High-Current Differential Line Driver for Powerline Communications

Absolute Maximum Ratings

AVDD to AGND0.3V to +22V
DVDD to DGND0.3V to +6V
DGND to AGND0.3V to +0.3V
OUT+, OUT- to AGND0.3V to (V _{AVDD} + 0.3V)
IN+, IN- to AGND (Gain = $6V/V$) $A_{VDD}/2 \pm 4.0V$
(Gain = 12V/V)A _{VDD} /2 ±2.0V
(Gain = 15V/V)A _{VDD} /2 ±1.6V
(Gain = 18V/V) A _{VDD} /2 ±1.33V
STATUS1, STATUS0, G1, G0,
TXEN, MODE, ILSET to DGND0.3V to (V _{DVDD} + 0.3V)
Continuous Current into STATUS1, STATUS0, G1, G0
±20mA

Package Thermal Characteristics (Note 1)

TQFN Multilayer Board	
lunction to Ambient Thermal F	5

Junction-to-Ambient Thermal Resistance (θ _{JA})	33°C/W
Junction-to-Case Thermal Resistance (0 _{JC})	2°C/W

Continuous Power Dissipation (T _A = +70°C) TQFN (Four-Laver Board)	
(derate 33.3mW/°C above +70°C)	
TQFN (Single Layer)	
(derate 20.8mW/°C above +70°C)	1666.7mW
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	+260°C

TQFN Single-Layer Board	
Junction-to-Ambient Thermal Resistance (θ_{JA})	48°C/W
Junction-to-Case Thermal Resistance (θ _{JC})	2°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Electrical Characteristics

 $(V_{DVDD} = 3V, V_{AVDD} = 12V \text{ or } 18V, V_{AGND} = V_{DGND} = 0V, V_{IN+} = V_{IN-} = V_{AVDD}/2, G0 = V_{DVDD}, G1 = V_{DVDD}, TXEN = V_{DVDD}, MODE = V_{DVDD}, R_{SET} = 29k\Omega, R_{LOAD} = 50\Omega$ differential from OUT+ to OUT-, $T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS					
POWER SUPPLY		1										
Analog Supply Voltage Range	V _{AVDD}				8		20	V				
Digital Supply Voltage Range	V _{DVDD}				2.7		5.5	V				
			MODE Low,	T _A = +25°C		22	28					
			TXEN High				30					
			MODE Low,	T _A = +25°C		1.9	3					
	luce	V _{AVDD} = 12V	TXEN Low				3.5					
			MODE High, TXEN High	T _A = +25°C		41	50	-				
							55					
							MODE High,	T _A = +25°C		2	3	
Quiescent Current			TXEN Low				3.5	mA				
	IAVDD		MODE Low,	T _A = +25°C		22	28					
			TXEN High				30					
			MODE Low,	T _A = +25°C		2.1	3.0					
			TXEN Low				3.5					
		V _{AVDD} = 18V	MODE High,	T _A = +25°C		43	50					
			TXEN High				55					
			MODE High,	T _A = +25°C		2.3	3.0					
			TXEN Low				3.5					

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Electrical Characteristics (continued)

 $(V_{DVDD} = 3V, V_{AVDD} = 12V \text{ or } 18V, V_{AGND} = V_{DGND} = 0V, V_{IN+} = V_{IN-} = V_{AVDD}/2, G0 = V_{DVDD}, G1 = V_{DVDD}, TXEN = V_{DVDD}, MODE = V_{DVDD}, R_{SET} = 29k\Omega, R_{LOAD} = 50\Omega$ differential from OUT+ to OUT-, $T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 2)

