#### **ABSOLUTE MAXIMUM RATINGS**

V <sub>CC</sub> to GND V <sub>L</sub> to GND V+ to GND V- to GND V+ +   V- (Note 1)	0.3V to (V <sub>CC</sub> + 0.3V) 
Input Voltages	+ 10V
T_IN, SHDN to GND	
Output Voltages	
T_OUT to GND R_OUT	

Short-Circuit Duration T_OUT to GND	Continuous
Continuous Power Dissipation (TA = +70	
20-Pin TSSOP (derate 7.0mW/°C abo	ve +70°C)559mW
Operating Temperature Ranges	
MAX3386ECUP	0°C to +70°C
MAX3386EEUP	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Note 1: V+ and V- can have maximum magnitudes of 7V, but their absolute difference cannot exceed 13V.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### DC ELECTRICAL CHARACTERISTICS

(VCC = VL = +3.0V to +5.5V; C1–C4 = 0.1 $\mu$ F, tested at +3.3V ±10%; C1 = 0.047 $\mu$ F, C2–C4 = 0.33 $\mu$ F, tested at +5.0V ±10%; TA = T<sub>MIN</sub> to T<sub>MAX</sub>; unless otherwise noted. Typical values are at V<sub>CC</sub> = V<sub>L</sub> = +3.3V, T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
DC CHARACTERISTICS (VCC =	= +3.3V or +5	5V, T <sub>A</sub> = +25°C)		1			
Shutdown Supply Current		SHDN = GND, all inputs at V <sub>CC</sub> or GND			1	10	μΑ
Supply Current		SHDN = V <sub>CC</sub> , no load			0.3	1	mA
LOGIC INPUTS	·						
Innest Logic Throubold Loss		T IN CLIDA	V <sub>L</sub> = 3.3V or 5.0V			0.8	.,
Input Logic Threshold Low		T_IN, SHDN	V <sub>L</sub> = 2.5V			0.6	V
			V <sub>L</sub> = 5.0V	2.4			V
Innest Logio Throphold High		T IN CLIDA	V <sub>L</sub> = 3.3V	2.0			
Input Logic Threshold High		T_IN, SHDN	V <sub>L</sub> = 2.5V	1.4			
			V <sub>L</sub> = 1.8V		0.9		
Transmitter Input Hysteresis			•		0.5		V
Input Leakage Current		T_IN, SHDN			±0.01	±1	μΑ
RECEIVER OUTPUTS				1			
Output Leakage Currents		R_OUT, receivers disabled			±0.05	±10	μΑ
Output Voltage Low		I <sub>OUT</sub> = 1.6mA				0.4	V
Output Voltage High		I <sub>OUT</sub> = -1mA		V <sub>L</sub> - 0.6	V <sub>L</sub> - 0.1		V
RECEIVER INPUTS	'						
Input Voltage Range				-25		+25	V
Input Threshold Low		T <sub>A</sub> = +25°C	V <sub>L</sub> = 5.0V	0.8	1.2		V
			V∟ = 3.3V	0.6	1.5		
Input Threshold High		T <sub>A</sub> = +25°C	V <sub>L</sub> = 5.0V		1.8	2.4	V
			V <sub>L</sub> = 3.3V		1.5	2.4	
Input Hysteresis			1		0.5		V
Input Resistance		T <sub>A</sub> = +25°C		3	5	7	kΩ

#### DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = V_L = +3.0V \text{ to } +5.5V; C1-C4 = 0.1\mu\text{F}, \text{ tested at } +3.3V \pm 10\%; C1 = 0.047\mu\text{F}, C2-C4 = 0.33\mu\text{F}, \text{ tested at } +5.0V \pm 10\%; T_A = T_{MIN} \text{ to } T_{MAX}; \text{ unless otherwise noted. Typical values are at } V_{CC} = V_L = +3.3V, T_A = +25^{\circ}\text{C.})$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
TRANSMITTER OUTPUTS						
Output Voltage Swing		All transmitter outputs loaded with $3\text{k}\Omega$ to ground	±5	±5.4		V
Output Resistance		$V_{CC} = V_{+} = V_{-} = 0$ , transmitter output = $\pm 2V$	300	10M		Ω
Output Short-Circuit Current		VT_OUT = 0			±60	mA
Output Leakage Current		$V_{T\_OUT} = \pm 12V$ , transmitters disabled; $V_{CC} = 0$ or 3.0V to 5.5V			±25	μΑ
ESD PROTECTION						
R_IN, T_OUT ESD Protection		Human Body Model		±15		
		IEC 1000-4-2 Air-Gap Discharge method		±25		kV
		IEC 1000-4-2 Contact Discharge method		±8		]

#### **TIMING CHARACTERISTICS**

 $(V_{CC} = V_L = +3V \text{ to } +5.5V; C1-C4 = 0.1\mu\text{F}, \text{ tested at } +3.3V \pm 10\%; C1 = 0.047\mu\text{F}, C2-C4 = 0.33\mu\text{F}, \text{ tested at } +5.0V \pm 10\%; T_A = T_{MIN} \text{ to } T_{MAX}; \text{ unless otherwise noted. Typical values are at } V_{CC} = V_L = +3.3V, T_A = +25^{\circ}\text{C.})$ 

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Maximum Data Rate		$R_L = 3k\Omega$ , $C_L = 1000pF$ , one transmitter switching		250			kbps
Receiver Propagation Delay	tphL	Receiver input to receiver output,			0.15		LIC.
t <sub>PLH</sub> C <sub>L</sub> = 150pF				0.15		μs	
Receiver Output Enable Time					200		ns
Receiver Output Disable Time					200		ns
Time to Exit Shutdown		V <sub>T_OUT</sub>   > 3.7V			100		μs
Transmitter Skew	tphl - tplh	(Note 2)			100		ns
Receiver Skew	tphl - tplh				50		ns
Transition-Region Slew		$V_{CC} = 3.3V$ , $T_A = +25^{\circ}C$ , $R_I = 3k\Omega$ to $7k\Omega$ .	C <sub>L</sub> = 150pF to 1000pF	6		30	V/µs
Rate		measured from +3V to -3V or -3V to +3V	C <sub>L</sub> = 150pF to 2500pF	4		30	ν/μ5

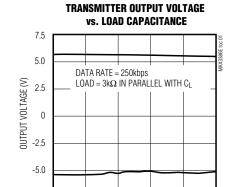
Note 2: Transmitter skew is measured at the transmitter zero crosspoint.

#### **Typical Operating Characteristics**

 $(V_{CC} = V_L = +3.3V, T_A = +25^{\circ}C, unless otherwise noted.)$ 

-7.5

1000

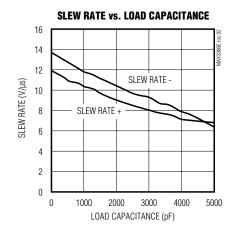


2000

3000

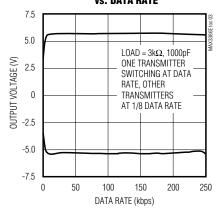
4000

5000

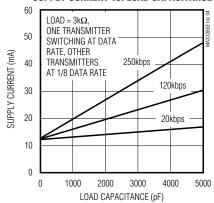


## TRANSMITTER OUTPUT VOLTAGE vs. Data rate

LOAD CAPACITANCE (pF)



