

# LTC3588-2

## ABSOLUTE MAXIMUM RATINGS (Note 1)

$V_{IN}$		$V_{OUT}$ .....	-0.3V to [Lesser of ( $V_{IN} + 0.3V$ ) or 6V]
Low Impedance Source .....	-0.3V to 18V*	PGOOD.....	-0.3V to [Lesser of ( $V_{OUT} + 0.3V$ ) or 6V]
Current Fed, $I_{SW} = 0A$ .....	25mA†	$I_{PZ1}, I_{PZ2}$ .....	$\pm 50mA$
PZ1, PZ2 .....	0V to $V_{IN}$	$I_{SW}$ .....	350mA
D0, D1 .....	-0.3V to [Lesser of ( $V_{IN2} + 0.3V$ ) or 6V]	Operating Junction Temperature Range	
CAP .....	[Higher of -0.3V or ( $V_{IN} - 6V$ )] to $V_{IN}$	(Notes 2, 3) .....	-40 to 125°C
$V_{IN2}$ .....	-0.3V to [Lesser of ( $V_{IN} + 0.3V$ ) or 6V]	Storage Temperature Range .....	-65 to 125°C
		Lead Temperature (Soldering, 10 sec)	
		MSE Only .....	300°C

\*  $V_{IN}$  has an internal 20V clamp

† For  $t < 1ms$  and Duty Cycle  $< 1\%$ ,  
Absolute Maximum Continuous Current = 5mA

## PIN CONFIGURATION

<p>TOP VIEW</p> <p>DD PACKAGE 10-LEAD (3mm × 3mm) PLASTIC DFN</p> <p><math>T_{JMAX} = 125^{\circ}C</math>, <math>\theta_{JA} = 43^{\circ}C/W</math>, <math>\theta_{JC} = 7.5^{\circ}C/W</math> EXPOSED PAD (PIN 11) IS GND, MUST BE SOLDERED TO PCB</p>	<p>TOP VIEW</p> <p>MSE PACKAGE 10-LEAD PLASTIC MSOP</p> <p><math>T_{JMAX} = 125^{\circ}C</math>, <math>\theta_{JA} = 45^{\circ}C/W</math>, <math>\theta_{JC} = 10^{\circ}C/W</math> EXPOSED PAD (PIN 11) IS GND, MUST BE SOLDERED TO PCB</p>
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## ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LTC3588EDD-2#PBF	LTC3588EDD-2#TRPBF	LFYK	10-Lead (3mm × 3mm) Plastic DFN	-40°C to 125°C
LTC3588IDD-2#PBF	LTC3588IDD-2#TRPBF	LFYK	10-Lead (3mm × 3mm) Plastic DFN	-40°C to 125°C
LTC3588EMSE-2#PBF	LTC3588EMSE-2#TRPBF	LTFYM	10-Lead Plastic MSOP	-40°C to 125°C
LTC3588IMSE-2#PBF	LTC3588IMSE-2#TRPBF	LTFYM	10-Lead Plastic MSOP	-40°C to 125°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. \*The temperature grade is identified by a label on the shipping container.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

This product is only offered in trays. For more information go to: <http://www.linear.com/packaging/>

## ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating junction temperature range, otherwise specifications are for  $T_A = 25^\circ\text{C}$  (Note 2).  $V_{IN} = 18\text{V}$  unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS	
V <sub>IN</sub>	Input Voltage Range	Low Impedance Source on V <sub>IN</sub>	●	18.0			V	
I <sub>Q</sub>	V <sub>IN</sub> Quiescent Current UVLO Buck Enabled, Sleeping Buck Enabled, Not Sleeping	V <sub>IN</sub> = 12V, Not PGOOD V <sub>IN</sub> = 18V I <sub>SW</sub> = 0A (Note 4)			830 1500 150	1400 2500 250	nA nA μA	
V <sub>UVLO</sub>	V <sub>IN</sub> Undervoltage Lockout Threshold	V <sub>IN</sub> Rising	●		16.0	17.0	V	
		V <sub>IN</sub> Falling	●	13.0	14.0	V		
V <sub>SHUNT</sub>	V <sub>IN</sub> Shunt Regulator Voltage	I <sub>VIN</sub> = 1mA		18.8	20.0	21.2	V	
I <sub>SHUNT</sub>	Maximum Protective Shunt Current	1ms Duration		25			mA	
	Internal Bridge Rectifier Loss ( V <sub>PZ1</sub> – V <sub>PZ2</sub>   – V <sub>IN</sub> )	I <sub>BRIDGE</sub> = 10μA		350	400	450	mV	
	Internal Bridge Rectifier Reverse Leakage Current	V <sub>REVERSE</sub> = 18V		20			nA	
	Internal Bridge Rectifier Reverse Breakdown Voltage	I <sub>REVERSE</sub> = 1μA		V <sub>SHUNT</sub>	30		V	
V <sub>OUT</sub>	Regulated Output Voltage	3.45V Output Selected Sleep Threshold	●	3.346	3.466	3.554	V	
		Wake-Up Threshold	●		3.434	V		
		4.1V Output Selected Sleep Threshold	●	3.979	4.116	4.221	V	
		Wake-Up Threshold	●		4.084	V		
		4.5V Output Selected Sleep Threshold	●	4.354	4.516	4.646	V	
		Wake-Up Threshold	●		4.484	V		
		5.0V Output Selected Sleep Threshold	●	4.825	5.016	5.175	V	
		Wake-Up Threshold	●		4.984	V		
			PGOOD Falling Threshold	As a Percentage of the Selected V <sub>OUT</sub>		83	92	%
		I <sub>VOUT</sub>	Output Quiescent Current	V <sub>OUT</sub> = 5.0V		125	250	nA
I <sub>PEAK</sub>	Buck Peak Switch Current			200	260	350	mA	
I <sub>BUCK</sub>	Available Buck Output Current			100		mA		
R <sub>P</sub>	Buck PMOS Switch On-Resistance			1.1		Ω		
R <sub>N</sub>	Buck NMOS Switch On-Resistance			1.3		Ω		
	Max Buck Duty Cycle		●	100		%		
V <sub>IH</sub> (D0, D1)	D0/D1 Input High Voltage		●	1.2		V		
V <sub>IL</sub> (D0, D1)	D0/D1 Input Low Voltage		●		0.4	V		
I <sub>IH</sub> (D0, D1)	D0/D1 Input High Current				10	nA		
I <sub>IL</sub> (D0, D1)	D0/D1 Input Low Current				10	nA		

**Note 1:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

**Note 2:** The LTC3588E-2 is tested under pulsed load conditions such that  $T_J \approx T_A$ . The LTC3588E-2 is guaranteed to meet specifications from  $0^\circ\text{C}$  to  $85^\circ\text{C}$  junction temperature. Specifications over the  $-40^\circ\text{C}$  to  $125^\circ\text{C}$  operating junction temperature range are assured by design, characterization and correlation with statistical process controls. The LTC3588I-2 is guaranteed over the  $-40^\circ\text{C}$  to  $125^\circ\text{C}$  operating junction

