# Block Diagrams

# 1. S-8351 Series

(1) A Type (With shutdown function)





(2) B and C Types (Without shutdown function,  $V_{DD}$  /  $V_{OUT}$  non-separate type)





(3) D Type ( $V_{DD}$  /  $V_{OUT}$  separate type)



Figure 3

# ABLIC Inc.

# 2. S-8352 Series

(1) A Type (With shutdown function)



(2) B and C Type (Without Shutdown function,  $V_{DD} / V_{OUT}$  non-separate type)





(3) D Type ( $V_{DD}$  /  $V_{OUT}$  separate type)



# Product Name Structure

The product types, output voltage, and packages for the S-8351/8352 Series can be selected at the user's request. Please refer to the "3. Product Name" for the definition of the product name, "4. Package" regarding the package drawings and "5. Product Name List" for the full product names.

# 1. Function List

# (1) Built-in Power MOS FET Type

Product Name	Controll system	Duty ratio [%]	Switching frequency [kHz]	Shutdown function	V <sub>DD</sub> / V <sub>OUT</sub> separate type	Package	Application
S-8351AxxMC	PFM	50 / 75	100	Yes	-	SOT-23-5	Applications requiring shutdown function
S-8351BxxMA	PFM	50 / 75	100	-	-	SOT-23-3	Applications not requiring shutdown function
S-8351CxxMA	PFM	75	100	-	-	SOT-23-3	Applications not requiring shutdown function
S-8351CxxUA	PFM	75	100	-	-	SOT-89-3	Applications not requiring shutdown function
S-8351DxxMC	PFM	50 / 75	100	_	Yes	SOT-23-5	Applications in which output voltage is adjusted by external resistor

Table 1

# (2) External Power MOS FET Type

### Table 2

Product Name	Controll System	Duty Ratio [%]	Switching Frequency [kHz]	Shutdown Function	V <sub>DD</sub> / V <sub>OUT</sub> Separate Type	Package	Application
S-8352AxxMC	PFM	50 / 75	100	Yes	-	SOT-23-5	Applications requiring shutdown function
S-8352BxxMA	PFM	50 / 75	100	-	-	SOT-23-3	Applications not requiring shutdown function
S-8352CxxMA	PFM	75	100	-	-	SOT-23-3	Applications not requiring shutdown function
S-8352CxxUA	PFM	75	100	-	-	SOT-89-3	Applications not requiring shutdown function
S-8352DxxMC	PFM	50 / 75	100	_	Yes	SOT-23-5	Applications in which output voltage is adjusted by external resistor

# 2. Package and Function List by Product Type

Table 3

Series Name	Туре	Package Name (Abbreviation)	Shutdown Function Yes / No	V <sub>DD</sub> / V <sub>OUT</sub> Separate Type Yes / No
	A (Duty ratio 50% / 75% auto-switching type) A = 100 kHz	MC	Yes	No
S-8351 Series,	B (Duty ratio 50% / 75% auto-switching type) B = 100 kHz	MA	No	No
S-8352 Series	C (Duty ratio 75% fixed type) C = 100 kHz	MA / UA	No	No
	D (Duty ratio 50% / 75% auto-switching type) D = 100 kHz	MC	No	Yes

### 3. Product Name

# (1) SOT-23-3



\*1. Refer to the tape specifications.

\*2. Refer to the Table 4, 5 in the "5. Product Name List".

#### STEP-UP, BUILT-IN / EXTERNAL FET PFM CONTROL SWITCHING REGULATOR / SWITCHING REGULATOR CONTROLLER S-8351/8352 Series Rev.3.1\_00



\*1. Refer to the tape specifications.

\*2. Refer to the Table 4, 5 in the "5. Product Name List".

# 4. Package

Deskare Name	Drawing Code					
Package Name	Package	Таре	Reel			
SOT-23-3	MP003-A-P-SD	MP003-A-C-SD	MP003-A-R-SD			
SOT-23-5	MP005-A-P-SD	MP005-A-C-SD	MP005-A-R-SD			
SOT-89-3	UP003-A-P-SD	UP003-A-C-SD	UP003-A-R-SD			

#### 5. Product Name List

#### (1) S-8351 Series

Table 4

Output voltage	S-8351AxxMC Series	S-8351BxxMA Series	S-8351CxxMA Series	S-8351CxxUA Series	S-8351DxxMC Series
1.5 V	_	_	_	_	S-8351D15MC-J8AT2x
2.0 V	S-8351A20MC-J2FT2x	S-8351B20MA-J4FT2G	-	-	S-8351D20MC-J8FT2x
2.2 V	S-8351A22MC-J2HT2x	-	-	-	-
2.5 V	S-8351A25MC-J2KT2x	S-8351B25MA-J4KT2G	-	S-8351C25UA-J6KT2x	-
2.6 V	S-8351A26MC-J2LT2x	_	_	-	_
2.7 V	S-8351A27MC-J2MT2x	S-8351B27MA-J4MT2G	_	-	_
2.8 V	S-8351A28MC-J2NT2x	_	_	-	_
3.0 V	S-8351A30MC-J2PT2x	S-8351B30MA-J4PT2G	_	S-8351C30UA-J6PT2x	S-8351D30MC-J8PT2x
3.1 V	-	-	-	S-8351C31UA-J6QT2x	-
3.2 V	S-8351A32MC-J2RT2x	-	-	S-8351C32UA-J6RT2x	-
3.3 V	S-8351A33MC-J2ST2x	S-8351B33MA-J4ST2G	S-8351C33MA-J6ST2G	S-8351C33UA-J6ST2x	-
3.5 V	S-8351A35MC-J2UT2x	-	-	S-8351C35UA-J6UT2x	-
4.0 V	S-8351A40MC-J2ZT2x	-	-	-	S-8351D40MC-J8ZT2x
4.5 V	S-8351A45MC-J3ET2x	S-8351B45MA-J5ET2G	-	-	-
4.6 V	-	S-8351B46MA-J5FT2G	-	-	-
4.7 V	S-8351A47MC-J3GT2x	-	-	-	-
5.0 V	S-8351A50MC-J3JT2x	S-8351B50MA-J5JT2G	-	S-8351C50UA-J7JT2x	S-8351D50MC-J9JT2x
5.5 V	S-8351A55MC-J3OT2x	S-8351B55MA-J5OT2G	_	-	_
5.6 V	S-8351A56MC-J3PT2x	_	_	_	-
6.0 V	S-8351A60MC-J3TT2x	-	-	-	S-8351D60MC-J9TT2x

