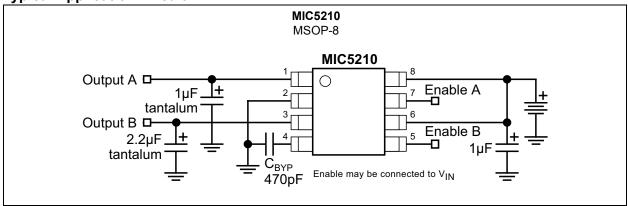
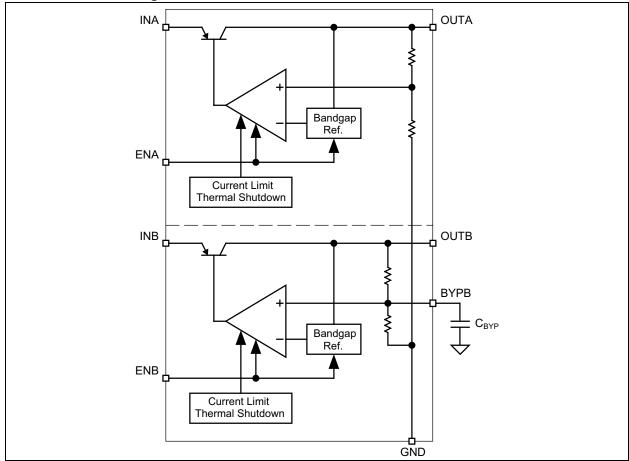
# **Typical Application Circuit**



#### **Functional Block Diagram**



#### 1.0 ELECTRICAL CHARACTERISTICS

### **Absolute Maximum Ratings †**

Supply Input Voltage (V <sub>IN</sub> )	–20V to +20V
Enable Input Voltage (V <sub>EN</sub> )	–20V to +20V
Power Dissipation (P <sub>D</sub> )	Internally Limited

#### **Operating Ratings ‡**

Supply Input Voltage (V <sub>IN</sub> )	+2.5V to +16V
Enable Input Voltage (V <sub>EN</sub> )	0V to +16V

**<sup>†</sup> Notice:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability.

**‡ Notice:** The device is not guaranteed to function outside its operating ratings.

TABLE 1-1: ELECTRICAL CHARACTERISTICS

**Electrical Characteristics:**  $V_{IN} = V_{OUT} + 1V$ ;  $I_L = 100 \ \mu\text{A}$ ;  $C_L = 1.0 \ \mu\text{F}$ ;  $V_{EN} \ge 2.0V$ ;  $T_J = +25 \ ^{\circ}\text{C}$ , **bold** values indicate  $-40 \ ^{\circ}\text{C} < T_J < +125 \ ^{\circ}\text{C}$ , unless noted.

Parameter	Symbol	Min.	Тур.	Max.	Units	Conditions	
Output Valtage Assurage	Vo	-1	_	1	0/	Variation from specified V <sub>OUT</sub>	
Output Voltage Accuracy		-2	_	2	%		
Output Voltage Temperature Coefficient	$\Delta V_{O}/\Delta T$	_	40	_	ppm/°C	Note 1	
Line Regulation	$\Delta V_{O}/\Delta V_{O}$	_	0.004	0.012	%/V	V <sub>IN</sub> = V <sub>OUT</sub> +1V to +16V	
Line Regulation		_	_	0.05	707 V		
Load Regulation	70/70	_	0.02	0.2	%	L = 0.1 mA to 150 mA (Note 2)	
Load Regulation		_	_	0.5	/0	$I_L = 0.1 \text{ mA to } 150 \text{ mA (Note 2)}$	
			10	50	mV	I <sub>1</sub> = 100 μA	
		_	_	70	IIIV		
		_	110	150	mV	I <sub>I</sub> = 50 mA	
Dropout Voltage (Note 3)	V <sub>IN</sub> – V <sub>O</sub>	_		230	IIIV	- 50 IIIA	
Diopout Voltage (Note 3)		_	140	250	mV	L = 100 mA	
		_	_	300		I <sub>L</sub> = 100 mA	
			165	275	mV	I <sub>L</sub> = 150 mA	
		_	_	350	111 V		
Quiescent Current	1	_	0.01	1	μA	V <sub>EN</sub> ≤ 0.4V (shutdown)	
Quiescent Current	I <sub>GND</sub>	_		5	μΛ	V <sub>EN</sub> ≤ 0.18V (shutdown)	
	I <sub>GND</sub>	_	80	125	μA	V <sub>FN</sub> ≥ 2.0V, I <sub>I</sub> = 100 μA	
Ground Pin Current (Note 4), per regulator		_	_	150		VEN = 2.0 V, IL = 100 μA	
			350	600	μA	I <sub>I</sub> = 50 mA	
		_	_	800	μΛ	- 30	
		_	600	1000	^	I <sub>L</sub> = 100 mA	
			_	1500	μA		
			1300	1900	μA	I <sub>L</sub> = 150 mA	
			_	2500	μΛ	- 150 IIIA	
Ripple Rejection	PSRR		75	_	dB	Frequency = 100 Hz, $I_L$ = 100 $\mu$ A	

