nominal) = 2.5 V/3.3 V and 0.15 V for V_{DDEXT}/V_{VDDMEM} (nominal) = 1.8 V.

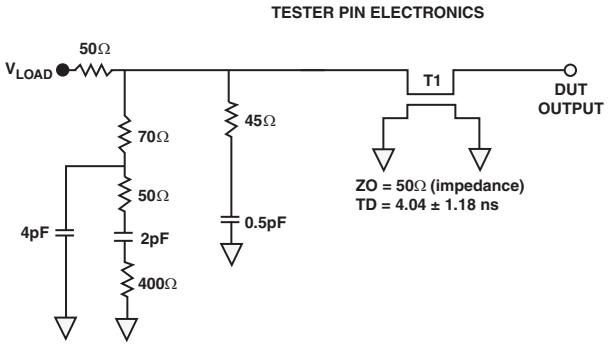
The time $t_{DIS_MEASURED}$ is the interval from when the reference signal switches to when the output voltage decays ΔV from the measured output high or output low voltage.

Example System Hold Time Calculation

To determine the data output hold time in a particular system, first calculate t_{DECAY} using the equation given above. Choose ΔV to be the difference between the ADSP-BF51x processor's output voltage and the input threshold for the device requiring the hold time. C_L is the total bus capacitance (per data line), and I_L is the total leakage or three-state current (per data line). The hold time is t_{DECAY} plus the various output disable times as specified in the Timing Specifications (for example t_{DSDAT} for an SDRAM write cycle as shown in SDRAM Interface Timing).

Capacitive Loading

Output delays and holds are based on standard capacitive loads of an average of 6 pF on all balls (see Figure 54). V_{LOAD} is equal to $(V_{DDEXT}/V_{DDMEM})/2$. The graphs of Figure 55 through Figure 66 show how output rise time varies with capacitance. The delay and hold specifications given should be derated by a factor derived from these figures. The graphs in these figures may not be linear outside the ranges shown.



NOTES:
THE WORST CASE TRANSMISSION LINE DELAY IS SHOWN AND CAN BE USED FOR THE OUTPUT TIMING ANALYSIS TO REFLECT THE TRANSMISSION LINE EFFECT AND MUST BE CONSIDERED. THE TRANSMISSION LINE (TD) IS FOR LOAD ONLY AND DOES NOT AFFECT THE DATA SHEET TIMING SPECIFICATIONS.

ANALOG DEVICES RECOMMENDS USING THE IBIS MODEL TIMING FOR A GIVEN SYSTEM REQUIREMENT. IF NECESSARY, A SYSTEM MAY INCORPORATE EXTERNAL DRIVERS TO COMPENSATE FOR ANY TIMING DIFFERENCES.

Figure 54. Equivalent Device Loading for AC Measurements (Includes All Fixtures)

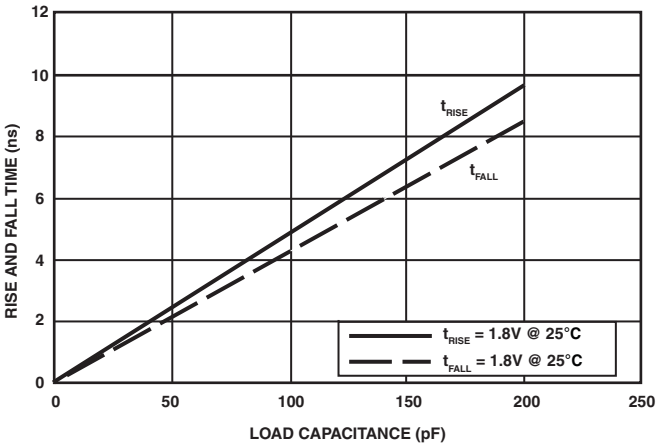


Figure 55. Driver Type A Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (1.8V V_{DDEXT}/V_{DDMEM})

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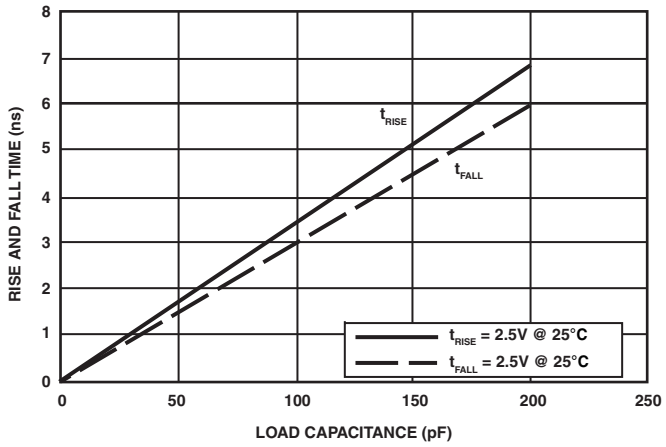

Figure 56. Driver Type A Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (2.5V V_{DDEXT}/V_{DDMEM})

