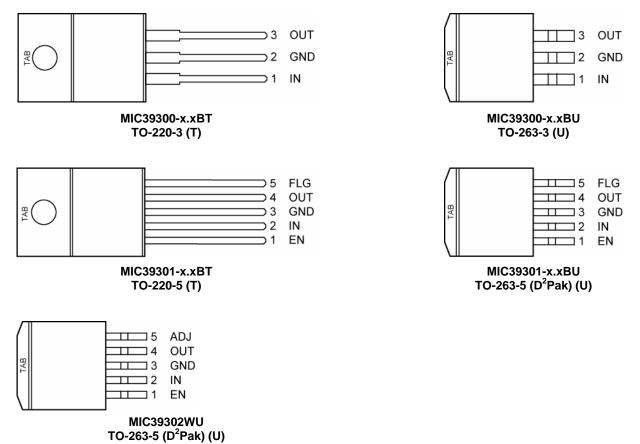
Ordering Information

Part Number		Voltage	Junction Temp Range	Package	
Standard	RoHS Compliant*		Kange		
MIC39300-1.8BT	MIC39300-1.8WT	1.8V	–40°C to +125°C	3-Pin TO-220	
MIC39300-1.8BU	MIC39300-1.8WU	1.8V	–40°C to +125°C	3-Pin TO-263	
MIC39300-2.5BT	MIC39300-2.5WT	2.5V	–40°C to +125°C	3-Pin TO-220	
MIC39300-2.5BU	MIC39300-2.5WU	2.5V	–40°C to +125°C	3-Pin TO-263	
MIC39301-1.8BT	MIC39301-1.8WT	1.8V	–40°C to +125°C	5-Pin TO-220	
MIC39301-1.8BU	MIC39301-1.8WU	1.8V	–40°C to +125°C	5-Pin TO-263	
MIC39301-2.5BT	MIC39301-2.5WT	2.5V	–40°C to +125°C	5-Pin TO-220	
MIC39301-2.5BU	MIC39301-2.5WU	2.5V	–40°C to +125°C	5-Pin TO-263	
	MIC39302WU	Adjustable	–40°C to +125°C	5-Pin TO-263	

Note:

* RoHS compliant with 'high-melting solder' exemption.

Pin Configuration



Pin Description

Pin Number MIC39300	Pin Number MIC39301	Pin Number MIC39302	Pin Name	Pin Function
—	1	1	EN	Enable (Input): TTL/CMOS compatible input. Logic high = enable; logic low or open = shutdown.
1	2	2	IN	Unregulated Input: +16V maximum supply.
2, TAB	3, TAB	3, TAB	GND	Ground: Ground pin and TAB are internally connected.
3	4	4	OUT	Regulator Output
_	5	_	FLG	Error Flag (Ouput): Open-collector indicates an output fault condition. Active low.
_	_	5	ADJ	Adjustable Regulator Feedback Input: Connect to the resistor voltage divider that is placed from OUT to GND in order to set the output voltage.

Absolute Maximum Ratings⁽¹⁾

Supply Voltage (V _{IN})	–20V to +20V
Enable Voltage (V _{EN})	+20V
Storage Temperature (T _S)	–65°C to +150°C Lead
Temperature (soldering, 5 sec.)	
ESD Rating	Note 3

Operating Ratings⁽²⁾

Supply Voltage (V _{IN})	+2.5V to +16V
Enable Voltage (V _{EN})	+16V
Maximum Power Dissipation (P _{D(max)})	Note 4
Junction Temperature (T _J)	
Package Thermal Resistance	
TO-263 (θ _{JC})	2°C/W
TO-220 (θ _{JC})	2°C/W

Electrical Characteristics⁽⁵⁾

 T_J = 25°C, **bold** values indicate –40°C ≤ T_J ≤ +125°C, unless noted.

