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REVISION HISTORY

10/12—Rev. 0 to Rev. A

Changes to Figure 1	1
Changes to Figure 10	7
Updated Outline Dimensions	10

4/11—Revision 0: Initial Version

SPECIFICATIONS

All minimum and maximum specifications are guaranteed. Typical specifications are not guaranteed. $T_A = 25^\circ\text{C}$, $V_S = AV_{CC} = V_{DD} = 5\text{ V}$, $V_{\text{RATIO}} = AV_{CC}$, angular rate = $0^\circ/\text{sec}$, bandwidth = 80 Hz ($C_{\text{OUT}} = 0.01\text{ }\mu\text{F}$), $I_{\text{OUT}} = 100\text{ }\mu\text{A}$, $\pm 1\text{ g}$, unless otherwise noted.

Table 1.

Parameter	Conditions	Min	Typ	Max	Unit
SENSITIVITY ¹	Clockwise rotation is positive output				
Measurement Range ²	Full-scale range over specifications range	± 250	± 300		$^\circ/\text{sec}$
Initial and Over Temperature	-40°C to $+105^\circ\text{C}$		7.0		$\text{mV}/^\circ/\text{sec}$
Temperature Drift ³			± 2		%
Nonlinearity	Best fit straight line		0.01		% of FS
NULL ¹					
Null	-40°C to $+105^\circ\text{C}$		2.5		V
Calibrated Null ⁴	-40°C to $+105^\circ\text{C}$		± 0.1		$^\circ/\text{sec}$
Temperature Drift	-40°C to $+105^\circ\text{C}$		± 1		$^\circ/\text{sec}$
Linear Acceleration Effect	Any axis		0.03		$^\circ/\text{sec}/\text{g}$
Vibration Rectification	25 g rms, 50 Hz to 5 kHz		0.0002		$^\circ/\text{s}/\text{g}^2$
NOISE PERFORMANCE					
Rate Noise Density	$T_A \leq 25^\circ\text{C}$		0.02		$^\circ/\text{sec}/\sqrt{\text{Hz}}$
Resolution Floor	$T_A = 25^\circ\text{C}$ 1 minute to 1 hour in-run		20		$^\circ/\text{hr}$
FREQUENCY RESPONSE					
Bandwidth ⁵	$\pm 3\text{ dB}$ user adjustable up to specification		2000		Hz
Sensor Resonant Frequency		15	17	19	kHz
SELF-TEST ¹					
ST1 RATEOUT Response	ST1 pin from Logic 0 to Logic 1		-45		$^\circ/\text{sec}$
ST2 RATEOUT Response	ST2 pin from Logic 0 to Logic 1		45		$^\circ/\text{sec}$
ST1 to ST2 Mismatch ⁶		-5	± 2	+5	%
Logic 1 Input Voltage		3.3			V
Logic 0 Input Voltage				1.7	V
Input Impedance	To common	40	50	100	k Ω
TEMPERATURE SENSOR ¹					
V_{OUT} at 25°C	Load = 10 M Ω	2.35	2.5	2.65	V
Scale Factor ⁷	25°C , $V_{\text{RATIO}} = 5\text{ V}$		9		$\text{mV}/^\circ\text{C}$
Load to V_S			25		k Ω
Load to Common			25		k Ω
TURN-ON TIME ⁴	Power on to $\pm 0.5^\circ/\text{sec}$ of final with CP5 = 100 nF			50	ms
OUTPUT DRIVE CAPABILITY					
Current Drive	For rated specifications			200	μA
Capacitive Load Drive				1000	pF
POWER SUPPLY					
Operating Voltage (V_S)		4.75	5.00	5.25	V
Quiescent Supply Current			3.5	4.5	mA
TEMPERATURE RANGE					
Specified Performance		-40		+105	$^\circ\text{C}$

¹ Parameter is linearly ratiometric with V_{RATIO} .

² Measurement range is the maximum range possible, including output swing range, initial offset, sensitivity, offset drift, and sensitivity drift at 5 V supplies.

³ From $+25^\circ\text{C}$ to -40°C or $+25^\circ\text{C}$ to $+105^\circ\text{C}$.

