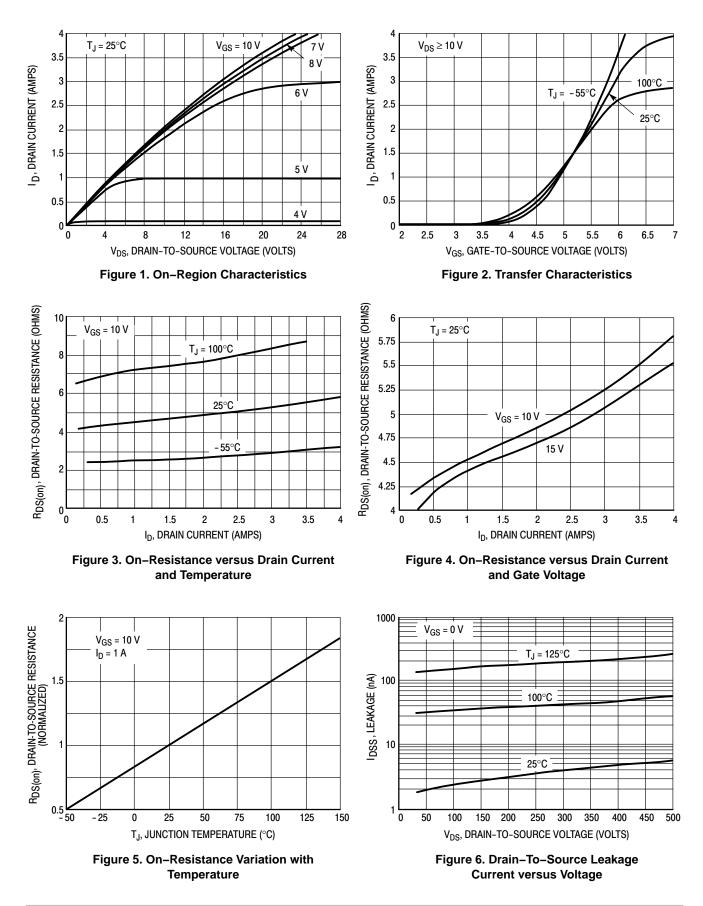
## **ELECTRICAL CHARACTERISTICS** (T<sub>J</sub> = $25^{\circ}$ C unless otherwise noted)

