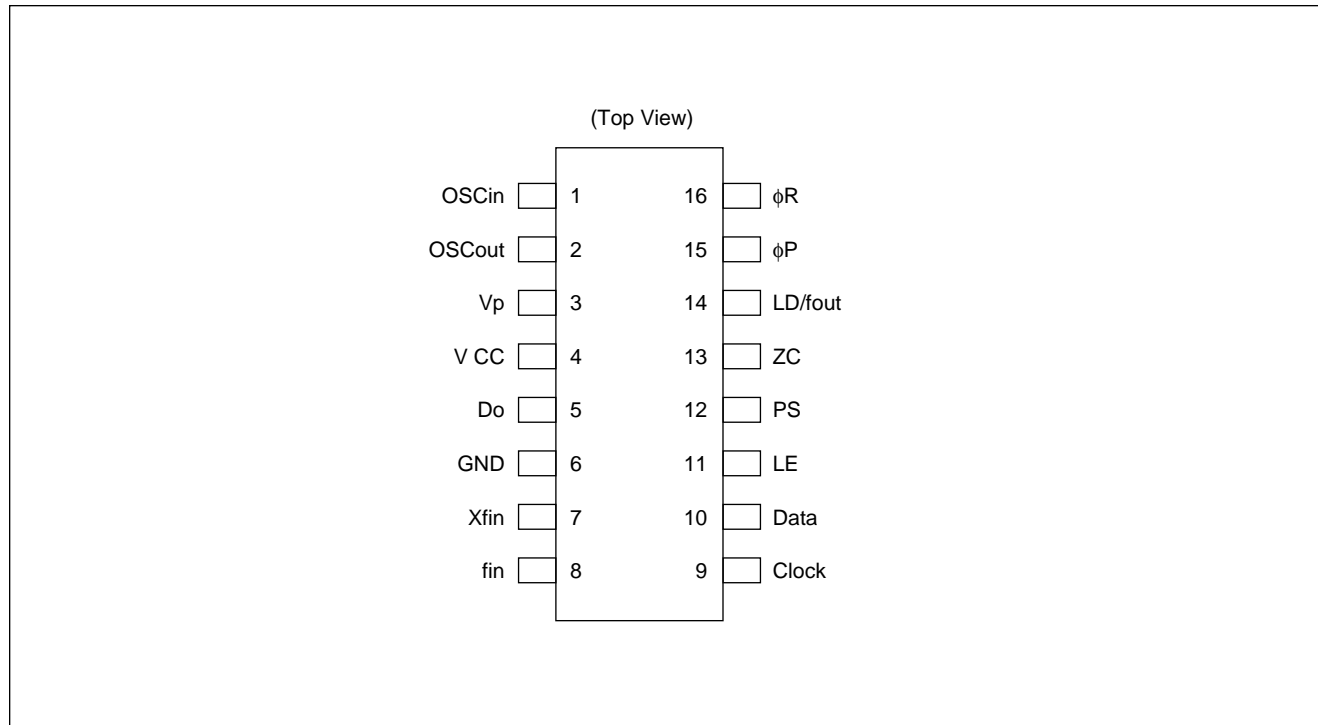


# MB15E07

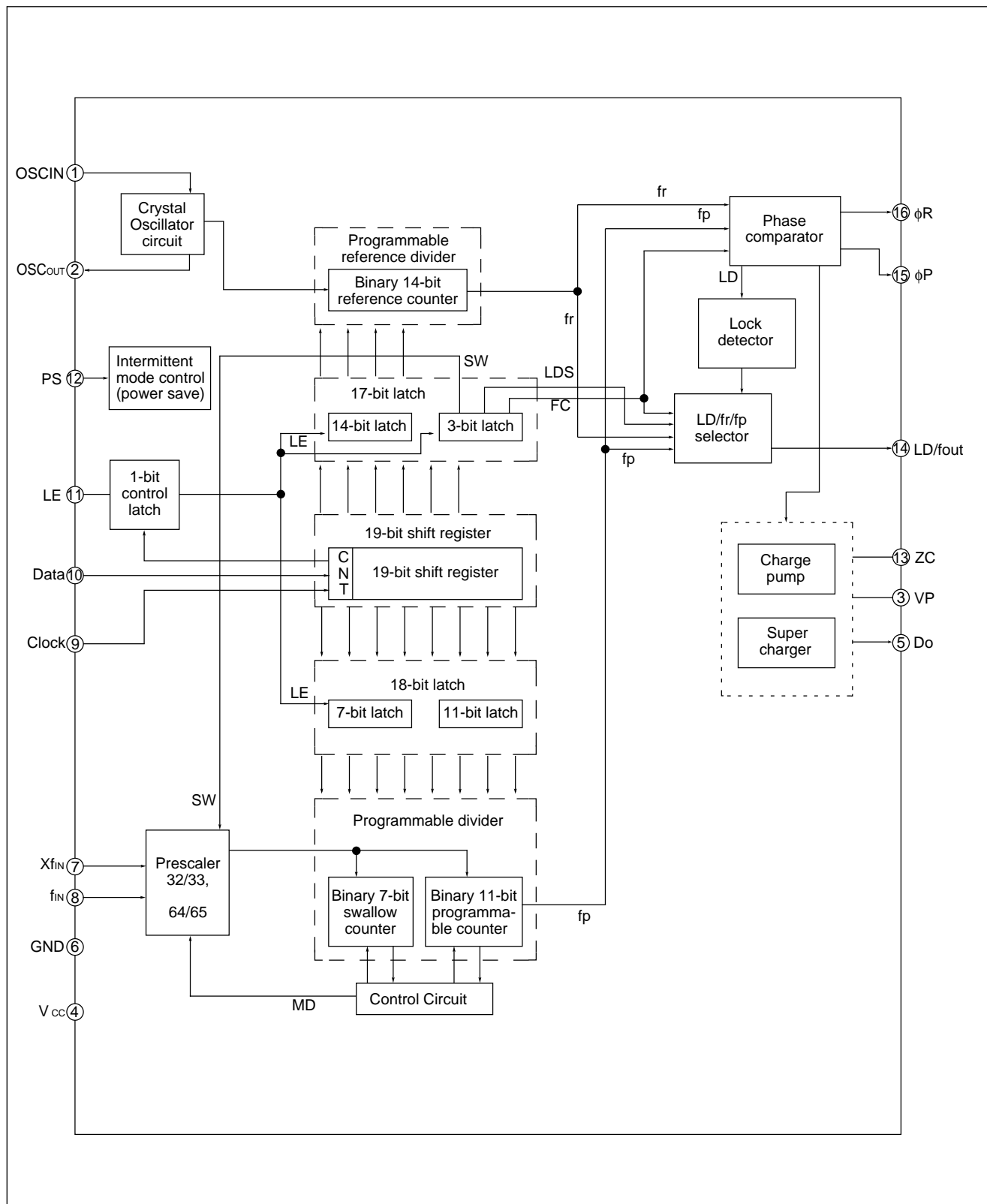
## ■ PIN ASSIGNMENT



## ■ PIN DESCRIPTIONS

Pin no.	Pin name	I/O	Descriptions
1	OSC <sub>IN</sub>	I	Programmable reference divider input. Oscillator input. Connection for an crystal or a TCXO. TCXO should be connected with a coupling capacitor.
2	OSC <sub>OUT</sub>	O	Oscillator output. Connection for an external crystal.
3	V <sub>P</sub>	–	Power supply voltage input for the charge pump.
4	V <sub>CC</sub>	–	Power supply voltage input.
5	D <sub>o</sub>	O	Charge pump output. Phase of the charge pump can be reversed by FC bit.
6	GND	–	Ground.
7	Xfin	I	Prescaler complementary input, and should be grounded via a capacitor.
8	fin	I	Prescaler input. Connection with an external VCO should be done with AC coupling.
9	Clock	I	Clock input for the 19-bit shift register. Data is shifted into the shift register on the rising edge of the clock. ( <i>Open is prohibited.</i> )
10	Data	I	Serial data input using binary code. The last bit of the data is a control bit. ( <i>Open is prohibited.</i> ) Control bit = “H” ;Data is transmitted to the programmable reference counter. Control bit = “L” ;Data is transmitted to the programmable counter.
11	LE	I	Load enable signal input ( <i>Open is prohibited.</i> ) When LE is high, the data in the shift register is transferred to a latch, according to the control bit in the serial data.
12	PS	I	Power saving mode control. This pin must be set at “L” at Power-ON. ( <i>Open is prohibited.</i> ) PS = “H” ; Normal mode PS = “L” ; Power saving mode
13	ZC	I	Forced high-impedance control for the charge pump (with internal pull up resistor.) ZC = “H” ; Normal Do output. ZC = “L” ; Do becomes high impedance.
14	LD/fout	O	Lock detect signal output(LD)/phase comparator monitoring output (fout). The output signal is selected by LDS bit in the serial data. LDS = “H” ; outputs fout (fr/fp monitoring output) LDS = “L” ; outputs LD (“H” at locking, “L” at unlocking.)
15	φP	O	Phase comparator output for an external charge pump. Nch open drain output.
16	φR	O	Phase comparator output for an external charge pump. CMOS output.