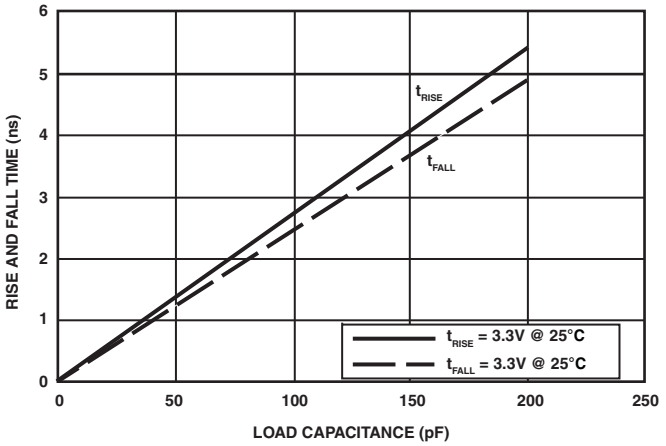

Figure 57. Driver Type A Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (3.3V V_{DDEXT}/V_{DDMEM})

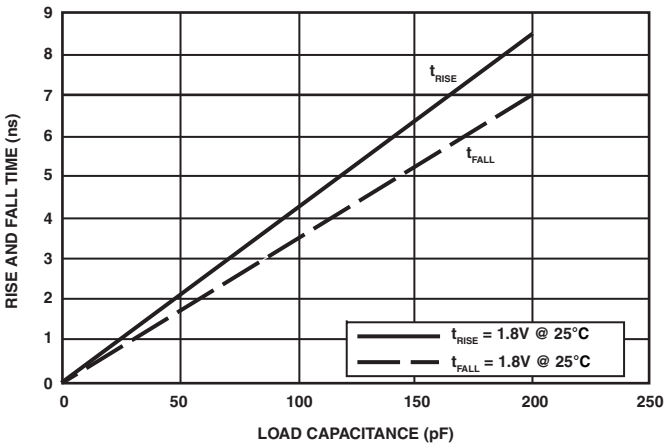

Figure 58. Driver Type B Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (1.8V V_{DDEXT}/V_{DDMEM})

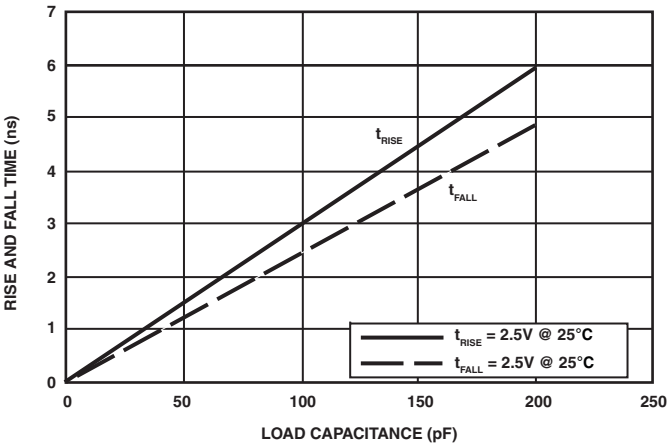

Figure 59. Driver Type B Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (2.5V V_{DDEXT}/V_{DDMEM})

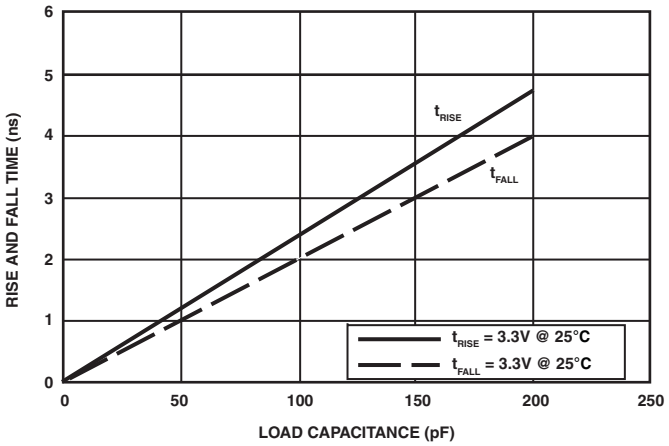

Figure 60. Driver Type B Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (3.3V V_{DDEXT}/V_{DDMEM})

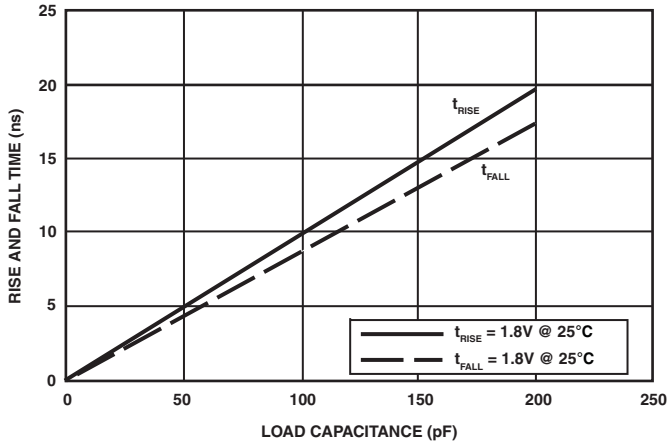

Figure 61. Driver Type C Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (1.8V V_{DDEXT}/V_{DDMEM})