C	Symbol	Min	Тур	Max	Unit	
OFF CHARACTERISTICS						
Drain–Source Breakdown Voltage ( $V_{GS} = 0 \text{ Vdc}, I_D = 250 \mu \text{Adc}$ ) Temperature Coefficient (Positive	V <sub>(BR)DSS</sub>	500 -	_ 564		Vdc mV/°C	
Zero Gate Voltage Drain Current ( $V_{DS} = 500 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}$ ) ( $V_{DS} = 500 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}, T_J$	I <sub>DSS</sub>			10 100	μAdc	
Gate–Body Leakage Current ( $V_{GS}$ =	I <sub>GSS</sub>	-	-	100	nAdc	
ON CHARACTERISTICS (Note 1)						
Gate Threshold Voltage (V <sub>DS</sub> = V <sub>GS</sub> , I <sub>D</sub> = 250 μAdc) Temperature Coefficient (Negative	V <sub>GS(th)</sub>	2.0 _	3.0 4.0	4.0 _	Vdc mV/°C	
Static Drain-Source On-Resistance	R <sub>DS(on)</sub>	-	4.5	6.0	Ω	
$      Drain-Source On-Voltage (V_{GS} = 10 \\ (I_D = 2.0 \text{ Adc}) \\ (I_D = 1.0 \text{ Adc},  T_J = 125^{\circ}\text{C}) $	V <sub>DS(on)</sub>		9.5 -	14.4 12.6	Vdc	
Forward Transconductance ( $V_{DS} = 1$	9 <sub>FS</sub>	0.5	-	-	mhos	
DYNAMIC CHARACTERISTICS						
Input Capacitance		C <sub>iss</sub>	-	845	1183	pF
Output Capacitance	(V <sub>DS</sub> = 25 Vdc, V <sub>GS</sub> = 0 Vdc, f = 1.0 MHz)	C <sub>oss</sub>	-	100	140	
Reverse Transfer Capacitance		C <sub>rss</sub>	_	26	52	
SWITCHING CHARACTERISTICS (I	Note 2)					
Turn-On Delay Time		t <sub>d(on)</sub>	-	12	24	ns
Rise Time	(V <sub>DD</sub> = 250 Vdc, I <sub>D</sub> = 2.0 Adc,	t <sub>r</sub>	-	14	28	
Turn-Off Delay Time	$V_{GS} = 10 \text{ Vdc}, R_G = 9.1 \Omega$	t <sub>d(off)</sub>	-	21	42	
Fall Time		t <sub>f</sub>	-	19	38	
Gate Charge (See Figure 8)		Q <sub>T</sub>	-	19	27	nC
	(V <sub>DS</sub> = 400 Vdc, I <sub>D</sub> = 2.0 Adc,	Q <sub>1</sub>	-	3.7	-	
	$V_{GS} = 10 \text{ Vdc})$	Q <sub>2</sub>	-	7.9	-	
		Q <sub>3</sub>	-	9.9	-	
SOURCE-DRAIN DIODE CHARACT	ERISTICS					
Forward On–Voltage (Note 1)	$(I_{S} = 2.0 \text{ Adc}, V_{GS} = 0 \text{ Vdc})$ $(I_{S} = 2.0 \text{ Adc}, V_{GS} = 0 \text{ Vdc}, T_{J} = 125^{\circ}\text{C})$	V <sub>SD</sub>		2.3 1.85	3.5 -	Vdc
Reverse Recovery Time		t <sub>rr</sub>	-	223	-	ns
(See Figure 14)	$(I_{S} = 2.0 \text{ Adc}, V_{GS} = 0 \text{ Vdc},$	ta	-	161	-	
	$dI_{S}/dt = 100 \text{ A}/\mu\text{s}$	t <sub>b</sub>	-	62	-	
Reverse Recovery Stored Charge	Q <sub>RR</sub>	-	1.92	-	μC	
INTERNAL PACKAGE INDUCTANC	E			•	•	•
Internal Drain Inductance (Measured from contact screw on tak (Measured from the drain lead 0.25"	LD		3.5 4.5		nH	
Internal Source Inductance (Measured from the source lead 0.25	LS	-	7.5	-	nH	

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.
1. Pulse Test: Pulse Width ≤ 300 µs, Duty Cycle ≤ 2%.
2. Switching characteristics are independent of operating junction temperature.

# **TYPICAL ELECTRICAL CHARACTERISTICS**



### POWER MOSFET SWITCHING

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals ( $\Delta t$ ) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain–gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ( $I_{G(AV)}$ ) can be made from a rudimentary analysis of the drive circuit so that

 $t = Q/I_{G(AV)}$ 

During the rise and fall time interval when switching a resistive load,  $V_{GS}$  remains virtually constant at a level known as the plateau voltage,  $V_{SGP}$ . Therefore, rise and fall times may be approximated by the following:

 $t_r = Q_2 \ x \ R_G / (V_{GG} - V_{GSP})$ 

 $t_f = Q_2 \ x \ R_G / V_{GSP}$ 

where

 $V_{GG}$  = the gate drive voltage, which varies from zero to  $V_{GG}$  $R_G$  = the gate drive resistance

and  $Q_2$  and  $V_{GSP}$  are read from the gate charge curve.

