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1 Block diagram and pin description

Figure 1. Block diagram

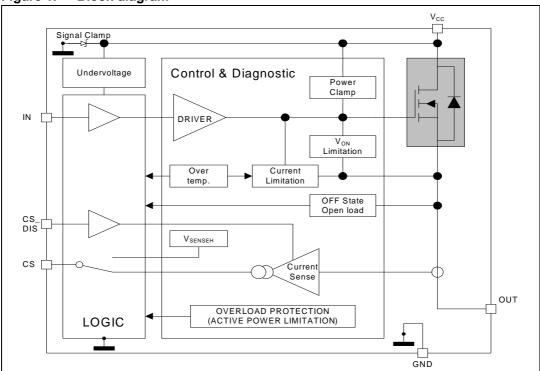


Table 1. Pin function

Name	Function
V _{CC}	Battery connection.
OUTPUT	Power output.
GND	Ground connection. Must be reverse battery protected by an external diode/resistor network.
INPUT	Voltage controlled input pin with hysteresis, CMOS compatible; it controls output switch state.
CURRENT SENSE	Analog current sense pin; it delivers a current proportional to the load current.
CS_DIS	Active high CMOS compatible pin, to disable the current sense pin.

Vcc 16 Vcc GND NC NC NC **INPUT** NC C SENSE OUTPUT OUTPUT NC OUTPUT CS DIS Vcc 8 Vcc GAPGCFT00526

Figure 2. Configuration diagram (top view)

Table 2. Suggested connections for unused and not connected pins

	1								
Connection / pin	Current sense	N.C.	Output	Input	CS_DIS				
Floating	Not allowed	Х	X	Х	Х				
To ground	Through 1 KΩ resistor	Х	Through 22 KΩ resistor	Through 10 KΩ resistor	Through 10 KΩ resistor				

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2 Electrical specifications

V_{CC} V_F V_{CC} V

GND

I_{GND}

INPUT

CURRENT SENSE

 V_{SENSE}

Figure 3. Current and voltage conventions

Note:

 $V_F = V_{OUT} - V_{CC}$ during reverse battery condition.

 V_{IN}

2.1 Absolute maximum ratings

Stressing the device above the rating listed in *Table 3* may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

Table 3. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V _{CC}	DC supply voltage	41	V
-V _{CC}	Reverse DC supply voltage	0.3	V
-I _{GND}	DC reverse ground pin current	200	mA
I _{OUT}	DC output current	Internally limited	А
-I _{OUT}	Reverse DC output current	20	Α
I _{IN}	DC input current	-1 to 10	mA
I _{CSD}	DC current sense disable input current	-1 to 10	mA
-I _{CSENSE}	DC reverse CS pin current	200	mA
V _{CSENSE}	Current sense maximum voltage	V _{CC} - 41 to +V _{CC}	V
E _{MAX}	Maximum switching energy (single pulse) L = 3 mH; $R_L = 0 \Omega$; $V_{bat} = 13.5 V$; $T_{jstart} = 150 ^{\circ}C$; $I_{OUT} = I_{limL}(Typ.)$	104	mJ

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Table 3. Absolute maximum ratings (continued)

Symbol	Parameter	Value	Unit
	Electrostatic discharge		
	(Human Body Model: $R = 1.5 \text{ K}\Omega$; $C = 100 \text{ pF}$)		
V _{ESD}	- INPUT	4000	V
	- CURRENT SENSE	2000	V
	- CS_DIS	4000	V
	– OUTPUT	5000	V
	- V _{CC}	5000	V
V _{ESD}	Charge device model (CDM-AEC-Q100-011)	750	V
T _j	Junction operating temperature	-40 to 150	°C
T _{stg}	Storage temperature	-55 to 150	°C

2.2 Thermal data

Table 4. Thermal data

Symbol	Parameter	Typ. value	Unit
R _{thj-pcb}	Thermal resistance junction-pcb ⁽¹⁾	18.5	°C/W
R _{thj-amb}	Thermal resistance junction - ambient on two layers pcb	See Figure 36	°C/W
R _{thj-amb}	Thermal resistance junction - ambient on two layers pcb ⁽²⁾	34.5	°C/W

- 1. The measure is done in accordance with the JESD 51-8.
- 2. Four Layers PCB characteristics:
- Cu thickness: 70 um outer layers, 35 um inner layers
- Board finish thickness 1.6 mm +/- 10%
- Thermal vias separation 1.2 mm
- Thermal via diameter 0.3 mm +/- 0.08 mm
- Cu thickness on vias 0.025 mm
- Device soldered at about 2 cm from the PCB edge with two sqcm of exposed copper.

2.3 Electrical characteristics

Values specified in this section are for 8 V < V_{CC} < 28 V; -40°C < T_j < 150°C, unless otherwise stated.

Table 5. Power section

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{CC}	Operating supply voltage		4.5	13	28	V
V _{USD}	Undervoltage shutdown			3.5	4.5	V
V _{USDhyst}	Undervoltage shutdown hysteresis			0.5		٧

Table 5. Power section (continued)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
		I _{OUT} = 2 A; T _j = 25°C		50		$m\Omega$
R _{ON}	On-state resistance	I _{OUT} = 2 A; T _j = 150°C			100	$m\Omega$
		$I_{OUT} = 2 \text{ A}; V_{CC} = 5 \text{ V}; T_j = 25^{\circ}\text{C}$			65	$m\Omega$
V _{clamp}	Clamp voltage	I _S = 20 mA	41	46	52	V
1.	Supply current	Off-state; $V_{CC} = 13 \text{ V}$; $T_j = 25^{\circ}\text{C}$; $V_{IN} = V_{OUT} = V_{SENSE} = V_{CSD} = 0 \text{ V}$		2 ⁽¹⁾	5 ⁽¹⁾	μΑ
I _S		On-state; $V_{CC} = 13 \text{ V}$; $V_{IN} = 5 \text{ V}$; $I_{OUT} = 0 \text{ A}$		1.5	3	mA
1	Off. state output current	$V_{IN} = V_{OUT} = 0 \text{ V}; V_{CC} = 13 \text{ V};$ $T_j = 25^{\circ}\text{C}$	0	0.01	3	μΑ
L(off1)	Off-state output current	$V_{IN} = V_{OUT} = 0 \text{ V}; V_{CC} = 13 \text{ V};$ $T_j = 125^{\circ}\text{C}$	0		5	μΑ
V _F	Output - V _{CC} diode voltage	-l _{OUT} = 2 A; T _j = 150°C			0.7	V

^{1.} PowerMOS leakage included.