#### TABLE 1-1: ELECTRICAL CHARACTERISTICS (CONTINUED)

**Electrical Characteristics:**  $V_{IN} = V_{OUT} + 1V$ ;  $I_L = 100 \ \mu\text{A}$ ;  $C_L = 1.0 \ \mu\text{F}$ ;  $V_{EN} \ge 2.0V$ ;  $T_J = +25 \ ^{\circ}\text{C}$ , **bold** values indicate  $-40 \ ^{\circ}\text{C} < T_J < +125 \ ^{\circ}\text{C}$ , unless noted.

Parameter	Symbol	Min.	Тур.	Max.	Units	Conditions
Current Limit	I <sub>LIMIT</sub>		320	500	mA	V <sub>OUT</sub> = 0V
Thermal Regulation	$\Delta V_{O}/\Delta P_{D}$		0.05	_	%/W	Note 5
Output Noise (Regulator B only)	e <sub>no</sub>	l	260	_	nV/√Hz	$I_L$ = 50 mA, $C_L$ = 2.2 $\mu$ F, 470 pF from BYPB to GND
Enable Input Logic-Low	V <sub>IL</sub>		_	0.4	V	Regulator shutdown
Voltage		_	_	0.18		Regulator strutuowii
Enable Input Logic-High Voltage	V <sub>IH</sub>	2.0	_	_	V	Regulator enabled
	I <sub>IL</sub>	_	0.01	-1	μΑ	V <sub>IL</sub> ≤ 0.4V
Enable Input Current		_	_	-2		V <sub>IL</sub> ≤ 0.18V
Enable Input Current		_	5	20		V <sub>IH</sub> ≥ 2.0V
	I <sub>IH</sub>	_	_	25	μA	V <sub>IH</sub> ≥ 2.0V

- **Note 1:** Output voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
  - 2: Regulation is measured at constant junction temperature using low duty cycle pulse testing. Parts are tested for load regulation in the load range from 0.1 mA to 150 mA. Changes in output voltage due to heating effects are covered by the thermal regulation specification.
  - **3:** Dropout Voltage is defined as the input to output differential at which the output voltage drops 2% below its nominal value measured at 1V differential.
  - **4:** Ground pin current is the regulator quiescent current plus pass transistor base current. The total current drawn from the supply is the sum of the load current plus the ground pin current.
  - 5: Thermal regulation is defined as the change in output voltage at a time "t" after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 150 mA load pulse at V<sub>IN</sub> = 16V for t = 10 ms.