PARAMETER	SYMBOL	CONE	DITIONS	MIN	ТҮР	MAX	UNITS	
			T _A = +25°C		28	130		
Quiescent Current DVDD	IDVDD					170	μA	
Enable Time	t _{ENABLE}	From TXEN transition	, low to high		2.4		μs	
Disable Time	t _{DIS}	From TXEN transition	, high to low		60		μs	
INPUT	1							
Input Common-Mode Voltage		External AC coupled,	internal bias level		V _{AVDD} /2		V	
		G1 = 0, G0 = 0, Gain	= 6V/V		20			
lanut Danistanaa		G1 = 0, G0 = 1, Gain	= 12V/V		13.3		1.0	
Input Resistance	R _{IN}	G1 = 1, G0 = 0, Gain	= 15V/V		11.4		kΩ	
		G1 = 1, G0 = 1, Gain	= 18V/V		10			
Input Capacitance	C _{IN}				1		pF	
GAIN AND FREQUENCY RES	PONSE							
Gain Accuracy				-2		+2	%	
Gain Error Drift					± 20		ppm/°C	
Output Slew Rate	SR	10V output step			80		V/µs	
Full Dawer Dandwidth		V _{OUTDIFF} = 120dBµV, Gain = 18			3.9		N41 I-	
Full-Power Bandwidth	BW _{FP}	V _{OUTDIFF} = 120dBµV, Gain = 6V/V			8.5		MHz	
AC PSRR		f = 50kHz, V _{AVDD} /(OUT+ - OUT-)			20		dB	
LINEARITY		-						
In-Band/Out-of-Band Intermodulation Products (MODE Low)		f _{IN1} = 50kHz, f _{IN2} = 55kHz, V _{LOAD} = 125.6dBµV _{RMS} (Note 4)			51		dBµV	
In-Band/Out-of-Band Intermodulation Products (MODE High)		$f_{IN1} = 200 \text{kHz}, f_{IN2} = 250 \text{kHz},$ $V_{LOAD} = 125.6 \text{dB} \mu V_{\text{RMS}},$ $R_{\text{LOADDIFF}} = 50 \Omega$			51		dBµV	
OUTPUT								
Output \/alta as Lliah		Output a structure t	I _{SOURCE} = 1.5A		VAV	_{/DD} - 1.8	Ň	
Output Voltage High		Output saturated ISOURCE = 300mA			V _{AV}	_{DD} – 0.5	- V	
	ISINK = 1.5A			+1.8				
Output Voltage Low		Output saturated	I _{SINK} = 300mA	+0.5			V	
Drive Capability		V _{AVDD} = 12V, R _{LOAD}	= 2Ω differential		125.6		dBµV _{RMS}	

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Electrical Characteristics (continued)

 $(V_{DVDD} = 3V, V_{AVDD} = 12V \text{ or } 18V, V_{AGND} = V_{DGND} = 0V, V_{IN+} = V_{IN-} = V_{AVDD}/2, G0 = V_{DVDD}, G1 = V_{DVDD}, TXEN = V_{DVDD}, MODE = V_{DVDD}, R_{SET} = 29k\Omega, R_{LOAD} = 50\Omega$ differential from OUT+ to OUT-, $T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 2)

PARAMETER	SYMBOL	CONE	DITIONS	MIN	TYP	MAX	UNITS
Output Impedance	ZO	Single ended	TXEN = 1, f _C = 100kHz		500		mΩ
			TXEN = 0		35		kΩ
		G1 = 0, G0 = 0			6		
Gain		G1 = 0, G0 = 1			12		V/V
Galli		G1 = 1, G0 = 0			15		V/V
		G1 = 1, G0 = 1			18		
Input Noise-Voltage Density	e _n	f = 100kHz, G1 = 0, G	0 = 0, differential output		120		nV/√Hz
PROTECTION							
Overtemperature Shutdown Threshold	T _{OVTS}				+160		°C
Overtemperature Shutdown Threshold Hysteresis					15		°C
Overtemperature Warning Threshold	T _{OVTW}				+150		°C
Output Current Limit	ILIM	R _{RSET} = 29kΩ			2		А
DIGITAL INPUT CHARACTERI	STICS (G1, G	0, TXEN, MODE)					
Input Voltage Range				0		V _{DVDD}	V
Input Voltage High	VIH			0.7 x V _D	VDD		V
Input Voltage Low	VIL				0.3	x V _{DVDD}	V
Input Capacitance	C _{IN}				10		pF
Input Hysteresis Voltage	V _{HYS}				0.15		V
DIGITAL OUTPUT CHARACTE		ATUS1, STATUSO)		•			
Output Voltage Low (Active)	V _{ST}	I _{SINK} = 3mA				0.4	V
Output Leakage (Inactive)	I _{STZ}				± 0.1	± 1	μA
Output Capacitance (Inactive)	C _{STZ}				10		pF