Table 6. Switching ($V_{CC} = 13 \text{ V}; T_j = 25 \text{ °C}$)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
t _{d(on)}	Turn-on delay time	$R_L = 6.5 \Omega$ (see <i>Figure 6</i>)	_	20	_	μs
t _{d(off)}	Turn-off delay time	$R_L = 6.5 \Omega$ (see <i>Figure 6</i>)	_	40		μs
(dV _{OUT} /dt) _{on}	Turn-on voltage slope	$R_L = 6.5 \Omega$	_	See Figure 26	_	V/µs
(dV _{OUT} /dt) _{off}	Turn-off voltage slope	$R_L = 6.5 \Omega$	_	See Figure 28		V/µs
W _{ON}	Switching energy losses during t _{on}	$R_L = 6.5 \Omega$ (see <i>Figure 6</i>)	_	0.20		mJ
W _{OFF}	Switching energy losses during t _{off}	$R_L = 6.5 \Omega$ (see <i>Figure 6</i>)	_	0.3	_	mJ

Table 7. Logic inputs

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{IL}	Input low level voltage				0.9	V
I _{IL}	Low level input current	V _{IN} = 0.9 V	1			μΑ
V _{IH}	Input high level voltage		2.1			V
I _{IH}	High level input current	V _{IN} = 2.1 V			10	μΑ
V _{I(hyst)}	Input hysteresis voltage		0.25			V
V	Innut clamp voltage	I _{IN} = 1 mA	5.5		7	V
V _{ICL}	Input clamp voltage	I _{IN} = -1 mA		-0.7		V

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Table 7. Logic inputs (continued)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{CSDL}	CS_DIS low level voltage				0.9	V
I _{CSDL}	Low level CS_DIS current	V _{CSD} = 0.9 V	1			μA
V_{CSDH}	CS_DIS high level voltage		2.1			V
I _{CSDH}	High level CS_DIS current	V _{CSD} = 2.1 V			10	μΑ
V _{CSD(hyst)}	CS_DIS hysteresis voltage		0.25			V
V	CS DIS clamp voltage	I _{CSD} = 1 mA	5.5		7	V
V _{CSCL}	CS_DIS clamp voltage	I _{CSD} = -1 mA		-0.7		V

Table 8. Protections and diagnostics (1)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
	DC short circuit current	V _{CC} = 13 V	19	27	38	Α
llimH	DC Short circuit current	5 V < V _{CC} < 28 V			38	Α
I _{limL}	Short circuit current during thermal cycling	$V_{CC} = 13V; T_R < T_j < T_{TSD}$		7		Α
T _{TSD}	Shutdown temperature		150	175	200	°C
T _R	Reset temperature		T _{RS} + 1	T _{RS} + 5		°C
T _{RS}	Thermal reset of status		135			°C
T _{HYST}	Thermal hysteresis (T _{TSD} - T _R)			7		°C
V _{DEMAG}	Turn- Off output voltage clamp	I _{OUT} = 2 A; V _{IN} = 0; L = 6 mH	V _{CC} -41	V _{CC} -46	V _{CC} -52	V
V _{ON}	Output voltage drop limitation	$I_{OUT} = 0.1 \text{ A};$ $T_j = -40^{\circ}\text{C to } 150^{\circ}\text{C}$ (see <i>Figure 8</i>)		25		mV

To ensure long term reliability under heavy overload or short circuit conditions, protection and related diagnostic signals must be used together with a proper software strategy. If the device is subjected to abnormal conditions, this software must limit the duration and number of activation cycles.

Table 9. Current sense (8 V < V_{CC} < 18 V)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
K ₀	I _{OUT} /I _{SENSE}	$I_{OUT} = 0.05 \text{ A}; V_{SENSE} = 0.5 \text{ V}; V_{CSD} = 0 \text{ V}; T_j = -40^{\circ}\text{C} \text{ to } 150^{\circ}\text{C}$	1170	2000	3090	
К ₁	I _{OUT} /I _{SENSE}	$I_{OUT} = 1 \text{ A}; V_{SENSE} = 4 \text{ V};$ $V_{CSD} = 0 \text{ V};$ $T_j = -40^{\circ}\text{C to } 150^{\circ}\text{C}$ $T_j = 25^{\circ}\text{C to } 150^{\circ}\text{C}$		2000 2000	2750 2465	
dK ₁ /K ₁ ⁽¹⁾	Current sense ratio drift	I _{OUT} = 1 A; V _{SENSE} = 4 V; V _{CSD} = 0 V; T _J = -40 °C to 150 °C	-10		10	%