#### (2) S-8352 Series

#### Table 5

Output	S-8352AxxMC	S-8352BxxMA	S-8352CxxUA	S-8352DxxMC
voltage	Series	Series	Series	Series
2.0 V	_	-	-	S-8352D20MC-K8FT2x
2.5 V	S-8352A25MC-K2KT2x	-	-	-
3.0 V	S-8352A30MC-K2PT2x	S-8352B30MA-K4PT2G	S-8352C30UA-K6PT2x	S-8352D30MC-K8PT2x
3.1 V	_	_	S-8352C31UA-K6QT2x	_
3.2 V	S-8352A32MC-K2RT2x	_	S-8352C32UA-K6RT2x	_
3.3 V	S-8352A33MC-K2ST2x	-	S-8352C33UA-K6ST2x	S-8352D33MC-K8ST2x
3.5 V	S-8352A35MC-K2UT2x	-	-	-
3.7 V	S-8352A37MC-K2WT2x	-	-	-
4.0 V	S-8352A40MC-K2ZT2x	_	_	_
4.6 V	S-8352A46MC-K3FT2x	_	_	_
4.7 V	S-8352A47MC-K3GT2x	_	_	_
5.0 V	S-8352A50MC-K3JT2x	S-8352B50MA-K5JT2G	S-8352C50UA-K7JT2x	_
5.4 V	S-8352A54MC-K3NT2x	_	_	_
5.6 V	_	_	S-8352C56UA-K7PT2x	_

**Remark 1.** Please contact the ABLIC Inc. marketing department for products with an output voltage other than those specified above.

- 2. x: G or U
- 3. Please select products of environmental code = U for Sn 100%, halogen-free products.

# Pin Configurations

SOT-23-3

Top view



Figure 7

# Table 6 S-8351 Series B and C Types (Without shutdown function, $V_{\text{DD}}$ / $V_{\text{OUT}}$ non-separate type)

Pin No.	Symbol	Pin Description
1	VOUT	Output voltage pin and IC power supply pin
2	VSS	GND pin
3	CONT	External inductor connection pin (Open-drain output)

Table 7 S-8352 Series B and C Types (Without shutdown function,  $V_{DD}$  /  $V_{OUT}$  non-separate type)

Pin No.	Symbol	Pin Description
1	VOUT	Output voltage pin and IC power supply pin
2	VSS	GND pin
3	EXT	External transistor connection pin (CMOS output)

SOT-23-5

Top view



Figure 8

Table 8 S-8351 Series A Type (With shutdown function,  $V_{\text{DD}}$  /  $V_{\text{OUT}}$  non-separate type)

Pin No.	Symbol	Pin Description
1	ON/OFF	Shutdown pin "H": Normal operation (Step-up operating) "L": Step-up stopped (Entire circuit stopped)
2	VOUT	Output voltage pin and IC power supply pin
3	NC <sup>*1</sup>	No connection
4	VSS	GND pin
5	CONT	External inductor connection pin (Open-drain output)

\*1. The NC pin indicates electrically open.

Table 9 S-8352 Series A Type (With shutdown function,  $V_{DD}$  /  $V_{OUT}$  non-separate type)

Pin No.	Symbol	Pin Description
1	ON/OFF	Shutdown pin "H": Normal operation (Step-up operating) "L": Step-up stopped (Entire circuit stopped)
2	VOUT	Output voltage pin and IC power supply pin
3	NC <sup>*1</sup>	No connection
4	VSS	GND pin
5	EXT	External transistor connection pin (CMOS output)

\*1. The NC pin indicates electrically open.

Pin No.	Symbol	Pin Description			
1	VOUT	Output voltage pin			
2	VDD	IC power supply pin			
3	NC <sup>*1</sup>	No connection			
4	VSS	GND pin			
5	CONT	External inductor connection pin (Open-drain output			

 $\label{eq:stable} \begin{array}{ll} \mbox{Table 10} & \mbox{S-8351 Series D Type} \\ \mbox{(Without shutdown function, $V_{DD}$ / $V_{OUT}$ separate type)} \end{array}$ 

**\*1.** The NC pin indicates electrically open.

Table 11 S-8352 Series D Type (Without shutdown function,  $V_{DD}$  /  $V_{OUT}$  separate type)

Pin No.	Symbol	Pin Description
1	VOUT	Output voltage pin
2	VDD	IC power supply pin
3	NC <sup>*1</sup>	No connection
4	VSS	GND pin
5	EXT	External transistor connection pin (CMOS output)

\*1. The NC pin indicates electrically open.

SOT-89-3 Top view



Figure 9

# Table 12 S-8351 Series C Type (Without shutdown function, $V_{\text{DD}}$ / $V_{\text{OUT}}$ non-separate type)

Pin No.	Symbol	Pin Description
1	VSS	GND pin
2	VOUT	Output voltage pin and IC power supply pin
3	CONT	External inductor connection pin (Open-drain output)

Table 13 S-8352 Series C Type (Without shutdown function,  $V_{DD}$  /  $V_{OUT}$  non-separate type)

Pin No.	Symbol	Pin Description
1	VSS	GND pin
2	VOUT	Output voltage pin and IC power supply pin
3	EXT	External transistor connection pin (CMOS output)

Table 14

# Absolute Maximum Ratings

			(Ta = 25°C unless othe	erwise specified)
Ite	m	Symbol	Absolute maximum rating	Unit
VOUT pin voltage		V <sub>OUT</sub>	$V_{SS}$ – 0.3 to $V_{SS}$ + 12	V
ON/OFF pin voltage	e <sup>*1</sup>	$V_{ON/\overline{OFF}}$	$V_{SS}{-}0.3$ to $V_{SS}{+}12$	V
VDD pin voltage *2		V <sub>DD</sub>	$V_{SS}$ – 0.3 to $V_{SS}$ + 12	V
CONT pin voltage		V <sub>CONT</sub>	$V_{SS}$ – 0.3 to $V_{SS}$ + 12	V
EVT nin voltage	D type		$V_{SS} - 0.3$ to $V_{DD} + 0.3$	V
	ge Others V <sub>EXT</sub> V		$V_{SS}$ – 0.3 to $V_{OUT}$ + 0.3	V
CONT pin current		I <sub>CONT</sub> 300		mA
EXT pin current	irrent I <sub>EXT</sub> ±50		mA	
	COT 22 2		150 (When not mounted on board)	mW
	501-23-3		430 <sup>*3</sup>	mW
Dever dissinction	COT 22 5	$\begin{array}{c c c c c c c c c c c c c c c c c c c $	mW	
Power dissipation	SOT-23-5	PD	600 <sup>*3</sup>	mW
			500 (When not mounted on board)	mW
	SOT-89-3		1000 <sup>*3</sup>	mW
Operating ambient te	mperature	T <sub>opr</sub>	- 40 to + 85	
Storage temperature		T <sub>stg</sub>	- 40 to + 125	°C

\*1. With shutdown function

\*2. For V<sub>DD</sub> / V<sub>OUT</sub> separate type

**\*3.** When mounted on board

[Mounted board]