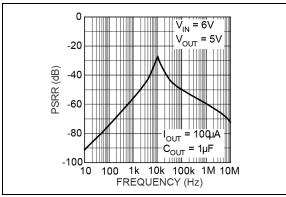
#### **TEMPERATURE SPECIFICATIONS (Note 1)**

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions
Temperature Ranges						
Junction Operating Temperature Range	T <sub>J</sub>	-40	_	+85	°C	_
Storage Temperature Range	T <sub>S</sub>	-60	_	+150	°C	_
Lead Temperature	_	_	_	+260	°C	Soldering, 5s
Package Thermal Resistances						
Thermal Resistance MSOP-8	$\theta_{JA}$	_	200	_	°C/W	Note 2

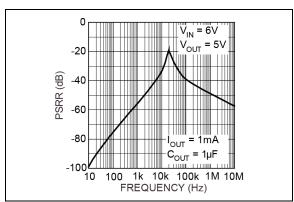
- Note 1: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air (i.e., T<sub>A</sub>, T<sub>J</sub>, θ<sub>JA</sub>). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +85°C rating. Sustained junction temperatures above +85°C can impact the device reliability.
  - 2: Absolute maximum ratings indicate limits beyond which damage to the component may occur. Electrical specifications do not apply when operating the device outside of its operating ratings. The maximum allowable power dissipation is a function of the maximum junction temperature,  $T_{J(max)}$ , the junction-to-ambient thermal resistance,  $\theta_{JA}$ , and the ambient temperature,  $T_A$ . The maximum allowable power dissipation at any ambient temperature is calculated using:  $P_{D(max)} = (T_{J(max)} T_A)/\theta_{JA}$ . Exceeding the maximum allowable power dissipation will result in excessive die temperature, and the regulator will go into thermal shutdown. The  $\theta_{JA}$  of the 8-pin MSOP (MM) is 200°C/W mounted on a PC board (see "Thermal Considerations" section for further details).

#### 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.



**FIGURE 2-1:** Power Supply Rejection Ratio.



**FIGURE 2-2:** Power Supply Rejection Ratio.

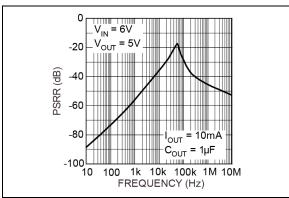
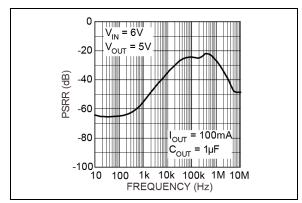
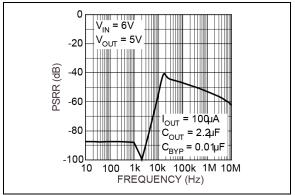


FIGURE 2-3: Power Supply Rejection Ratio.



**FIGURE 2-4:** Power Supply Rejection Ratio.



**FIGURE 2-5:** Power Supply Rejection Ratio.

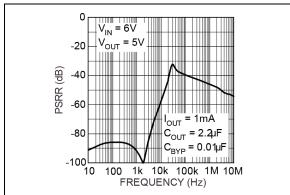


FIGURE 2-6: Power Supply Rejection Ratio.

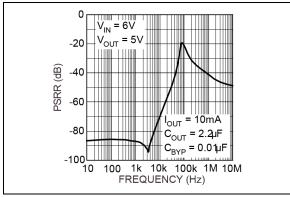


FIGURE 2-7: Power Supply Rejection Ratio.

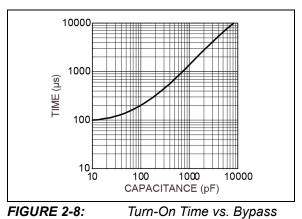


FIGURE 2-8: Capacitance.

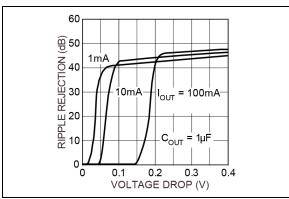


FIGURE 2-9: Power Supply Ripple Rejection vs. Voltage Drop.

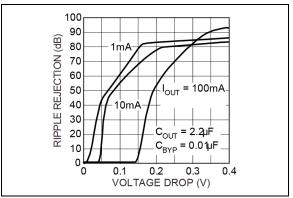
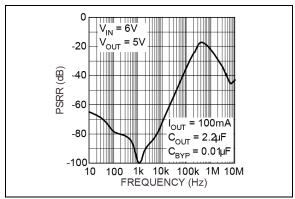
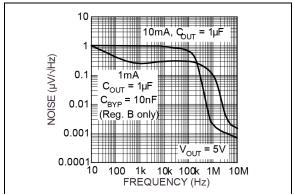


FIGURE 2-10: Power Supply Ripple Rejection vs. Voltage Drop.