#### SUPPLY CURRENT vs. LOAD CAPACITANCE



### **Pin Description**

PIN	NAME	FUNCTION	
1	C1+	Positive Terminal of the Voltage-Doubler Charge-Pump Capacitor	
2	V+	+5.5V Supply Generated by the Charge Pump	
3	C1-	Negative Terminal of the Voltage-Doubler Charge-Pump Capacitor	
4	C2+	Positive Terminal of the Inverting Charge-Pump Capacitor	
5	C2-	Negative Terminal of the Inverting Charge-Pump Capacitor	
6	V-	-5.5V Generated by the Charge Pump	
7	T1IN		
8	T2IN	TTL/CMOS Transmitter Inputs	
9	T3IN		
10	R2OUT	TTL/CMOS Pagaivar Outputa Suring between 0 and V	
11	R1OUT	TTL/CMOS Receiver Outputs. Swing between 0 and V <sub>L</sub> .	
12	VL	Logic-Level Supply. All CMOS inputs and outputs are referenced to this supply.	
13	R2IN	RS-232 Receiver Inputs	
14	R1IN	no-232 neceiver iliputs	
15	T3OUT		
16	T2OUT	RS-232 Transmitter Outputs	
17	T1OUT		
18	GND	Ground	
19	Vcc	+3.0V to +5.5V Supply Voltage	
20	SHDN	Shutdown Input. 0 = shutdown, 1 = normal operation.	



#### Detailed Description

#### Dual Charge-Pump Voltage Converter

The MAX3386E's internal power supply consists of a regulated dual charge pump that provides output voltages of +5.5V (doubling charge pump) and -5.5V (inverting charge pump), regardless of the input voltage (VCC) over a +3.0V to +5.5V range. The charge pumps operate in a discontinuous mode: if the output voltages are less than 5.5V, the charge pumps are enabled; if the output voltages exceed 5.5V, the charge pumps are disabled. Each charge pump requires a flying capacitor (C1, C2) and a reservoir capacitor (C3, C4) to generate the V+ and V- supplies.

#### **RS-232 Transmitters**

The transmitters are inverting level translators that convert CMOS-logic levels to 5.0V EIA/TIA-232 levels.

The MAX3386E's transmitters guarantee a 250kbps data rate with worst-case loads of  $3k\Omega$  in parallel with 1000pF, providing compatibility with PC-to-PC communication software (such as LapLink<sup>TM</sup>). Transmitters can be paralleled to drive multiple receivers or mice. Figure 1 shows a complete system connection.

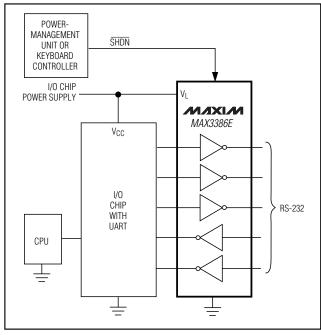


Figure 1. Interface Under Control of PMU

These RS-232 output stages are turned off (high impedance) when the device is in shutdown mode. When the power is off, the MAX3386E permits the outputs to be driven up to  $\pm 12V$ .

The transmitter inputs do not have pull-up resistors. Connect unused inputs to GND or  $V_{\rm L}$ .

#### **RS-232 Receivers**

The receivers convert RS-232 signals to CMOS-logic output levels. The MAX3386E's receivers have inverting three-state outputs, which depend on the shutdown state of the device.

#### **Shutdown Mode**

Supply current falls to less than 1 $\mu$ A when the MAX3386E is placed in shutdown mode (SHDN logic low). When shut down, the device's charge pumps are turned off, V+decays to V<sub>CC</sub>, V- is pulled to ground, and the transmitter outputs are disabled (high impedance). The time required to exit shutdown is typically 100 $\mu$ s, as shown in Figure 2. Connect SHDN to V<sub>CC</sub> if the shutdown mode is not used. In shutdown mode, the receiver outputs are high impedance (Table 1).