Symbol	Parameter	Condition	Min	Тур	Max	Units
V _{OUT}	Output Voltage	10mA 10mA $\leq I_{OUT} \leq$ 3A, V_{OUT} + 1V $\leq V_{IN} \leq$ 8V	-1 -2		1 2	% %
	Line Regulation	I_{OUT} = 10mA, V_{OUT} + 1V \leq $V_{IN} \leq$ 8V		0.06	0.5	%
	Load Regulation	$V_{IN} = V_{OUT} + 1V$, $10mA \le I_{OUT} \le 3A$		0.2	1	%
$\Delta V_{OUT} / \Delta T$	Output Voltage Temp. Coefficient			20	100	ppm/°C
	Note 6					
V _{DO}	Dropout Voltage	I_{OUT} = 100mA, ΔV_{OUT} = -1%		65	200	mV
	Note 7, Note 10	I _{OUT} = 750mA, ΔV _{OUT} = -1%		185		mV
		I _{OUT} = 1.5A, ΔV _{OUT} = –1%		250		mV
		$I_{OUT} = 3A, \Delta V_{OUT} = -1\%$		385	550	mV
I _{GND}	Ground Current	I _{OUT} = 750mA, V _{IN} = V _{OUT} + 1V		10	20	mA
	Note 8	I _{OUT} = 1.5A, V _{IN} = V _{OUT} + 1V		17		mA
		$I_{OUT} = 3A, V_{IN} = V_{OUT} + 1V$		45		mA
I _{GND(do)}	Dropout Ground Pin Current	$V_{IN} \le V_{OUT}(nominal) - 0.5V, I_{OUT} = 10mA$		6		mA
I _{OUT(lim)}	Current Limit	V _{OUT} = 0V, VIN = VOUT + 1V		4.5		А
Enable Inp	out (MIC39301)					
V _{EN}	Enable Input Voltage	logic low (off)			0.8	V
		logic high (on)	2.5			V
l _{in}	Enable Input Current	V _{EN} = 2.5V		15	30 75	μA μA
		V _{EN} = 0.8V			2	μA
					4	μA
$I_{OUT(shdn)}$	Shutdown Output Current	Note 9		10	20	μA
Flag Outp	ut (MIC39301)	<u>.</u>				•
$I_{FLG(leak)}$	Output Leakage Current	V _{OH} = 16V		0.01	1 2	μA μA
$V_{FLG(do)}$	Output Low Voltage	V _{IN} = 2.50V, I _{OL} , = 250µA, Note 10		220	300 400	mV mV
V _{FLG}	Low Threshold	% of V _{OUT}	93			%
	High Threshold	% of V _{OUT}			99.2	%
		1		1		1

Hysteresis

%

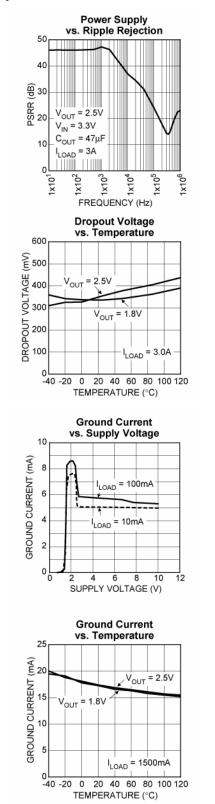
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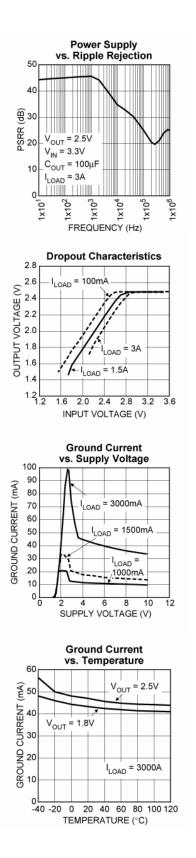
Symbol	Parameter	Condition	Min	Тур	Max	Units	
Reference (Adjust Pin) – MIC39302 only							
V _{ADJ}	Reference Voltage		1.228	1.240	1.252	V	
			1.215		1.265	V	
V _{TC}	Reference Voltage Temp. Coefficient	Note 11		20		ppm/°C	
I _{ADJ}	Adjust Pin Bias Current			40	80	nA	
					120	nA	
I _{TC}	Adjust Pin Bias Current Temp. Coefficient			0.1		nA/°C	