(1) Board size : 114.3 mm  $\times$  76.2 mm  $\times$  t1.6 mm

(2) Board name : JEDEC STANDARD51-7

Caution The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

### (1) When mounted on board



#### (2) When not mounted on board



Figure 10 Power Dissipation of Packages

# Electrical Characteristics

# (1) S-8351 Series

# Table 15 (1 / 2)

(1a = 25°C unless otherwise specified)								
Item	Symbol	Condition	1	Min.	Тур.	Max.	Unit	Measurement circuit
Output voltage	V <sub>OUT</sub>			$\begin{array}{c} V_{OUT(S)} \\ \times 0.976 \end{array}$	V <sub>OUT(S)</sub>	$\begin{array}{c} V_{\text{OUT(S)}} \\ \times  1.024 \end{array}$	V	1
Input voltage	V <sub>IN</sub>	_			_	10	V	1
Operation start voltage	V <sub>ST1</sub>	I <sub>OUT</sub> = 1 mA		-	_	0.9	V	1
Oscillation start voltage	V <sub>ST2</sub>	No external parts, Voltage applied to $V_{OUT}$ , CONT pin pulled up to $V_{OUT}$ with 300 $\Omega$ resistor		-	-	0.8	V	2
			S-8351x15 to 29	-	8.5	-	μA	1
Input current during no- load	I <sub>IN</sub>	I <sub>OUT</sub> = 0 mA	S-8351x30 to 49	-	9.0	-	μA	1
			S-8351x50 to 65	-	9.5	-	μA	1
			S-8351x15 to 19	-	9.6	16.0	μA	2
			S-8351x20 to 29	-	15.7	26.2	μA	2
Current concumption 1	1		S-8351x30 to 39	-	23.2	38.6	μA	2
Current consumption 1	ISS1	$v_{OUT} = v_{OUT(S)} \times 0.95$	S-8351x40 to 49	-	32.0	53.3	μA	2
			S-8351x50 to 59	-	42.1	70.2	μA	2
			S-8351x60 to 65	-	54.9	91.5	μA	2
			S-8351x15 to 19	-	2.3	3.5	μA	2
			S-8351x20 to 29	-	2.5	3.8	μA	2
		N/ N/ 05N/	S-8351x30 to 39	-	2.7	4.1	μA	2
Current consumption 2	I <sub>SS2</sub>	$v_{OUT} = v_{OUT(S)} + 0.5 v$	S-8351x40 to 49	_	2.9	4.4	μA	2
			S-8351x50 to 59	-	3.1	4.7	μA	2
			S-8351x60 to 65	_	3.3	5.1	μA	2
Current consumption during shutdown (With shutdown function)	I <sub>SSS</sub>	V <sub>ON/OFF</sub> = 0 V		-	-	0.5	μA	2
· · · · · · · · · · · · · · · · · · ·			S-8351x15 to 19	50.2	91.2	-	mA	2
			S-8351x20 to 24	65.0	118.2	_	mA	2
			S-8351x25 to 29	78.5	142.7	_	mA	2
Switching current	sw	$V_{CONT} = 0.4 V$	S-8351x30 to 39	90.7	164.8	_	mA	2
Ũ		00111	S-8351x40 to 49	110.9	201.6	_	mA	2
			S-8351x50 to 59	125.7	228.6	_	mA	2
			S-8351x60 to 65	135.2	245.8	_	mA	2
Switching transistor leakage current	I <sub>SWQ</sub>	No external parts, $V_{CONT} = V_{OUT} = 10 V$ , $V_{ON/\overline{OFF}} = 0 V$		_	-	0.5	μA	2
CONT pin limit voltage	V <sub>CONTLMT</sub>	Apply to CONT pin, Confin	Apply to CONT pin, Confirm oscillation stop		0.9	-	V	2
Line regulation	$\Delta V_{OUT1}$	$V_{\rm IN} = V_{\rm OUT(S)} \times 0.4$ to $\times 0.6$		_	30	60	mV	1
Load regulation	$\Delta V_{OUT2}$	$I_{OUT} = 10 \mu\text{A to } V_{OUT(S)} / 250 \times 1.25$		-	30	60	mV	1
Output voltage temperature coefficient	$\frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}}$	$Ta = -40^{\circ}C$ to $+85^{\circ}C$		-	±50	-	ppm / °C	1
Oscillation frequency	f <sub>OSC</sub>	V <sub>OUT</sub> = V <sub>OUT(S)</sub> × 0.95, Measured waveform at CONT pin		90	100	110	kHz	2
Duty ratio 1	Duty1	$V_{OUT} = V_{OUT(S)} \times 0.95$ , Measured waveform at CC	ONT pin	70	75	80	%	2
Duty ratio 2 (For A, B, D type)	Duty2	Measured waveform at CC	ONT pin at light load	_	50	-	%	1

Table 15 (2 / 2)

				(1	a = 25°	C unles	s otherw	ise specified)
Item	Symbol	Condition		Min.	Тур.	Max.	Unit	Measurement circuit
	V <sub>SH</sub>	V <sub>OUT</sub> = V <sub>OUT(S)</sub> × 0.95, Measured oscillation at CONT pin		0.75	-	Ι	V	2
(With shutdown function)	V <sub>SL1</sub>	$V_{OUT} = V_{OUT(S)} \times 0.95$ ,	At V <sub>OUT</sub> ≥1.5 V	-	-	0.3	V	2
()	V <sub>SL2</sub>	CONT pin	At $V_{OUT}$ <1.5 V	-	-	0.2	V	2
ON/OFF pin input current	I <sub>SH</sub>	$V_{ON/\overline{OFF}} = 10 V$		- 0.1	-	0.1	μA	2
(With shutdown function)	I <sub>SL</sub>	$V_{ON/\overline{OFF}} = 0 V$		- 0.1	-	0.1	μA	2
L'étaione (			S-8351x30	-	86	-	%	1
Ellicency		EFFI – –		_	88	_	%	1

External parts

Coil: CDRH6D28-101 (100 µH) of Sumida Corporation

Diode: MA2Z748 (Shottky type) of Matsushita Electric Industrial Co., Ltd.

Capacitor: F93 (16 V, 47 µF tantalum type) of Nichicon Corporation

 $V_{IN} = V_{OUT(S)} \times 0.6$  applied,  $I_{OUT} = V_{OUT(S)} / 250 \Omega$ 

With shutdown function :  $ON/\overline{OFF}$  pin is connected to  $V_{OUT}$ 

For  $V_{DD}$  /  $V_{OUT}$  separate type : VDD pin is connected to VOUT pin

**Remark 1.** V<sub>OUT(S)</sub> specified above is the set output voltage value, and V<sub>OUT</sub> is the typical value of the actual output voltage.

2.  $V_{DD} / V_{OUT}$  separate type

A step-up operation is performed from  $V_{DD} = 0.8 \text{ V}$ . However, 1.8 V $\leq$ V<sub>DD</sub><10 V is recommended stabilizing the output voltage and oscillation frequency. (V<sub>DD</sub> $\geq$ 1.8 V must be applied for products with a set value of less than 1.9 V.)