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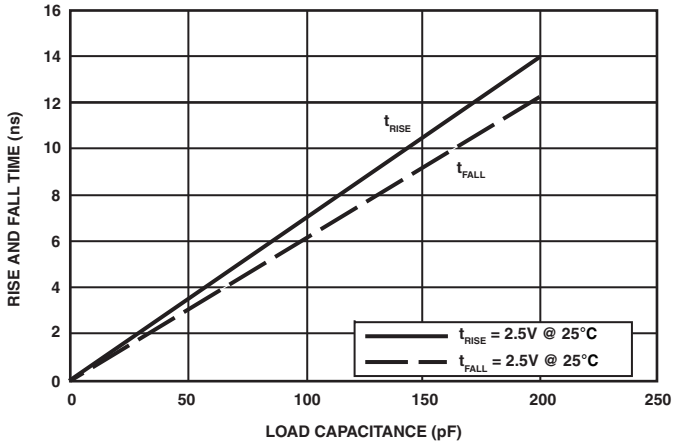

Figure 62. Driver Type C Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (2.5V V_{DDEXT}/V_{DDMEM})

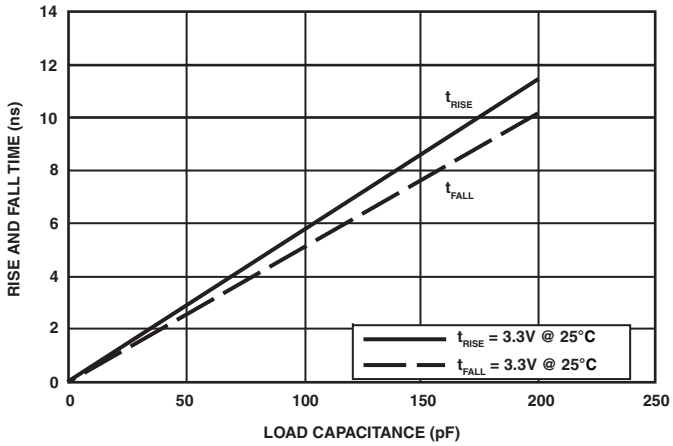

Figure 63. Driver Type C Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (3.3V V_{DDEXT}/V_{DDMEM})

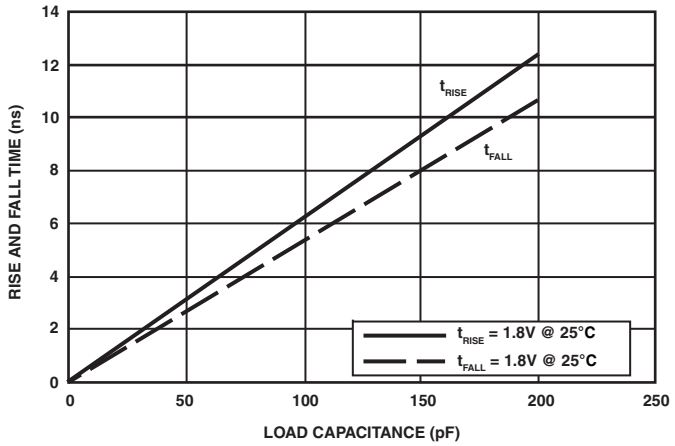

Figure 64. Driver Type D Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (1.8V V_{DDEXT}/V_{DDMEM})

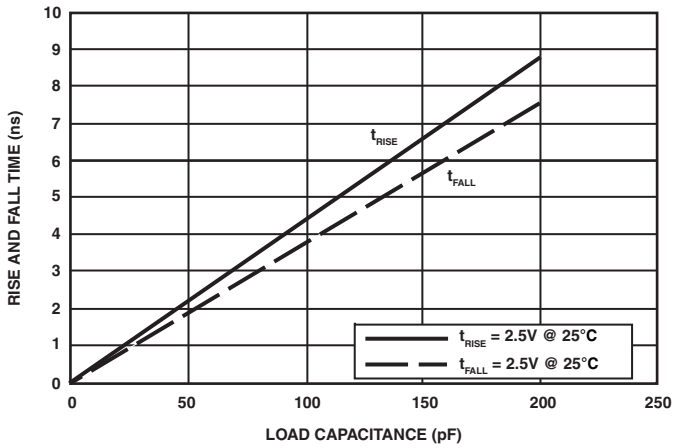

Figure 65. Driver Type D Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (2.5V V_{DDEXT}/V_{DDMEM})

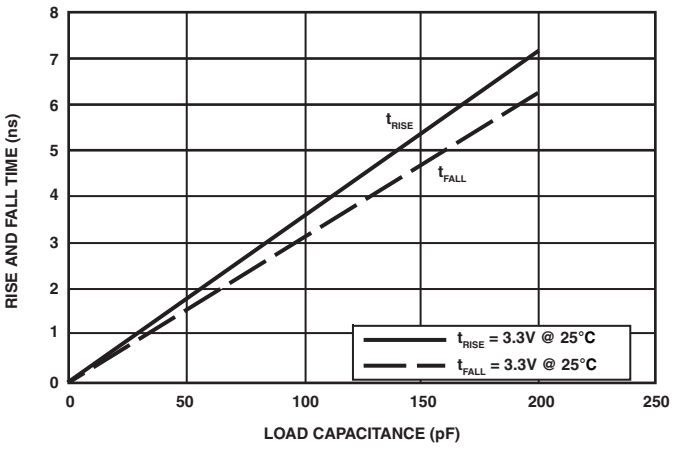

Figure 66. Driver Type D Typical Rise and Fall Times (10%–90%) vs. Load Capacitance (3.3V V_{DDEXT}/V_{DDMEM})

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THERMAL CHARACTERISTICS

To determine the junction temperature on the application printed circuit board use:

$$T_J = T_{CASE} + (\Psi_{JT} \times P_D)$$

where:

T_J = Junction temperature (°C)

T_{CASE} = Case temperature (°C) measured by customer at top center of package.