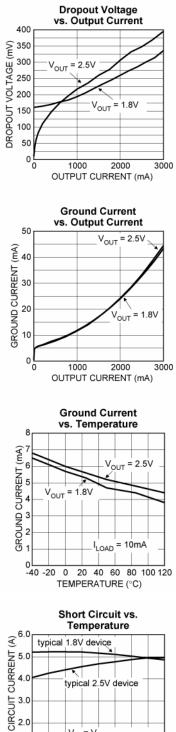
Notes:

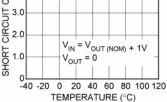
- 1. Exceeding the absolute maximum ratings may damage the device.
- 2. The device is not guaranteed to function outside its operating rating.
- 3. Devices are ESD sensitive. Handling precautions recommended.
- 4. $P_{D(max)} = (T_{J(max)} T_A) \div \theta_{JA}$, where θ_{JA} depends upon the printed circuit layout. See "Applications Information."
- 5. Specification for packaged product only.
- 6. Output voltage temperature coefficient is $\Delta V_{OUT(worst case)} \div (T_{J(max)} T_{J(min)})$ where $T_{J(max)}$ is +125°C and $T_{J(min)}$ is -40°C.
- V_{DO} = V_{IN} − V_{OUT} when V_{OUT} decreases to 99% of its nominal output voltage with V_{IN} = V_{OUT} + 1V. For output voltages below 2.5V, dropout voltage is the input-to-output voltage differential with the minimum input voltage being 2.5V. Minimum input operating voltage is 2.5V.
- 8. I_{GND} is the quiescent current. $I_{IN} = I_{GND} + I_{OUT}$.
- 9. $V_{EN} \le 0.8V$, $V_{IN} \le 8V$, and $V_{OUT} = 0V$
- 10. For a 1.8V device, $V_{IN} = 2.5V$.
- 11. Thermal regulation is defined as the change in output voltage at a time t after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 200mA load pulse at V_{IN} = 8V for t = 10ms.

Typical Characteristics

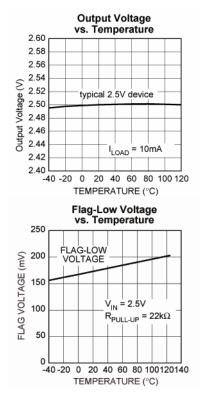


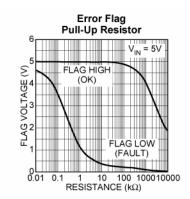


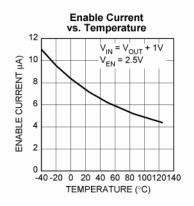




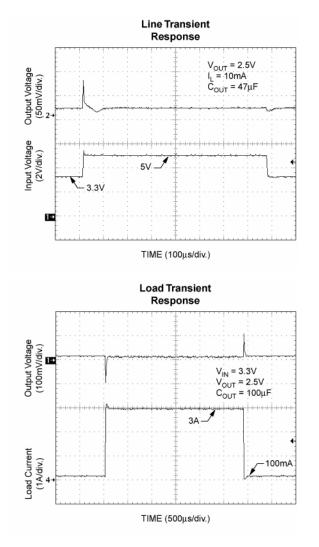
Typical Characteristics (cont)

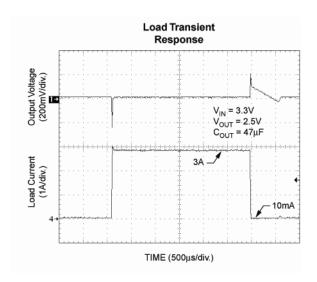




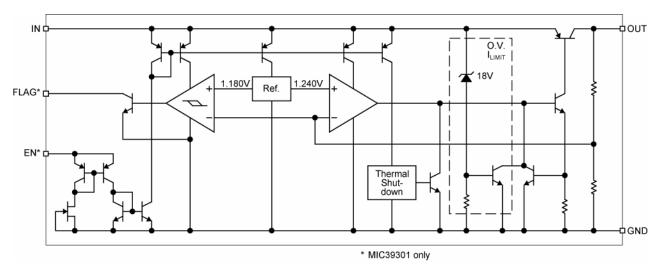


Functional Characteristics





Functional Diagram



Application Information

The MIC39300/1/2 are high-performance, low-dropout voltage regulators suitable for moderate to high-current voltage regulator applications. Its 550mV dropout voltage at full load makes it especially valuable in battery-powered systems and as a high-efficiency noise filter in post-regulator applications. Unlike older NPN-pass transistor designs, where the minimum dropout voltage is limited by the base-to-emitter voltage drop and collector-to-emitter saturation voltage, dropout performance of the PNP output of these devices is limited only by the low V_{CE} saturation voltage.