**FIGURE 2-11:** Power Supply Rejection Ratio.



**FIGURE 2-12:** Noise Performance.

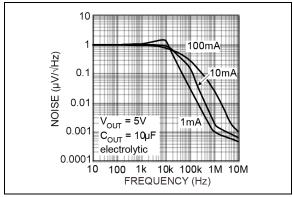


FIGURE 2-13: Noise Performance.

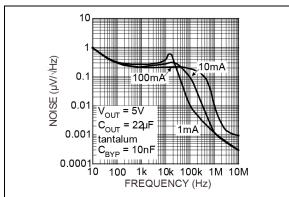
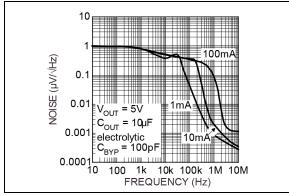
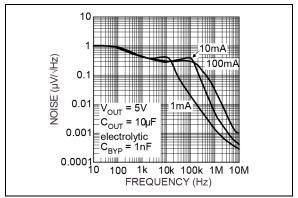


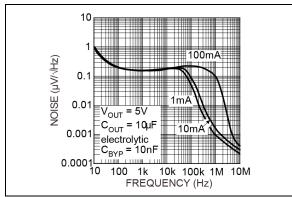
FIGURE 2-14: Noise Performance (Regulator B).



**FIGURE 2-15:** Noise Performance (Regulator B).



**FIGURE 2-16:** Noise Performance (Regulator B).



**FIGURE 2-17:** Noise Performance (Regulator B).

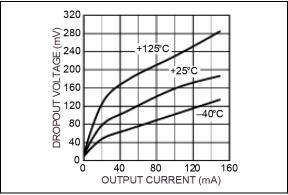


FIGURE 2-18: Dropout Voltage vs. Output Current.

## 3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in Table 3-1.

TABLE 3-1: PIN FUNCTION TABLE

Pin Number	Pin Name	Description
1	OUTA	Regulator Output A
2	GND	Ground
3	OUTB	Regulator Output B
4	BYPB	Reference Bypass B: Connect external 470 pF capacitor to GND to reduce output noise in regulator "B". May be left open.
5	ENB	Enable/Shutdown B (Input): CMOS-compatible input. Logic-high = enable, logic-low = shutdown. Do not leave floating.
6	INB	Supply Input B
7	ENA	Enable/Shutdown A (Input): CMOS-compatible input. Logic-high = enable, logic-low = shutdown. Do not leave floating.
8	INA	Supply Input A

#### 4.0 APPLICATION INFORMATION

#### 4.1 Enable/Shutdown

Forcing EN (enable/shutdown) high (greater than 2V) enables the regulator. EN is compatible with CMOS and TTL logic gates.

If the enable/shutdown feature is not required, connect EN to IN (supply input).

#### 4.2 Input Capacitor

A 1  $\mu F$  capacitor should be placed from IN to GND if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

#### 4.3 Reference Bypass Capacitor

BYPB (reference bypass) is connected to the internal voltage reference of regulator B. A 470 pF capacitor ( $C_{BYP}$ ) connected from BYPB to GND quiets this reference, providing a significant reduction in output noise.  $C_{BYP}$  reduces the regulator phase margin; when using  $C_{BYP}$ , output capacitors of 2.2  $\mu$ F or greater are generally required to maintain stability.

The start-up speed of the MIC5210 is inversely proportional to the size of the reference bypass capacitor. Applications requiring a slow ramp-up of output voltage should consider larger values of  $C_{\rm BYP}$ . Likewise, if rapid turn-on is necessary, consider omitting  $C_{\rm BYP}$ .

If output noise is not a major concern, omit  $\mathsf{C}_{\mathsf{BYP}}$  and leave  $\mathsf{BYPB}$  open.