#### (2) S-8352 Series

(Ta = 25°C unless otherwise specified) Measurement Item Symbol Condition Min. Typ. Max. Unit circuit V<sub>OUT(S)</sub> V<sub>OUT(S)</sub> V<sub>OUT(S)</sub> Output voltage VOUT V 3 ×0.976 × 1.024 VIN ٧ Input voltage \_ 10 3  $I_{OUT} = 1 \text{ mA}$ Operation start voltage V<sub>ST1</sub> 0.9 ٧ 3 Oscillation start voltage No external parts, Voltage applied to VOUT 0.8 V 4 V<sub>ST2</sub> \_ \_ S-8352x15 to 19 7.4 12.3 4 μA \_ 12.0 20.0 S-8352x20 to 29 \_ μΑ 4 S-8352x30 to 39 17.8 29.6 μA 4 \_  $I_{SS1}$ Current consumption 1  $V_{OUT} = V_{OUT(S)} \times 0.95$ 24.7 4 S-8352x40 to 49 \_ 41.1 μA 4 S-8352x50 to 59 \_ 32.7 54.5 μA 43.0 71.6 μA 4 S-8352x60 to 65 2.3 4 3.5 μΑ S-8352x15 to 19 S-8352x20 to 29 2.5 3.8 μΑ 4 μA 2.7 4.1 4 S-8352x30 to 39 \_ Current consumption 2  $V_{OUT} = V_{OUT(S)} + 0.5 V$ I<sub>SS2</sub> 2.9 4.4 4 S-8352x40 to 49 \_ μΑ 4.7 4 S-8352x50 to 59 \_ 3.1 μA S-8352x60 to 65 \_ 3.3 5.1 μA 4 Current consumption during  $V_{ON/\overline{OFF}} = 0 V$ shutdown 0.5 4 \_ μA I<sub>SSS</sub> (With shutdown function) S-8352x15 to 19 - 3.5 -6.3 4 \_ mΑ S-8352x20 to 24 - 9.4 - 5.2 mΑ 4 S-8352x25 to 29 -6.8 - 12.3 mΑ 4  $V_{EXT} = V_{OUT} - 0.4 V$ S-8352x30 to 39 - 8.2 - 14.9 mΑ 4 **I**EXTH S-8352x40 to 49 - 10.7 - 19.4 4 mΑ \_ 22.8 4 S-8352x50 to 59 12.5 mΑ S-8352x60 to 65 - 13.9 - 25.2 mΑ 4 EXT pin output current S-8352x15 to 19 3.8 6.9 mΑ 4 S-8352x20 to 24 5.6 10.2 mΑ 4 S-8352x25 to 29 7.3 13.3 mΑ 4  $V_{EXT} = 0.4 V$ S-8352x30 to 39 8.9 16.2 mΑ 4 **I**EXTL \_ S-8352x40 to 49 11.6 21.1 mΑ 4 \_ S-8352x50 to 59 13.7 25.0 4 mΑ -S-8352x60 to 65 15.3 27.8 4 \_ mΑ  $V_{IN} = V_{OUT(S)} \times 0.4 \text{ to} \times 0.6$ Line regulation 60  $\Delta V_{OUT1}$ 30 mV 3 \_  $I_{OUT} = 10 \ \mu A \text{ to } V_{OUT(S)} / 100 \times 1.25$ 30 60 3 Load regulation  $\Delta V_{OUT2}$ \_ mV ΔVουτ Output voltage temperature  $Ta = -40^{\circ}C$  to  $+85^{\circ}C$ 3 ±50 ppm / °C \_ \_ ΔTa • Vout coefficient  $V_{OUT} = V_{OUT(S)} \times 0.95,$ Oscillation frequency  $\mathbf{f}_{\mathrm{OSC}}$ 90 100 4 110 kHz Measured waveform at EXT pin  $V_{OUT} = V_{OUT(S)} \times 0.95,$ Duty ratio 1 Duty1 70 75 80 % 4 Measured waveform at EXT Duty ratio 2 Dutv2 Measured waveform at EXT pin at light load \_ 50 % 3 (For A, B, D type)

				(T	a = 25°	C unles	s otherw	ise specified)
Item	Symbol	Condition		Min.	Тур.	Max.	Unit	Measurement circuit
ON/OFF pin input voltage (With shutdown function)	V <sub>SH</sub>	$V_{OUT} = V_{OUT(S)} \times 0.95$ , Measured oscillation at EXT		0.75	_	_	V	4
	V <sub>SL1</sub>	$V_{OUT} = V_{OUT(S)} \times 0.95$ ,	At V <sub>OUT</sub> ≥1.5 V			0.3	V	4
	V <sub>SL2</sub>	EXT pin	At V <sub>OUT</sub> <1.5 V			0.2	V	4
ON/OFF pin input current	current $I_{SH}$ $V_{ON/\overline{OFF}} = 10 V$			- 0.1	_	0.1	μA	4
(With shutdown function)	I <sub>SL</sub>	V <sub>ON/OFF</sub> = 0 V		- 0.1		0.1	μA	4
Efficiency			S-8352x30	<u> </u>	83	<u> </u>	%	3
Eniciency			S-8352x50	- '	85	- I	%	3

Table 16 (2 / 2)

External parts

Coil:	CDRH6D28-101 (100 μH) from Sumida Corporation
Diode:	MA2Z748 (Shottky type) from Matsushita Electric Industrial Co., Ltd
Capacitor:	F93 (16 V, 47 $\mu$ F tantalum type) from Nichicon Corporation
Transistor:	CPH3210 from Sanyo Electric Co., Ltd.
Base resistor (Rb):	1 kΩ
Base capacitor (Cb):	2200 pH (ceramic type)
$V_{a} = 0.6$ applied $I_{a}$	$v_{\rm T} = V_{\rm OUT(0)} / 100  \Omega$

 $V_{IN} = V_{OUT(S)} \times 0.6$  applied,  $I_{OUT} = V_{OUT(S)}$  / 100  $\Omega$ 

With shutdown function :  $ON/\overline{OFF}$  pin is connected to  $V_{OUT}$ 

For  $V_{\text{DD}}$  /  $V_{\text{OUT}}$  separate type : VDD pin is connected to VOUT pin

**Remark 1.**  $V_{OUT(S)}$  specified above is the set output voltage value, and  $V_{OUT}$  is the typical value of the actual output voltage.

**2.**  $V_{DD}$  /  $V_{OUT}$  separate type

A step-up operation is performed from  $V_{DD} = 0.8 \text{ V}$ . However, 1.8 V $\leq$ V<sub>DD</sub><10 V is recommended stabilizing the output voltage and oscillation frequency. (V<sub>DD</sub> $\geq$ 1.8 V must be applied for products with a set value of less than 1.9 V.)

