



ON Semiconductor®

## FMS6143

# Low-Cost Three-Channel 4th-Order Standard Definition Video Filter Driver

## Features

- Three 4th-order 8MHz (SD) filters
- Drives single, AC- or DC-coupled, video loads ( $2V_{pp}$ ,  $150\Omega$ )
- Drives dual, AC- or DC-coupled, video loads ( $2V_{pp}$ ,  $75\Omega$ )
- Transparent input clamping
- AC- or DC-coupled inputs
- AC- or DC-coupled outputs
- DC-coupled outputs eliminate AC-coupling capacitors
- 5V only
- Robust 8kV ESD protection
- Lead-free SOIC-8 package

## Applications

- Cable set-top boxes
- Satellite set-top boxes
- DVD players
- HDTV
- Personal Video Recorders (PVR)
- Video On Demand (VOD)

## Description

The FMS6143 Low-Cost Video Filter (LCVF) is intended to replace passive LC filters and drivers with a low-cost integrated device. Three 4th-order filters provide improved image quality compared to typical 2nd or 3rd-order passive solutions.

The FMS6143 may be directly driven by a DC-coupled DAC output or an AC-coupled signal. Internal diode clamps and bias circuitry may be used if AC-coupled inputs are required (see *Applications* section for details).

The outputs can drive AC- or DC-coupled single ( $150\Omega$ ) or dual ( $75\Omega$ ) loads. DC-coupling the outputs removes the need for output coupling capacitors. The input DC-levels are offset approximately +280mV at the output (see the *Applications* section for details).

## Related Application Notes

AN-8002 — : <http://www.onsemi.com/pub/Collateral/AN-8002.PDF>

AN-6024 — <http://www.onsemi.com/pub/Collateral/AN-6024.pdf.pdf>

AN-6041 — <http://www.onsemi.com/pub/Collateral/AN-6041.pdf.pdf>

## Functional Block Diagram

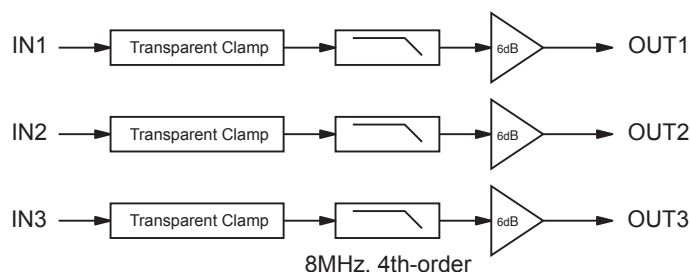


Figure 1. AC-Coupled Inputs and Outputs

## Ordering Information

Part Number	Operating Temperature Range	Eco Status	Package	Packaging Method
FMS6143CSX	-40 to +85°C	RoHS	SOIC-8	Tape and Reel

## Pin Configuration

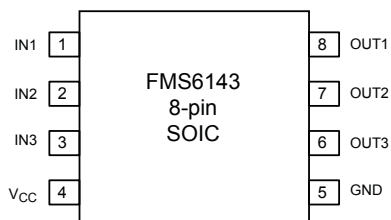


Figure 2. AC-Coupled Inputs and Outputs

## Pin Assignments

Pin #	Name	Type	Description
1	IN1	Input	Video input, Channel 1
2	IN2	Input	Video input, Channel 2
3	IN3	Input	Video input, Channel 3
4	V <sub>CC</sub>	Input	+5V supply, do not float
5	GND	Output	Must be tied to ground, do not float
6	OUT3	Output	Filtered output, Channel 3
7	OUT2	Output	Filtered output, Channel 2
8	OUT1	Output	Filtered output, Channel 1

## Absolute Maximum Ratings

The “Absolute Maximum Ratings” are those values beyond which the safety of the device cannot be guaranteed. The device should not be operated at these limits. The parametric values defined in the Electrical Characteristics tables are not guaranteed at the absolute maximum ratings. The “Recommended Operating Conditions” table defines the conditions for actual device operation.

Symbol	Parameter	Min.	Max.	Unit
$V_{CC}$	DC Supply Voltage	-0.3	6.0	V
	Analog and Digital I/O	-0.3	$V_{CC} + 0.3$	V
	Output Channel - Any One Channel (Do Not Exceed)		50	mA

### Note:

- Functional operation under any of these conditions is NOT implied. Performance and reliability are guaranteed only if operating conditions are not exceeded.