#### 4.4 Output Capacitor

An output capacitor is required between OUT and GND to prevent oscillation. The minimum size of the output capacitor is dependent upon whether a reference bypass capacitor is used. 1.0  $\mu F$  minimum is recommended when CBYP is not used. 2.2  $\mu F$  minimum is recommended when  $C_{BYP}$  is 470 pF (see Typical Application Circuit). Larger values improve the regulator's transient response. The output capacitor value may be increased without limit.

The output capacitor should have an ESR (effective series resistance) of about  $5\Omega$  or less and a resonant frequency above 1 MHz. Ultralow-ESR capacitors may cause a low-amplitude oscillation and/or underdamped transient response. Most tantalum or aluminum electrolytic capacitors are adequate; film types will work, but are more expensive. Because many aluminum electrolytic capacitors have electrolytes that freeze at about  $-30^{\circ}\text{C}$ , solid tantalum capacitors are recommended for operation below  $-25^{\circ}\text{C}$ .

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.47  $\mu$ F for current below 10 mA or 0.33  $\mu$ F for currents below 1 mA.

#### 4.5 No-Load Stability

The MIC5210 will remain stable and in regulation with no load (other than the internal voltage divider) unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications.

#### 4.6 Dual-Supply Operation

When used in dual supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.

#### 4.7 Thermal Considerations

Multilayer boards having a ground plane, wide traces near the pads, and large supply bus lines provide better thermal conductivity.

The MIC5210-xxYMM (8-pin MSOP) has a thermal resistance of 200°C/W when mounted on a FR4 board with minimum trace widths and no ground plane.

For additional heat sink characteristics, please refer to Application Hint 17, "Designing P.C. Board Heat Sinks."

# 4.7.1 THERMAL EVALUATION EXAMPLES

For example, at 50°C ambient temperature, the maximum package power dissipation is:

#### **EQUATION 4-1:**

$$P_{D(MAX)} = (125^{\circ}C - 50^{\circ}C)/(200^{\circ}C/W) = 375mW$$

If the intent is to operate the 5V version from a 6V supply at the full 150mA load for both outputs in a 50°C maximum ambient temperature, make the following calculation:

#### **EQUATION 4-2:**

$$P_{D(EACHREG)} = (V_{IN} - V_{OUT}) \times I_{OUT} + (V_{IN} \times I_{GND})$$

#### **EQUATION 4-3:**

$$P_{D(EACHREG)} = (6V - 5V) \times 150mA + (6V \times 2.5mA)$$
  
= 165mW

#### **EQUATION 4-4:**

$$P_{D(BOTHREG)} = 2Regulators \times 165mW = 330mW$$

The actual total power dissipation of 330 mW is below the 375 mW package maximum; therefore, the regulator can be used.

Note that both regulators cannot always be used at their maximum current rating. For example, in a 5V input to 3.3V output application at +50°C, if one regulator supplies 150 mA, the other regulator is limited to a much lower current. The first regulator dissipates:

#### **EQUATION 4-5:**

$$P_D = (5V - 3.3V) \times 150mA + (5V \times 2.5mA)$$
  
= 267.5mW

Then, the load that the remaining regulator can dissipate must not exceed 375 mW - 267.5 mW = 107.5 mW.

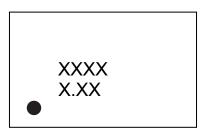
This means, using the same 5V input and 3.3V output voltage, the second regulator is limited to about 60 mA.

Taking advantage of the extremely low-dropout voltage characteristics of the MIC5210, power dissipation can be reduced by using the lowest possible input voltage to minimize the input-to-output voltage drop.