# Measurement Circuits















\*1. For  $V_{\text{DD}}$  /  $V_{\text{OUT}}$  separate type

\*2. With shutdown function

# Operation

### 1. Step-up DC-DC Converter

The S-8351/8352 Series is a DC-DC converter that uses a pulse frequency modulation method (PFM) and features low current consumption. This series is an especially efficient DC-DC converter at an output current of 100  $\mu$ A or lower.

In conventional fixed-duty PFM DC-DC converters, although a low duty ratio allows a lower ripple voltage when the current load is light, the efficiency is decreased when the output load current is large. Conversely, a high duty ratio increases the output load current and efficiency, but increases the ripple voltage when the output load current is low. In the A, B, and D types, the duty ratio is automatically switched 75% when the output load current is high to secure the load drive conclusion of 50% when the output load current is high to secure

the load drive capability and 50% when the output load current is low to control the load drive capability to decrease pulse skipping. This suppresses a drop in the ripple frequency, enabling control of the increase in the ripple voltage. The C type adopts a 75% fixed-duty PFM method. The ripple voltage increases more than that of the duty switching type with the load is low, but the efficiency is better.

In the A, B, and D types, the duty ratio is not rapidly changed, but rather smoothly switched in the intermediate area between 50% and 75%. Therefore, fluctuation of the ripple voltage caused by duty switching is minimized. **Figures 15, 16** show the ripple voltage characteristics versus the output current.



Figure 15 Output Current (I<sub>OUT</sub>) vs. Ripple Voltage (V<sub>rp-p</sub>) Characteristics



These figures show that the ripple voltage decreases as the output load current ( $I_{OUT}$ ) changes from large to small. The ripple voltage becomes particularly small when  $I_{OUT}$  is in the coil current discontinuous region of 20 mA or less.

# 2. ON/OFF Pin (Shutdown Pin) (A Type)

ON/OFF pin stops or starts step-up operation.

Setting the  $ON/\overline{OFF}$  pin to the "L" level stops operation of all the internal circuits and reduces the current consumption significantly.

DO NOT use the  $ON/\overline{OFF}$  pin in a floating state because it has the structure shown in **Figure 17** and is not pulled up or pulled down internally. DO NOT apply a voltage of between 0.3 V and 0.75 V to the  $ON/\overline{OFF}$  pin because applying such a voltage increases the current consumption. If the shutdown pin is not used, connect it to the VOUT pin.

The ON/OFF pin does not have hysteresis.

Table '	17
---------	----

ON/OFF pin	CR oscillation circuit	Output voltage
"H"	Operation	Fixed
"L"	Stop	≅V <sub>IN</sub> *1

\*1. Voltage obtained by subtracting the voltage drop due to the DC resistance of the inductor and the diode forward voltage from V<sub>IN</sub>.



Figure 17 ON/OFF Pin Structure

# 3. Operation

The following are the basic equations [(1) through (7)] of the step-up switching regulator. (Refer to Figure 18.)





Voltage at CONT pin at the moment M1 is turned ON (V\_A)  $^{\star 1}$  :

$$V_{A} = V_{S}^{*2}$$
(1)  
\*1. Current flowing through L (L) is zero

\*2. Non-saturated voltage of M1.

The change in  ${\rm I}_{\rm L}$  over time :

$$\frac{dIL}{dt} = \frac{VL}{L} = \frac{VIN - VS}{L}$$
(2)

Integration of equation (2) ( $I_L$ ) :

$$I_{L} = \left(\frac{V_{IN} - V_{S}}{L}\right) \bullet t$$
(3)

 $I_L$  flows while M1 is ON (t<sub>ON</sub>). The time of t<sub>ON</sub> is determined by the oscillation frequency of OSC.

The peak current ( $I_{PK}$ ) after  $t_{ON}$  :

$$I_{PK} = \left(\frac{V_{IN} - V_S}{L}\right) \bullet t_{ON}$$
(4)

The energy stored in L is represented by  $1/2 \bullet L (I_{PK})^2$ .

When M1 is turned OFF (t<sub>OFF</sub>), the energy stored in L is emitted through a diode to the output capacitor.

Then, the reverse voltage (V<sub>L</sub>) is generated :

$$V_{L} = (V_{OUT} + V_{D}^{-1}) - V_{IN}$$
\*1. Diode forward voltage (5)

The voltage at CONT pin rises only by  $V_{OUT}+V_D$ .

The char	ige in	the current $(I_L)$ flo	owing through the diode into $V_{OUT}$ during $t_{OFF}$ :	
dl	V <sub>L</sub>	$V_{OUT} + V_D - V_{IN}$		(6)
dt –	L		,	(0)

Integration of the equation (6) is as follows :

$$I_{L} = I_{PK} - \left(\frac{V_{OUT} + V_{D} - V_{IN}}{L}\right) \bullet t$$
(7)

During  $t_{ON}$ , the energy is stored in L and is not transmitted to  $V_{OUT}$ . When receiving the output current ( $I_{OUT}$ ) from  $V_{OUT}$ , the energy of the capacitor ( $C_L$ ) is consumed. As a result, the pin voltage of  $C_L$  is reduced, and goes to the lowest level after M1 is turned ON ( $t_{ON}$ ). When M1 is turned OFF, the energy stored in L is transmitted through the diode to  $C_L$ , and the voltage of  $C_L$  rises rapidly.  $V_{OUT}$  is a time function, and therefore indicates the maximum value (ripple voltage ( $V_{P-P}$ ) ) when the current flowing through into  $V_{OUT}$  and load current ( $I_{OUT}$ ) match.

Next, the ripple voltage is determined as follows.

 $I_{OUT}$  vs. t<sub>1</sub> (time) from when M1 is turned OFF (after t<sub>ON</sub>) to when V<sub>OUT</sub> reaches the maximum level :

$$I_{OUT} = I_{PK} - \left(\frac{v_{OUT} + v_D - v_{IN}}{L}\right) \bullet t_1$$

$$\therefore t_1 = (I_{PK} - I_{OUT}) \bullet \left(\frac{L}{V_{OUT} + V_D - V_{IN}}\right)$$
(8)
(9)

When M1 is turned OFF ( $t_{OFF}$ ),  $I_L = 0$  (when the energy of the inductor is completely transmitted). Based on equation (7) :

When substituting equation (10) for equation (9) :

$$t_{1} = t_{OFF} - \left(\frac{I_{OUT}}{I_{PK}}\right) \bullet t_{OFF}$$
(11)

Electric charge  $\Delta Q_1$  which is charged in  $C_L$  during  $t_1$ :

$$\Delta Q_{1} = \int_{0}^{t1} I_{L} dt = I_{PK} \bullet \int_{0}^{t1} dt - \frac{V_{OUT} + V_{D} - V_{IN}}{L} \bullet \int_{0}^{t1} t dt = I_{PK} \bullet t_{1} - \frac{V_{OUT} + V_{D} - V_{IN}}{L} \bullet \frac{1}{2} t_{1}^{2}$$
(12)