**Table 1. Shutdown Logic Truth Table** 

SHDN	TRANSMITTER OUTPUTS	RECEIVER OUTPUTS	CHARGE PUMP	
L	High-Z	High-Z	Inactive	
Н	Active	Active	Active	

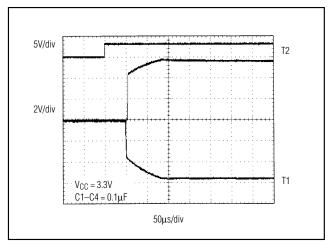


Figure 2. Transmitter Outputs when Exiting Shutdown

LapLink is a trademark of Traveling Software.

#### **V<sub>L</sub> Logic Supply Input**

Unlike other RS-232 interface devices where the receiver outputs swing between 0 and V<sub>CC</sub>, the MAX3386E features a separate logic supply input (V<sub>L</sub>) that sets V<sub>OH</sub> for the receiver outputs and sets thresholds for the transmitted and shutdown inputs. This feature allows a great deal of flexibility in interfacing to many different types of systems with different logic levels. Connect this input to the host logic supply (1.8V  $\leq$  V<sub>L</sub>  $\leq$  V<sub>CC</sub>). Also see the *Typical PDA/Cell-Phone Application* section.

#### ±25kV ESD Protection

As with all Maxim devices, ESD-protection structures are incorporated on all pins to protect against electrostatic discharges (ESDs) encountered during handling and assembly. The MAX3386E's driver outputs and receiver inputs have extra protection against static electricity. Maxim has developed state-of-the-art structures to protect these pins against an ESD of ±25kV without damage. The ESD structures withstand high ESD in all states: normal operation, shutdown, and powered down. After an ESD event, Maxim's "E" version devices keep working without latchup, whereas competing RS-232 products can latch and must be powered down to remove latchup. ESD protection can be tested in various ways. The transmitter outputs and receiver inputs of this product family are characterized for protection to the following limits:

- 1) ±15kV using the Human Body Model
- 2) ±8kV using the Contact Discharge method specified in IEC 1000-4-2
- 3) ±25kV using IEC 1000-4-2's Air-Gap Discharge method

#### **ESD Test Conditions**

ESD performance depends on a variety of conditions. Contact Maxim for a reliability report that documents test setup, methodology, and results.

#### **Human Body Model**

Figure 3a shows the Human Body Model, and Figure 3b shows the current waveform it generates when discharged into a low impedance. This model consists of a 100pF capacitor charged to the ESD voltage of interest, which is then discharged into the test device through a  $1.5 \mathrm{k}\Omega$  resistor.

#### IEC 1000-4-2

The IEC 1000-4-2 standard covers ESD testing and performance of finished equipment; it does not specifically refer to ICs. The MAX3386E helps you design equipment that meets Level 4 (the highest level) of IEC

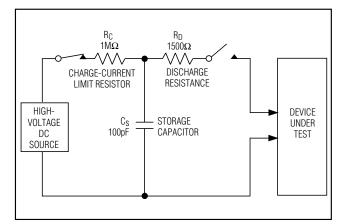


Figure 3a. Human Body ESD Test Model

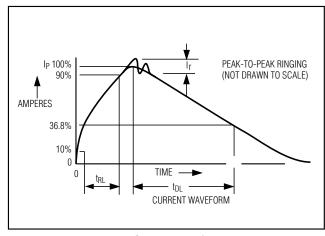


Figure 3b. Human Body Current Waveform

1000-4-2, without the need for additional ESD-protection components.

The major difference between tests done using the Human Body Model and IEC 1000-4-2 is higher peak current in IEC 1000-4-2, because series resistance is lower in the IEC 1000-4-2 model. Hence, the ESD withstand voltage measured to IEC 1000-4-2 is generally lower than that measured using the Human Body Model. Figure 4a shows the IEC 1000-4-2 model, and Figure 4b shows the current waveform for the ±8kV IEC 1000-4-2 Level 4 ESD Contact-Discharge test.

The air-gap test involves approaching the device with a charged probe. The contact-discharge method connects the probe to the device before the probe is energized.