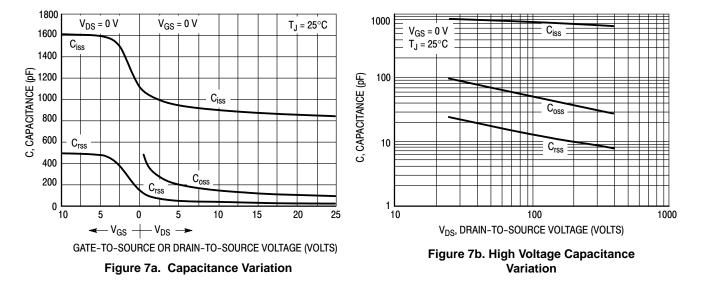
During the turn–on and turn–off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in an RC network. The equations are:

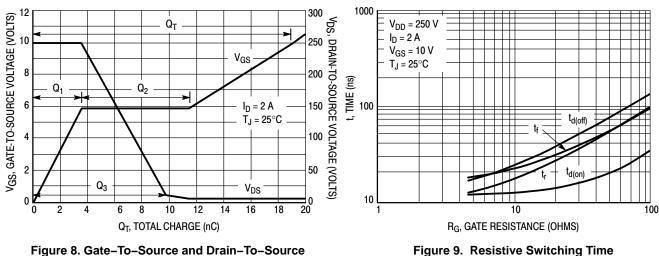
$$\begin{split} t_{d(on)} &= R_G \; C_{iss} \; In \; [V_{GG}/(V_{GG}-V_{GSP})] \\ t_{d(off)} &= R_G \; C_{iss} \; In \; (V_{GG}/V_{GSP}) \end{split}$$

The capacitance ( $C_{iss}$ ) is read from the capacitance curve at a voltage corresponding to the off–state condition when calculating  $t_{d(on)}$  and is read at a voltage corresponding to the on–state when calculating  $t_{d(off)}$ .

At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by Ldi/dt, but since di/dt is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

The resistive switching time variation versus gate resistance (Figure 9) shows how typical switching performance is affected by the parasitic circuit elements. If the parasitics were not present, the slope of the curves would maintain a value of unity regardless of the switching speed. The circuit used to obtain the data is constructed to minimize common inductance in the drain and gate circuit loops and is believed readily achievable with board mounted components. Most power electronic loads are inductive; the data in the figure is taken with a resistive load, which approximates an optimally snubbed inductive load. Power MOSFETs may be safely operated into an inductive load; however, snubbing reduces switching losses.





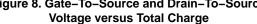


Figure 9. Resistive Switching Time Variation versus Gate Resistance

DRAIN-TO-SOURCE DIODE CHARACTERISTICS

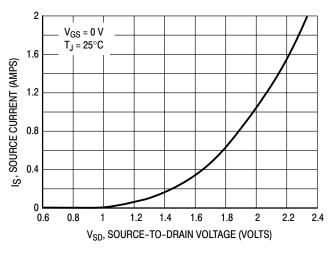


Figure 10. Diode Forward Voltage versus Current

### SAFE OPERATING AREA

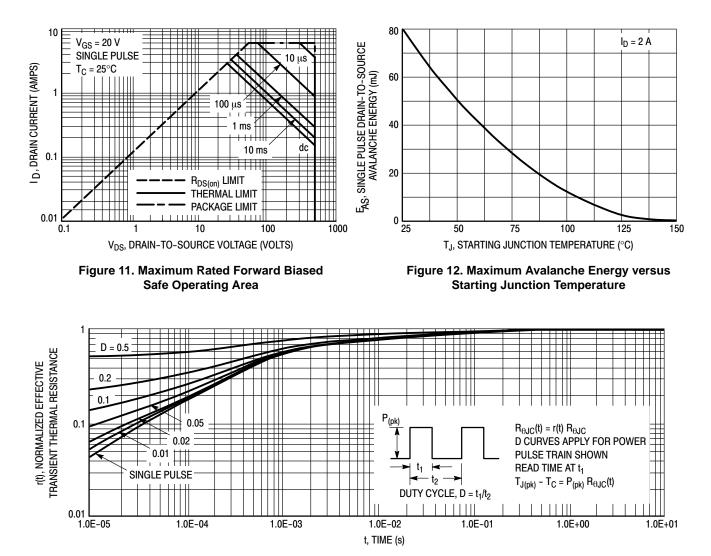
The Forward Biased Safe Operating Area curves define the maximum simultaneous drain–to–source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature ( $T_C$ ) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance–General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current ( $I_{DM}$ ) nor rated voltage ( $V_{DSS}$ ) is exceeded and the transition time ( $t_r$ , $t_f$ ) do not exceed 10  $\mu$ s. In addition the total power averaged over a complete switching cycle must not exceed ( $T_{J(MAX)} - T_C$ )/( $R_{\theta JC}$ ).