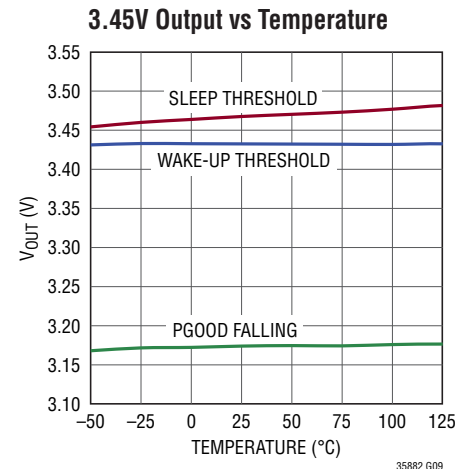
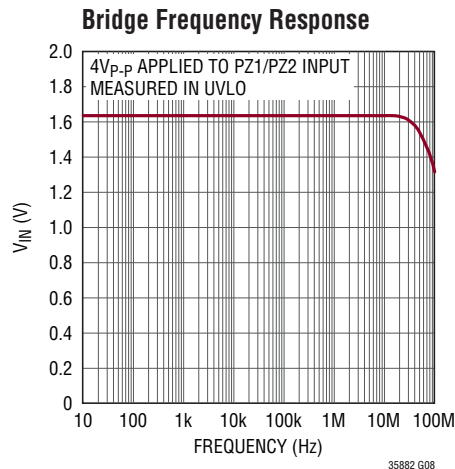
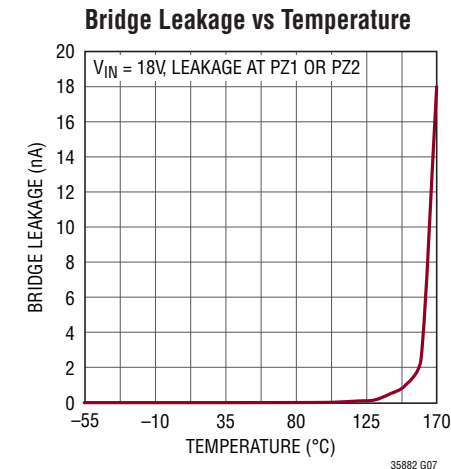
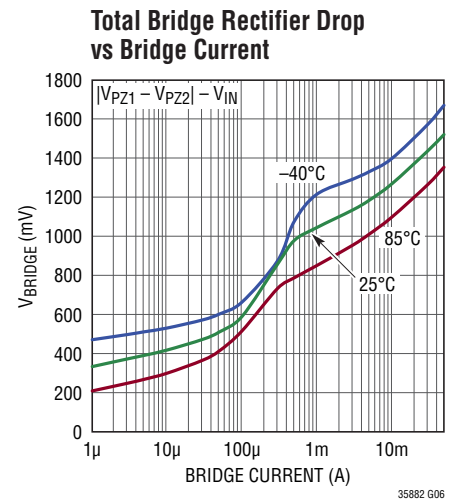
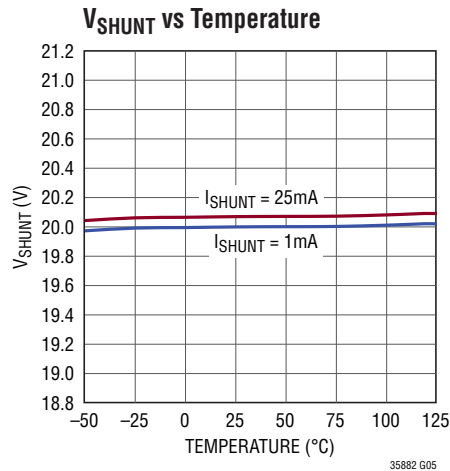
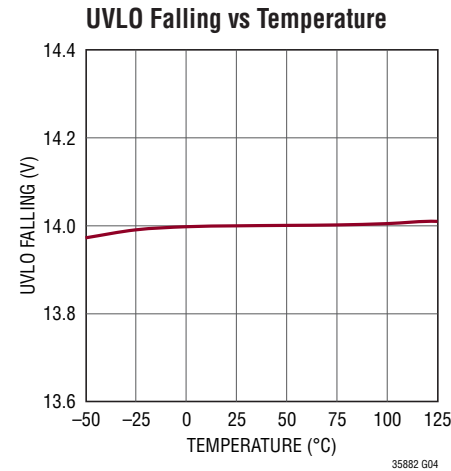
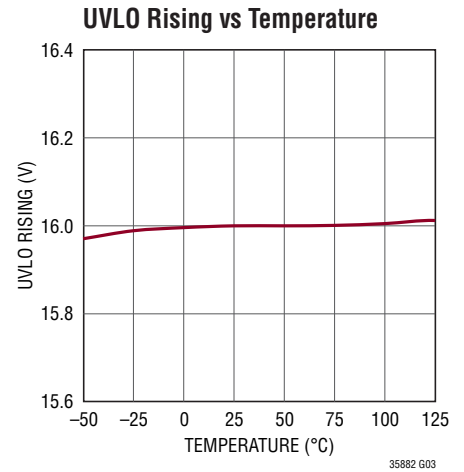
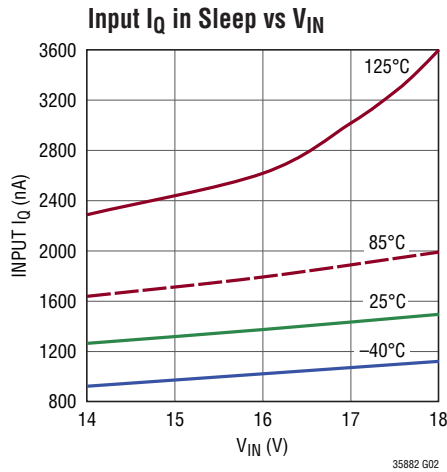
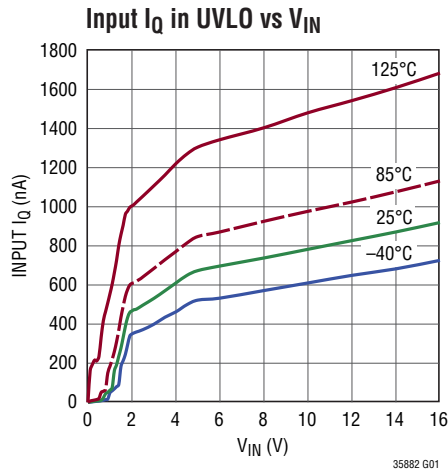
temperature range. Note that the maximum ambient temperature consistent with these specifications is determined by specific operating conditions in conjunction with board layout, the rated package thermal impedance and other environmental factors.

**Note 3:** The junction temperature ( $T_J$ , in  $^\circ\text{C}$ ) is calculated from the ambient temperature ( $T_A$ , in  $^\circ\text{C}$ ) and power dissipation ( $P_D$ , in Watts) according to the formula:  $T_J = T_A + (P_D \cdot \theta_{JA})$ , where  $\theta_{JA}$  (in  $^\circ\text{C}/\text{W}$ ) is the package thermal impedance.

**Note 4:** Dynamic supply current is higher due to gate charge being delivered at the switching frequency.

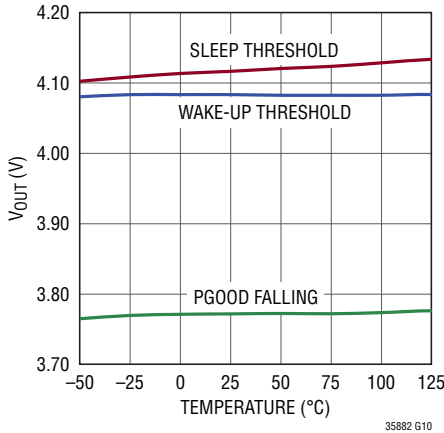
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# TYPICAL PERFORMANCE CHARACTERISTICS