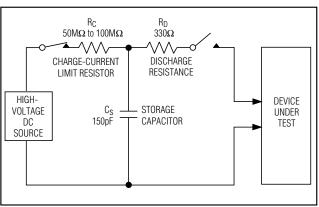


Figure 4a. IEC 1000-4-2 ESD Test Model

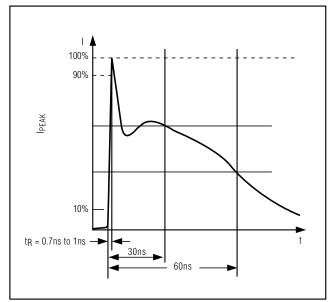


Figure 4b. IEC 1000-4-2 ESD Generator Current Waveform

## Table 2. Minimum Required Capacitor Values

Vcc (V)	C1 (μF)	C2, C3, C4 (μF)
3.0 to 3.6	0.1	0.1
4.5 to 5.5	0.047	0.33
3.0 to 5.5	0.22	1

#### **Machine Model**

The Machine Model for ESD tests all pins using a 200pF storage capacitor and zero discharge resistance. Its objective is to emulate the stress caused by contact that occurs with handling and assembly during manufacturing. All pins require this protection during manufacturing, not just RS-232 inputs and outputs. Therefore, after PC board assembly, the Machine Model is less relevant to I/O ports.

### \_Applications Information

#### **Capacitor Selection**

The capacitor type used for C1–C4 is not critical for proper operation; polarized or nonpolarized capacitors can be used. The charge pump requires 0.1µF capacitors for 3.3V operation. For other supply voltages, see Table 2 for required capacitor values. Do not use values smaller than those listed in Table 2. Increasing the capacitor values (e.g., by a factor of 2) reduces ripple on the transmitter outputs and slightly reduces power consumption. C2, C3, and C4 can be increased without changing C1's value. However, do not increase C1 without also increasing the values of C2, C3, and C4 to maintain the proper ratios (C1 to the other capacitors).

When using the minimum required capacitor values, make sure the capacitor value does not degrade excessively with temperature. If in doubt, use capacitors with a larger nominal value. The capacitor's equivalent series resistance (ESR), which usually rises at low temperatures, influences the amount of ripple on V+ and V-.

#### **Power-Supply Decoupling**

In most circumstances, a  $0.1\mu F$  bypass capacitor is adequate. In applications that are sensitive to power-supply noise, decouple V<sub>CC</sub> to ground with a capacitor of the same value as charge-pump capacitor C1. Connect bypass capacitors as close to the IC as possible.

#### Operation Down to 2.7V

Transmitter outputs will meet EIA/TIA-562 levels of  $\pm 3.7V$  with supply voltages as low as  $\pm 2.7V$ .

#### Transmitter Outputs when Exiting Shutdown

Figure 2 shows two transmitter outputs when exiting shutdown mode. As they become active, the two transmitter outputs are shown going to opposite RS-232 levels (one transmitter input is high; the other is low). Each transmitter is loaded with  $3k\Omega$  in parallel with 2500pF. The transmitter outputs display no ringing or undesirable transients as they come out of shutdown. Note that the transmitters are enabled only when the magnitude of V- exceeds approximately 3V.

#### **High Data Rates**

The MAX3386E maintains the RS-232 ±5.0V minimum transmitter output voltage even at high data rates. Figure 5 shows a transmitter loopback test circuit. Figure 6 shows a loopback test result at 120kbps, and Figure 7 shows the same test at 250kbps. For Figure 6, all transmitters were driven simultaneously at 120kbps into RS-232 loads in parallel with 1000pF. For Figure 7, a single transmitter was driven at 250kbps, and all transmitters were loaded with an RS-232 receiver in parallel with 1000pF.

## Interconnection with 3V and 5V Logic

The MAX3386E can directly interface with various 5V logic families, including ACT and HCT CMOS. The logic voltage power-supply pin  $V_L$  sets the output voltage level of the receivers and the input thresholds of the transmitters.