Ψ_{JT} = From Table 47

P_D = Power dissipation (see Total Power Dissipation for the method to calculate P_D)

Values of θ_{JA} are provided for package comparison and printed circuit board design considerations. θ_{JA} can be used for a first order approximation of T_J by the equation:

$$T_J = T_A + (\theta_{JA} \times P_D)$$

where:

T_A = Ambient temperature (°C)

Values of θ_{JC} are provided for package comparison and printed circuit board design considerations when an external heat sink is required.

Values of θ_{JB} are provided for package comparison and printed circuit board design considerations.

In Table 47, airflow measurements comply with JEDEC standards JESD51-2 and JESD51-6, and the junction-to-board measurement complies with JESD51-8. The junction-to-case measurement complies with MIL-STD-883 (Method 1012.1). All measurements use a 2S2P JEDEC test board.

The LQFP_EP package requires thermal trace squares and thermal vias to an embedded ground plane in the PCB. The paddle must be connected to ground for proper operation to data sheet specifications. Refer to JEDEC standard JESD51-5 for more information.

Table 46. Thermal Characteristics for SQ-176-2 Package

Parameter	Condition	Typical	Unit
θ_{JA}	0 Linear m/s Airflow	17.4	°C/W
θ_{JMA}	1 Linear m/s Airflow	14.8	°C/W
θ_{JMA}	2 Linear m/s Airflow	14.0	°C/W
θ_{JC}	Not Applicable	7.8	°C/W
Ψ_{JT}	0 Linear m/s Airflow	0.28	°C/W
Ψ_{JT}	1 Linear m/s Airflow	0.39	°C/W
Ψ_{JT}	2 Linear m/s Airflow	0.48	°C/W

Table 47. Thermal Characteristics for BC-168-1 Package

Parameter	Condition	Typical	Unit
θ_{JA}	0 Linear m/s Airflow	30.5	°C/W
θ_{JMA}	1 Linear m/s Airflow	27.6	°C/W
θ_{JMA}	2 Linear m/s Airflow	26.3	°C/W
θ_{JC}	Not Applicable	11.1	°C/W
Ψ_{JT}	0 Linear m/s Airflow	0.20	°C/W
Ψ_{JT}	1 Linear m/s Airflow	0.35	°C/W
Ψ_{JT}	2 Linear m/s Airflow	0.45	°C/W

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176-LEAD LQFP_EP LEAD ASSIGNMENT

Table 48 lists the LQFP_EP leads by lead number.

Table 48. 176-Lead LQFP_EP Pin Assignment (Numerical by Lead Number)

Lead No.	Signal	Lead No.	Signal	Lead No.	Signal	Lead No.	Signal
1	GND	45	GND	89	GND	133	GND
2	GND	46	GND	90	GND	134	GND
3	PF9	47	PG1	91	A12	135	$\overline{PG}$
4	PF8	48	PG0	92	A11	136	V_{DDEXT}
5	PF7	49	V_{DDEXT}	93	A10	137	GND
6	PF6	50	TDO	94	A9	138	V_{DDINT}
7	V_{DDEXT}	51	$\overline{EMU}$	95	V_{DDMEM}	139	GND
8	V_{PPOTP}	52	TDI	96	A8	140	RTXO
9	V_{DDOTP}	53	TCK	97	A7	141	RTXI
10	PF5	54	$\overline{TRST}$	98	V_{DDINT}	142	V_{DDRTC}
11	PF4	55	TMS	99	GND	143	CLKIN
12	PF3	56	D15	100	V_{DDINT}	144	XTAL
13	PF2	57	D14	101	A6	145	V_{DDEXT}
14	V_{DDINT}	58	D13	102	A5	146	$\overline{RESET}$
15	GND	59	V_{DDMEM}	103	A4	147	$\overline{NMI}$
16	V_{DDEXT}	60	D12	104	V_{DDMEM}	148	V_{DDEXT}
17	V_{DDEXT}	61	D11	105	A3	149	GND
18	PF1	62	D10	106	A2	150	CLKBUF
19	PF0	63	V_{DDINT}	107	A1	151	GND
20	PG15	64	D9	108	$\overline{ABE}$1	152	V_{DDINT}
21	PG14	65	D8	109	$\overline{ABE}$0	153	PH7
22	GND	66	D7	110	SA10	154	PH6
23	V_{DDINT}	67	GND	111	GND	155	PH5
24	V_{DDEXT}	68	V_{DDMEM}	112	V_{DDMEM}	156	PH4
25	PG13	69	D6	113	$\overline{SWE}$	157	GND
26	PG12	70	D5	114	$\overline{SCAS}$	158	V_{DDEXT}
27	PG11	71	D4	115	$\overline{SRAS}$	159	PH3
28	PG10	72	D3	116	V_{DDINT}	160	PH2
29	V_{DDEXT}	73	D2	117	GND	161	PH1
30	V_{DDINT}	74	D1	118	$\overline{SMS}$	162	PH0
31	PG9	75	V_{DDMEM}	119	SCKE	163	GND
32	PG8	76	D0	120	$\overline{AMS}$1	164	V_{DDINT}
33	PG7	77	A19	121	$\overline{ARE}$	165	PF15
34	PG6	78	A18	122	$\overline{AWE}$	166	PF14
35	V_{DDEXT}	79	V_{DDINT}	123	$\overline{AMS}$0	167	PF13
36	PG5	80	A17	124	V_{DDMEM}	168	PF12
37	PG4	81	A16	125	CLKOUT	169	GND
38	PG3	82	V_{DDMEM}	126	V_{DDEXT}	170	V_{DDEXT}
39	PG2	83	GND	127	NC[1]	171	PF11
40	BMODE2	84	A15	128	V_{DDEXT}	172	SDA
41	BMODE1	85	A14	129	V_{DDEXT}	173	SCL
42	BMODE0	86	A13	130	EXT_WAKE	174	PF10
43	GND	87	GND	131	GND	175	GND
44	GND	88	GND	132	GND	176	GND
						GND	177*