When substituting equation (12) for equation (9) :

$$\Delta Q1 = I_{PK} - \frac{1}{2} (I_{PK} - I_{OUT}) \bullet t_1 = \frac{I_{PK} + I_{OUT}}{2} \bullet t_1$$
(13)

A rise in voltage (V\_{P-P}) due to  $\Delta Q_1$ :

$$V_{P-P} = \frac{\Delta Q_1}{C_L} = \frac{1}{C_L} \bullet \left( \frac{I_{PK} + I_{OUT}}{2} \right) \bullet t_1 \quad \dots$$
(14)

When taking into consideration  $I_{OUT}$  to be consumed during  $t_1$  and the Equivalent Series Resistance (R<sub>ESR</sub>) of C<sub>L</sub> :

$$V_{P-P} = \frac{\Delta Q_1}{C_L} = \frac{1}{C_L} \bullet \left(\frac{I_{PK} + I_{OUT}}{2}\right) \bullet t1 + \left(\frac{I_{PK} + I_{OUT}}{2}\right) \bullet R_{ESR} - \frac{I_{OUT} \bullet t_1}{C_L}$$
(15)

When substituting equation (11) for equation (15) :

$$V_{P-P} = \frac{(I_{PK} - I_{OUT})^2}{2I_{PK}} \bullet \frac{t_{OFF}}{C_L} + \left(\frac{I_{PK} + I_{OUT}}{2}\right) \bullet R_{ESR}$$
(16)

Therefore to reduce the ripple voltage, it is important that the capacitor connected to the output pin has a large capacity and a small R<sub>ESR</sub>.

# External Parts Selection

#### 1. Inductor

To minimize the loss due to inductor direct current resistance, select an inductor with the smallest possible direct current resistance (less than 1  $\Omega$ ). Set the inductance value (L value) to around 22  $\mu$ H to 1 mH.

To make the average value of the output voltage (V<sub>OUT</sub>) constant, it is necessary to supply the energy corresponding to the output current ( $I_{OUT}$ ) from the inductor. The amount of charge required for  $I_{OUT}$  is  $I_{OUT} \times (t_{ON} + t_{OFF})$ . Because the inductor can supply energy only during t<sub>OFF</sub>, the charge is obtained by integrating equation (7) in the "3.

**Operation**" in the "**IDENTIFY OPERATION**" with  $0 \rightarrow t_{OFF}$ , namely,  $\frac{I_{PK}}{2} \bullet t_{OFF}$ . Thus,

$$\frac{I_{PK}}{2} \bullet t_{OFF} = I_{OUT} \times (t_{ON} + t_{OFF})$$

$$\therefore I_{PK} = 2 \bullet \frac{t_{ON} + t_{OFF}}{t_{OFF}} \bullet I_{OUT}$$
(17)
(18)

When the oscillation duty ratio of OSC is 75%, IPK = 8 • IOUT. Therefore, an IPK current which is eight times IOUT flows into transistor (M1).

The S-8351 Series includes a switching current controller which monitors the current flowing into the CONT pin by the voltage (CONT control voltage) and controls the current. This controller prevents destruction of the IC due to excess current.

If an inductor with a large L value is selected, both I<sub>PK</sub> and I<sub>OUT</sub> decrease. Since the energy stored in the inductor is

equal to  $\frac{1}{2}L \bullet (I_{PK})^2$ , the energy decreases because  $I_{PK}$  decreases in steps of squares offsetting the increase of L

value. As a result, stepping up at a low voltage becomes difficult and the minimum operating input voltage becomes high. However, the direct current resistance loss of L value and the M1 transistor decreases by the amount IPK decreased, and the inductance efficiency improves.

On the other hand, if an inductor with a smaller L value is selected, both IPK and IOUT increase. Accordingly, the minimum operating input voltage becomes low but the inductance efficiency deteriorates.

#### Caution An excessively large IPK may cause magnetic saturation for some core materials, leading to the destruction of the IC. Use a core with material that satisfies $I_{sat}^{*1} > I_{PK}$

\*1. Level of current that causes magnetic saturation.

#### 2. Diode

Use an external diode that meets the following requirements :

- Low forward voltage : V<sub>F</sub> < 0.3 V
- High switching speed : 500 ns max.
- Reverse voltage :  $V_{OUT} + V_F$  or more
- Current rate : I<sub>PK</sub> or more

# 3. Capacitor (C<sub>IN</sub>, C<sub>L</sub>)

A capacitor on the input side ( $C_{IN}$ ) improves the efficiency by reducing the power impedance and stabilizing the input current. Select a  $C_{IN}$  value according to the impedance of the power supply used.

A capacitor on the output side ( $C_L$ ) is used for smoothing the output voltage. For step-up types, the output voltage flows intermittently to the load current, so step-up types need a larger capacitance than step-down types. Therefore, select an appropriate capacitor in accordance with the ripple voltage, which increases in case of a higher output voltage or a higher load current. The capacitor value should be 10  $\mu$ F or more.

A capacitor at the output side ( $C_L$ ) is used for smoothing the ripple voltage. Select an appropriate capacitor with a small equivalent series resistance ( $R_{ESR}$ ) and a large capacitance. The capacitor value should be 10  $\mu$ F or mpre. A tantalum electrolytic capacitor and an organic semiconductor capacitor are especially recommended because of their superior low-temperature and leakage current characteristics.

### 4. External Transistor (S-8352 Series)

For the S-8352 Series, connecting an external transistor increases the output current. An enhancement (N-channel) MOS FET type or a bipolar (NPN) type can be used as the external transistor.

#### 4.1 Enhancement (N-channel) MOS FET Type

Figure 19 is a circuit example using a MOS FET transistor (N-channel).



\*1. For A type.

Figure 19 Circuit Example Using MOS FET (N-channel) Type

An N-channel power MOS FET should be used for the MOS FET. In particular, the EXT pin can drive a MOS FET with a gate capacitance of around 1000 pF. Because the gate voltage and current of the external power MOS FET are supplied from the stepped-up output voltage ( $V_{OUT}$ ), the MOS FET is driven more effectively.

A large current may flow during startup, depending on the MOS FET selection. The S-8352 Series does not feature overcurrent protection for the external MOS FET, so perform sufficient evaluation using the actual devices. Also recommend to use a MOS FET with an input capacitance of 700 pF or less.

Since the ON-resistance of the MOS FET might depend on the difference between the output voltage ( $V_{OUT}$ ) and the threshold voltage of the MOS FET, and affect the output current as well as the efficiency, the threshold voltage should be low. When the output voltage is as low as 2.0 V, like in the S-8352A20, the circuit operates only when the MOS FET has a threshold voltage lower than the output voltage.