A trade-off for the low dropout voltage is a varying base drive requirement. Micrel's Super β eta PNP[®] process reduces this drive requirement to only 2% to 5% of the load current.

The MIC39300/1/2 regulators are fully protected from damage due to fault conditions. Current limiting is provided. This limiting is linear; output current during overload conditions is constant. Thermal shutdown disables the device when the die temperature exceeds the maximum safe operating temperature. Transient protection allows device (and load) survival even when the input voltage spikes above and below nominal. The output structure of these regulators allows voltages in excess of the desired output voltage to be applied without reverse current flow.

Thermal Design

Linear regulators are simple to use. The most complicated design parameters to consider are thermal characteristics. Thermal design requires four applicationspecific parameters:

- Maximum ambient temperature (TA)
- Output Current (IOUT)
- Output Voltage (VOUT)
- Input Voltage (VIN)
- Ground Current (IGND)

Calculate the power dissipation of the regulator from these numbers and the device parameters from this datasheet, where the ground current is taken from the data sheet.

 $P_{D} = (V_{IN} - V_{OUT}) I_{OUT} + V_{IN} \times I_{GND}$

The heat sink thermal resistance is determined by:

$$\theta_{SA} = \frac{T_{J(max)} - T_{A}}{P_{D}} - \left(\theta_{JC} + \theta_{CS}\right)$$

where $T_{J(max)} \le 125^{\circ}C$ and θ_{CS} is between 0° and 2°C/W.

The heat sink may be significantly reduced in applications where the minimum input voltage is known and is large compared with the dropout voltage. Use a series input resistor to drop excessive voltage and distribute the heat between this resistor and the regulator. The low dropout properties of Micrel Super β eta PNP[®] regulators allow significant reductions in regulator power dissipation and the associated heat sink without compromising performance. When this technique is employed, a capacitor of at least 1.0µF is needed directly between the input and regulator ground.

Refer to "Application Note 9" for further details and examples on thermal design and heat sink specification.

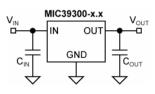


Figure 1. Capacitor Requirements

Output Capacitor

The MIC39300/1/2 requires an output capacitor to maintain stability and improve transient response. Proper capacitor selection is important to ensure proper operation. The MIC39300/1/2 output capacitor selection is dependent upon the ESR (equivalent series resistance) of the output capacitor to maintain stability. When the output capacitor is 47µF or greater, the output capacitor should have less than 1Ω of ESR. This will improve transient response as well as promote stability. Ultralow ESR capacitors, such as ceramic chip capacitors may promote instability. These very low ESR levels may cause an oscillation and/or underdamped transient response. A low-ESR solid tantalum capacitor works extremely well and provides good transient response and stability over temperature. Aluminum electrolytics can also be used, as long as the ESR of the capacitor is $< 1\Omega$.

The value of the output capacitor can be increased without limit. Higher capacitance values help to improve transient response and ripple rejection and reduce output noise.

Input Capacitor

An input capacitor of 1μ F or greater is recommended when the device is more than 4 inches away from the bulk ac supply capacitance, or when the supply is a battery. Small, surface mount, ceramic chip capacitors can be used for the bypassing. Larger values will help to improve ripple rejection by bypassing the input to the regulator, further improving the integrity of the output voltage.

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Transient Response and 3.3V to 2.5V and 2.5V to 1.8V Conversions

The MIC39300/1/2 has excellent transient response to variations in input voltage and load current. The device has been designed to respond quickly to load current variations and input voltage variations. Large output capacitors are not required to obtain this performance. A standard 47μ F output capacitor, preferably tantalum, is all that is required. Larger values help to improve performance even further.