Note 2: Min/max values are 100% production tested at T_A = +25°C. Specifications over temperature are guaranteed by design.

Note 3: Linearity specification limits are quasi-peak.

Note 4: The device is loaded with the network shown in Figure 3 and Figure 4.

Note 5: The device is driven with a typical CENELEC OFDM signal.

Note 6: The device is driven with a typical ARIB OFDM signal. It is loaded with the network shown in Figure 5. Signals are measured at the measurement port but calculated as signals over the entire resistive load.

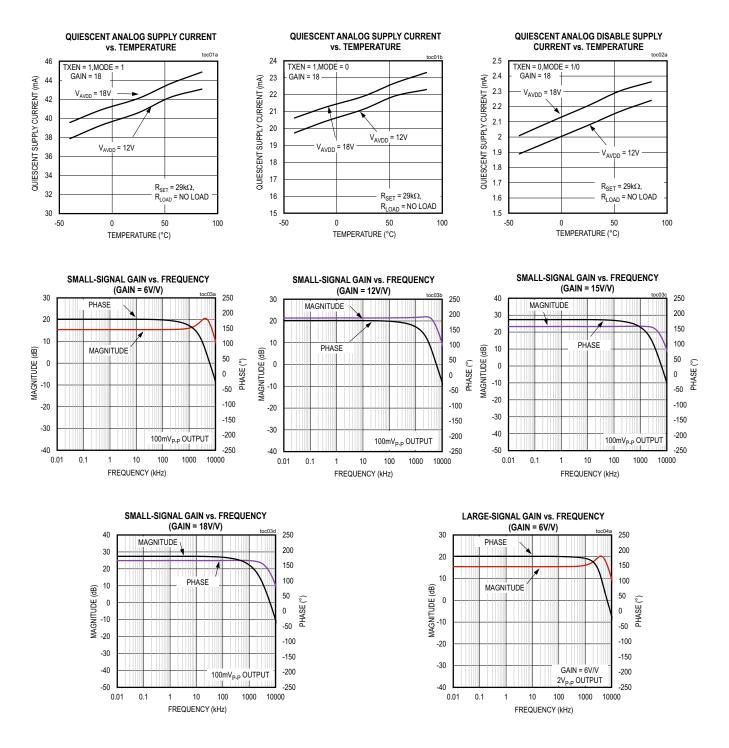
Note 7: External DC blocking capacitors are required.

Note 8: Measured with 200Hz bandwidth up to 150kHz and 9kHz bandwidth above 150kHz as specified in EN50065-1:2011.

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Typical Operating Characteristics

 $(V_{DVDD} = 3V, V_{AVDD} = 12V \text{ or } 18V, V_{AGND} = V_{DGND} = 0V, IN+ = IN- = V_{AVDD}/2, G0 = V_{DVDD}, G1 = V_{DVDD}, T_{XEN} = V_{DVDD}, MODE = V_{DVDD}, R_{SET} = 29k\Omega, R_{LOAD} = 50\Omega$ differential from OUT+ to OUT-, $T_A = +25^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