#### 4.2 Bipolar (NPN) Type

A circuit example using the CPH3210 ( $h_{FE}$  = 200 to 560) from Sanyo Electric Co., Ltd. as a bipolar transistor (NPN) is shown in **Figure 24 to 26** in the "**Standard Circuits**". The  $h_{FE}$  value and  $R_b$  value of the bipolar transistor determine the driving capacity to increase the output current using a bipolar transistor. A peripheral circuit example of the transistor is shown in **Figure 20**.



\*1. V<sub>DD</sub> for D type.

Figure 20 External Transistor Peripheral Circuit

The recommended  $R_b$  value is around 1 k $\Omega$ . Actually, calculate the necessary base current ( $I_b$ ) from the bipolar transistor ( $h_{FE}$ ) using  $I_b = \frac{I_{PK}}{h_{FE}}$ , and select the smaller  $R_b$  value than  $R_b = \frac{V_{OUT} - 0.7}{I_b} - \frac{0.4}{|I_{EXTH}|}$ <sup>\*1</sup>.

A small  $R_b$  value can increase the output current, but the efficiency decreases. Since a current may flow on the pulse and the voltage may drop due to wiring resistance or other factors in the actual circuit, therefore the optimum  $R_b$  value should be determined by experiment.

Connecting the speed-up capacitor ( $C_b$ ) in parallel with the  $R_b$  resistance as shown in **Figure 20**, decreases switching loss and improves the efficiency.

The 
$$C_b$$
 value is calculated according to  $C_b \le \frac{1}{2\pi \bullet R_b \bullet f_{osc} \bullet 0.7}$ 

Select a  $C_b$  value after performing sufficient evaluation since the optimum  $C_b$  value differs depending upon the characteristics of the bipolar transistor.

\*1. For D type, 
$$R_b = \frac{V_{DD} - 0.7}{I_b} - \frac{0.4}{|I_{EXTH}|}$$

# 5. V<sub>DD</sub> / V<sub>OUT</sub> Separate Type (For S-8351/8352 Series D Type)

The D type provides separate internal circuit power supply (VDD pin) and output voltage setting pin (VOUT pin) in the IC, making it ideal for the following applications.

- (1) Changing the output voltage value using an external resistor
- (2) Setting a high output voltage value, such as +15 V
- Cautions 1. This IC starts a step-up operation at  $V_{DD} = 0.8 \text{ V}$ , but set  $1.8 \le V_{DD} \le 10 \text{ V}$  to stabilize the output voltage and frequency of the oscillator. (Input a voltage of 1.8 V or more at the VDD pin for all products with a setting less than 1.9 V.) An input voltage of 1.8 V or more at the VDD pin allows connection of the VDD pin to either the input voltage VIN pin or output VOUT pin.
  - 2. Choose external resistors  $R_A$  and  $R_B$  so as to not affect the output voltage, considering that there is impedance between the VOUT pin and VSS pin in the IC chip. The internal resistance between the VOUT pin and VSS pin is as follows :
    - (1) S-835xx18 : 2.1 MΩ to 14.8 MΩ
    - (2) S-835xx20 : 1.4 M $\Omega$  to 14.8 M $\Omega$
    - (3) S-835xx30 : 1.4 M $\Omega$  to 14.2 M $\Omega$
    - (4) S-835xx50 : 1.4 M $\Omega$  to 12.1 M $\Omega$
  - 3. Attach a capacitor ( $C_c$ ) in parallel to the  $R_A$  resistance when an unstable event such as oscillation of the output voltage occurs. Calculate  $C_c$  using the following equation :

$$\mathbf{C}_{\mathsf{C}}[\mathsf{F}] = \frac{1}{2 \bullet \pi \bullet \mathsf{R}_{\mathsf{A}} \bullet 20 \,\mathsf{kHz}}$$

# Standard Circuits

# 1. S-8351 Series

(1) A type









Figure 22

(3) D type





Caution The above connection diagrams and constants will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constants.

2. S-8352 Series

# (1) A type



Figure 24

(2) B and C types



Figure 25

(3) D type





Caution The above connection diagrams and constants will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constants.

# Precautions

- · Mount the external capacitors, the diode, and the coil as close as possible to the IC.
- · Characteristics ripple voltage and spike noise occur in IC containing switching regulators. Moreover, rush current flows at the time of a power supply injection. Because they largely depend on the coil and the capacitor and impedance used, fully check them using an actually mounted model.
- · Make sure that the dissipation of the switching transistor (especially at a high temperature) does not exceed the allowable power dissipation of the package.
- The performance of this IC varies depending on the design of the PCB patterns, peripheral circuits and external parts. Thoroughly test all settings with your device. Also, try to use the recommended external parts. If not, contact an ABLIC Inc. sales person.
- When the impedance of the power supply is high, the shutdown pin is switched from "L" to "H", or VIN is connected to the power supply, note that the power supply voltage drops temporarily because a rush current flows into the power supply.
- · Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection IC.
- ABLIC Inc. claims no responsibility for any and all disputes arising out of or in connection with any infringement of the products including this IC upon patents owned by third party.

# Characteristics (Typical Data)

1. Input voltage (VIN) vs. Powe Supply Input Current at No Load (IN)



2. Output Voltage (V<sub>OUT</sub>) vs. Current Consumption 1 (I<sub>SS1</sub>)





3. Temperature (Ta) vs. Current Consumption 1 (I<sub>SS1</sub>)



4. Output Voltage (V<sub>OUT</sub>) vs. Current Consumption 2 (I<sub>SS2</sub>) 5. Temperature (Ta) vs. Current Consumption 2 (I<sub>SS2</sub>)







6. Temperature (Ta) vs. Oscillation Frequency (fosc)



7. Temperature (Ta) vs. Duty Ratio 1 (Duty1)



9. Output Voltage (V<sub>OUT</sub>) vs. Switching Current (I<sub>sw</sub>)



11. Output Voltage (Vout) vs. EXT Pin Output Current "H" (IEXTH) 12. Temperature (Ta) vs. EXT Pin Output Current "H" (IEXTH)



Temperature (Ta) vs. Duty Ratio 2 (Duty2) 8.



10. Temperature (Ta) vs. Switching Current (I<sub>sw</sub>)





13. Output Voltage (V<sub>OUT</sub>) vs. EXT Pin Output Current "L" (I<sub>EXTL</sub>) 14. Temperature (Ta) vs. EXT Pin Output Current "L" (I<sub>EXTL</sub>)



15. Temperature (Ta) vs. Operation Start Voltage (V<sub>ST1</sub>) 16. Temperature (Ta) vs. Retention Voltage (V<sub>HLD</sub>)







#### 17. Transient Response Characteristics

The conditions for external parts are the same as those specified in the electrical characteristics.