Table 9. Current sense (8 V < V_{CC} < 18 V) (continued)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
K ₂	lout/Isense	$I_{OUT} = 2 \text{ A}; V_{SENSE} = 4 \text{ V};$ $V_{CSD} = 0 \text{ V};$ $T_j = -40^{\circ}\text{C to } 150^{\circ}\text{C}$ $T_j = 25^{\circ}\text{C to } 150^{\circ}\text{C}$	1765 1765	2000 2000	2315 2155	
$dK_2/K_2^{(1)}$	Current sense ratio drift	I _{OUT} = 2 A; V _{SENSE} = 4 V; V _{CSD} = 0 V; T _J = -40 °C to 150 °C	-7		7	%
K ₃	l _{OUT} /l _{SENSE}	$I_{OUT} = 4 \text{ A}; V_{SENSE} = 4 \text{ V};$ $V_{CSD} = 0 \text{ V};$ $T_j = -40^{\circ}\text{C to } 150^{\circ}\text{C}$ $T_j = 25^{\circ}\text{C to } 150^{\circ}\text{C}$	1840 1840	2000 2000	2135 2080	
$dK_3/K_3^{(1)}$	Current sense ratio drift	I _{OUT} = 4 A; V _{SENSE} = 4 V; V _{CSD} = 0 V; T _J = -40 °C to 150 °C	-4		4	%
		$I_{OUT} = 0 \text{ A; } V_{SENSE} = 0 \text{ V;}$ $V_{CSD} = 5 \text{ V; } V_{IN} = 0 \text{ V;}$ $T_j = -40^{\circ}\text{C to } 150^{\circ}\text{C}$	0		1	μΑ
I _{SENSE0}	Analog sense leakage current	$V_{CSD} = 0 \text{ V}; V_{IN} = 5 \text{ V};$ $T_j = -40^{\circ}\text{C to } 150^{\circ}\text{C}$	0		2	μΑ
		I _{OUT} = 2 A; V _{SENSE} = 0 V; V _{CSD} = 5 V; V _{IN} = 5 V; T _j = -40°C to 150°C	0		1	μΑ
I _{OL}	Open-load on-state current detection threshold	V _{IN} = 5 V; 8 V < V _{CC} < 18 V; I _{SENSE} = 5 μA	4		20	mA
V _{SENSE}	Max analog sense output voltage	I _{OUT} = 4 A; V _{CSD} = 0 V	5			٧
V _{SENSEH}	Analog sense output voltage in fault condition ⁽²⁾	V_{CC} = 13 V; R_{SENSE} = 3.9 K Ω		8		٧
I _{SENSEH}	Analog sense output current in fault condition ⁽²⁾	V _{CC} = 13 V; V _{SENSE} = 5 V		9		mA
^t DSENSE1H	Delay response time from falling edge of CS_DIS pin	V _{SENSE} < 4 V; 0.5 A < I _{OUT} < 4 A; I _{SENSE} = 90% of I _{SENSE} max (see <i>Figure 4</i>)		50	100	μs
^t DSENSE1L	Delay response time from rising edge of CS_DIS pin	V _{SENSE} < 4 V; 0.5 A < I _{OUT} < 4 A; I _{SENSE} = 10% of I _{SENSE} max (see <i>Figure 4</i>)		5	20	μs
[†] DSENSE2H	Delay response time from rising edge of INPUT pin	V _{SENSE} < 4 V; 0.5 A < I _{OUT} < 4 A; I _{SENSE} = 90% of I _{SENSE} max (see <i>Figure 4</i>)		80	250	μs

Table 9. Current sense (8 V < V_{CC} < 18 V) (continued)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
$\Delta t_{ extsf{DSENSE2H}}$	Delay response time between rising edge of output current and rising edge of current sense	V _{SENSE} < 4 V; I _{SENSE} = 90% of I _{SENSEMAX} ; I _{OUT} = 90% of I _{OUTMAX} ; I _{OUTMAX} = 2 A (see <i>Figure 7</i>)			40	□□µs
t _{DSENSE2L}	Delay response time from falling edge of INPUT pin	V _{SENSE} < 4 V; 0.5 A < I _{OUT} < 4 A; I _{SENSE} = 10% of I _{SENSE} max (see <i>Figure 4</i>)		100	250	μs

- 1. Parameter guaranteed by design; it is not tested.
- 2. Fault condition includes: power limitation, overtemperature and open-load off-state detection.

Table 10. Open-load detection (8 V < V_{CC} < 18 V)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{OL}	Open-load off-state voltage detection threshold	V _{IN} = 0 V	2	See Figure 5	4	V
t _{DSTKON}	Output short circuit to V _{CC} detection delay at turn Off	See Figure 5	180		1200	μs
I _{L(off2)r}	Off-state output current at V _{OUT} = 4 V	$V_{IN} = 0 \text{ V}; V_{SENSE} = 0 \text{ V}; V_{OUT} \text{ rising from } 0 \text{ V to } 4 \text{ V}$	-120		0	μΑ
I _{L(off2)f}	Off-state output current at V _{OUT} = 2 V	$V_{IN} = 0 \text{ V};$ $V_{SENSE} = V_{SENSEH}; V_{OUT}$ falling from V_{CC} to 2 V	-50		90	μΑ
t _{d_vol}	Delay response from output rising edge to V _{SENSE} rising edge in open-load	$V_{OUT} = 4 \text{ V}; V_{IN} = 0 \text{ V};$ $V_{SENSE} = 90\% \text{ of } V_{SENSEH}$			20	μs

Figure 4. Current sense delay characteristics

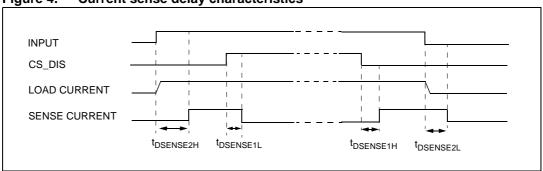


Figure 5. Open-load off-state delay timing

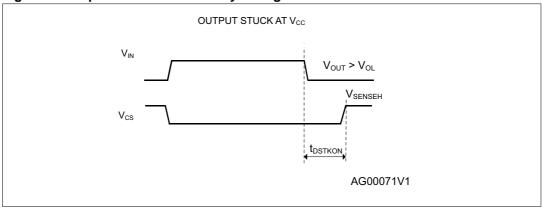
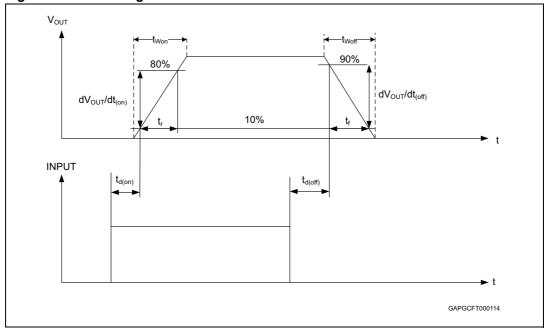