⁴ Based on characterization.

⁵ Adjusted by external capacitor, C_{OUT} . Reducing bandwidth below 0.01 Hz does not result in further noise improvement.

⁶ Self-test mismatch is described as $(\text{ST2} + \text{ST1})/((\text{ST2} - \text{ST1})/2)$.

⁷ Scale factor for a change in temperature from 25°C to 26°C . V_{TEMP} is ratiometric to V_{RATIO} . See the Temperature Output and Calibration section for more information.

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Acceleration (Any Axis, 0.5 ms)	
Unpowered	10,000 g
Powered	10,000 g
V _{DD} , AV _{CC}	−0.3 V to +6.0 V
V _{RATIO}	AV _{CC}
ST1, ST2	AV _{CC}
Output Short-Circuit Duration (Any Pin to Common)	Indefinite
Operating Temperature Range	−55°C to +125°C
Storage Temperature Range	−65°C to +150°C

Stresses above those listed under the Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Drops onto hard surfaces can cause shocks of greater than 10,000 g and can exceed the absolute maximum rating of the device. Care should be exercised in handling to avoid damage.

RATE SENSITIVE AXIS

This is a z-axis rate-sensing device (also called a yaw rate-sensing device). It produces a positive going output voltage for clockwise rotation about the axis normal to the package top, that is, clockwise when looking down at the package lid.

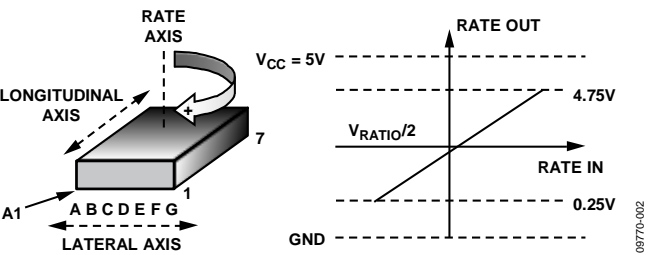


Figure 2. RATEOUT Signal Increases with Clockwise Rotation

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

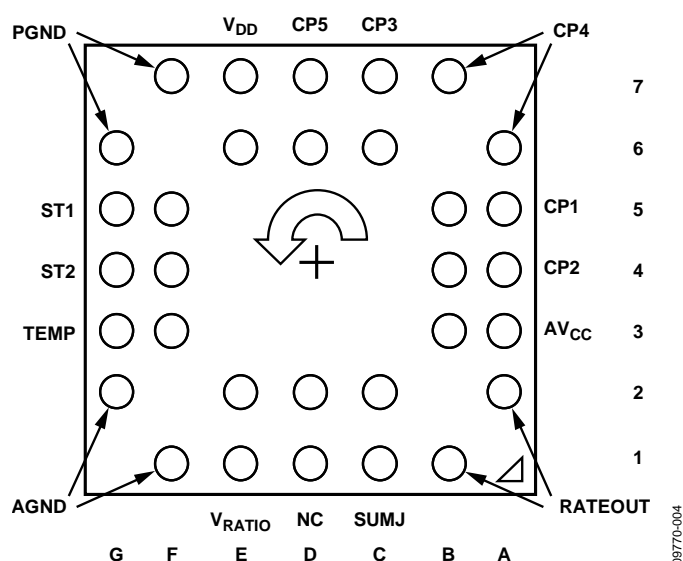


Figure 3. Pin Configuration

Table 3. Pin Function Descriptions

Pin No.	Mnemonic	Description
6D, 7D	CP5	HV Filter Capacitor, 100 nF.
6A, 7B	CP4	Charge Pump Capacitor, 22 nF.
6C, 7C	CP3	Charge Pump Capacitor, 22 nF.
5A, 5B	CP1	Charge Pump Capacitor, 22 nF.
4A, 4B	CP2	Charge Pump Capacitor, 22 nF.
3A, 3B	AV _{CC}	Positive Analog Supply.
1B, 2A	RATEOUT	Rate Signal Output.
1C, 2C	SUMJ	Output Amp Summing Junction.
1D, 2D	NC	No Connection. Do not connect to these pins.
1E, 2E	V _{RATIO}	Reference Supply for Ratiometric Output.
1F, 2G	AGND	Analog Supply Return.
3F, 3G	TEMP	Temperature Voltage Output.
4F, 4G	ST2	Self-Test for Sensor 2.
5F, 5G	ST1	Self-Test for Sensor 1.
6G, 7F	PGND	Charge Pump Supply Return.
6E, 7E	V _{DD}	Positive Charge Pump Supply.