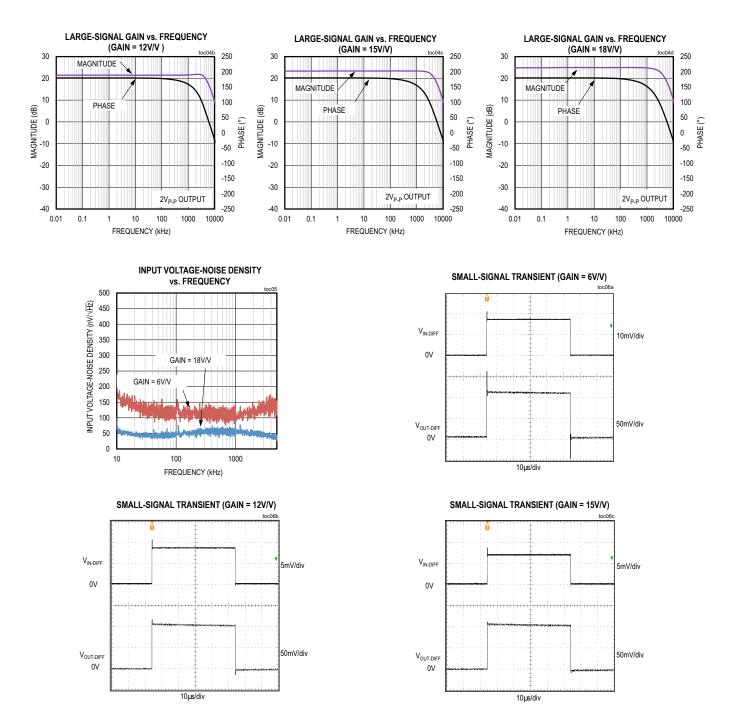


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Typical Operating Characteristics (continued)

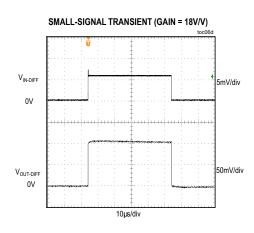
 $(V_{DVDD} = 3V, V_{AVDD} = 12V \text{ or } 18V, V_{AGND} = V_{DGND} = 0V, IN+ = IN- = V_{AVDD}/2, G0 = V_{DVDD}, G1 = V_{DVDD}, T_{XEN} = V_{DVDD}, MODE = V_{DVDD}, R_{SET} = 29k\Omega, R_{LOAD} = 50\Omega$ differential from OUT+ to OUT-, $T_A = +25^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

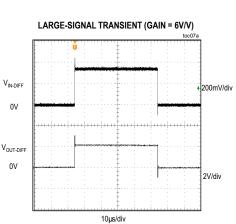


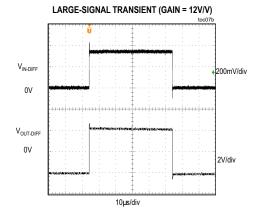
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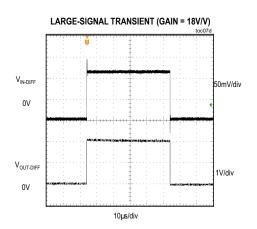
Typical Operating Characteristics (continued)

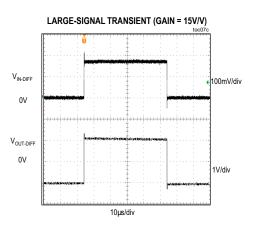
 $(V_{DVDD} = 3V, V_{AVDD} = 12V \text{ or } 18V, V_{AGND} = V_{DGND} = 0V, IN+ = IN- = V_{AVDD}/2, G0 = V_{DVDD}, G1 = V_{DVDD}, T_{XEN} = V_{DVDD}, MODE = V_{DVDD}, R_{SET} = 29k\Omega, R_{LOAD} = 50\Omega$ differential from OUT+ to OUT-, $T_A = +25^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