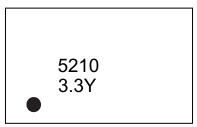
#### 5.0 PACKAGING INFORMATION

#### 5.1 Package Marking Information

8-Lead MSOP\*



Example



**Legend:** XX...X Product code or customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

e3 Pb-free JEDEC® designator for Matte Tin (Sn)

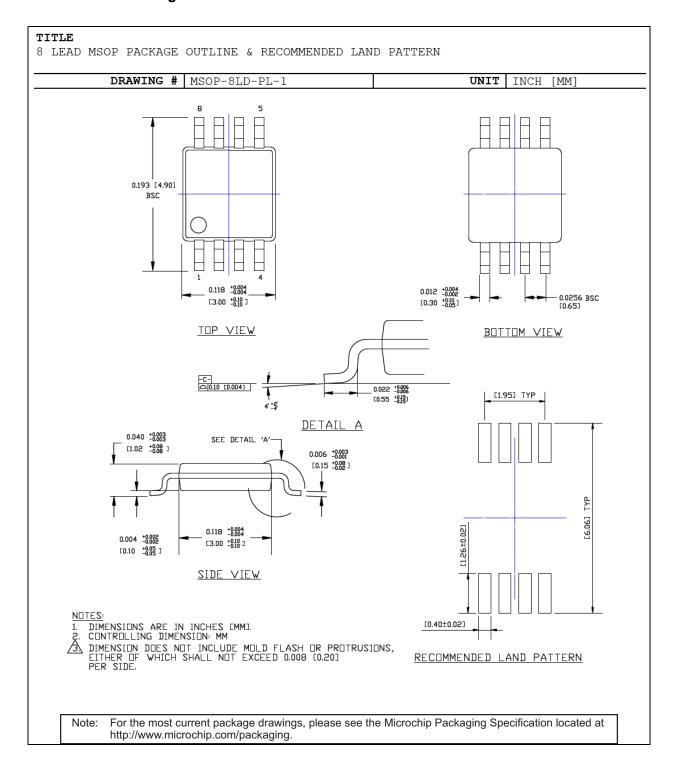
This package is Pb-free. The Pb-free JEDEC designator (©3) can be found on the outer packaging for this package.

•, ▲, ▼ Pin one index is identified by a dot, delta up, or delta down (triangle mark).

**Note**: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information. Package may or may not include the corporate logo.

Underbar (\_) and/or Overbar (¯) symbol may not be to scale.

#### 8-Lead MSOP Package Outline and Recommended Land Pattern



NOTES:

## APPENDIX A: REVISION HISTORY

## Revision A (March 2019)

- Converted Micrel document MIC5210 to Microchip data sheet DS20006096A.
- · Minor text changes throughout.

NOTES:

## PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, contact your local Microchip representative or sales office.

DART	NO VV V VV	Examples:
PART   Devi	— T T T	a) MIC5210-2.8YMM: Dual 150 mA LDO Regulator, 2.8V, -40°C to +125°C, 8-Lead MSOP, 100/Tube
Device:	MIC5210: Dual 150 mA LDO Regulator	b) MIC5210-3.0YMM: Dual 150 mA LDO Regulator, 3.0V, -40°C to +125°C, 8-Lear MSOP, 100/Tube
Voltage:	2.8 = 2.8V 3.0 = 3.0V 3.3 = 3.3V	c) MIC5210-3.3YMM: Dual 150 mA LDO Regulator, 3.3V, -40°C to +125°C, 8-Lear MSOP, 100/Tube
Temperature:	5.0 = 5.0V Y = $-40^{\circ}$ C to +125°C	d) MIC5210-5.0YMM: Dual 150 mA LDO Regulator, 5.0V, -40°C to +125°C, 8-Lead MSOP, 100/Tube
Package:	MM = 8-Lead MSOP	e) MIC5210-2.8YMM-TR: Dual 150 mA LDO Regulator, 2.8V, -40°C to +125°C, 8-Lear MSOP, 2,500/Reel
Media Type:	     TR = 2,500/Reel	f) MIC5210-3.0YMM-TR: Dual 150 mA LDO Regulator, 3.0V, -40°C to +125°C, 8-Lear MSOP, 2,500/Reel
		g) MIC5210-3.3YMM-TR: Dual 150 mA LDO Regulator, 3.3V, -40°C to +125°C, 8-Lear MSOP, 2,500/Reel
		h) MIC5210-5.0YMM-TR: Dual 150 mA LDO Regulator, 5.0V, -40°C to +125°C, 8-Lead MSOP, 2,500/Reel
		Note 1: Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.

NOTES:

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