TYPICAL PERFORMANCE CHARACTERISTICS

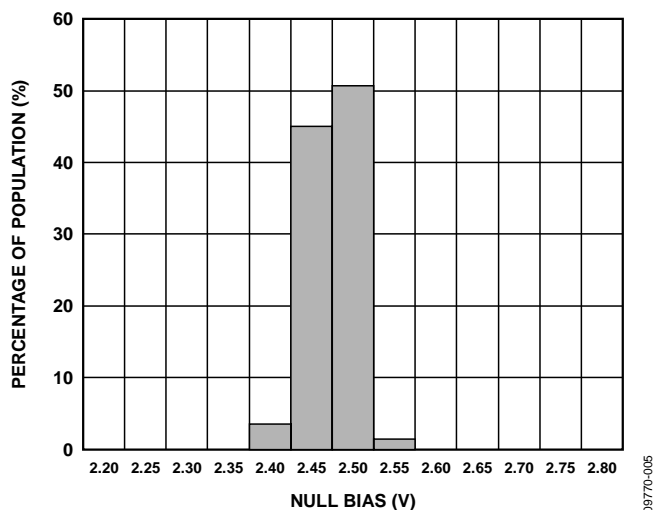


Figure 4. Null Bias at 25°C

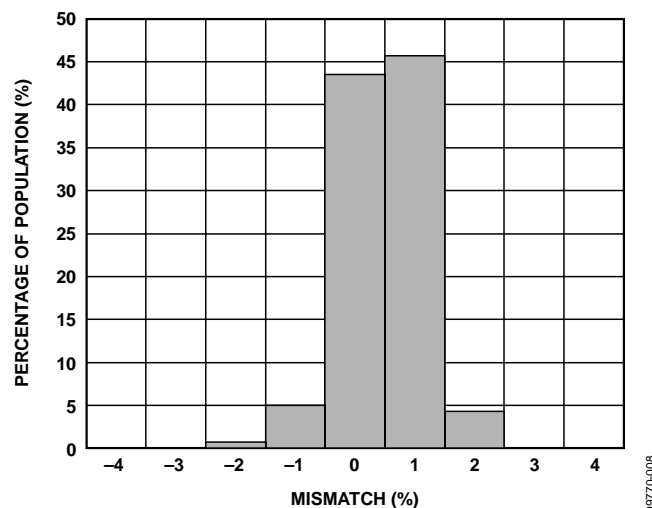


Figure 7. Self-Test Mismatch at 25°C ($V_{RATIO} = 5 V$)

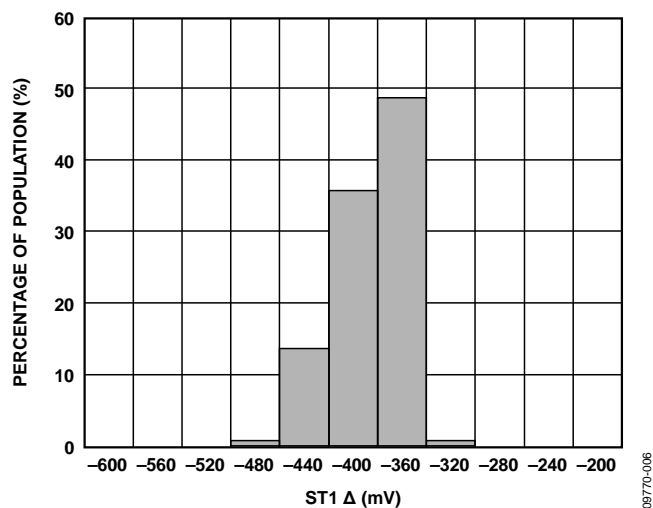


Figure 5. ST1 Output Change at 25°C ($V_{RATIO} = 5 V$)