Figure 6. Switching characteristics



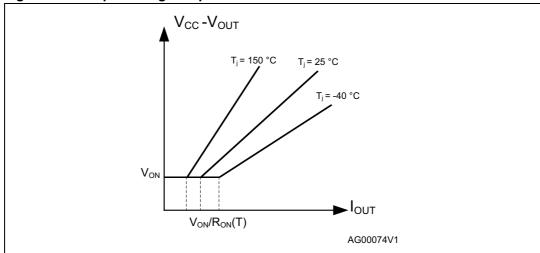
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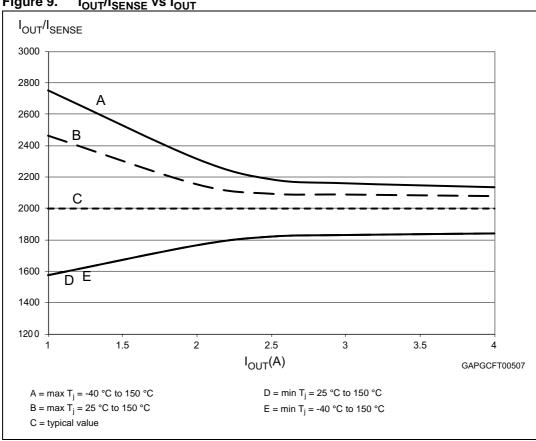
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IOUT
IOUT
IOUT
IOUT
ISENSE
ISENSE
ISENSEMAX

GAPGCFT000115

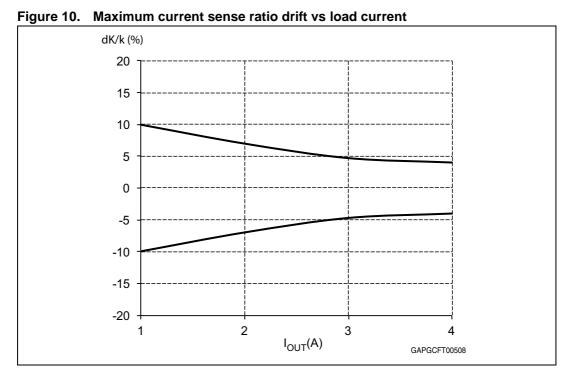
Figure 7. Delay response time between rising edge of output current and rising edge of current sense (CS enabled)







I_{OUT}/I_{SENSE} vs I_{OUT} Figure 9.



Note:

Parameter guaranteed by design; it is not tested.



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Table 11. Truth table

Conditions	Input	Output	Sense (V _{CSD} = 0 V) ⁽¹⁾
Normal operation	L	L	0
Normal operation	Н	Н	Nominal
Overtemperature	L	L	0
Overtemperature	Н	L	V _{SENSEH}
Undervoltage	L	L	0
Ondervoltage	Н	L	0
	Н	Х	Nominal
Overload	н	(no power limitation) Cycling (power limitation)	V _{SENSEH}
Short circuit to GND	L	L	0
(power limitation)	Н	L	V _{SENSEH}
Open-load off-state (with external pull-up)	L	Н	V _{SENSEH}
Short circuit to V _{CC} (external pull-up disconnected)	L H	н н	V _{SENSEH} < Nominal
Negative output voltage clamp	L	L	0

If the V_{CSD} is high, the SENSE output is at a high impedance, its potential depends on leakage currents and external circuit.

Table 12. Electrical transient requirements (part 1)

ISO 7637-2: 2004(E)	Test levels ⁽¹⁾		NUMBEROL		cle/pulse ion time	Delays and Impedance	
Test pulse	III	IV	test times	Min.	Max.	impedance	
1	-75V	-100V	5000 pulses	0.5s	5s	2 ms, 10Ω	
2a	+37V	+50V	5000 pulses	0.2s	5s	50μs, 2Ω	
3a	-100V	-150V	1h	90ms	100ms	0.1μs, 50Ω	
3b	+75V	+100V	1h	90ms	100ms	0.1μs, 50Ω	
4	-6V	-7V	1 pulse			100ms, 0.01Ω	
5b ⁽²⁾	+65V	+87V	1 pulse			400ms, 2Ω	

^{1.} The above test levels must be considered referred to V_{CC} = 13.5 V except for pulse 5b.

Table 13. Electrical transient requirements (part 2)

ISO 7637-2: 2004E	Test level results		
Test pulse	III	VI	
1	С	С	
2a	С	С	
3a	С	С	
3b	С	С	
4	С	С	
5b ⁽¹⁾	С	С	

^{1.} Valid in case of external load dump clamp: 40V maximum referred to ground.

Table 14. Electrical transient requirements (part 3)

Class	Contents
С	All functions of the device performed as designed after exposure to disturbance.
E	One or more functions of the device did not perform as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

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^{2.} Valid in case of external load dump clamp: 40 V maximum referred to ground.

2.4 Waveforms

Figure 11. Normal operation

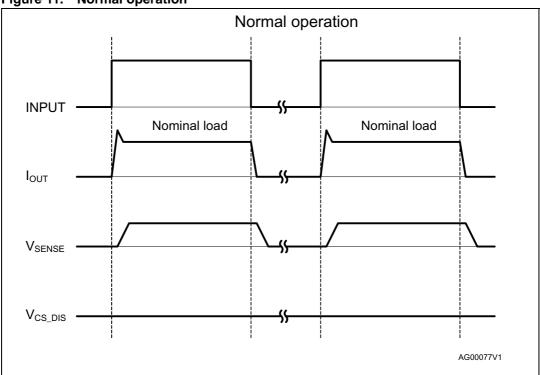
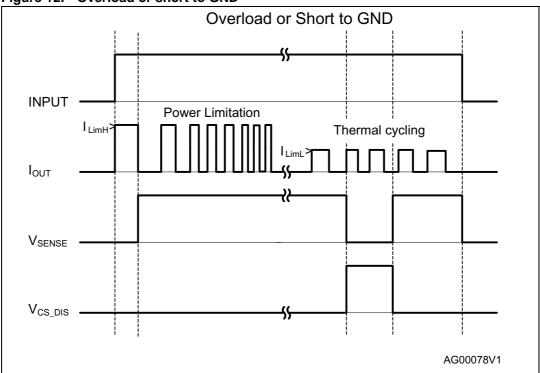