### (1) Power-on (Ta = $25^{\circ}$ C, R<sub>L</sub> = $250 \Omega$ )









(2) Power Supply Voltage Fluctuation (Ta = 25°C,  $R_L$  = 250  $\Omega$ )

S-8351A30MC

S-8351A30MC











# STEP-UP, BUILT-IN / EXTERNAL FET PFM CONTROL SWITCHING REGULATOR / SWITCHING REGULATOR CONTROLLER Rev.3.1\_00 S-8351/8352 Series



#### STEP-UP, BUILT-IN / EXTERNAL FET PFM CONTROL SWITCHING REGULATOR / SWITCHING REGULATOR CONTROLLER S-8351/8352 Series Rev.3.1\_00



# Reference Data

Use this reference data to choose the external parts. This reference data makes it possible to choose the recommended external part based on the application and characteristics data.

# 1. External Parts for Reference Data

Table 18

Condition	Product Name	Output Voltage V	Power MOS FET	Coil
1	S-8351A30MC	3.0	Built-in	CDRH6D28-470
2	S-8351A30MC	3.0	Built-in	CDRH6D28-101
3	S-8351A30MC	3.0	Built-in	CXLP120-101
4	S-8351A50MC	5.0	Built-in	CDRH6D28-101
5	S-8351A50MC	5.0	Built-in	CDRH125-221
6	S-8351A50MC	5.0	Built-in	CXLP120-470
7	S-8352A30MC	3.0	External	CDRH6D28-220
8	S-8352A30MC	3.0	External	CDRH6D28-101
9	S-8352A30MC	3.0	External	CXLP120-470
10	S-8352A50MC	5.0	External	CDRH6D28-220
11	S-8352A50MC	5.0	External	CDRH6D28-101
12	S-8352A50MC	5.0	External	CXLP120-101

The properties of the external parts are shown below.

	Table 19	Evaluation	coil
--	----------	------------	------

Part	Product Name	Manufacturer	Characteristics
Coil	CDRH6D28-220	Sumida Corporation	22 μH, DCR <sup>*1</sup> = 0.128 Ω, I <sub>MAX</sub> <sup>*2</sup> = 1200 mA
	CDRH6D28-470	Sumida Corporation	47 μH, DCR <sup>*1</sup> = 0.238 Ω, I <sub>MAX</sub> <sup>*2</sup> = 800 mA
	CDRH6D28-101	Sumida Corporation	100 μH, DCR <sup>*1</sup> = 0.535 Ω, $I_{MAX}$ <sup>*2</sup> = 540 mA
	CDRH125-221	Sumida Corporation	220 μH, DCR <sup>*1</sup> = 0.4 Ω, $I_{MAX}$ <sup>*2</sup> = 800 mA
	CXLP120-470	Sumitomo Special Metals Co., Ltd	47 μH, DCR <sup>*1</sup> = 0.95 Ω, $I_{MAX}$ <sup>*2</sup> = 450 mA
	CXLP120-101	Sumitomo Special Metals Co., Ltd	100 μH, DCR <sup>*1</sup> = 2.5 Ω, I <sub>MAX</sub> <sup>*2</sup> = 200 mA

\*1. Direct current resistance

\*2. Maximum allowable current

Table 20	Properties of External I	Parts
----------	--------------------------	-------

Part	Product Name	Manufacturer	Characteristics
Diode	MA2Z748	Matsushita Electronic Components Co., Ltd.	$V_{F}^{*1} = 0.4V, I_{F}^{*2} = 0.3A$ (Shottky type)
Capacitor (Output capacitance)	F93	Nichicon Corporation	16V, 47μF (Tantalum type)
Transistor (NPN)	CPH3210	Sanyo Electric Co.,Ltd.	$V_{CBO}^{*3} = 40V, V_{CEO}^{*4} = 30V$ hFE <sup>*5</sup> = 200 min. (V <sub>CE</sub> = 2V, I <sub>C</sub> = 500mA) fT <sup>*6</sup> = 290 MHz typ. (V <sub>CE</sub> = 10V, I <sub>C</sub> = 500mA)

\*1. Forward voltage, \*2. Forward current, \*3. Collector-to-base voltage, \*4. Collector-to-emitter voltage, \*5. DC current gain, \*6. Gain-bandwidth product

Caution The above values shown in the characteristics column of Table 19 and 20 are based on the materials provided by each manufacture. However, consider the characteristics of the original materials when using the above products.

### 2. Step-up Characteristics (Ta = 25°C)

The data of the step-up characteristics ((a) Input voltage ( $V_{IN}$ ) vs. Output voltage ( $V_{OUT}$ ) characteristics (Input voltage stepped up), (b) Input voltage ( $V_{IN}$ ) vs. Output voltage ( $V_{OUT}$ ) characteristics (Input voltage stepped down), (c) Output current ( $I_{OUT}$ ) vs. Output voltage ( $V_{OUT}$ ) characteristics, (d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ ) characteristics under conditions of 1 to 12 in **Table 18** is shown below.

#### Condition 1 S-8351A30MC

(a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>)

(Input voltage raising)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



#### Condition 2 S-8351A30MC

(a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage raising)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)







(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



 (b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



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#### Condition 3 S-8351A30MC

(a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



# Condition 4 S-8351A50MC

(a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage raising)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)







(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



 (b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



#### Condition 5 S-8351A50MC

(a) Input voltage ( $V_{IN}$ ) vs. Output voltage ( $V_{OUT}$ )



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



# Condition 6 S-8351A50MC

(a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage raising)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



(b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



 (b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



#### Condition 7 S-8352A30MC

(a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



#### Condition 8 S-8352A30MC

(a) Input voltage ( $V_{IN}$ ) vs. Output voltage ( $V_{OUT}$ ) (Input voltage raising)



(c) Output current ( $I_{OUT}$ ) vs.Output voltage ( $V_{OUT}$ )







(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



(b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



#### Condition 9 S-8352A30MC

(a) Input voltage ( $V_{IN}$ ) vs. Output voltage ( $V_{OUT}$ )



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



# Condition 10 S-8352A50MC

(a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage raising)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



(b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



 (b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



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#### Condition 11 S-8352A50MC

(a) Input voltage ( $V_{IN}$ ) vs. Output voltage ( $V_{OUT}$ )



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



### Condition 12 S-8352A50MC

 (a) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage raising)



(c) Output current (I<sub>OUT</sub>) vs.Output voltage (V<sub>OUT</sub>)



(b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )



 (b) Input voltage (V<sub>IN</sub>) vs. Output voltage (V<sub>OUT</sub>) (Input voltage falling)



(d) Output current ( $I_{OUT}$ ) vs. Efficiency ( $\eta$ )









No. MP003-A-P-SD-1.2

TITLE	SOT233-A-PKG Dimensions	
No.	MP003-A-P-SD-1.2	
ANGLE	$\oplus$	
UNIT	mm	
ABLIC Inc.		











No. MP005-A-P-SD-1.3

TITLE	SOT235-A-PKG Dimensions	
No.	MP005-A-P-SD-1.3	
ANGLE	$\oplus$	
UNIT	mm	
ABLIC Inc.		











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