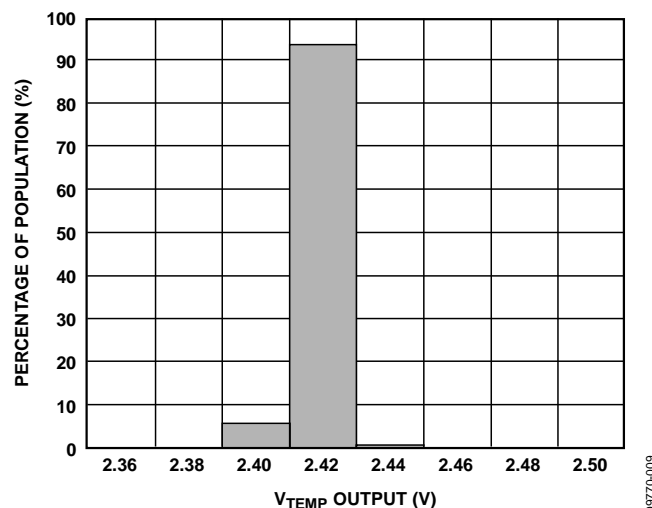


Figure 8. V_{TEMP} Output at 25°C ($V_{RATIO} = 5 V$)

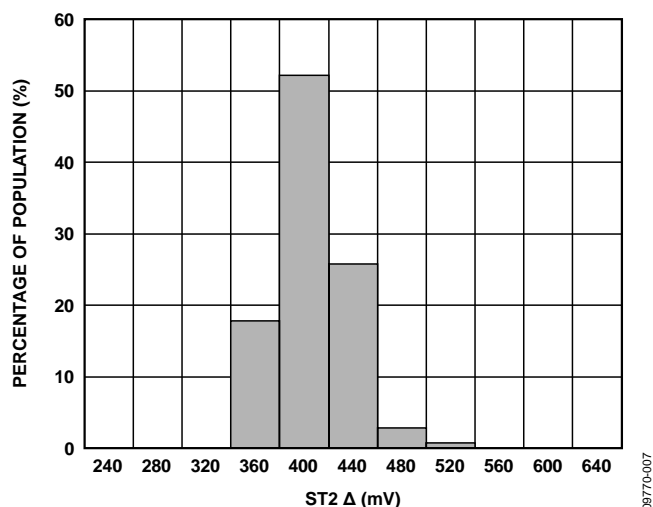


Figure 6. ST2 Output Change at 25°C ($V_{RATIO} = 5 V$)