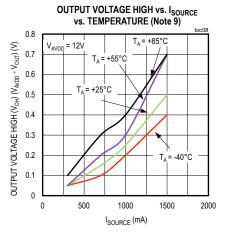










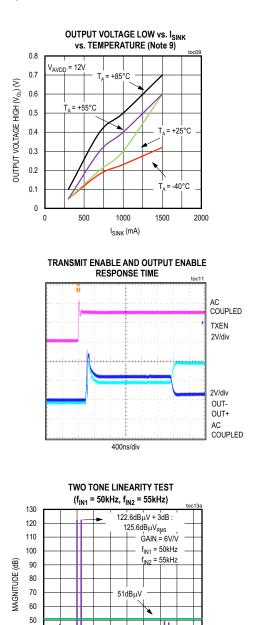


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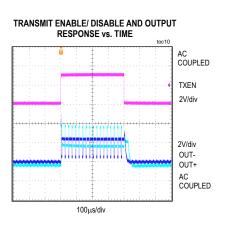
Typical Operating Characteristics (continued)

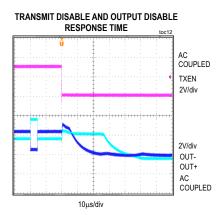
 $(V_{DVDD} = 3V, V_{AVDD} = 12V \text{ or } 18V, V_{AGND} = V_{DGND} = 0V, IN+ = IN- = V_{AVDD}/2, G0 = V_{DVDD}, G1 = V_{DVDD}, T_{XEN} = V_{DVDD}, MODE = V_{DVDD}, R_{SET} = 29k\Omega, R_{LOAD} = 50\Omega$ differential from OUT+ to OUT-, $T_A = +25^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

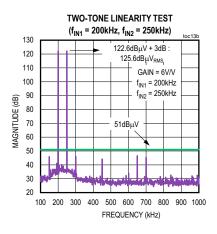


90 110 130 150 170 190

FREQUENCY (kHz)







Note 9: Output voltage high and output voltage low tests were performed by providing an AC input pulse with 1% duty cycle to saturate the outputs and obtain the results.

40

30

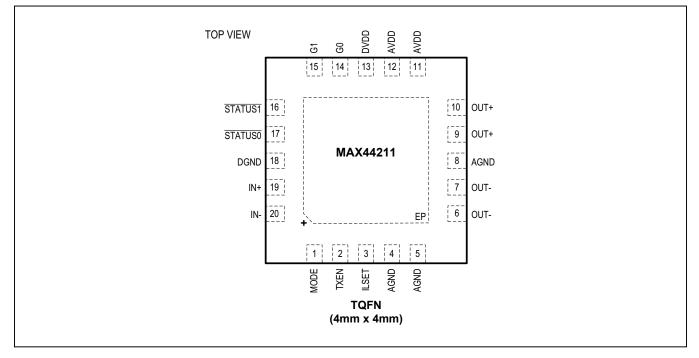
20

10

30 50 70

High-Current Differential Line Driver for Powerline Communications

Pin Configuration



Pin Description

PIN	NAME	FUNCTION
1	MODE	MODE Input. Leave MODE input unconnected to select low quiescent current mode to support CENELEC applications. Pull MODE input high to select higher quiescent current mode to support FCC and ARIB applications. See the <i>Bias Selection</i> section for more details.
2	TXEN	Transmit Enable. Pull TXEN high to enable the amplifier outputs. Tie TXEN to DGND to disable the amplifier outputs. See the <i>Output Enable</i> section.
3	ILSET	Current-Limit Setting Input. Connect a resistor between ILSET and AGND to set the current limit for the outputs. See the <i>Protection and Diagnostics</i> section.
4, 5, 8	AGND	Analog Ground
6, 7	OUT-	Negative Signal Output
9, 10	OUT+	Positive Signal Output
11, 12	AVDD	Amplifier Analog Power Supply. Bias AVDD to AGND between 8V to 20V.
13	DVDD	Digital Power Supply. Bias DVDD to DGND between 2.7V to 5.5V.
14	G0	Variable Gain Amplifier Gain Select Input 0. See Table 1 for details
15	G1	Variable Gain Amplifier Gain Select Input 1. See Table 1 for details.

High-Current Differential Line Driver for Powerline Communications

PIN	NAME	FUNCTION
16	STATUS1	Open-Drain Active-Low Status Output1. See Table 2 for details.
17	STATUS0 Open-Drain Active-Low Status Output0. See Table 2 for details.	
18	DGND	Digital Ground
19	IN+	Positive Signal Input
20	IN- Negative Signal Input	
_	EP	Exposed Pad. Internally connected to AGND. Connect the EP to a copper plane to enhance thermal dissipation.