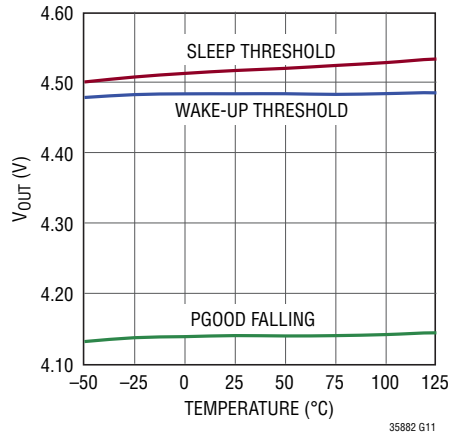


# TYPICAL PERFORMANCE CHARACTERISTICS

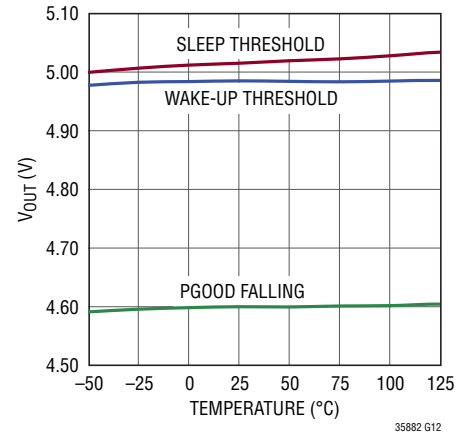
## 4.1V Output vs Temperature



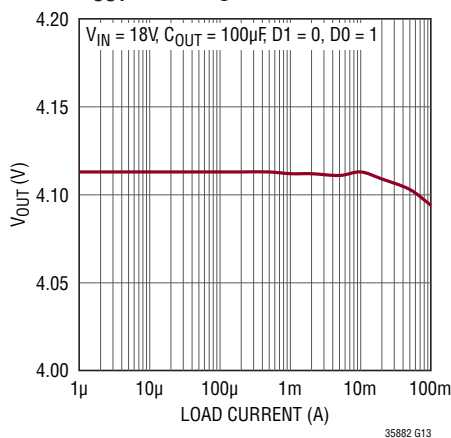
## 4.5V Output vs Temperature



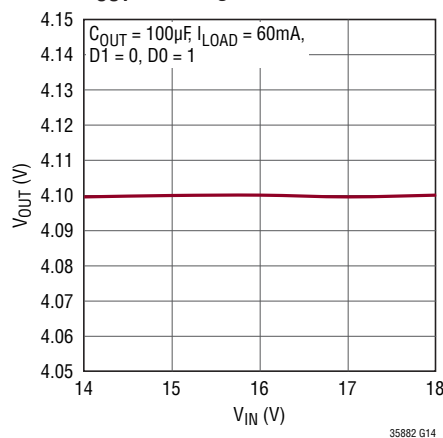
## 5.0V Output vs Temperature



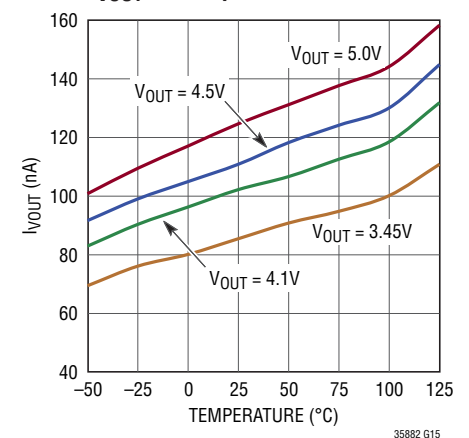
## VOUT Load Regulation



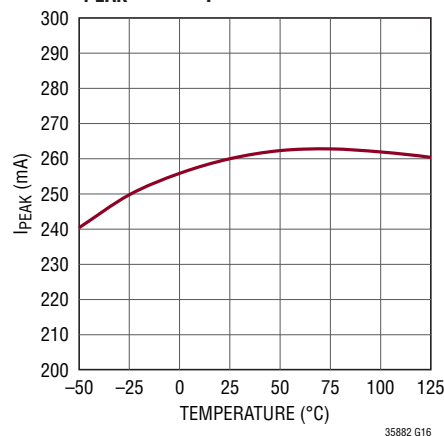
## VOUT Line Regulation



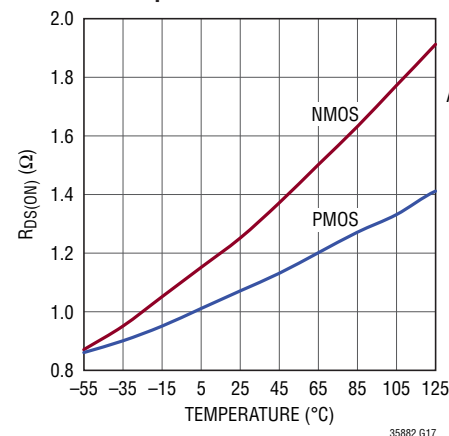
## IOUT vs Temperature



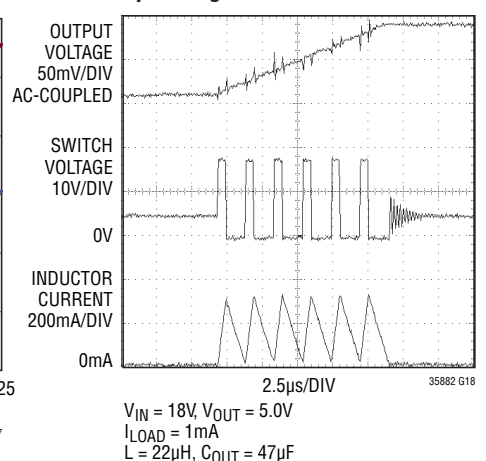
## IPEAK vs Temperature



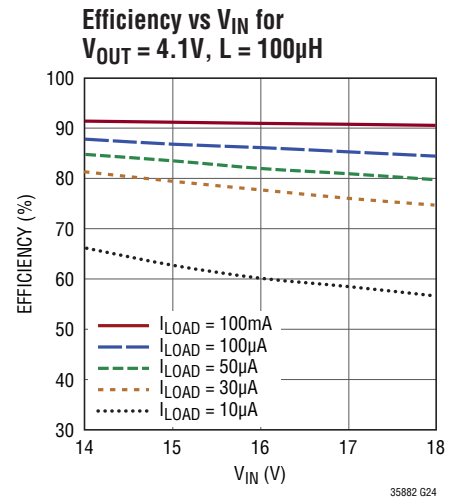
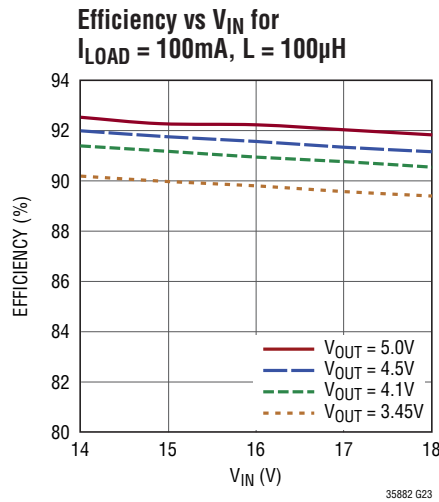
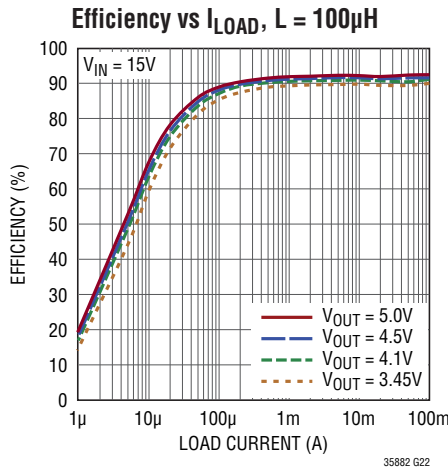
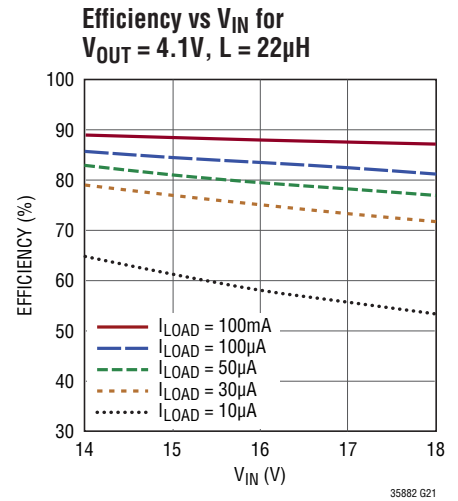
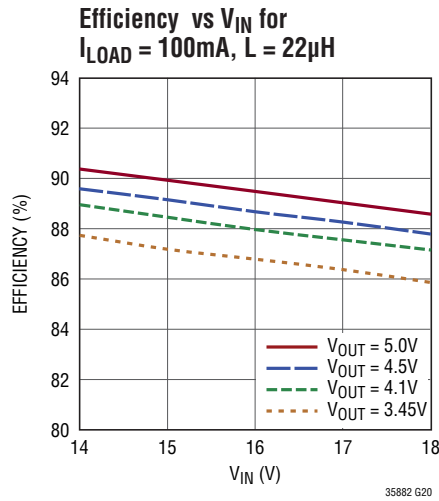
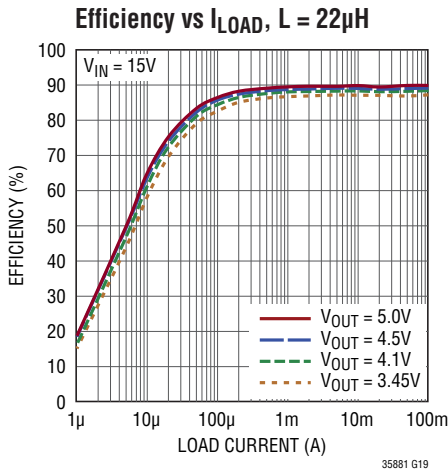
## RDS(ON) of PMOS/NMOS vs Temperature