Figure 12. Overload or short to GND



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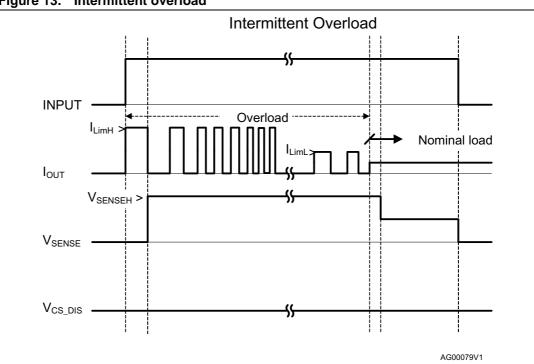
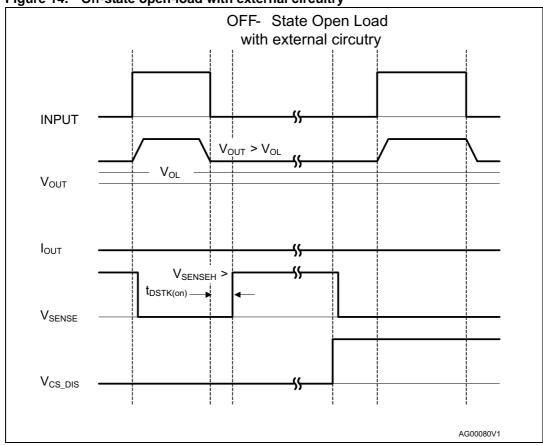


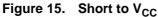
Figure 13. Intermittent overload





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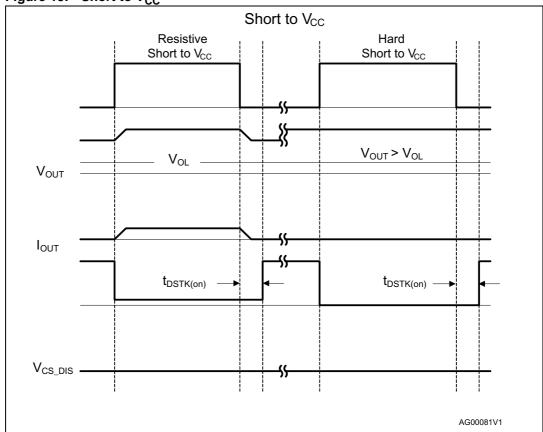
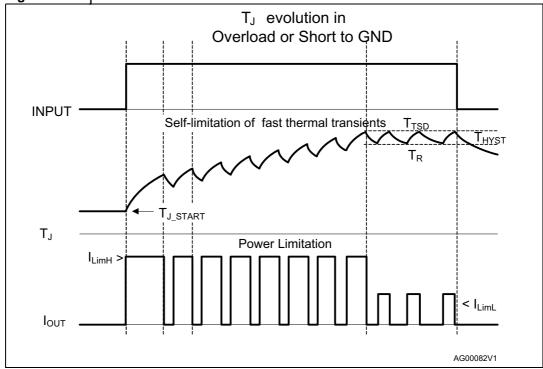


Figure 16. T_i evolution in over load or short to GND

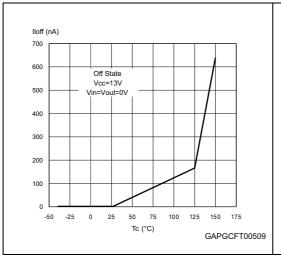


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2.5 Electrical characteristics curves

Figure 17. Off-state output current

Figure 18. High level input current



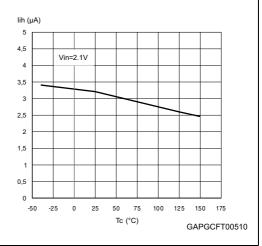
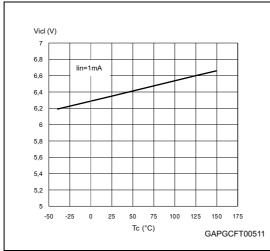


Figure 19. Input clamp level

Figure 20. Input low level



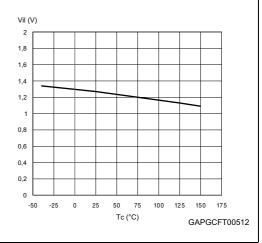
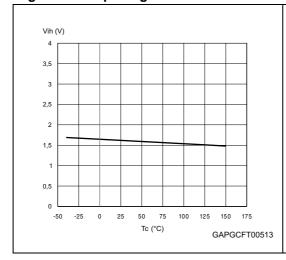
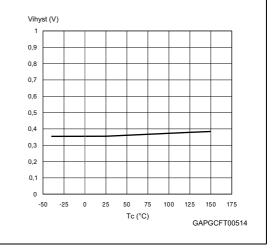


Figure 21. Input high level

Figure 22. Input hysteresis voltage

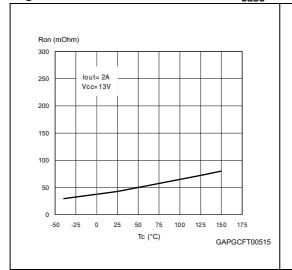




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Figure 23. On-state resistance vs T_{case} Figure 24. On-state resistance vs V_{CC}



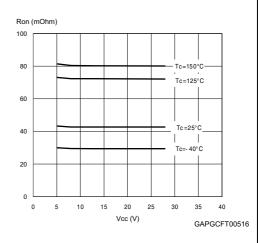
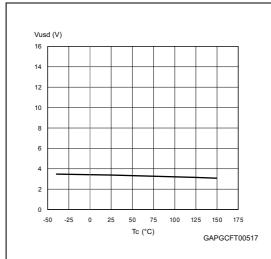