#### **Typical PDA/Cell-Phone Application**

The MAX3386E is designed with PDA applications in mind. Two transmitters and two receivers handle standard full-duplex communication protocol, while an extra transmitter allows a ring indicator (RI) signal to alert the UART on the PC. Without the ring indicator transmitter, solutions for these applications would require software-intensive polling of the cradle inputs.

The RI signal is generated when a PDA, cellular phone, or other "cradled" device is plugged into its cradle. This generates a logic-low signal to RI transmitter input, creating +6V at the ring indicate pin. The PC's UART RI input is the only pin that can generate an interrupt from signals arriving through the RS-232 port. The interrupt routine for this UART will then service the RS-232 full-duplex communication between the PDA and the PC.

As cell phone design becomes more like that of PDAs, cell phones will require similar docking ability and communication protocol. Cell phones operate on a single lithium-ion (Li+) battery and work with a power-supply voltage of +2.7V to +4V. The baseband logic coming from the phone connector can be as low as 1.8V at the transceivers. To prevent forward biasing of a device internal to the cell phone, the MAX3386E comes with a logic power-supply pin (VL) that limits the logic levels presented to the phone. The receiver outputs will sink to zero for low outputs, but will not exceed V<sub>L</sub> for logic highs. The input logic levels for the transmitters are also altered, scaled by the magnitude of the VL input. The device will work with VL as low as 1.8V before the charge-pump noise will begin to cause the transmitter outputs to oscillate. This is useful with cell phones and other power-efficient devices with core logic voltage levels that go as low as 1.8V.

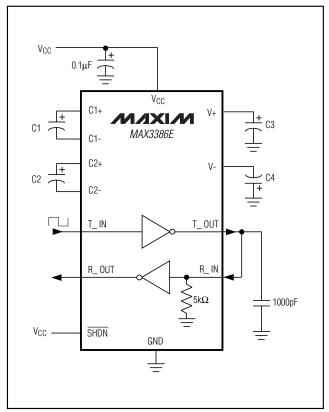


Figure 5. Loopback Test Circuit

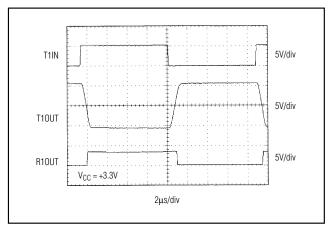


Figure 6. Loopback Test Results at 120kbps

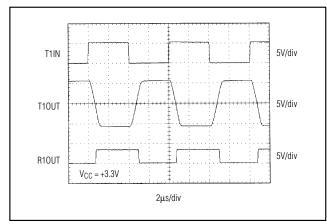
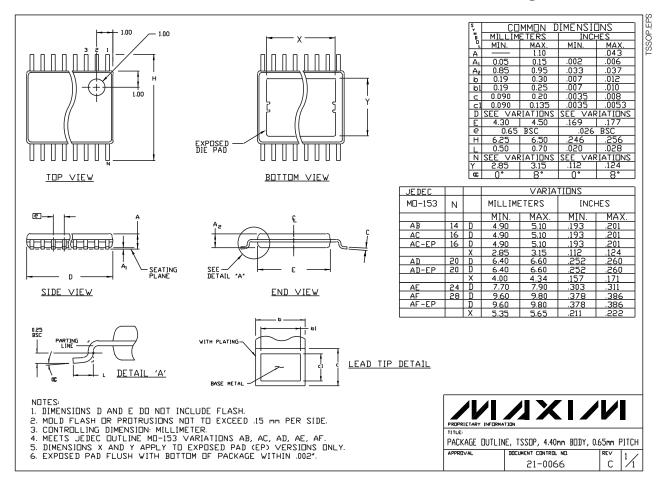


Figure 7. Loopback Test Results at 250kbps

**Chip Information** 

TRANSISTOR COUNT: 1267

#### Package Information



Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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