## Operating Waveforms



# TYPICAL PERFORMANCE CHARACTERISTICS



## PIN FUNCTIONS

**PZ1 (Pin 1):** Input connection for piezoelectric element or other AC source (used in conjunction with PZ2).

**PZ2 (Pin 2):** Input connection for piezoelectric element or other AC source (used in conjunction with PZ1).

**CAP (Pin 3):** Internal rail referenced to  $V_{IN}$  to serve as gate drive for buck PMOS switch. A  $1\mu\text{F}$  capacitor should be connected between CAP and  $V_{IN}$ . This pin is not intended for use as an external system rail.

**$V_{IN}$  (Pin 4):** Rectified Input Voltage. A capacitor on this pin serves as an energy reservoir and input supply for the buck regulator. The  $V_{IN}$  voltage is internally clamped to a maximum of 20V (typical).

**SW (Pin 5):** Switch Pin for the Buck Switching Regulator. A  $22\mu\text{H}$  or larger inductor should be connected from SW to  $V_{OUT}$ .

**$V_{OUT}$  (Pin 6):** Sense pin used to monitor the output voltage and adjust it through internal feedback.

**$V_{IN2}$  (Pin 7):** Internal low voltage rail to serve as gate drive for buck NMOS switch. Also serves as a logic high rail for output voltage select bits D0 and D1. A  $4.7\mu\text{F}$  capacitor should be connected from  $V_{IN2}$  to GND. This pin is not intended for use as an external system rail.

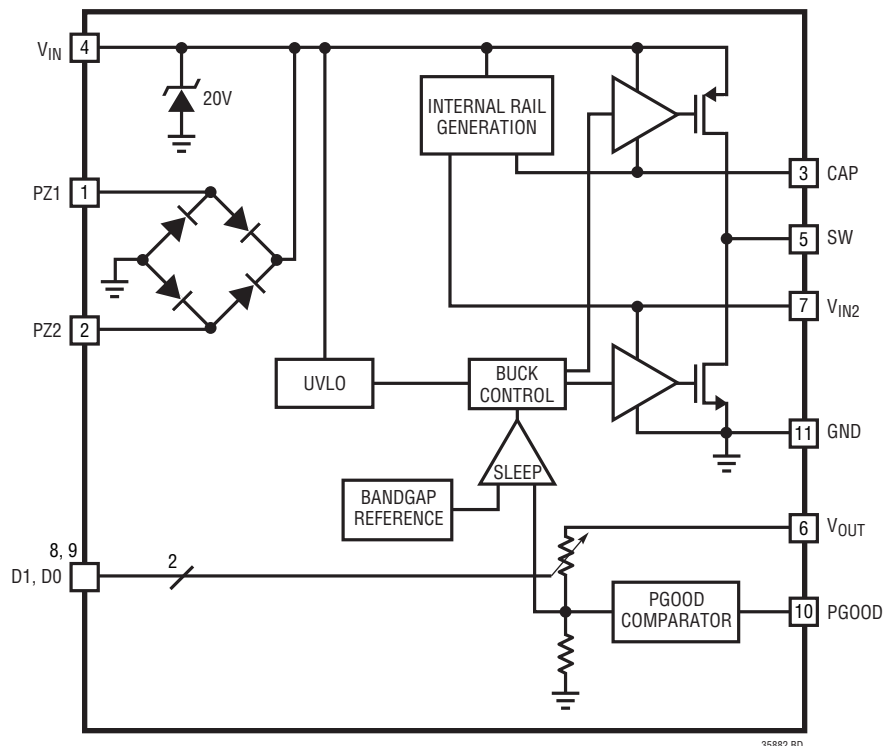
**D1 (Pin 8):** Output Voltage Select Bit. D1 should be tied high to  $V_{IN2}$  or low to GND to select desired  $V_{OUT}$  (see Table 1).

**D0 (Pin 9):** Output Voltage Select Bit. D0 should be tied high to  $V_{IN2}$  or low to GND to select desired  $V_{OUT}$  (see Table 1).

**PGOOD (Pin 10):** Power good output is logic high when  $V_{OUT}$  is above 92% of the target value. The logic high is referenced to the  $V_{OUT}$  rail.

**GND (Exposed Pad Pin 11):** Ground. The Exposed Pad should be connected to a continuous ground plane on the second layer of the printed circuit board by several vias directly under the LTC3588-2.

## BLOCK DIAGRAM



## OPERATION

The LTC3588-2 is an ultralow quiescent current power supply designed specifically for energy harvesting and/or low current step-down applications. The part is designed to interface directly to a piezoelectric or alternative A/C power source, rectify a voltage waveform and store harvested energy on an external capacitor, bleed off any excess power via an internal shunt regulator, and maintain a regulated output voltage by means of a nanopower high efficiency synchronous buck regulator.

### Internal Bridge Rectifier

The LTC3588-2 has an internal full-wave bridge rectifier accessible via the differential PZ1 and PZ2 inputs that rectifies AC inputs such as those from a piezoelectric element. The rectified output is stored on a capacitor at the  $V_{IN}$  pin and can be used as an energy reservoir for the buck converter. The low-loss bridge rectifier has a total drop of about 400mV with typical piezo generated currents ( $\sim 10\mu A$ ). The bridge is capable of carrying up to 50mA. One side of the bridge can be operated as a single-ended DC input. PZ1 and PZ2 should never be shorted together when the bridge is in use.

### Undervoltage Lockout (UVLO)

When the voltage on  $V_{IN}$  rises above the UVLO rising threshold the buck converter is enabled and charge is transferred from the input capacitor to the output capacitor. A wide ( $\sim 2V$ ) UVLO hysteresis window allows a portion of the energy stored on the input capacitor to be transferred to the output capacitor by the buck. When the input capacitor voltage is depleted below the UVLO falling threshold the buck converter is disabled. Extremely low quiescent current (830nA typical,  $V_{IN} = 12V$ ) in UVLO allows energy to accumulate on the input capacitor in situations where energy must be harvested from low power sources.

### Internal Rail Generation

Two internal rails, CAP and  $V_{IN2}$ , are generated from  $V_{IN}$  and are used to drive the high side PMOS and low side NMOS of the buck converter, respectively. Additionally the  $V_{IN2}$  rail serves as logic high for output voltage select bits D0 and D1. The  $V_{IN2}$  rail is regulated at 4.8V above GND while the CAP rail is regulated at 4.8V below  $V_{IN}$ . These are not intended to be used as external rails. Bypass capacitors

are connected to the CAP and  $V_{IN2}$  pins to serve as energy reservoirs for driving the buck switches. When  $V_{IN}$  is below 4.8V,  $V_{IN2}$  is equal to  $V_{IN}$  and CAP is held at GND. Figure 1 shows the ideal  $V_{IN}$ ,  $V_{IN2}$  and CAP relationship.