Figure 25. Undervoltage shutdown

Figure 26. Turn-On voltage slope



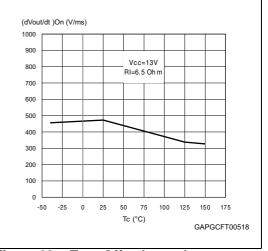
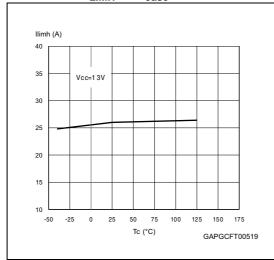


Figure 27. I_{LIMH} vs T_{case}

Figure 28. Turn-Off voltage slope



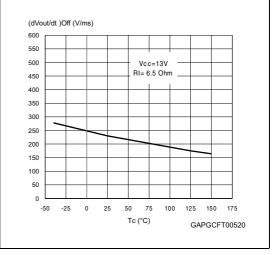
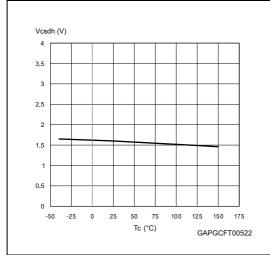


Figure 29. CS_DIS high level voltage

Figure 30. CS_DIS clamp voltage



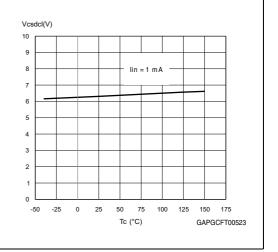
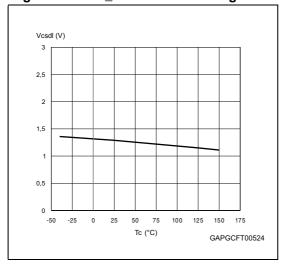


Figure 31. CS_DIS low level voltage



3 Application information

HSV

Application schematic

CS_DIS

VCC

Reprot

IINPUT

OUTPUT

Reprot

Current sense

GND

Cext

RSENSE

VGND

RGND

DGND

Figure 32. Application schematic

3.1 GND protection network against reverse battery

This section provides two solutions to implement a ground protection network against reverse battery.

3.1.1 Solution 1: resistor in the ground line (R_{GND} only)

This can be used with any type of load.

The following description shows how to select the R_{GND} resistor:

- 1. $R_{GND} \le 600 \text{ mV} / (I_{S(on)max})$
- 2. $R_{GND} \ge (-V_{CC}) / (-I_{GND})$

where -I_{GND} is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device datasheet.

Power dissipation in R_{GND} (when $V_{CC} < 0$ during reverse battery situations) is:

$$P_D = (-V_{CC})^2 / R_{GND}$$

This resistor can be shared amongst several different HSDs. Please note that the value of this resistor should be calculated with formula (1) where $I_{S(on)max}$ becomes the sum of the maximum on-state currents of the different devices.

Please note that, if the microprocessor ground is not shared by the device ground, then the R_{GND} produces a shift ($I_{S(on)max} * R_{GND}$) in the input thresholds and in the status output

values. This shift varies depending on how many devices are ON in case of several high side drivers sharing the same R_{GND} .

If the calculated power dissipation requires the use of a large resistor, or several devices have to share the same resistor, then ST suggests to utilize Section 3.1.2: Solution 2: diode (D_{GND}) in the ground line.

3.1.2 Solution 2: diode (D_{GND}) in the ground line

Note that a resistor ($R_{GND} = 1 \text{ k}\Omega$) should be inserted in parallel to D_{GND} if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network will produce a shift (≈600 mV) in the input threshold and in the status output values if the microprocessor ground is not common to the device ground. This shift will not vary if more than one HSD shares the same diode/resistor network.

3.2 Load dump protection

 D_{ld} is necessary (Voltage Transient Suppressor) if the load dump peak voltage exceeds the V_{CC} maximum DC rating. The same applies if the device is subject to transients on the V_{CC} line which are greater than the ones shown in the ISO 7637-2: 2004(E) table.

3.3 MCU I/O protection

If a ground protection network is used and negative transients are present on the V_{CC} line, the control pins are pulled negative. ST suggests to insert a resistor (R_{prot}) in line to prevent the microcontroller I/O pins from latching up.

The value of these resistors is a compromise between the leakage current of microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os:

 $-V_{CCpeak}/I_{latchup} \le R_{prot} \le (V_{OH\mu C}-V_{IH}-V_{GND})/I_{IHmax}$

Calculation example:

For $V_{CCpeak} = -100 \text{ V}$ and $I_{latchup} \ge 20 \text{ mA}$; $V_{OHuC} \ge 4.5 \text{ V}$

 $5 \text{ k}\Omega \leq R_{\text{prot}} \leq 180 \text{ k}\Omega$

Recommended values: $R_{prot} = 10 \text{ k}\Omega$, $C_{EXT} = 10 \text{ nF}$.

3.4 Current sense and diagnostic

The current sense pin performs a double function (see *Figure 33: Current sense and diagnostic*):

Current mirror of the load current in normal operation, delivering a current proportional to the load current according to a known ratio K_X.
 The current I_{SENSE} can be easily converted to a voltage V_{SENSE} by means of an external resistor R_{SENSE}. Linearity between I_{OUT} and V_{SENSE} is ensured up to 5V minimum (see parameter V_{SENSE} in *Table 9: Current sense (8 V < VCC < 18 V)*). The

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current sense accuracy depends on the output current (refer to current sense electrical characteristics *Table 9: Current sense (8 V < VCC < 18 V)*).

- Diagnostic flag in fault conditions, delivering a fixed voltage V_{SENSEH} up to a maximum current I_{SENSEH} in case of the following fault conditions (refer to Table 11: Truth table):
 - Power limitation activation
 - Overtemperature
 - Short to V_{CC} in off-state
 - Open-load in off-state with additional external components.