* Pin no. 177 is the GND supply (see Figure 68) for the processor; this pad must be **robustly** connected to GND.

[1] Do not make any electrical connection to this pin.

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Figure 67 shows the top view of the LQFP_EP lead configuration. Figure 68 shows the bottom view of the LQFP_EP lead configuration.

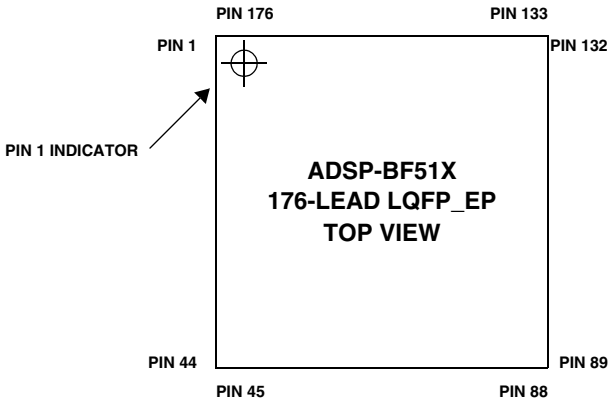

Figure 67. 176-Lead LQFP_EP Lead Configuration (Top View)

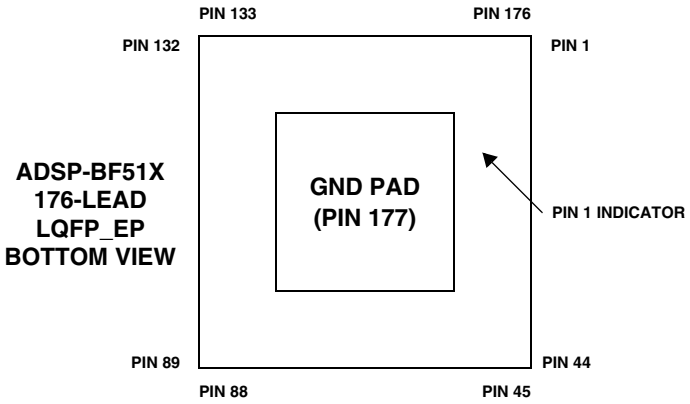

Figure 68. 176-Lead LQFP_EP Lead Configuration (Bottom View)

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168-BALL CSP_BGA BALL ASSIGNMENT

Table 49 lists the CSP_BGA by ball number.

Table 49. 168-Ball CSP_BGA Ball Assignment (Numerical by Ball Number)

Ball No.	Signal Name	Ball No.	Signal Name	Ball No.	Signal Name	Ball No.	Signal Name	Ball No.	Signal Name	Ball No.	Signal Name
A1	GND	C1	PF4	E10	V_{DDINT}	H1	PG12	K6	V_{DDMEM}	N1	BMODE1
A2	SCL	C2	PF7	E12	V_{DDMEM}	H2	PG13	K7	V_{DDMEM}	N2	PG1
A3	SDA	C3	PF8	E13	$\overline{ARE}$	H3	PG11	K8	V_{DDMEM}	N3	TDO
A4	PF13	C4	PF10	E14	$\overline{AWE}$	H5	V_{DDEXT}	K9	V_{DDMEM}	N4	$\overline{TRST}$
A5	PF15	C5	V_{DDEXT}	F1	PF0	H6	GND	K10	V_{DDMEM}	N5	TMS
A6	PH2	C6	V_{DDEXT}	F2	PF1	H7	GND	K12	A8	N6	D13
A7	PH1	C7	PF11	F3	V_{DDINT}	H8	GND	K13	A2	N7	D9
A8	PH5	C8	V_{DDEXT}	F5	V_{DDEXT}	H9	GND	K14	A1	N8	D5
A9	PH6	C9	V_{DDINT}	F6	GND	H10	V_{DDINT}	L1	PG5	N9	D1
A10	PH7	C10	V_{DDEXT}	F7	GND	H12	A3	L2	PG3	N10	A18
A11	CLKBUF	C11	RTXI	F8	GND	H13	$\overline{ABE}$0	L3	PG2	N11	A16
A12	XTAL	C12	RTXO	F9	GND	H14	$\overline{SCAS}$	L12	A9	N12	A14
A13	CLKIN	C13	$\overline{PG}$	F10	V_{DDINT}	J1	PG10	L13	A6	N13	A11
A14	GND	C14	NC[1]	F12	$\overline{SMS}$	J2	V_{DDEXT}	L14	A4	N14	A7
B1	V_{DDOTP}	D1	PF3	F13	SCKE	J3	PG9	M1	PG4	P1	GND
B2	GND	D2	PF5	F14	$\overline{AMS}$1	J5	V_{DDMEM}	M2	BMODE2	P2	TDI
B3	PF9	D3	VPPOTP	G1	PG15	J6	GND	M3	BMODE0	P3	TCK
B4	PF12	D12	V_{DDEXT}	G2	PG14	J7	GND	M4	PG0	P4	D15
B5	PF14	D13	CLKOUT	G3	V_{DDINT}	J8	GND	M5	$\overline{EMU}$	P5	D14
B6	PH0	D14	$\overline{AMS}$0	G5	V_{DDEXT}	J9	GND	M6	D12	P6	D11
B7	PH3	E1	V_{DDEXT}	G6	GND	J10	V_{DDINT}	M7	D10	P7	D8
B8	PH4	E2	PF2	G7	GND	J12	A15	M8	D2	P8	D7
B9	V_{DDEXT}	E3	PF6	G8	GND	J13	$\overline{ABE}$1	M9	D0	P9	D6
B10	$\overline{RESET}$	E5	V_{DDEXT}	G9	GND	J14	SA10	M10	A17	P10	D4
B11	$\overline{NMI}$	E6	V_{DDEXT}	G10	V_{DDINT}	K1	PG6	M11	A13	P11	D3
B12	V_{DDRTC}	E7	V_{DDINT}	G12	$\overline{SWE}$	K2	PG8	M12	A12	P12	A19
B13	V_{DDEXT}	E8	V_{DDINT}	G13	$\overline{SRAS}$	K3	PG7	M13	A10	P13	GND
B14	EXT_WAKE	E9	V_{DDINT}	G14	GND	K5	V_{DDMEM}	M14	A5	P14	GND