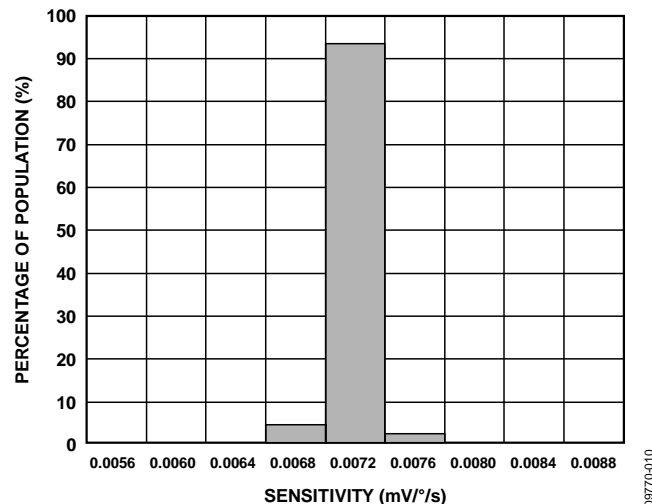


Figure 9. Sensitivity at 25°C

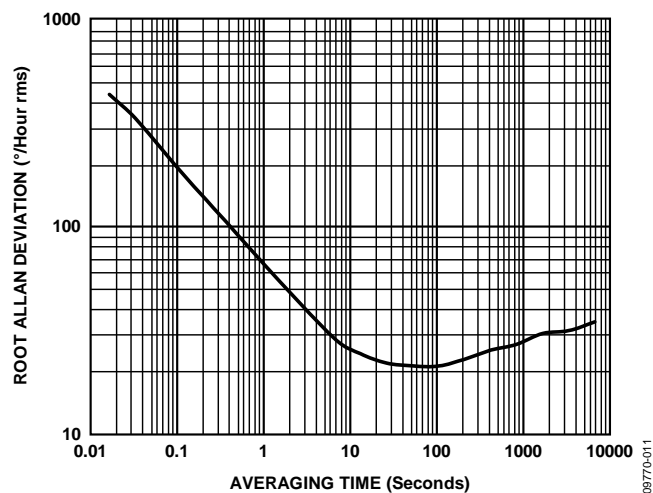


Figure 10. Typical Root Allan Deviation at 25°C vs. Averaging Time

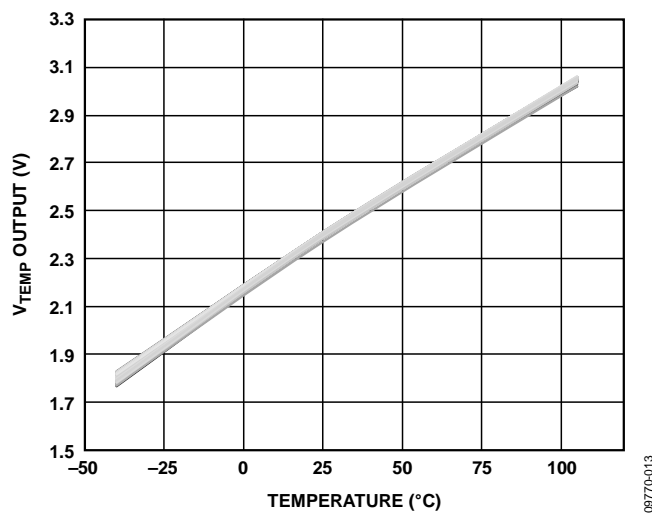


Figure 12. V_{TEMP} Output over Temperature, 256 Parts ($V_{RATIO} = 5$ V)

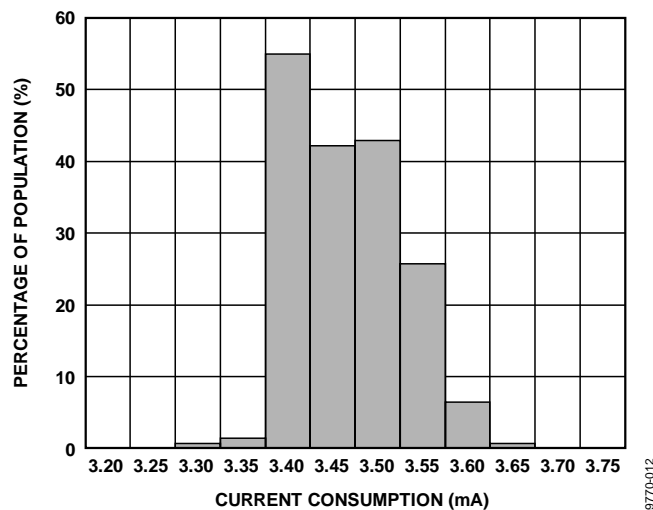


Figure 11. Current Consumption at 25°C ($V_{RATIO} = 5$ V)

THEORY OF OPERATION

The ADXR642 operates on the principle of a resonator gyro. Figure 13 shows a simplified version of one of four polysilicon sensing structures. Each sensing structure contains a dither frame that is electrostatically driven to resonance. This produces the necessary velocity element to produce a Coriolis force when experiencing angular rate. The ADXR642 is designed to sense a z-axis (yaw) angular rate.

When the sensing structure is exposed to angular rate, the resulting Coriolis force couples into an outer sense frame, which contains movable fingers that are placed between fixed pickoff fingers. This forms a capacitive pickoff structure that senses Coriolis motion. The resulting signal is fed to a series of gain and demodulation stages that produce the electrical rate signal output. The quad sensor design rejects linear and angular acceleration, including external *g*-forces and vibration. This is achieved by mechanically coupling the four sensing structures such that external *g*-forces appear as common-mode signals that can be removed by the fully differential architecture implemented in the ADXR642.