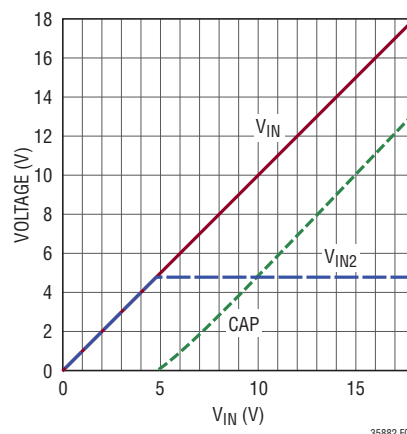


Figure 1. Ideal  $V_{IN}$ ,  $V_{IN2}$  and CAP Relationship

### Buck Operation

The buck regulator uses a hysteretic voltage algorithm to control the output through internal feedback from the  $V_{OUT}$  sense pin. The buck converter charges an output capacitor through an inductor to a value slightly higher than the regulation point. It does this by ramping the inductor current up to 260mA through an internal PMOS switch and then ramping it down to 0mA through an internal NMOS switch. This efficiently delivers energy to the output capacitor. The ramp rate is determined by  $V_{IN}$ ,  $V_{OUT}$ , and the inductor value. If the input voltage falls below the UVLO falling threshold before the output voltage reaches regulation, the buck converter will shut off and will not be turned on until the input voltage again rises above the UVLO rising threshold. During this time the output voltage will be loaded by approximately 100nA. When the buck brings the output voltage into regulation the converter enters a low quiescent current sleep state that monitors the output voltage with a sleep comparator. During this operating mode load current is provided by the buck output capacitor. When the output voltage falls below the regulation point the buck regulator wakes up and the cycle repeats. This hysteretic method of providing a regulated output reduces losses associated with FET switching and maintains an output at light loads. The buck delivers a minimum of 100mA of average current to the output when it is switching.

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## OPERATION

When the sleep comparator signals that the output has reached the sleep threshold the buck converter may be in the middle of a cycle with current still flowing through the inductor. Normally both synchronous switches would turn off and the current in the inductor would freewheel to zero through the NMOS body diode. The LTC3588-2 keeps the NMOS switch on during this time to prevent the conduction loss that would occur in the diode if the NMOS were off. If the PMOS is on when the sleep comparator trips the NMOS will turn on immediately in order to ramp down the current. If the NMOS is on it will be kept on until the current reaches zero.

Though the quiescent current when the buck is switching is much greater than the sleep quiescent current, it is still a small percentage of the average inductor current which results in high efficiency over most load conditions. The buck operates only when sufficient energy has been accumulated in the input capacitor and the length of time the converter needs to transfer energy to the output is much less than the time it takes to accumulate energy. Thus, the buck operating quiescent current is averaged over a long period of time so that the total average quiescent current is low. This feature accommodates sources that harvest small amounts of ambient energy.

Four selectable voltages are available by tying the output select bits, D0 and D1, to GND or  $V_{IN2}$ . Table 1 shows the four D0/D1 codes and their corresponding output voltages.

**Table 1. Output Voltage Selection**

D1	D0	$V_{OUT}$	$V_{OUT}$ QUIESCENT CURRENT ( $I_{VOUT}$ )
0	0	3.45V	86nA
0	1	4.1V	101nA
1	0	4.5V	111nA
1	1	5.0V	125nA

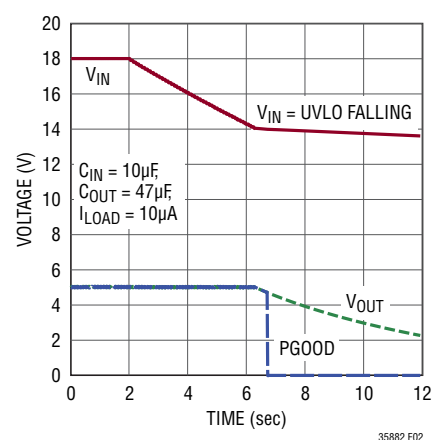
The internal feedback network draws a small amount of current from  $V_{OUT}$  as listed in Table 1.

### Power Good Comparator

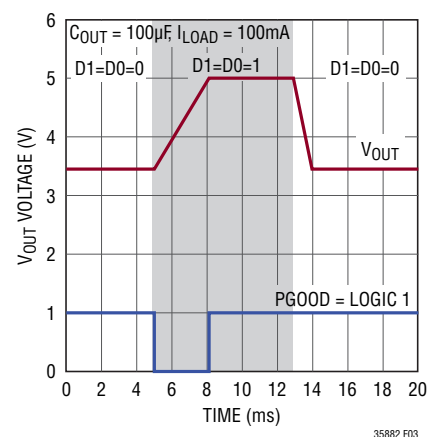
A power good comparator produces a logic high referenced to  $V_{OUT}$  on the PGOOD pin the first time the converter reaches the sleep threshold of the programmed  $V_{OUT}$ , signaling that the output is in regulation. The PGOOD pin will remain high until  $V_{OUT}$  falls to 92% of the desired

regulation voltage. Several sleep cycles may occur during this time. Additionally, if PGOOD is high and  $V_{IN}$  falls below the UVLO falling threshold, PGOOD will remain high until  $V_{OUT}$  falls to 92% of the desired regulation point. This allows output energy to be used even if the input is lost. Figure 2 shows the behavior for  $V_{OUT} = 5V$  and a  $10\mu A$  load. At  $t = 2s$   $V_{IN}$  becomes high impedance and is discharged by the quiescent current of the LTC3588-2 and through servicing  $V_{OUT}$  which is discharged by its own leakage current.  $V_{IN}$  crosses UVLO falling but PGOOD remains high until  $V_{OUT}$  decreases to 92% of the desired regulation point. The PGOOD pin is designed to drive a microprocessor or other chip I/O and is not intended to drive higher current loads such as an LED.

The D0/D1 inputs can be switched while in regulation as shown in Figure 3. If  $V_{OUT}$  is programmed to a voltage with a PGOOD falling threshold above the old  $V_{OUT}$ , PGOOD will



**Figure 2. PGOOD Operation During Transition to UVLO**



**Figure 3. PGOOD Operation During D0/D1 Transition**

OPERATION

transition low until the new regulation point is reached. When  $V_{OUT}$  is programmed to a lower voltage, PGOOD will remain high through the transition.

Energy Storage

Harvested energy can be stored on the input capacitor or the output capacitor. The high UVLO threshold takes advantage of the fact that energy storage on a capacitor is proportional to the square of the capacitor voltage. After the output voltage is brought into regulation any excess energy is stored on the input capacitor and its voltage increases. When a load exists at the output the buck can efficiently transfer energy stored at a high voltage to the

regulated output. While energy storage at the input utilizes the high voltage at the input, the load current is limited to what the buck converter can supply. If larger loads need to be serviced the output capacitor can be sized to support a larger current for some duration. For example, a current burst could begin when PGOOD goes high and would continuously deplete the output capacitor until PGOOD went low.

The output voltages available on the LTC3588-2 are particularly suited to Li-Ion and LiFePO<sub>4</sub> batteries as well as supercapacitors for applications where energy storage at the output is desired.

APPLICATIONS INFORMATION

Introduction

The LTC3588-2 harvests ambient vibrational energy through a piezoelectric element in its primary application. Common piezoelectric elements are PZT (lead zirconate titanate) ceramics, PVDF (polyvinylidene fluoride) polymers, or other composites. Ceramic piezoelectric elements exhibit a piezoelectric effect when the crystal structure of the ceramic is compressed and internal dipole movement produces a voltage. Polymer elements comprised of long-chain molecules produce a voltage when flexed as molecules repel each other. Ceramics are often used under direct pressure while a polymer can be flexed more

readily. A wide range of piezoelectric elements are available and produce a variety of open-circuit voltages and short-circuit currents. Typically the open-circuit voltage and short-circuit currents increase with available vibrational energy as shown in Figure 4. Piezoelectric elements can be placed in series or in parallel to achieve desired open-circuit voltages.

The LTC3588-2 is well-suited to a piezoelectric energy harvesting application. The 20V input protective shunt can accommodate a variety of piezoelectric elements. The low quiescent current of the LTC3588-2 enables efficient energy accumulation from piezoelectric elements which can have short-circuit currents on the order of tens of microamps. Piezoelectric elements can be obtained from manufacturers listed in Table 2.