## Reliability Information

Symbols	Parameter	Min.	Typ.	Max.	Unit
$T_J$	Junction Temperature			150	°C
$T_{STG}$	Storage Temperature Range	-65		+150	°C
$T_L$	Lead Temperature (Soldering, 10s)			+300	°C
$\theta_{JA}$	Thermal Resistance, JEDEC Standard Multi-layer Test Boards, Still Air		115		°C/W

## Electrostatic Discharge Information

Symbols	Parameter	Max.	Unit
ESD	Human Body Model, JESD22-A114	6.5	kV
	Charged Device Model, JESD22-C101	2.0	

Symbols	Parameter	Min.	Typ.	Max.	Unit
$T_A$	Operating Temperature Range	-40		+85	°C
$V_{CC}$	$V_{CC}$ Range	4.75	5.00	5.25	V

## DC Electrical Characteristics

$T_A = 25^\circ\text{C}$ ,  $V_{CC} = 5\text{V}$ ,  $R_{SOURCE} = 37.5\Omega$ ; all inputs are AC-coupled with  $0.1\mu\text{F}$ ; all outputs are AC-coupled with  $220\mu\text{F}$  into  $150\Omega$  loads; unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$I_{CC}$	Supply Current <sup>(1)</sup>	FMS6143 (No Load)		19	27	mA
$V_{IN}$	Video Input Voltage Range	Referenced to GND if DC-coupled		1.4		$V_{pp}$
PSRR	Power Supply Rejection	DC (All Channels)		-50		dB

## AC Electrical Characteristics

$T_A = 25^\circ\text{C}$ ,  $V_{IN} = 1V_{pp}$ ,  $V_{CC} = 5\text{V}$ ,  $R_{SOURCE} = 37.5\Omega$ ; all inputs are AC-coupled with  $0.1\mu\text{F}$ ; all outputs are AC-coupled with  $220\mu\text{F}$  into  $150\Omega$  loads; unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
AV	Channel Gain <sup>(1)</sup>	All Channels	6.0	6.2	6.4	dB
$f_{1dB}$	-1dB Bandwidth <sup>(1)</sup>	All Channels	5.6	6.5		MHz
$f_C$	-3dB Bandwidth	All Channels		7.7		MHz
$f_{SB}$	Attenuation (Stopband Reject)	All Channels at $f = 27\text{MHz}$		48		dB
DG	Differential Gain	All Channels		0.3		%
DP	Differential Phase	All Channels		0.6		°
THD	Output Distortion (All Channels)	$V_{OUT} = 1.8V_{pp}$ , 1MHz		0.4		%
$X_{TALK}$	Crosstalk (Channel-to-Channel)	at 1MHz		-60		dB
SNR	Signal-to-Noise Ratio	All Channels, NTC-7 Weighting: 100kHz to 4.2MHz		75		dB
$t_{pd}$	Propagation Delay	Delay from Input-to-Output, 4.5MHz		59		ns

### Note:

1. 100% tested at  $25^\circ\text{C}$ .

## Typical Performance Characteristics

$T_A = 25^\circ\text{C}$ ,  $V_{CC} = 5\text{V}$ ,  $R_{SOURCE} = 37.5\Omega$ ; all inputs AC-coupled with  $0.1\mu\text{F}$ ; all outputs are AC-coupled with  $220\mu\text{F}$  into  $150\Omega$  loads; unless otherwise noted.