Pin Description (continued)

Detailed Description

A common technique used to couple OFDM signals to the AC powerline is through a signal transformer. A line driver is needed to provide adequate levels of current and voltage to drive the varying loads that exist on today's powerlines.

The MAX44211 line driver is efficient low-distortion power amplifiers that drive high current to the low-impedance loads common in powerline communication (PLC) applications. The output stage is designed to linearly drive up to 1.5A peak current and a differential voltage of up to $36V_{P-P}$. They feature very intermodulation distortion to meet the demanding requirements of today's standards.

The MAX44211 also features two open-drain diagnostic outputs, STATUS1 and STATUS0. These act as flags to indicate the status of the part. Another significant feature of the MAX44211 is its thermal monitoring and shutdown capability. This allows the device to alert the host micro-controller of high temperature situations and then to automatically shut down to prevent damage should the temperature rise further.

Output Enable

Enable the MAX44211 output by pulling TXEN high. The amplifier outputs are fully enabled $2.4\mu s$ (typ) after TXEN is pulled high. If a signal is applied to the IN+ and IN-inputs during the startup time, it may be distorted and the linearity specifications may not be met.

Bias MODE Selection

The MAX44211 linearity can be improved at the expense of quiescent current. To meet CENELEC linearity requirements and save power, the MODE input is left unconnected. To improve linearity and meet ARIB or FCC requirements, the MODE input is pulled high.

Gain Selection and Output Connection

Set the MAX44211 overall gain using the G1 and G0 inputs (<u>Table 1</u>). The outputs, OUT+ and OUT- are internally biased at $V_{AVDD}/2$ to allow for maximum voltage swing. Therefore, the output should be A.C. coupled to the coupling transformer to avoid DC currents flowing in the transformer. In addition, the secondary side of the transformer should also be AC-coupled to avoid shorting the line input.

Protection and Diagnostics

The MAX44211 has two diagnostic status outputs: STATUS0 and STATUS1. These are open drain outputs that indicate the status of the device as shown in Table 2.

Both of the MAX44211 outputs are current limited. Set the output current limit according to the following equation.

$$I_{\text{LIM}} = \frac{60}{R_{\text{SET}} + 1}$$

While R_{SET} is in k Ω and I_{LIM} is in ampere *(see Typical Operating Circuit)*.

Do not use R_{SET} values below $23k\Omega$ or above $250k\Omega$. Note that the tolerance for the current limit is $\pm 30\%$ so care must be taken to ensure that the limit is set high enough to avoid clipping at peak loads. In addition, if the current limit is set too high, the device may enter thermal shutdown mode.

Table 1. Voltage Gain Selection

G1	G0	GAIN (V/V)
0	0	6
0	1	12
1	0	15
1	1	18

If the device tries to drive current in excess of the programmable, threshold, it will limit at the threshold level. This will be indicated by the STATUS1 and STATUS0 outputs as shown in <u>Table 2</u>.External protection for the line driver is required. Schottky diodes like B320A or B350A protect the outputs from the back EMF generated by the coupler/isolator connected to the AC mains. TVS diodes on the primary and the secondary side of the coupler help suppress the high-voltage transient spikes from the AC mains to the MAX44211 outputs.

Table 2. Diagnostic Status LineDefinitions

STATUS1	STATUS0	DEVICE STATUS
0	0	Overtemperature Shutdown Active
0	1	High Temperature Warning Active
1	0	Overcurrent Active
1	1	Normal Operation

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Temperature Protection

A typical PLC signal is shown in Figure 1 continuously driving a 2 Ω load. During this condition, the internal die temperature of the MAX44211 will eventually go beyond 160°C and the part enters into shutdown due to overheating. Once the part shuts down, the internal die temperature cools down enough to reach the 15°C hysteresis within milliseconds and the part then comes out of shutdown. If the load remains 2 Ω , similar conditions are seen and the process becomes cyclic. This internal temperature protection circuit regulates the temperature thereby avoiding a thermal breakdown of the MAX44211.