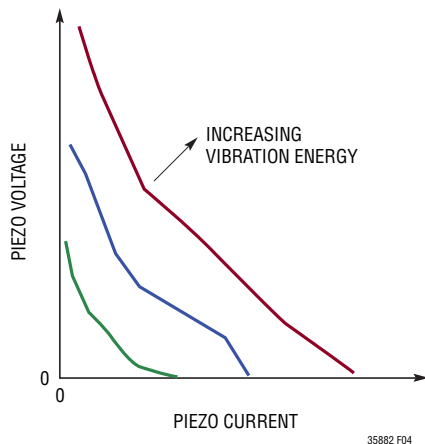


Figure 4. Typical Piezoelectric Load Lines

Table 2. Piezoelectric Element Manufacturers

Advanced Cerametrics	<a href="http://www.advancedcerametrics.com">www.advancedcerametrics.com</a>
Piezo Systems	<a href="http://www.piezo.com">www.piezo.com</a>
Measurement Specialties	<a href="http://www.meas-spec.com">www.meas-spec.com</a>
PI (Physik Instrumente)	<a href="http://www.pi-usa.us">www.pi-usa.us</a>
MIDE Technology Corporation	<a href="http://www.mide.com">www.mide.com</a>
Morgan Technical Ceramics	<a href="http://www.morganelectroceramics.com">www.morganelectroceramics.com</a>



## APPLICATIONS INFORMATION

effect during this time. For applications where the output must reach regulation on a single UVLO cycle, the energy required to charge the output capacitor must be taken into account when sizing  $C_{OUT}$ .

The duration for which the regulator sleeps depends on the load current and the size of the output capacitor. The sleep time decreases as the load current increases and/or as the output capacitor decreases. The DC sleep hysteresis window is  $\pm 16\text{mV}$  around the programmed output voltage. Ideally this means that the sleep time is determined by the following equation:

$$t_{\text{SLEEP}} = C_{\text{OUT}} \frac{32\text{mV}}{I_{\text{LOAD}}}$$

This is true for output capacitors on the order of  $100\mu\text{F}$  or larger, but as the output capacitor decreases towards  $10\mu\text{F}$  delays in the internal sleep comparator along with the load current may result in the  $V_{\text{OUT}}$  voltage slewing past the  $\pm 16\text{mV}$  thresholds. This will lengthen the sleep time and increase  $V_{\text{OUT}}$  ripple. A capacitor less than  $10\mu\text{F}$  is not recommended as  $V_{\text{OUT}}$  ripple could increase to an undesirable level.

If transient load currents above  $100\text{mA}$  are required then a larger capacitor can be used at the output. This capacitor will be continuously discharged during a load condition and the capacitor can be sized for an acceptable drop in  $V_{\text{OUT}}$ :

$$C_{\text{OUT}} = (I_{\text{LOAD}} - I_{\text{BUCK}}) \frac{t_{\text{LOAD}}}{V_{\text{OUT}}^+ - V_{\text{OUT}}^-}$$

Here  $V_{\text{OUT}}^+$  is the value of  $V_{\text{OUT}}$  when PGOOD goes high and  $V_{\text{OUT}}^-$  is the desired lower limit of  $V_{\text{OUT}}$ .  $I_{\text{BUCK}}$  is the average current being delivered from the buck converter, typically  $I_{\text{PEAK}}/2$ .

A standard surface mount ceramic capacitor can be used for  $C_{\text{OUT}}$ , though some applications may be better suited to a low leakage aluminum electrolytic capacitor or a supercapacitor. These capacitors can be obtained from manufacturers such as Vishay, Illinois Capacitor, AVX, or CAP-XX.

### Inductor

The buck is optimized to work with a  $22\mu\text{H}$  inductor. Inductor values greater than  $22\mu\text{H}$  may yield benefits in some applications. For example, a larger inductor will benefit high voltage applications by increasing the on-time of the PMOS switch and improving efficiency by reducing gate charge loss. Choose an inductor with a DC current rating greater than  $350\text{mA}$ . The DCR of the inductor can have an impact on efficiency as it is a source of loss. Trade-offs between price, size, and DCR should be evaluated. Table 3 lists several inductors that work well with the LTC3588-2.

**Table 3. Recommended Inductors for LTC3588-2**

INDUCTOR TYPE	L ( $\mu\text{H}$ )	MAX $I_{\text{DC}}$ (mA)	MAX DCR ( $\Omega$ )	SIZE in mm (L x W x H)	MANUFACTURER
A997AS-220M	22	390	0.440	$4.0 \times 4.0 \times 1.8$	Toko
LPS5030-223MLC	22	700	0.190	$4.9 \times 4.9 \times 3.0$	Coilcraft
LPS4012-473MLC	47	350	1.400	$4.0 \times 4.0 \times 1.2$	Coilcraft
SLF7045T	100	500	0.250	$7.0 \times 7.0 \times 4.8$	TDK

### $V_{\text{IN}2}$ and CAP Capacitors

A  $1\mu\text{F}$  capacitor should be connected between  $V_{\text{IN}}$  and CAP and a  $4.7\mu\text{F}$  capacitor should be connected between  $V_{\text{IN}2}$  and GND. These capacitors hold up the internal rails during buck switching and compensate the internal rail generation circuits.

### Additional Applications with Piezo Inputs

The versatile LTC3588-2 can be used in a variety of configurations. Figure 6 shows a single piezo source powering two LTC3588-2s simultaneously, providing capability for multiple rail systems. As the piezo provides input power both  $V_{\text{IN}}$  rails will initially come up together, but when one output starts drawing power, only its corresponding  $V_{\text{IN}}$  will fall as the bridges of each LTC3588-2 provide isolation. Input piezo energy will then be directed to this lower voltage capacitor until both  $V_{\text{IN}}$  rails are again equal. This configuration is expandable to any number of LTC3588-2s powered by a single piezo as long as the piezo can support the sum total of the quiescent currents from each LTC3588-2.