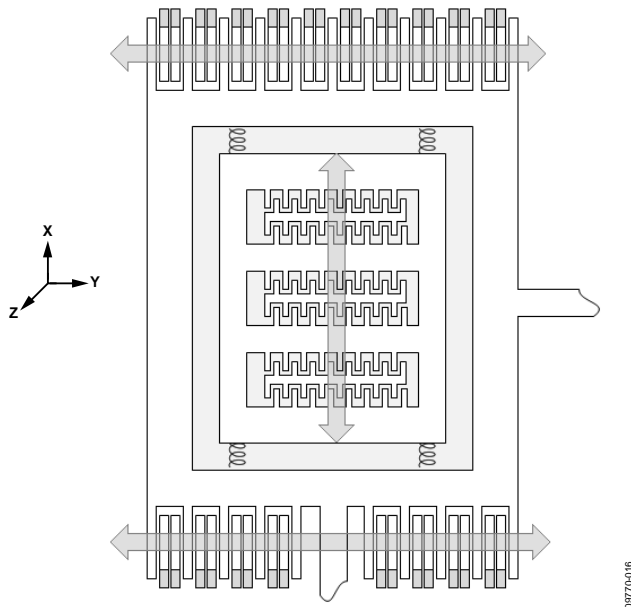


Figure 13. Simplified Gyro Sensing Structure—One Corner

The electrostatic resonator requires 18 V to 20 V for operation. Because only 5 V are typically available in most applications, a charge pump is included on chip. If an external 18 V to 20 V supply is available, the two capacitors on CP1 to CP4 can be omitted, and this supply can be connected to CP5 (Pin 6D, Pin 7D). CP5 should not be grounded when power is applied to the ADXR642. No damage occurs, but under certain conditions, the charge pump may fail to start up after the ground is removed without first removing power from the ADXR642.

SETTING BANDWIDTH

The external capacitor, C_{OUT} , is used in combination with the on-chip resistor, R_{OUT} , to create a low-pass filter to limit the bandwidth of the ADXR642 rate response. The -3 dB frequency set by R_{OUT} and C_{OUT} is

$$f_{OUT} = 1/(2 \times \pi \times R_{OUT} \times C_{OUT})$$

and can be well controlled because R_{OUT} has been trimmed during manufacturing to be $180 \text{ k}\Omega \pm 1\%$. Any external resistor applied between the RATEOUT pin (1B, 2A) and SUMJ pin (1C, 2C) results in

$$R_{OUT} = (180 \text{ k}\Omega \times R_{EXT}) / (180 \text{ k}\Omega + R_{EXT})$$

In general, an additional filter (in either hardware or software) is added to attenuate high frequency noise arising from demodulation spikes at the 18 kHz resonant frequency of the gyro. An R/C output filter consisting of a 3.3k series resistor and 22 nF shunt capacitor (2.2 kHz pole) is recommended. Figure 13 shows the effect of adding this filter to the output of an ADXR642 set to a 2000 Hz bandwidth.

TEMPERATURE OUTPUT AND CALIBRATION

It is common practice to temperature-calibrate gyros to improve their overall accuracy. The ADXR642 has a temperature proportional voltage output that provides input to such a calibration method. The temperature sensor structure is shown in Figure 14. The temperature output is characteristically nonlinear, and any load resistance connected to the TEMP output results in decreasing the TEMP output and its temperature coefficient. Therefore, buffering the output is recommended.

The voltage at TEMP (3F, 3G) is nominally 2.5 V at 25°C, and $V_{RATIO} = 5 \text{ V}$. The temperature coefficient is $\sim 9 \text{ mV}/^\circ\text{C}$ at 25°C. Although the TEMP output is highly repeatable, it has only modest absolute accuracy.