[1] Do not make any electrical connection to this pin.

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Figure 69 shows the top view of the CSP_BGA ball configuration. Figure 70 shows the bottom view of the CSP_BGA ball configuration.

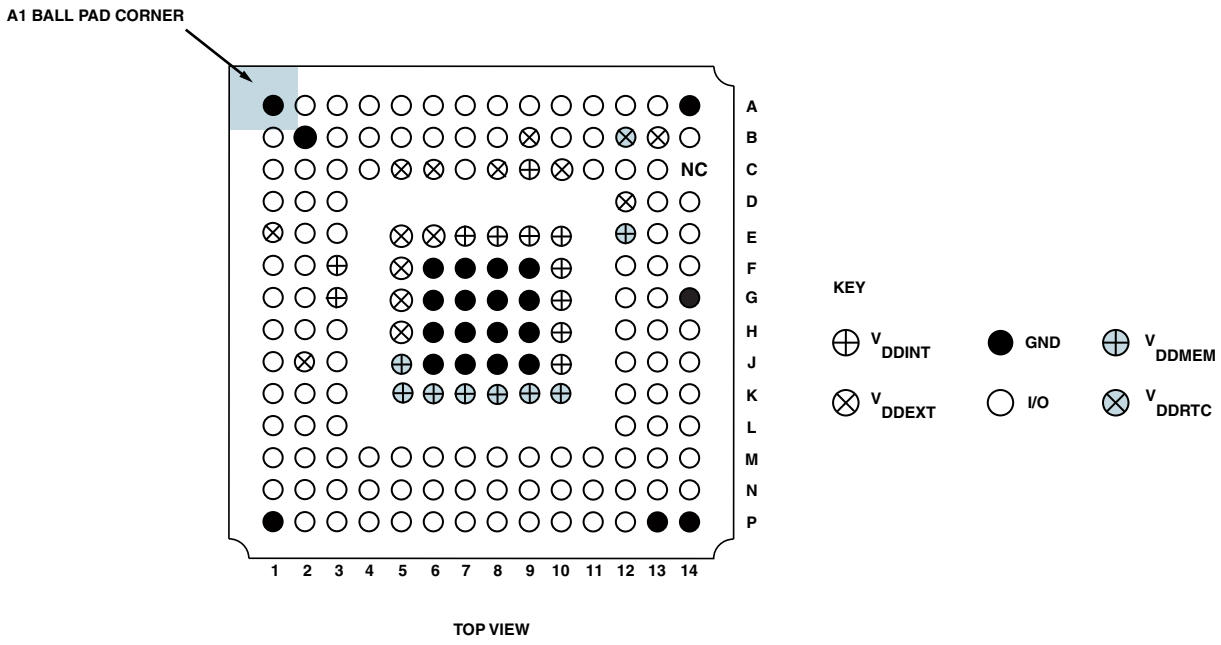

Figure 69. 168-Ball CSP_BGA Ball Configuration (Top View)