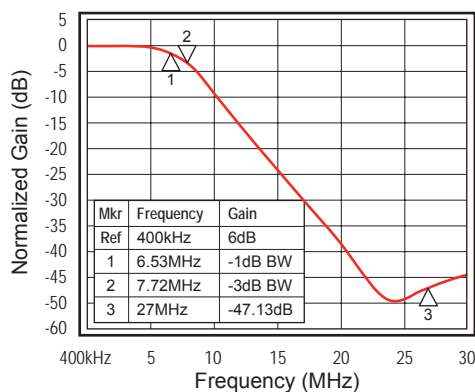


Figure 3. Frequency Response

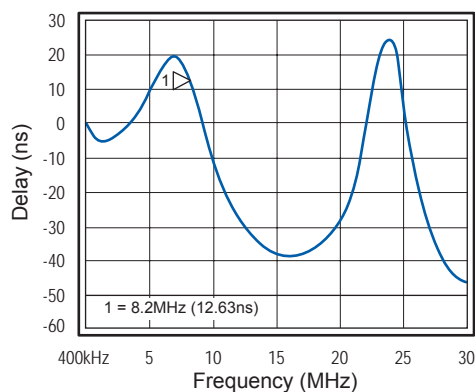


Figure 4. Group Delay vs. Frequency

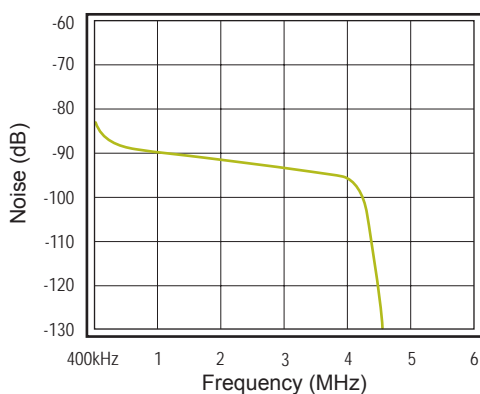


Figure 5. Noise vs. Frequency

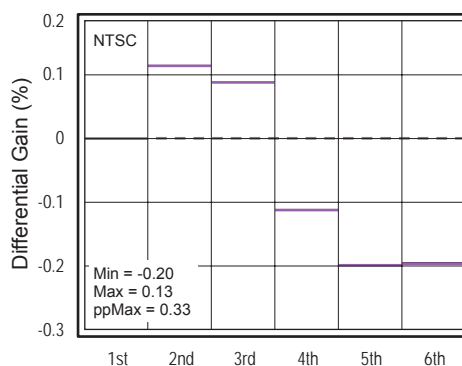


Figure 6. Differential Gain

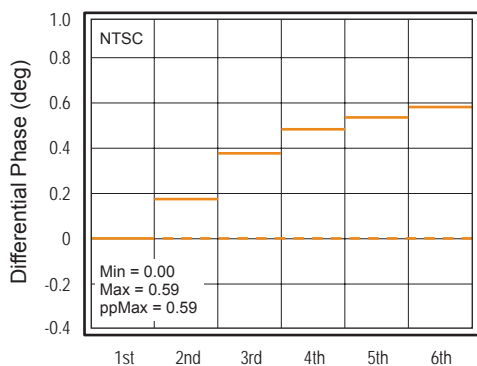


Figure 7. Differential Phase

## Typical Application Diagrams

The following circuit may be used for direct DC-coupled drive by DACs with an output voltage range of 0V to 1.4V. AC-coupled or DC-coupled outputs may be used with AC-coupled outputs, offering slightly lower power dissipation.