A Power MOSFET designated E–FET can be safely used in switching circuits with unclamped inductive loads. For reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non–linearly with an increase of peak current in avalanche and peak junction temperature.

Although many E–FETs can withstand the stress of drain–to–source avalanche at currents up to rated pulsed current ( $I_{DM}$ ), the energy rating is specified at rated continuous current ( $I_D$ ), in accordance with industry custom. The energy rating must be derated for temperature as shown in the accompanying graph (Figure 12). Maximum energy at currents below rated continuous  $I_D$  can safely be assumed to equal the values indicated.

# SAFE OPERATING AREA





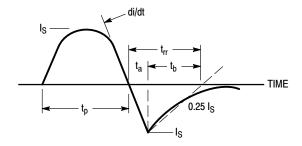
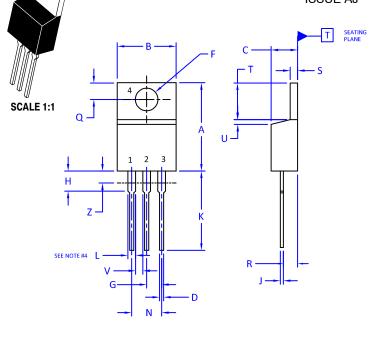


Figure 14. Diode Reverse Recovery Waveform

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TO-220 CASE 221A-09 ISSUE AJ



NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 2009.

2. CONTROLLING DIMENSION: INCHES

3. DIMENSION Z DEFINES A ZONE WHERE ALL BODY AND LEAD IRREGULARITIES ARE ALLOWED.

4. MAX WIDTH FOR F102 DEVICE = 1.35MM

	INCHES		MILLIMETERS		
DIM	MIN.	MAX.	MIN.	MAX.	
А	0.570	0.620	14.48	15.75	
В	0.380	0.415	9.66	10.53	
С	0.160	0.190	4.07	4.83	
D	0.025	0.038	0.64	0.96	
F	0.142	0.161	3.60	4.09	
G	0.095	0.105	2.42	2.66	
Н	0.110	0.161	2.80	4.10	
J	0.014	0.024	0.36	0.61	
К	0.500	0.562	12.70	14.27	
L	0.045	0.060	1.15	1.52	
Ν	0.190	0.210	4.83	5.33	
Q	0.100	0.120	2.54	3.04	
R	0.080	0.110	2.04	2.79	
S	0.045	0.055	1.15	1.41	
Т	0.235	0.255	5.97	6.47	
U	0.000	0.050	0.00	1.27	
V	0.045		1.15		
Z		0.080		2.04	

STYLE 1: PIN 1. 2. 3. 4.	COLLECTOR EMITTER	STYLE 2: PIN 1. 2. 3. 4.	EMITTER COLLECTOR	2. 3.	CATHODE ANODE GATE ANODE	STYLE 4: PIN 1. 2. 3. 4.	MAIN TERMINAL 1 MAIN TERMINAL 2 GATE MAIN TERMINAL 2
STYLE 5: PIN 1. 2. 3. 4.	DRAIN SOURCE	2. 3.	ANODE CATHODE ANODE CATHODE	2. 3.	CATHODE ANODE CATHODE ANODE	2. 3.	CATHODE ANODE EXTERNAL TRIP/DELAY ANODE
STYLE 9: PIN 1. 2. 3. 4.	COLLECTOR EMITTER	STYLE 10: PIN 1. 2. 3. 4.	GATE SOURCE DRAIN		DRAIN SOURCE GATE	STYLE 12 PIN 1. 2. 3. 4.	MAIN TERMINAL 1 MAIN TERMINAL 2 GATE NOT CONNECTED

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