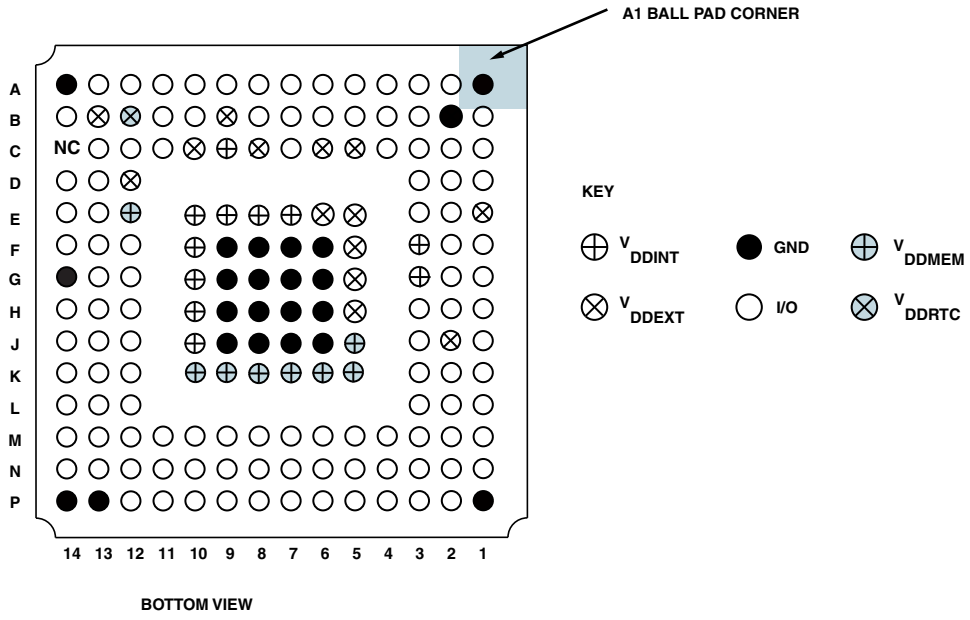

Figure 70. 168-Ball CSP_BGA Ball Configuration (Bottom View)

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OUTLINE DIMENSIONS

Dimensions in Figure 71 are shown in millimeters.

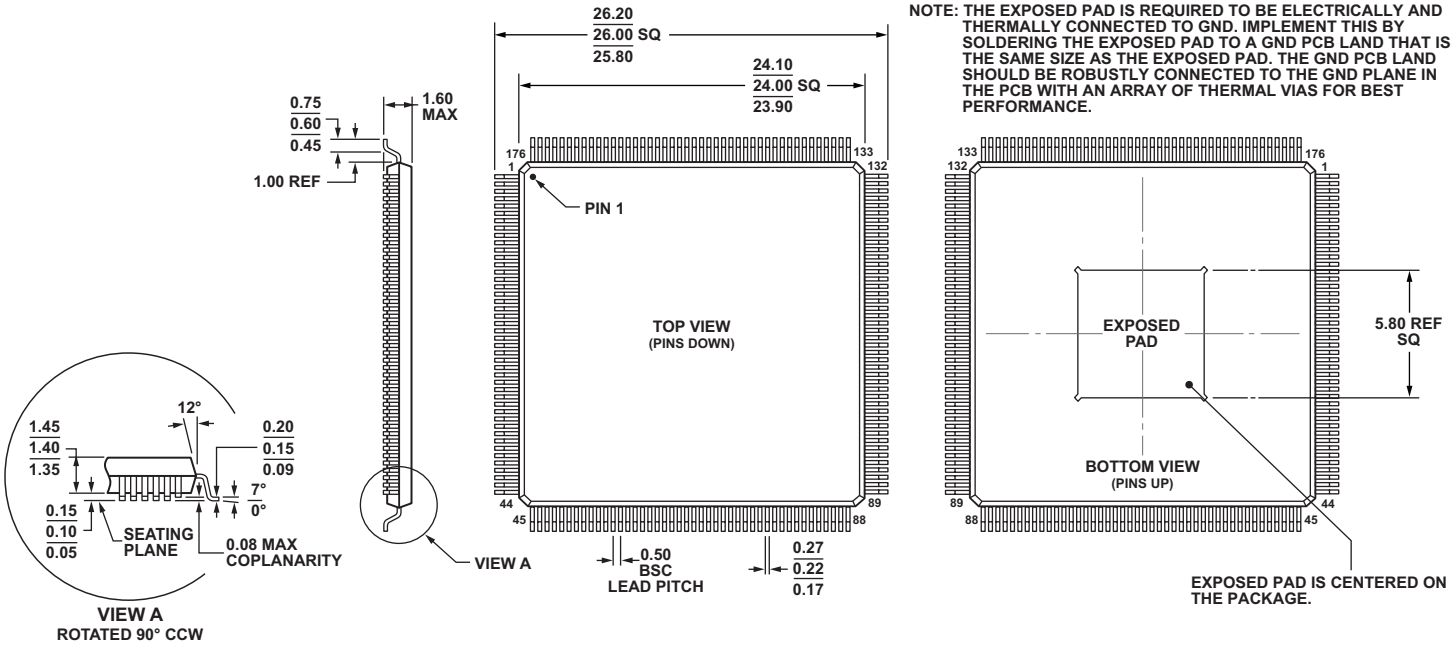

Figure 71. 176-Lead Low Profile Quad Flat Package [LQFP_EP]
(SQ-176-2)
Dimensions shown in millimeters