By virtue of its low-dropout voltage, this device does not saturate into dropout as readily as similar NPN-based designs. When converting from 3.3V to 2.5V or 2.5V to 1.8V, the NPN-based regulators are already operating in dropout, with typical dropout requirements of 1.2V or greater. To convert down to 2.5V without operating in dropout, NPN-based regulators require an input voltage of 3.7V at the very least. The MIC39300/1 regulator will provide excellent performance with an input as low as 3.0V or 2.5V. This gives the PNP-based regulators a distinct advantage over older, NPN-based linear regulators.

Minimum Load Current

The MIC39300/1/2 regulators are specified between finite loads. If the output current is too small, leakage currents dominate and the output voltage rises. A 10mA minimum load current is necessary for proper regulation.

Error Flag

The MIC39301 version features an error flag circuit which monitors the output voltage and signals an error condition when the voltage drops 5% below the nominal output voltage. The error flag is an open-collector output that can sink 10mA during a fault condition.

Low output voltage can be caused by a number of problems, including an overcurrent fault (device in current limit) or low input voltage. The flag is inoperative during overtemperature shutdown.

When the error flag is not used, it is best to leave it open. A pull-up resistor from FLG to either V_{IN} or V_{OUT} is required for proper operation.

Enable Input

The MIC39301/2 features an enable input for on/off control of the device. The enable input's shutdown state draws "zero" current (only microamperes of leakage). The enable input is TTL/CMOS compatible for simple logic interface, but can be connected to up to 20V. When enabled, it draws approximately 15µA.

Adjustable Regulator Design

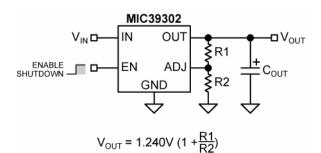


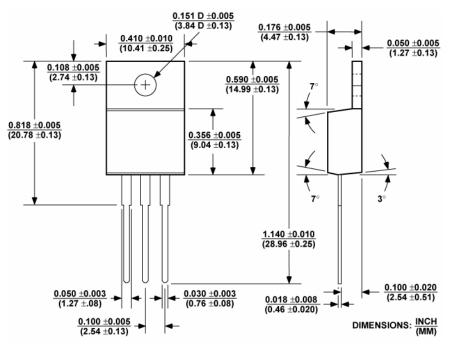
Figure 2. Adjustable Regulator with Resistors

The MIC39302 allows programming the output voltage anywhere between 1.24V and 15.5V. Two resistors are used. The resistor values are calculated by:

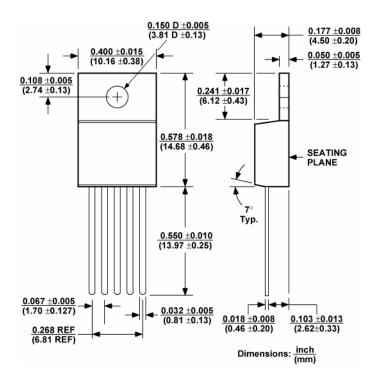
$$R1 = R2 \times \left(\frac{V_{OUT}}{1.240} - 1\right)$$

Where V_{OUT} is the desired output voltage. Figure 2 shows component definition. Applications with widely varying load currents may scale the resistors to draw the minimum load current required for proper operation (see *Minimum Load Current* section).

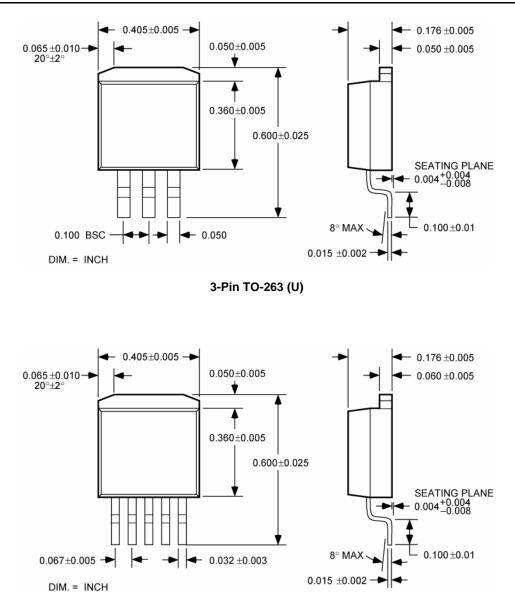
Package Information



3-Pin TO-220 (T)



5-Pin TO-220 (T)



5-Pin TO-263 (U)

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