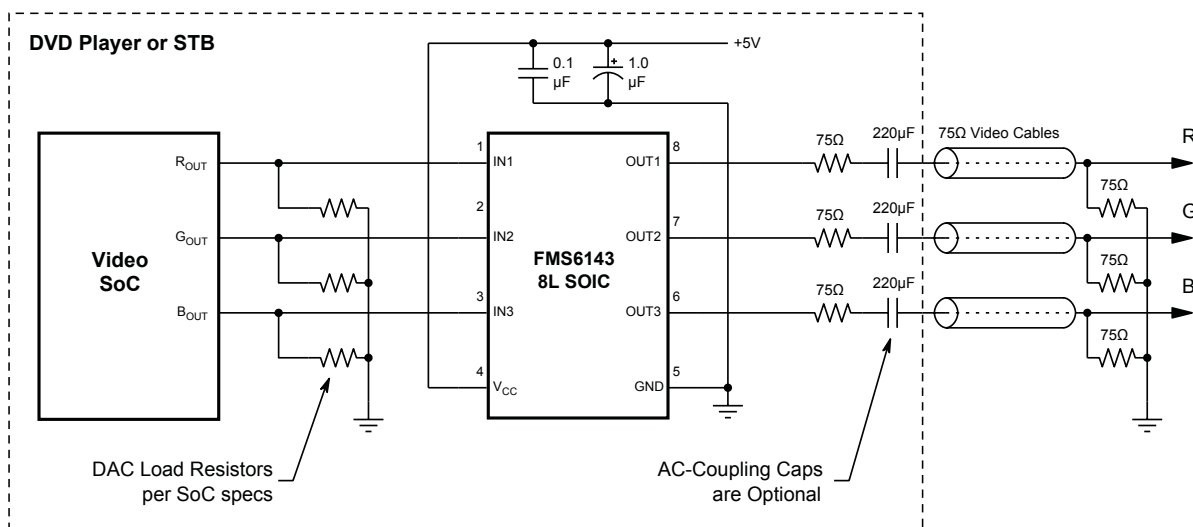
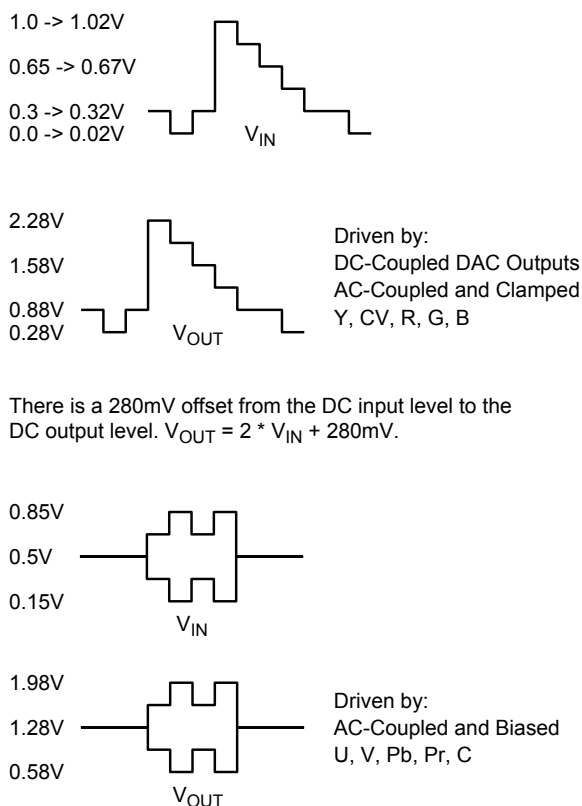


Figure 8. Typical Application Diagram

## Application Information

### Application Circuits

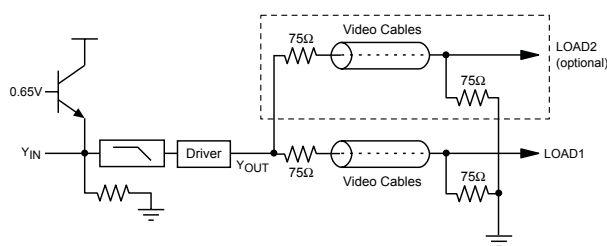
The FMS6143 Low Cost Video Filter (LCVF) provides 6dB gain from input to output. In addition, the input is slightly offset to optimize the output driver performance. The offset is held to the minimum required value to decrease the standing DC current into the load. Typical voltage levels are shown in the diagram below:



**Figure 9. Typical Voltage Levels**

The FMS6143 provides an internal diode clamp to support AC-coupled input signals. If the input signal does not go below ground, the input clamp does not operate. This allows DAC outputs to directly drive the FMS6143 without an AC coupling capacitor. When the input is AC-coupled, the diode clamp sets the sync tip (or lowest voltage) just below ground. The worst-case sync tip compression due to the clamp can not exceed 7mV. The input level set by the clamp, combined with the internal DC offset, keeps the output within its acceptable range.

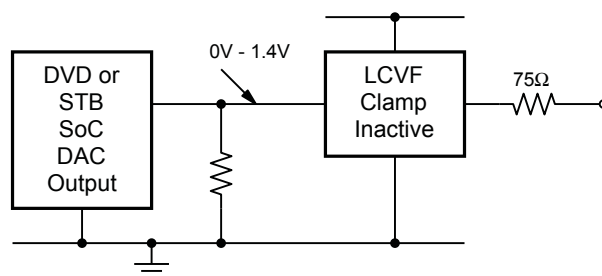
For symmetric signals like Chroma, U, V, Pb, and Pr, the average DC bias is fairly constant and the inputs can be AC-coupled with the addition of a pull-up resistor to set the DC input voltage. DAC outputs can also drive these same signals without the AC coupling capacitor. A conceptual illustration of the input clamp circuit is shown in Figure 10:



**Figure 10. Input Clamp Circuit**

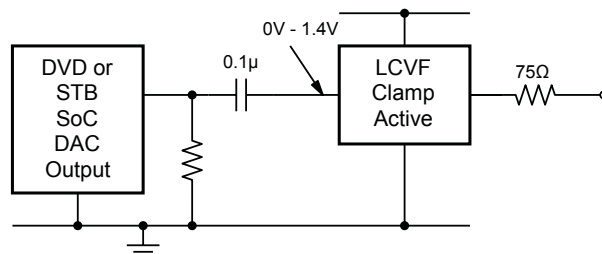
### I/O Configurations

For a DC-coupled DAC drive with DC-coupled outputs, use this configuration:



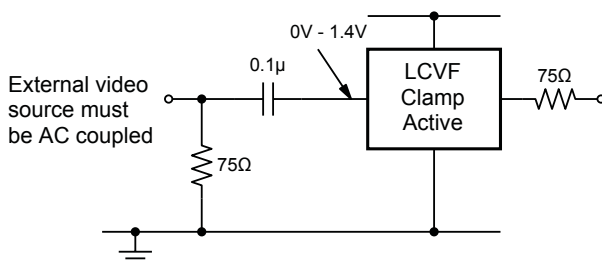
**Figure 11. DC-Coupled Inputs and Outputs**

Alternatively, if the DAC's average DC output level causes the signal to exceed the range of 0V to 1.4V, it can be AC coupled as follows:



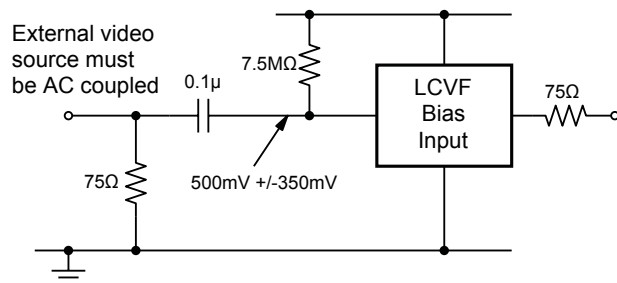
**Figure 12. AC-Coupled Inputs, DC-Coupled Outputs**

When the FMS6143 is driven by an unknown external source or a SCART switch with its own clamping circuitry, the inputs should be AC coupled like this:



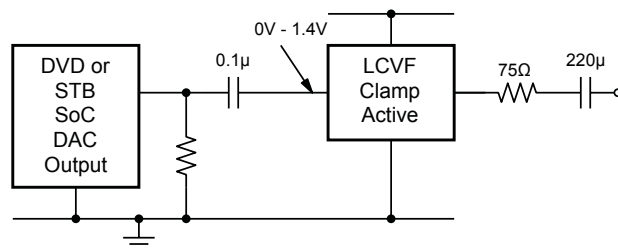
**Figure 13. SCART with DC-Coupled Outputs**

The same method can be used for biased signals, with the addition of a pull-up resistor to make sure the clamp never operates. The internal pull-down resistance is  $800\text{k}\Omega \pm 20\%$ , so the external resistance should be  $7.5\text{M}\Omega$  to set the DC level to  $500\text{mV}$ :

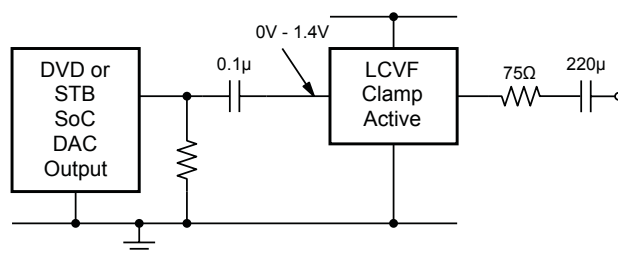


**Figure 14. Biased SCART with DC-Coupled Outputs**

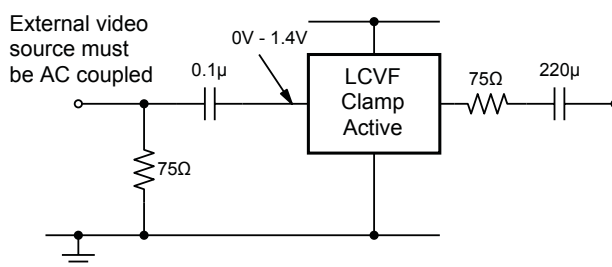
The same circuits can be used with AC-coupled outputs if desired.