## APPLICATIONS INFORMATION

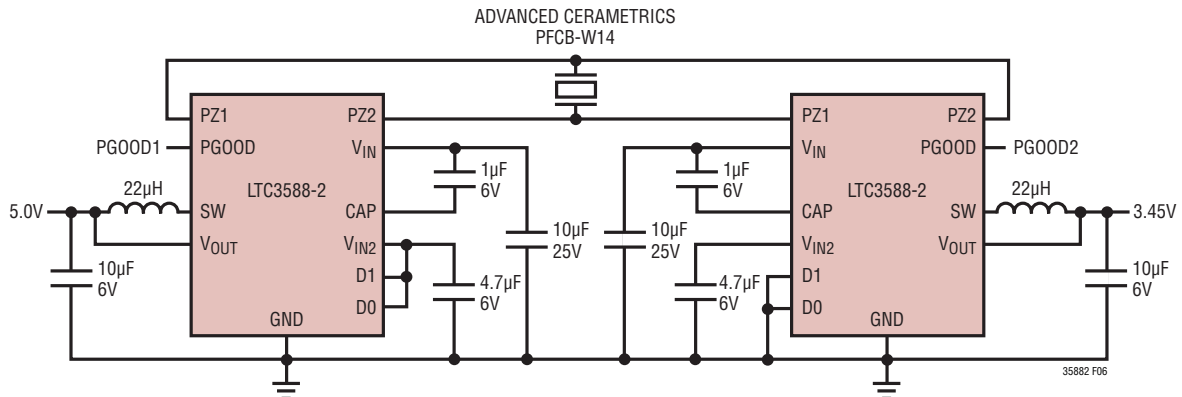


Figure 6. Dual Rail Power Supply with Single Piezo

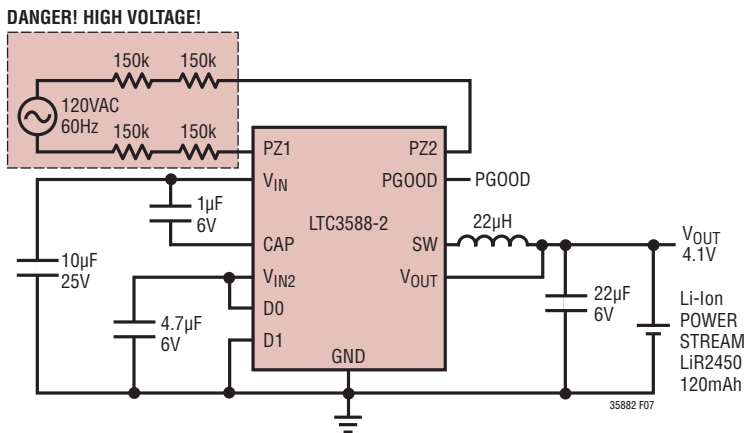


Figure 7. AC Line Powered 4.1V Li-Ion Battery Charger

**DANGER! HIGH VOLTAGE!**

DANGEROUS AND LETHAL POTENTIALS ARE PRESENT IN OFFLINE CIRCUITS! BEFORE PROCEEDING ANY FURTHER, THE READER IS WARNED THAT CAUTION MUST BE USED IN THE CONSTRUCTION, TESTING AND USE OF OFFLINE CIRCUITS. EXTREME CAUTION MUST BE USED IN WORKING WITH AND MAKING CONNECTIONS TO THESE CIRCUITS. REPEAT: OFFLINE CIRCUITS CONTAIN DANGEROUS, AC LINE-CONNECTED HIGH VOLTAGE POTENTIALS. USE CAUTION. ALL TESTING PERFORMED ON AN OFFLINE CIRCUIT MUST BE DONE WITH AN ISOLATION TRANSFORMER CONNECTED BETWEEN THE OFFLINE CIRCUIT'S INPUT AND THE AC LINE. USERS AND CONSTRUCTORS OF OFFLINE CIRCUITS MUST OBSERVE THIS PRECAUTION WHEN CONNECTING TEST EQUIPMENT TO THE CIRCUIT TO AVOID ELECTRIC SHOCK. REPEAT: AN ISOLATION TRANSFORMER MUST BE CONNECTED BETWEEN THE CIRCUIT INPUT AND THE AC LINE IF ANY TEST EQUIPMENT IS TO BE CONNECTED.

## Alternate Power Sources

The LTC3588-2 is not limited to use with piezoelectric elements but can accommodate a wide variety of input sources depending on the type of ambient energy available. Figure 7 shows the LTC3588-2 internal bridge rectifier connected to the AC line in series with four 150k current limiting resistors. This is a high voltage application and minimum spacing between the line, neutral, and any high voltage components should be maintained per the applicable UL specification. For general off-line applications refer to UL regulation 1012.

Figure 8 shows an application where copper panels are placed near a standard fluorescent room light to capacitively harvest energy from the electric field around the light. The

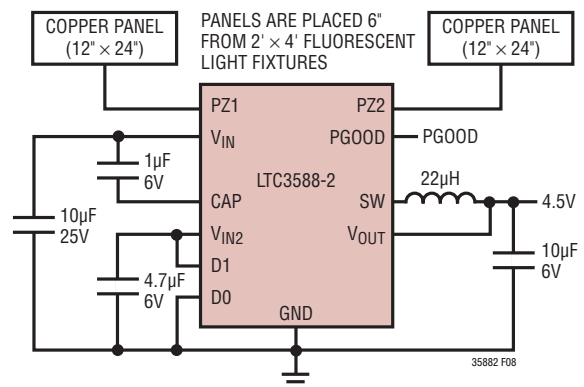


Figure 8. Electric Field Energy Harvester

## APPLICATIONS INFORMATION

frequency of the emission will be 120Hz for magnetic ballasts but could be higher if the light uses electronic ballast. The LTC3588-2 bridge rectifier can handle a wide range of input frequencies.

Figure 9 shows the LTC3588-2 powered by a 48V communications line. In this example, 1mA is the maximum current that is allowed to be drawn. The 28k current limiting resistor sets this current as the LTC3588-2 will shunt  $V_{IN}$  at 20V. The advantage of this scheme is that the current at the output is multiplied by the ratio of  $V_{IN}$  to  $V_{OUT}$  (less the loss in the buck converter). This is useful in cases where greater current is needed at the output than is available at the input. The high UVLO of 16V prevents any start-up issue as there is already a good multiplication factor at

that level. This same technique can be extended to AC source that also have limited current available at the input.

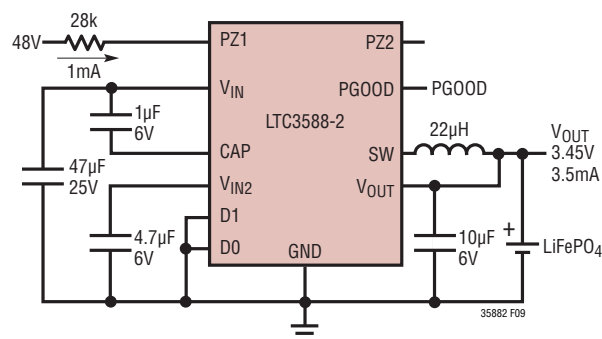


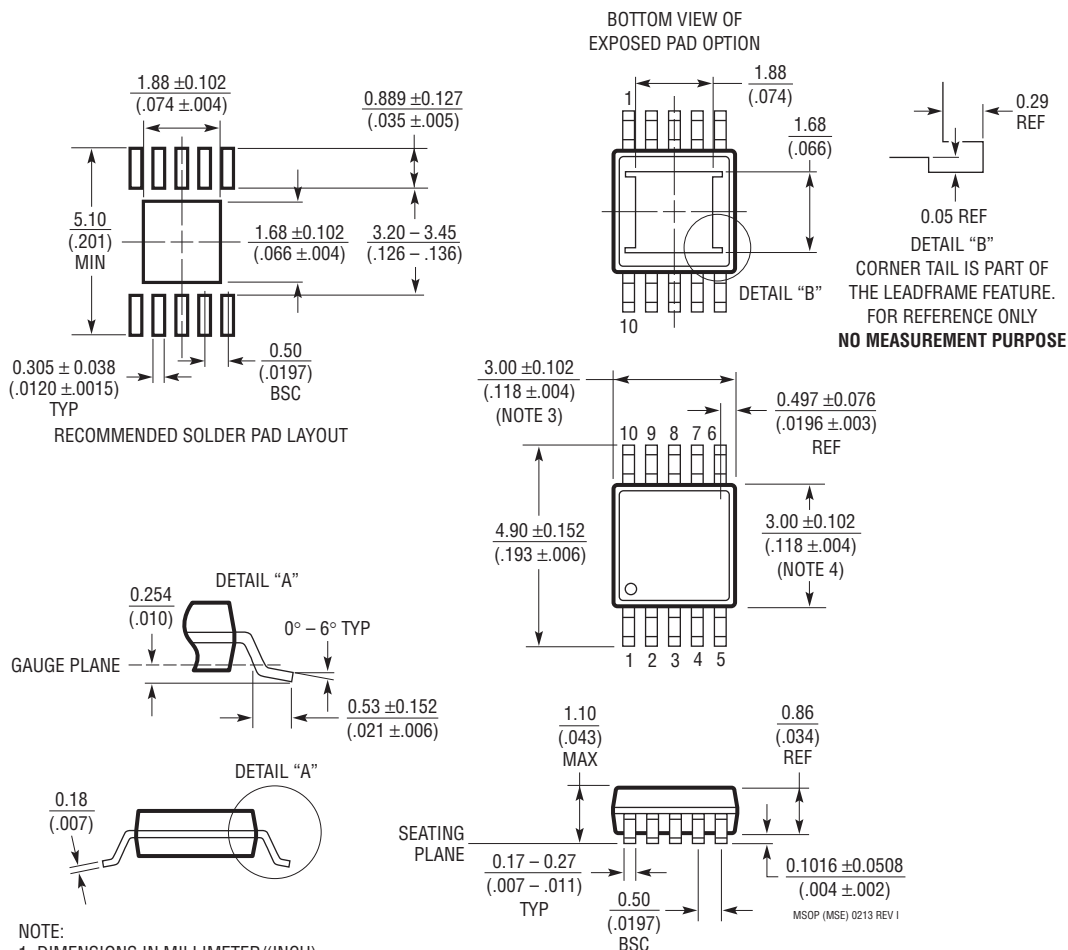
Figure 9. Current Fed 3.45V LiFePO<sub>4</sub> Battery Charger



## PACKAGE DESCRIPTION

Please refer to <http://www.linear.com/designtools/packaging/> for the most recent package drawings.