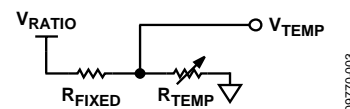


Figure 14. Temperature Sensor Structure

SUPPLY RATIOMETRICITY

The ADXR642 RATEOUT, ST1, ST2, and TEMP signals are ratiometric to the V_{RATIO} voltage; for example, the null voltage, rate sensitivity and temperature outputs are proportional to V_{RATIO} . Therefore, it is most easily used with a supply-ratiometric ADC, which results in self-cancellation of errors due to minor supply variations. There is some small, usually negligible, error due to nonratiometric behavior. Note that to guarantee full rate range, V_{RATIO} should not be greater than AV_{CC} .

MODIFYING THE MEASUREMENT RANGE

The ADXRS642 scale factor can be reduced to extend the measurement range to as much as $\pm 450^\circ/\text{sec}$ by adding a single $225\text{ k}\Omega$ resistor between the RATEOUT and SUMJ. If an external resistor is added between RATEOUT and SUMJ, C_{OUT} must be proportionally reduced to maintain the correct bandwidth.

NULL ADJUSTMENT

The nominal 2.5 V null is for a symmetrical swing range at RATEOUT (1B, 2A). However, a nonsymmetrical output swing may be suitable in some applications. Null adjustment is possible by injecting a suitable current to SUMJ (1C, 2C). Note that supply disturbances may reflect some null instability. Digital supply noise should be avoided, particularly in this case.

SELF-TEST FUNCTION

The ADXRS642 includes a self-test feature that actuates each of the sensing structures and associated electronics in the same manner, as if subjected to angular rate. It is activated by standard logic high levels applied to Input ST1 (5F, 5G), Input ST2 (4F, 4G), or both. ST1 causes the voltage at RATEOUT to change about -0.3 V , and ST2 causes an opposite change of $+0.3\text{ V}$. The self-test response follows the viscosity temperature dependence of the package atmosphere, approximately $0.25\%/^\circ\text{C}$. Activating both ST1 and ST2 simultaneously is not damaging. ST1 and ST2 are fairly closely matched ($\pm 2\%$), but actuating both simultaneously may result in a small apparent null bias shift proportional to the degree of self-test mismatch.

ST1 and ST2 are activated by applying a voltage equal to V_{RATIO} to the ST1 pin and the ST2 pin. The voltage applied to ST1 and ST2 must never be greater than AV_{CC} .

CONTINUOUS SELF-TEST

The on-chip integration of the ADXRS642 gives it higher reliability than is obtainable with any other high volume manufacturing method. Also, it is manufactured under a mature BiMOS process that has field-proven reliability. As an additional failure detection measure, power-on self-test can be performed. However, some applications may warrant continuous self-test while sensing rate. Details outlining continuous self-test techniques are also available in the [AN-768](#) Application Note.

MECHANICAL PERFORMANCE

The ADXRS642 excellent vibration rejection is demonstrated in Figure 15 and Figure 16. Figure 15 shows the ADXRS642 output response with and without 15 g rms 50 Hz to 5 kHz of random vibration. The bandwidth of the gyro was limited to 1600 Hz . Performance is similar regardless of the direction of the input vibration.

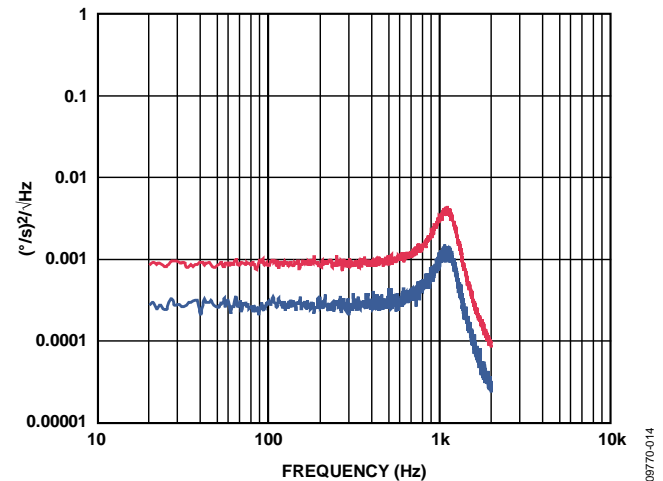


Figure 15. ADXRS642 Output Response with and Without Random Vibration (15 g rms , 50 Hz to 5 kHz)

Figure 16 demonstrates the ADXRS642 dc noise response to 5 g sine vibration over the 20 Hz to 5 kHz range. As can be seen, there are no sensitive frequencies present, and vibration rectification is vanishingly small. As in the previous example, the gyro bandwidth was set to 1600 Hz .