As shown in Figure 1, the input channel ((IN+) - (IN-)) depicts a typical PLC packet signal transmitted every 75ms. In the middle of the third packet transmission at about 175ms when driving a 125dBµV signal into a 2Ω load, the MAX44211 enters overtemperature shutdown for about 35ms. STAT0 and STAT1 go low and the outputs are disabled. After the internal die cools down by 15°C, the part turns on driving 125dBµV into 2Ω load enters into overtemperature warning immediately (STAT1: LOW and STAT0 : HIGH).

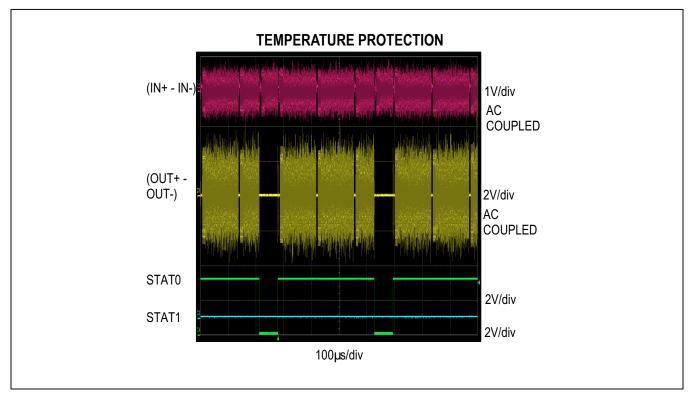


Figure 1. Temperature Protection

High-Current Differential Line Driver for Powerline Communications

Thermal Design

In PLC applications, the driver is required to drive high currents into potentially low-impedance AC lines. These conditions cause instantaneous power dissipation of several watts, resulting in heating of the driver. Thermal heat flow can be modeled in a similar way to current flow in an electrical circuit. See Figure 2.

Heat flows from the die through a thermal resistance, R_{JC} , to the case and through $R\theta_{CA}$ to the ambient outside world. These two thermal resistances are lumped together and specified as $R\theta_{JA}$. In addition to these thermal resistances, there are thermal capacitances. Therefore, the die will take a certain time to heat since the thermal capacitances need to be "charged." The die temperature is calculated using the following equation:

$$T_{J} = T_{A} + (R\theta_{JA} \times P_{D})$$

If the device were to be mounted on a four-layer board with a large copper area and dissipate 3.5W at an ambient temperature of 85°C, the steady-state die temperature would be 200.5°C. This would obviously cause a problem. However, the PLC application, the device is only required to be in transmit mode infrequently and for a small time. Therefore, since the thermal flow model is analogous to an electrical lowpass filter, the die temperature does not rise significantly.

It is recommended that the device be mounted on a fourlayer board where the exposed heat paddle is soldered to the board. Further, it is recommended that as many via holes as possible are positioned in the pad to allow heat to be conducted through the PCB to a large heatsink area on the reverse side of the board.

If, however, the devices do overheat through some system fault, they have a diagnostic monitor to avoid damage. The device have two temperature thresholds: Warning (150°C) and Shutdown (160°C). If the warning threshold is crossed, the device will indicate this by pulling the <u>STATUS1</u> output low while keeping <u>STATUS0</u> output

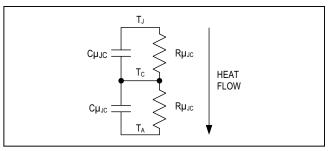


Figure 2. Thermal Flow Model

high. If the device heats further and crosses the shutdown threshold, both STATUS_ outputs assert low and the device will automatically shut down and remain in that condition until it cools below 145°C (shutdown threshold– shutdown hysteresis). When in thermal shutdown mode, both power amplifier outputs (OUT+, OUT-) are shut down and in a high-impedance state.