**Figure 15. DC-Coupled Inputs, AC-Coupled Outputs**



**Figure 16. AC-Coupled Inputs and Outputs**



**Figure 17. Biased SCART with AC-Coupled Outputs**

**NOTE:** The video tilt or line time distortion is dominated by the AC-coupling capacitor. The value may need to be increased beyond  $220\mu\text{F}$  to obtain satisfactory operation in some applications.

### Power Dissipation

The FMS6143 output drive configuration must be considered when calculating overall power dissipation. Care must be taken not to exceed the maximum die junction temperature. The following example can be used to calculate the power dissipation and internal temperature rise.

$$T_J = T_A + P_d \cdot \theta_{JA} \quad (1)$$

$$\text{where: } P_d = P_{CH1} + P_{CH2} + P_{CH3} \text{ and} \quad (2)$$

$$P_{CHx} = V_{CC} \cdot I_{CH} - (V_O^2 / R_L) \quad (3)$$

$$\text{where: } V_O = 2V_{IN} + 0.280\text{V} \quad (4)$$

$$I_{CH} = (I_{CC}/3) + (V_O / R_L) \quad (5)$$

$V_{IN}$  = RMS value of input signal

$I_{CC} = 19\text{mA}$

$V_{CC} = 5\text{V}$

$R_L$  = channel load resistance

Board layout can also affect thermal characteristics. Refer to the *Layout Considerations* section for details.

The FMS6143 is specified to operate with output currents typically less than  $50\text{mA}$ , more than sufficient for a dual ( $75\Omega$ ) video load. Internal amplifiers are current limited to a maximum of  $100\text{mA}$  and should withstand brief-duration short-circuit conditions. This capability is not guaranteed.



## Layout Considerations

General layout and supply bypassing play a major role in high-frequency performance and thermal characteristics. Fairchild offers a demonstration board to guide layout and aid device evaluation. The demo board is a four-layer board with full power and ground planes. Following this layout configuration provides optimum performance and thermal characteristics for the device. For the best results, follow the steps and recommended routing rules listed below.

### Recommended Routing/Layout Rules

- Use separate analog and digital power planes to supply power.
- Traces should run on top of the ground plane at all times.
- No trace should run over ground/power splits.
- Avoid routing at 90-degree angles.
- Minimize clock and video data trace length differences.
- Include 10 $\mu$ F and 0.1 $\mu$ F ceramic power supply bypass capacitors.
- Place the 0.1 $\mu$ F capacitor within 0.1 inches of the device power pin.
- Place the 10 $\mu$ F capacitor within 0.75 inches of the device power pin.
- For multi-layer boards, use a large ground plane to help dissipate heat.
- For two-layer boards, use a ground plane that extends beyond the device body at least 0.5 inches on all sides. Include a metal paddle under the device on the top layer.
- Minimize all trace lengths to reduce series inductance.

### Thermal Considerations

Since the interior of most systems, such as set-top boxes, TVs, and DVD players, are at +70°C; consideration must be given to providing an adequate heat sink for the device package for maximum heat dissipation. When designing a system board, determine how much power each device dissipates. Ensure that devices of high power are not placed in the same location, such as directly above (top plane) or below (bottom plane), each other on the PCB.

### PCB Thermal Layout Considerations

- Understand the system power requirements and environmental conditions.
- Maximize thermal performance of the PCB.

- Consider using 70 $\mu$ m of copper for high-power designs.
- Make the PCB as thin as possible by reducing FR4 thickness.
- Use vias in power pad to tie adjacent layers together.
- Remember that baseline temperature is a function of board area, not copper thickness.
- Modeling techniques provide a first-order approximation.

### Output Considerations

The FMS6143 outputs are DC offset from the input by 150mV. Therefore,  $V_{OUT} = 2 \cdot V_{IN} DC + 150mV$ . This offset is required to obtain optimal performance from the output driver and is held at the minimum value to decrease the standing DC current into the load. Since the FMS6143 has a 2x (6dB) gain, the output is typically connected via a 75 $\Omega$ -series back-matching resistor, followed by the 75 $\Omega$  video cable. Due to the inherent divide by two of this configuration, the blanking level at the load of the video signal is always less than 1V. When AC-coupling the output, ensure that the coupling capacitor of choice passes the lowest frequency content in the video signal and that line time distortion (video tilt) is kept as low as possible.