A logic level high the CS_DIS pin simultaneously sets all the current sense pins of the device in a high impedance state, thus disabling the current monitoring and diagnostic detection. This feature allows multiplexing of the microcontroller analog inputs by sharing the sense resistance and ADC line among different devices.

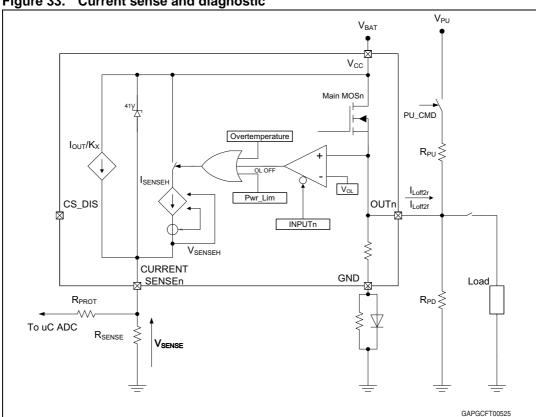


Figure 33. Current sense and diagnostic

3.4.1 Short to V_{CC} and off-state open-load detection

Short to V_{CC}

A short circuit between V_{CC} and output is indicated by the relevant current sense pin set to V_{SENSEH} during the device off-state. Little or no current is delivered by the current sense during the on-state depending on the nature of the short circuit.

Off-state open-load with external circuitry

Detection of an open-load in off mode requires an external pull-up resistor (R_{PU}) connecting the output to a positive supply voltage (V_{PU}).

It is preferable that V_{PU} is switched off during the module standby mode to avoid an increase in overall standby current consumption in normal conditions, that is, when the load is connected.

An external pull-down resistor (R_{PD}) connected between output and GND is mandatory to avoid misdetection in case of floating outputs in off-state (see *Figure 33: Current sense and diagnostic*).

 R_{PD} must be selected in order to ensure $V_{OUT} < V_{OLmin}$ unless pulled up by the external circuitry:

$$V_{OUT} \Big|_{Pull-up\ OFF} = R_{PD} \cdot I_{L(off\ 2)f} < V_{OL\min} = 2V$$

 $R_{PD} \le 22 \text{ K}\Omega$ is recommended.

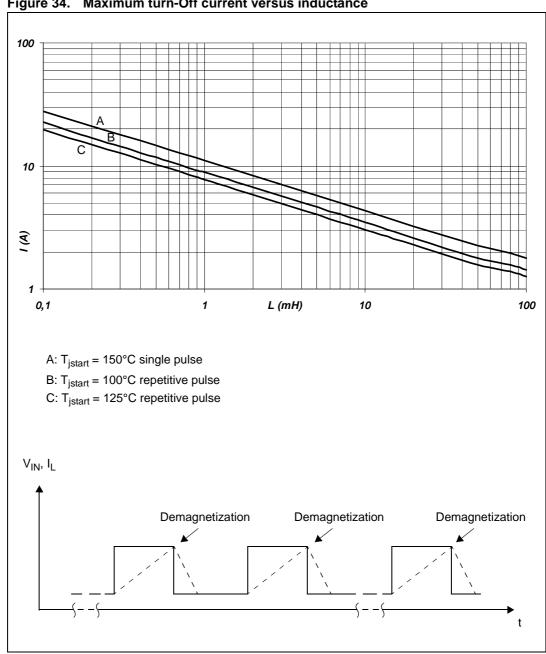
For proper open-load detection in off-state, the external pull-up resistor must be selected according to the following formula:

$$\left. V_{OUT} \right|_{Pull-up_ON} = \frac{R_{PD} \cdot V_{PU} - R_{PU} \cdot R_{PD} \cdot I_{L(off\ 2)r}}{R_{PU} + R_{PD}} > V_{OL\max} = 4V$$

For the values of V_{OLmin} , V_{OLmax} , $I_{L(off2)r}$ and $I_{L(off2)f}$ see *Table 10: Open-load detection* (8 V < VCC < 18 V).

Maximum demagnetization energy ($V_{CC} = 13.5V$) 3.5

Figure 34. Maximum turn-Off current versus inductance



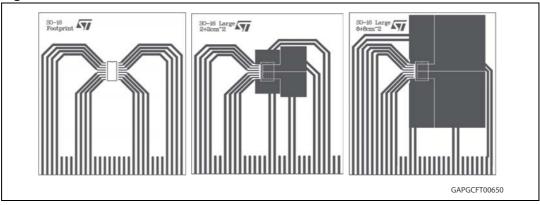
Note:

Values are generated with $R_L = 0 \Omega$. In case of repetitive pulses, T_{jstart} (at the beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.

4 Package and PCB thermal data

4.1 SO-16L thermal data

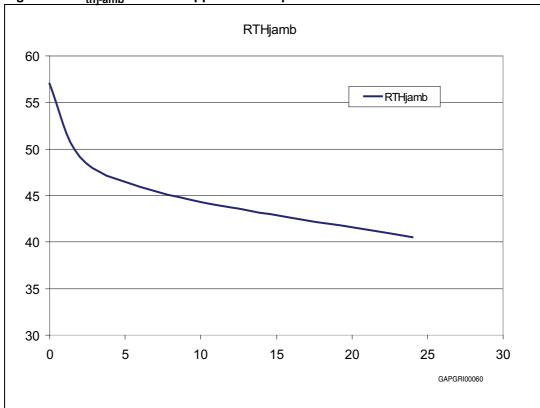
Figure 35. SO-16L PC board



Note:

Layout condition of R_{th} and Z_{th} measurements (PCB: Double layer, Thermal Vias, FR4 area = 77 mm x 86 mm, PCB thickness = 1.6 mm, Cu thickness = 70 μ m (front and back side), Copper areas: from minimum pad lay-out to 8 cm²).