### MSE Package 10-Lead Plastic MSOP, Exposed Die Pad (Reference LTC DWG # 05-08-1664 Rev I)



**NOTE:**

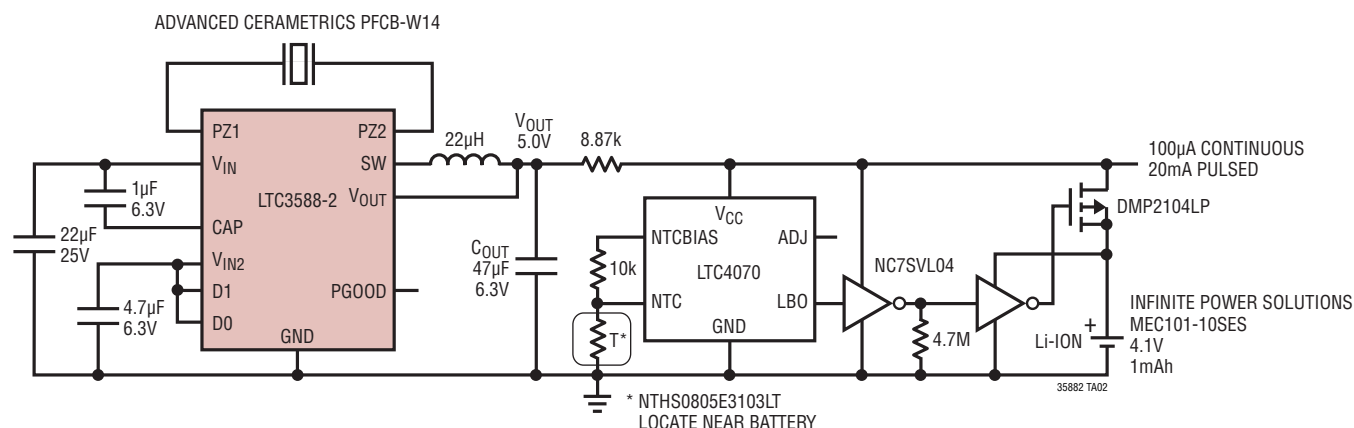
1. DIMENSIONS IN MILLIMETER/(INCH)
2. DRAWING NOT TO SCALE
3. DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.  
MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.152mm (0.006") PER SIDE
4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.  
INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (0.006") PER SIDE
5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (0.004") MAX
6. EXPOSED PAD DIMENSION DOES INCLUDE MOLD FLASH. MOLD FLASH ON E-PAD SHALL NOT EXCEED 0.254mm (0.010") PER SIDE.

## REVISION HISTORY

REV	DATE	DESCRIPTION	PAGE NUMBER
A	5/11	Add brackets to Absolute Maximum Ratings for $V_{OUT}$ and $PGOOD$ .	2
		Replace MS package description to the correct MSE package description.	15
		Add to Related Parts section and order parts by part number.	16
B	7/14	Clarified title and Description	1
		Clarified x-axis label on Figure 1	8
		Clarified Figure 8	14
		Clarified Related Parts list	18
C	8/15	Modified $C_{OUT}$ Equation	12

## TYPICAL APPLICATION

## Piezoelectric Shunt Charger for Small Li-Ion Cells or Thin Film Batteries



## RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
<a href="#">LT1389</a>	Nanopower Precision Shunt Voltage Reference	800nA Operating Current, 1.25V/2.5V/4.096V
<a href="#">LTC1540</a>	Nanopower Comparator with Reference	0.3µA I <sub>Q</sub> , Drives 0.01µF, Adjustable Hysteresis, 2V to 11V Input Range
<a href="#">LT3009</a>	3µA I <sub>Q</sub> , 20mA Low Dropout Linear Regulator	Low 3µA I <sub>Q</sub> , 1.6V to 20V Range, 20mA Output Current
<a href="#">LTC3105</a>	400mA Step-Up Converter with 250mV Start-Up and Maximum Power Point Control	High Efficiency Step-Up DC/DC Converter, V <sub>IN</sub> : 0.225V to 5V, Integrated Maximum Power Point Controller (MPPT), Photovoltaic Cells, Thermoelectric Generators (TEGs), and Fuel Cells, Burst Mode® Operation
<a href="#">LTC3108/</a> <a href="#">LTC3108-1</a>	Ultralow Voltage Step-Up Converter and Power Manager	V <sub>IN</sub> : 0.02V to 1V, V <sub>OUT</sub> = 2.2V, 2.35V, 3.3V, 4.1V, 5V, I <sub>Q</sub> = 6µA, 4mm × 3mm DFN-12, SSOP-16 Packages, LTC3108-1 V <sub>OUT</sub> = 2.2V, 2.5V, 3V, 3.7V, 4.5V
<a href="#">LTC3109</a>	Auto-Polarity, Ultralow Voltage Step-Up Converter and Power Manager	V <sub>IN</sub>  : 0.03V to 1V, V <sub>OUT</sub> = 2.2V, 2.35V, 3.3V, 4.1V, 5V, I <sub>Q</sub> = 7µA, 4mm × 4mm QFN-20, SSOP-20 Packages
<a href="#">LTC3388-1/</a> <a href="#">LTC3388-3</a>	20V High Efficiency Nanopower Step-Down Regulator	860nA I <sub>Q</sub> in Sleep, 2.7V to 20V Input, V <sub>OUT</sub> : 1.2V to 5V, Enable and Standby Pins
<a href="#">LTC3588-1</a>	Nanopower Energy Harvesting Power Supply	950nA I <sub>Q</sub> in Sleep, V <sub>OUT</sub> : 1.8V, 2.5V, 3.3V, 3.6V, Integrated Bridge Rectifier
<a href="#">LTC3631</a>	45V, 100mA, Synchronous Step-Down Regulator with 12µA I <sub>Q</sub>	4.5V to 45V Operating Range, Overvoltage Lockout Up to 60V
<a href="#">LTC3642</a>	45V, 50mA, Synchronous Step-Down Regulator with 12µA I <sub>Q</sub>	4.5V to 45V Operating Range, Overvoltage Lockout Up to 60V
<a href="#">LT3652</a>	Power Tracking 2A Battery Charger for Solar Power	MPPT for Solar Applications, V <sub>IN</sub> : 4.95V to 32V, Charge Rate Up to 2A, User Selectable Termination: C/10 or On-Board Timer, Resistor Programmable Float Voltage up to 14.4V, 3mm × 3mm DFN12 or MSOP-12
<a href="#">LT3970</a>	40V, 350mA Step-Down Regulator with 2.5µA I <sub>Q</sub>	Integrated Boost and Catch Diodes, 4.2V to 40V Operating Range
<a href="#">LT3971</a>	38V, 1.2A, 2MHz Step-Down Regulator with 2.8µA I <sub>Q</sub>	4.3V to 38V Operating Range, Low Ripple Burst Mode Operation
<a href="#">LT3991</a>	55V, 1.2A 2MHz Step-Down Regulator with 2.8µA I <sub>Q</sub>	4.3V to 55V Operating Range, Low Ripple Burst Mode Operation
<a href="#">LTC4070</a>	Li-Ion/Polymer Shunt Battery Charger System	450nA I <sub>Q</sub> , 1% Float Voltage Accuracy, 50mA Shunt Current 4V/4.1V/4.2V
<a href="#">LTC4071</a>	Li-Ion/Polymer Shunt Battery Charger System with Low Battery Disconnect	550nA I <sub>Q</sub> , 1% Float Voltage Accuracy, <10nA Low Battery Disconnect, 4V/4.1V/4.2V, 8-Lead 2mm × 3mm DFN and MSOP Packages
<a href="#">LTC3330</a>	Nanopower Buck-Boost DC/DC with Energy Harvesting Battery Life Extender	V <sub>IN</sub> : 2.7V to 20V, BAT: 1.8V to 5.5V, 750nA I <sub>Q</sub> , 5mm × 5mm QFN-32 Package
<a href="#">LTC3331</a>	Nanopower Buck-Boost DC/DC with Energy Harvesting Battery Charger	V <sub>IN</sub> : 2.7V to 20V, BAT: Up to 4.2V, Shunt Charger, Low Battery Disconnect, 950nA I <sub>Q</sub> , 5mm × 5mm QFN-32 Package

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