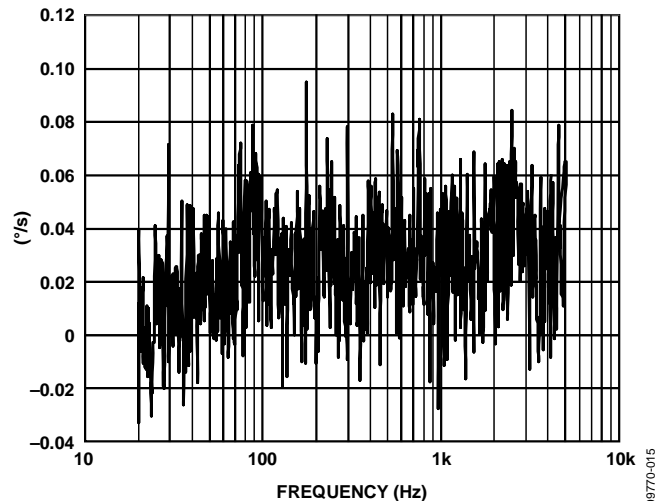
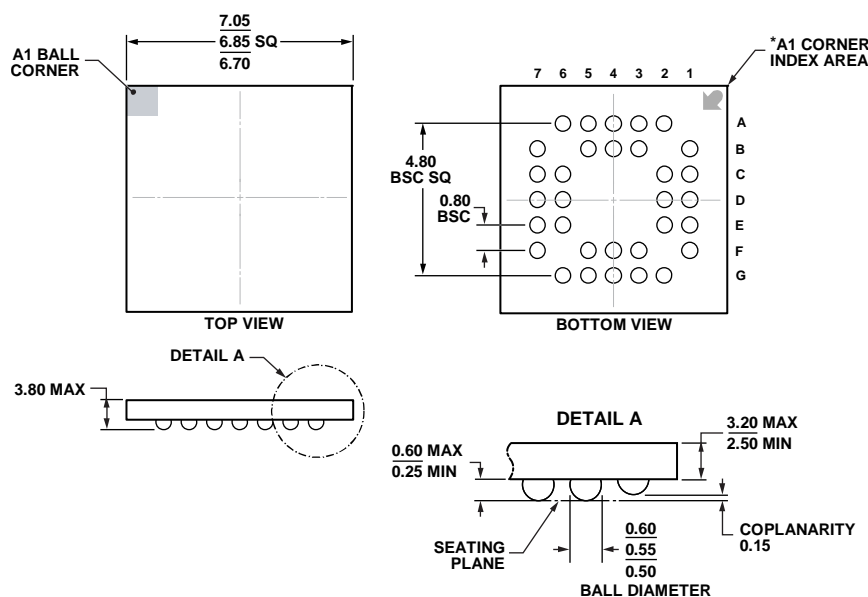


Figure 16. ADXRS642 Sine Vibration Noise Response (5 g , 20 Hz to 5 kHz)

OUTLINE DIMENSIONS



***BALL A1 IDENTIFIER IS GOLD PLATED AND CONNECTED TO THE D/A PAD INTERNALLY VIA HOLES.**

*Figure 17. 32-Lead Ceramic Ball Grid Array [CBGA]
(BG-32-3)*

Dimensions shown in millimeters

07-11-2012-B

ORDERING GUIDE

Model¹	Temperature Range	Package Description	Package Option
ADXR5642BBGZ	–40°C to +105°C	32-Lead Ceramic Ball Grid Array [CBGA]	BG-32-3
ADXR5642BBGZ-RL	–40°C to +105°C	32-Lead Ceramic Ball Grid Array [CBGA]	BG-32-3
EVAL-ADXR5642Z		Evaluation Board	

¹ Z = RoHS Compliant Part.

NOTES

NOTES