Figure 36. $R_{thj-amb}$ vs PCB copper area in open box free air condition



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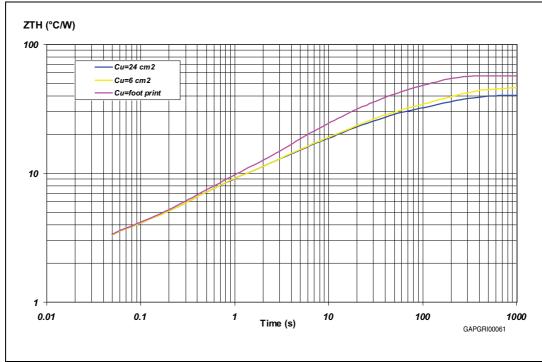


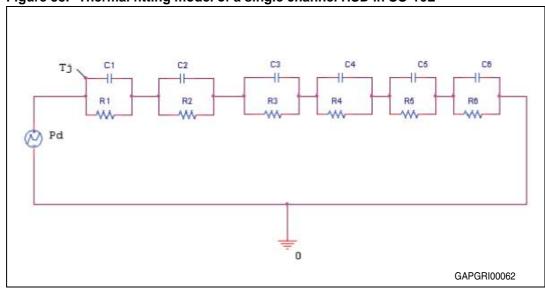
Figure 37. SO-16L thermal impedance junction ambient single pulse

Equation 1: pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp} (1 - \delta)$$

where $\delta = t_P/T$

Figure 38. Thermal fitting model of a single channel HSD in SO-16L (a)



a. The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.

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Table 15. Thermal parameter

Area/island (cm ²)	Footprint	2	8
R1 (°C/W)	0.7		
R2 (°C/W)	2.3		
R3 (°C/W)	4		
R4 (°C/W)	8	6	6
R5 (°C/W)	14	13	13
R6 (°C/W)	28	20	14.5
C1 (W.s/°C)	0.001		
C2 (W.s/°C)	0.01		
C3 (W.s/°C)	0.1		
C4 (W.s/°C)	0.5		
C5 (W.s/°C)	1	1.5	1.5
C6 (W.s/°C)	3	9	12

Package information VN5E050ASO-E

5 Package information

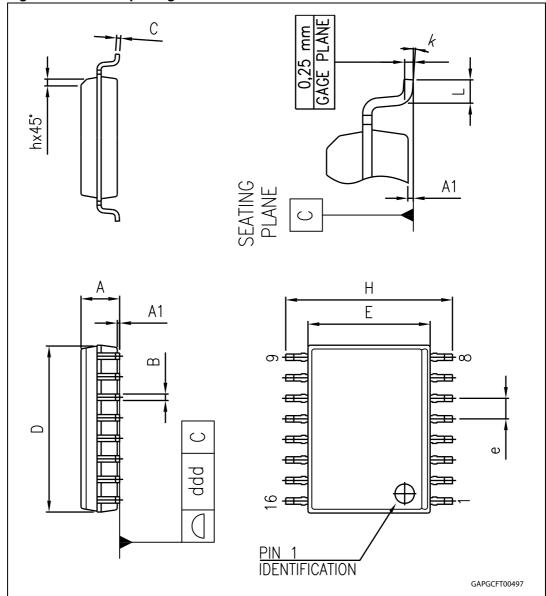
5.1 ECOPACK® packages

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com.

ECOPACK® is an ST trademark.

5.2 Package mechanical data

Figure 39. SO-16L package dimensions



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Table 16. SO-16L mechanical data

Symbol		Millimeters					
Symbol	Min	Тур	Max				
A	2.35		2.65				
A1	0.10		0.30				
В	0.33		0.51				
С	0.23		0.32				
D	10.10		10.50				
E	7.40		7.60				
е		1.27					
Н	10.00		10.65				
h	0.25		0.75				
L	0.40		1.27				
k	0°		8°				
ddd			0.10				

Package information VN5E050ASO-E

5.3 **Packing information**

Figure 40. SO-16L tube shipment (no suffix)

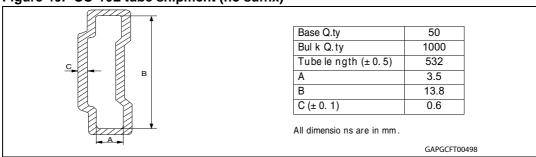
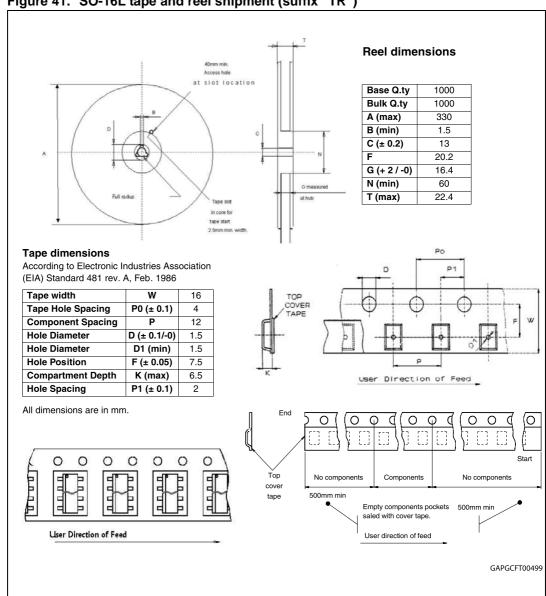


Figure 41. SO-16L tape and reel shipment (suffix "TR")



VN5E050ASO-E Order codes

6 Order codes

Table 17. Device summary

Package	Orde	Order codes	
	Tube	Tape and reel	
SO-16L	VN5E050ASO-E	VN5E050ASOTR-E	

Revision history VN5E050ASO-E

7 Revision history

Table 18. Document revision history

Date	Revision	Changes
14-Dec-2011	1	Initial release
16-Mar-2012	2	Added Section 4: Package and PCB thermal data and update Table 5.
25-June-2012	3	Update Table 4.
18-Sep-2012	4	Update Table 4.
18-Sep-2013	5	Updated disclaimer.

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