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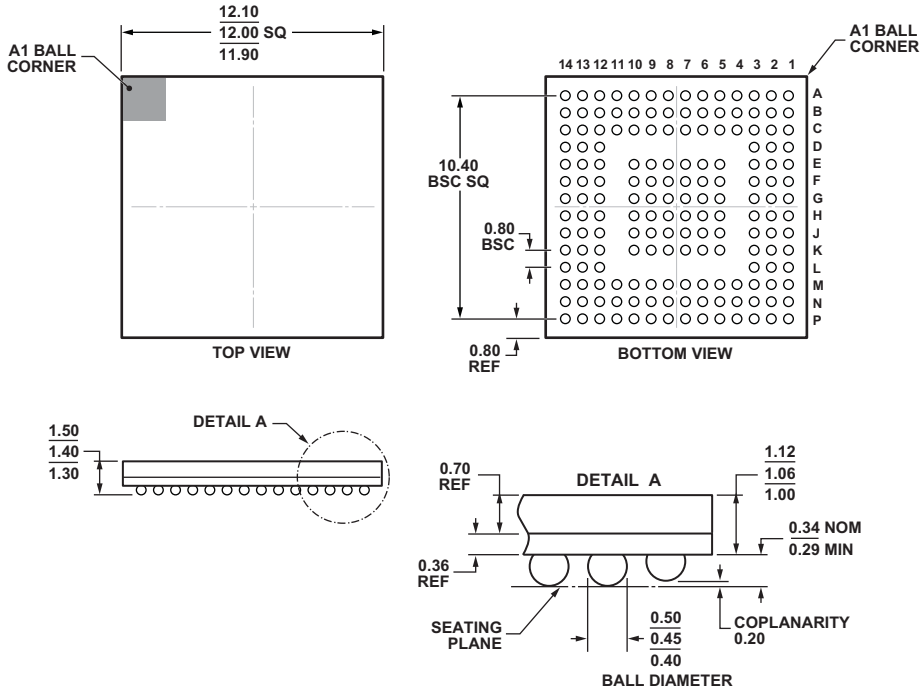

COMPLIANT TO JEDEC STANDARDS MO-275-GGAB-1.

Figure 72. 168-Ball Chip Scale Package Ball Grid Array [CSP_BGA]
(BC-168-1)
Dimensions shown in millimeters

SURFACE-MOUNT DESIGN

Table 50 is provided as an aid to PCB design. For industry standard design recommendations, refer to IPC-7351, *Generic Requirements for Surface Mount Design and Land Pattern Standard.*

Table 50. BGA Data for Use with Surface-Mount Design

Package	Package Ball Attach Type	Package Solder Mask Opening	Package Ball Pad Size
168-Ball CSP_BGA	Solder Mask Defined	0.35 mm diameter	0.48 mm diameter

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AUTOMOTIVE PRODUCTS

The ADBF512W and ADBF518W models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models and designers should review the product Specifications section of this data sheet carefully. Only the automotive grade products shown in Table 51 are available for use in automotive applications. Contact your local ADI account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

Table 51. Automotive Products

Automotive Models[1,2]	Temperature Range[3]	Processor Instruction Rate (Max)	Package Description	Package Option
ADBF512WBBCZ4xx	–40°C to +85°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADBF518WBBCZ4xx	–40°C to +85°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADBF512WBSWZ4xx	–40°C to +85°C	400 MHz	176-Lead LQFP_EP	SQ-176-2
ADBF518WBSWZ4xx	–40°C to +85°C	400 MHz	176-Lead LQFP_EP	SQ-176-2

[1] Z = RoHS Compliant Part.

[2] The use of xx designates silicon revision.

[3] Referenced temperature is ambient temperature. The ambient temperature is not a specification. Please see Operating Conditions for junction temperature (T_J) specification which is the only temperature specification.

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ORDERING GUIDE

Model[1]	Temperature Range[2]	Processor Instruction Rate (Max)	Package Description	Package Option
ADSP-BF512BBCZ-3	–40°C to +85°C	300 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF512BBCZ-4	–40°C to +85°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF512BSWZ-3	–40°C to +85°C	300 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF512BSWZ-4	–40°C to +85°C	400 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF512KBCZ-3	0°C to +70°C	300 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF512KBCZ-4	0°C to +70°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF512KSWZ-3	0°C to +70°C	300 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF512KSWZ-4	0°C to +70°C	400 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF514BBCZ-3	–40°C to +85°C	300 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF514BBCZ-4	–40°C to +85°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF514BSWZ-3	–40°C to +85°C	300 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF514BSWZ-4	–40°C to +85°C	400 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF514KBCZ-3	0°C to +70°C	300 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF514KBCZ-4	0°C to +70°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF514KSWZ-3	0°C to +70°C	300 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF514KSWZ-4	0°C to +70°C	400 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF516KSWZ-3	0°C to +70°C	300 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF516KBCZ-3	0°C to +70°C	300 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF516KSWZ-4	0°C to +70°C	400 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF516KBCZ-4	0°C to +70°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF516BBCZ-3	–40°C to +85°C	300 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF516BBCZ-4	–40°C to +85°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF516BSWZ-3	–40°C to +85°C	300 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF516BSWZ-4	–40°C to +85°C	400 MHz	176-Lead LQFP_EP	SQ-176-2
ADSP-BF518BBCZ-4	–40°C to +85°C	400 MHz	168-Ball CSP_BGA	BC-168-1
ADSP-BF518BSWZ-4	–40°C to +85°C	400 MHz	176-Lead LQFP_EP	SQ-176-2

[1] Z = RoHS compliant part.

[2] Referenced temperature is ambient temperature. The ambient temperature is not a specification. Please see Operating Conditions for junction temperature (T_J) specification which is the only temperature specification.

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