Power Supplies

The MAX44211 operates from separate analog and digital power supplies. DVDD is the digital supply and should be connected to the same supply as the host processor. The logic thresholds for the digital input lines are DVDD related and therefore no level translators are required to interface between the host and the MAX44211. DGND should be connected back to the host power supply.

The analog section including the power amplifier operates from a single unipolar supply, AVDD. AGND should be connected back to the analog supply.

AGND and DGND should be connected together in a star formation for best noise performance.

The exposed pad, EP, is internally connected to AGND and should be connected to a large copper area for best heat dissipation.

High-Current Differential Line Driver for Powerline Communications

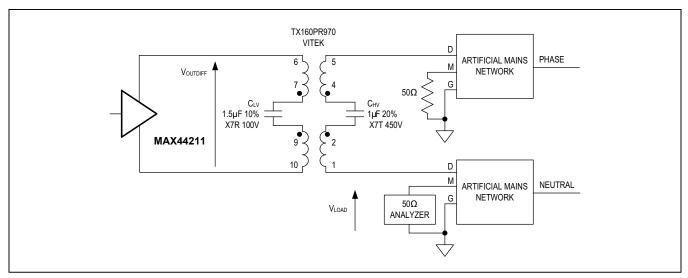
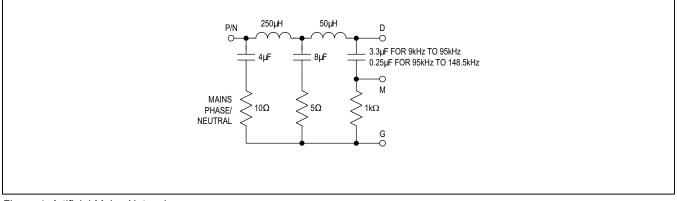


Figure 3. Test Structure for Cenelec





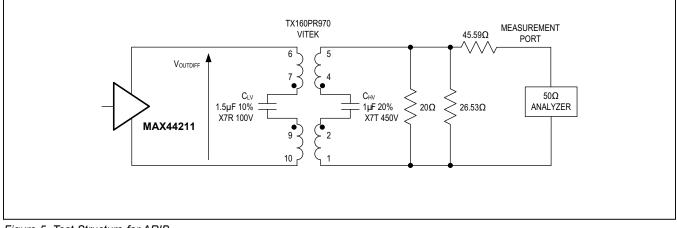
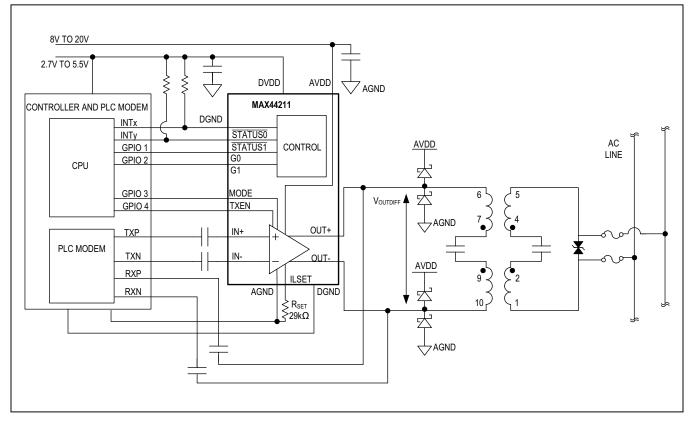


Figure 5. Test Structure for ARIB

High-Current Differential Line Driver for Powerline Communications

Typical Operating Circuit



Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	
MAX44211ETP+	-40°C to +85°C	20 TQFN-EP*	

+Denotes a lead(Pb)-free/RoHS-compliant package.

*EP = Exposed pad.

Chip Information

PROCESS: CMOS

Package Information

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
20 TQFN-EP	T2044+4C	<u>21-0139</u>	<u>90-0409</u>

High-Current Differential Line Driver for Powerline Communications

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	6/15	Initial release	—
1	9/17	Updated Absolute Maximum Ratings section	2

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