The selection of the coupling capacitor is a function of the subsequent circuit input impedance and the leakage current of the input being driven. To obtain the highest quality output video signal, the series termination resistor must be placed as close to the output pin as possible. This reduces the parasitic capacitance and inductance effect on the output driver. The distance from the device pin to the series termination resistor should be no greater than 0.1 inches.



Figure 18. Distance from Device Pin to Series Termination Resistor

## Physical Dimensions

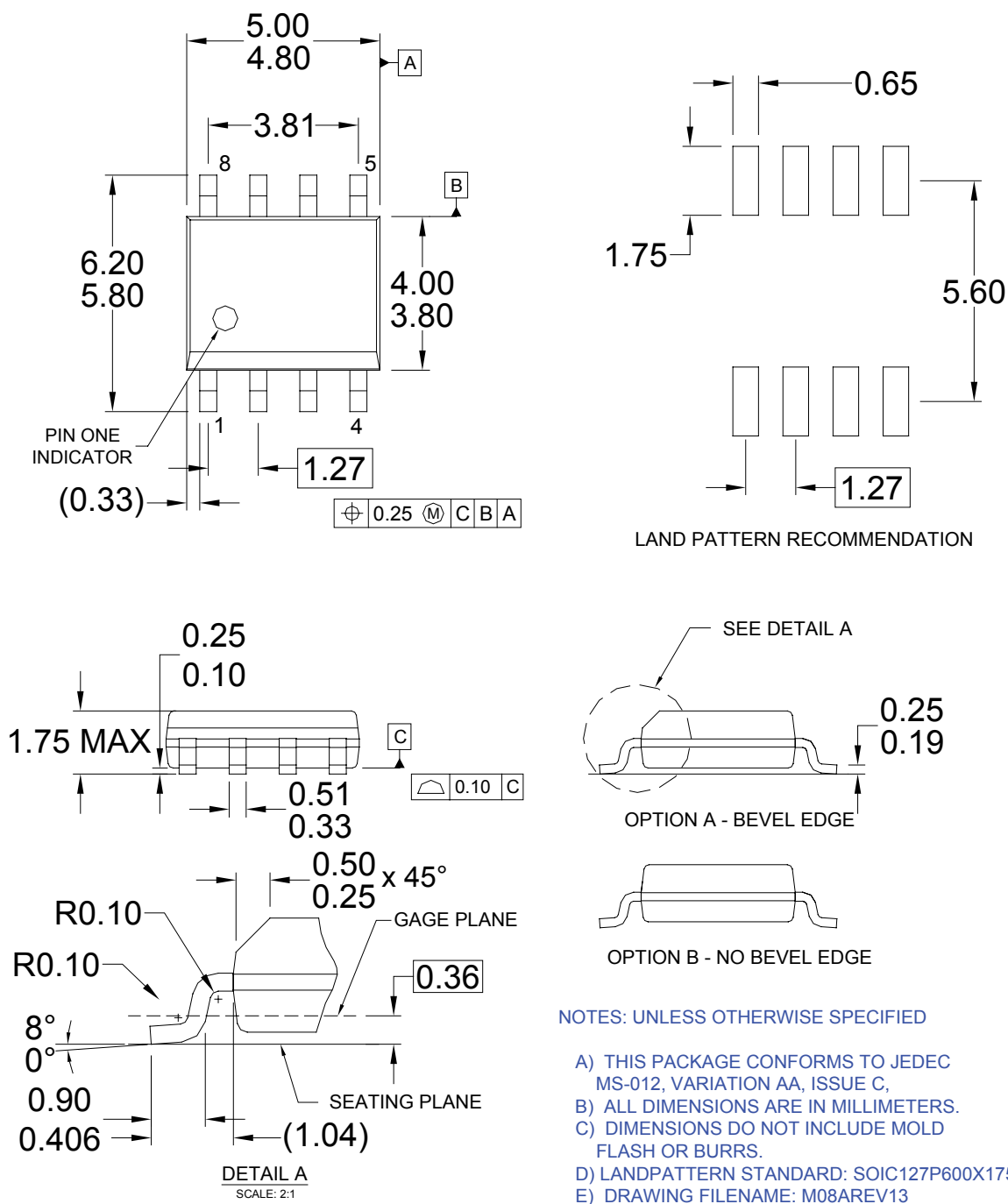


Figure 19. SOIC-8 Package

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