## ■ BLOCK DIAGRAM



## ■ ABSOLUTE MAXIMUM RATINGS

Parameter	Symbol	Rating	Unit	Remark
Power supply voltage	$V_{CC}$	-0.5 to +4.0	V	
	$V_P$	$V_{CC}$ to +6.0	V	
Input voltage	$V_I$	-0.5 to $V_{CC}$ +0.5	V	
Output voltage	$V_O$	-0.5 to $V_{CC}$ +0.5	V	
Storage temperature	$T_{stg}$	-55 to +125	°C	

Note: Permanent device damage may occur if the above **Absolute Maximum Ratings** are exceeded. Functional operation should be restricted to the conditions as detailed in the operational sections of this data sheet. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ■ RECOMMENDED OPERATING CONDITIONS

Parameter	Symbol	Value			Unit	Remark
		Min.	Typ.	Max.		
Power supply voltage	$V_{CC}$	2.7	3.0	3.6	V	
	$V_P$	$V_{CC}$	—	6.0	V	
Input voltage	$V_I$	GND	—	$V_{CC}$	V	
Operating temperature	$T_a$	-40	—	+85	°C	

## Handling Precautions

- This device should be transported and stored in anti-static containers.
- This is a static-sensitive device; take proper anti-ESD precautions. Ensure that personnel and equipment are properly grounded. Cover workbenches with grounded conductive mats.
- Always turn the power supply off before inserting or removing the device from its socket.
- Protect leads with a conductive sheet when handling or transporting PC boards with devices.

# MB15E07

## ■ ELECTRICAL CHARACTERISTICS

( $V_{CC} = 2.7$  to  $3.6$  V,  $T_a = -40$  to  $+85^{\circ}\text{C}$ )

Parameter		Symbol	Condition	Value			Unit
				Min.	Typ.	Max.	
Power supply current*1		$I_{CC}$	$f_{in} = 1800$ MHz, $f_{osc} = 12$ MHz, $P = 32/33$	–	8.0	–	mA
Power saving current		$I_{PS}$	ZC = “H” or open	–	0.1	10	$\mu\text{A}$
Operating frequency		$f_{in}$	$P = 32/33$	100	–	1800	MHz
			$P = 64/65$	100	–	2500	MHz
Crystal oscillator operating frequency		$f_{osc}$	min. 500 mVp-p	3	–	40	MHz
Input sensitivity	$f_{in}$	$V_{fin}$	50 $\Omega$ system (Refer to the test circuit.)	–10	–	+2	dBm
	OSCin	$V_{osc}$		500	–	$V_{CC}$	mVp-p
Input voltage	Data, Clock, LE, PS, ZC	$V_{IH}$		$V_{CC} \times 0.7$	–	–	V
		$V_{IL}$		–	–	$V_{CC} \times 0.3$	
Input current	Data, Clock, LE, PS	$I_{IH}$		–1.0	–	+1.0	$\mu\text{A}$
		$I_{IL}$		–1.0	–	+1.0	
	ZC	$I_{IH}$		–1.0	–	+1.0	$\mu\text{A}$
		$I_{IL}$	Pull up input	–100	–	0	
	OSCin	$I_{IH}$		0	–	+100	$\mu\text{A}$
		$I_{IL}$		–100	–	0	
Output voltage	$\phi P$	$V_{OL}$	Open drain output	–	–	0.4	V
	$\phi R$ , LD/fout	$V_{OH}$		$V_{CC} - 0.4$	–	–	V
		$V_{OL}$		–	–	0.4	
	Do	$V_{DOH}$		$V_p - 0.4$	–	–	V
		$V_{DOL}$		–	–	0.4	
High impedance cutoff current	Do	$I_{OFF}$		–	–	1.1	$\mu\text{A}$
Output current	$\phi P$	$I_{OL}$	Open drain output	–	–	1.0	mA
	$\phi R$ , LD/fout	$I_{OH}$		–	–	–1.0	mA
		$I_{OL}$		1.0	–	–	
	Do	$I_{DOH}$	$V_{CC} = 3.0$ V, $V_p = 5$ V, $V_{DOH} = 4.0$ V, $T_a = 25^{\circ}\text{C}$	–	–10.0	–	mA
		$I_{DOL}$	$V_{CC} = 3.0$ V, $V_p = 5$ V, $V_{DOL} = 1.0$ V, $T_a = 25^{\circ}\text{C}$	–	10.0	–	

\*1: Conditions;  $V_{CC} = 3.0$  V,  $T_a = 25^{\circ}\text{C}$ , in locking state.

## ■ FUNCTION DESCRIPTIONS

### Pulse Swallow Function

The divide ratio can be calculated using the following equation:

$$f_{VCO} = [(P \times N) + A] \times f_{osc} \div R \quad (A < N)$$

- $f_{VCO}$  : Output frequency of external voltage controlled oscillator (VCO)
- $N$  : Preset divide ratio of binary 11-bit programmable counter (5 to 2,047)
- $A$  : Preset divide ratio of binary 7-bit swallow counter ( $0 \leq A \leq 127$ )
- $f_{osc}$  : Output frequency of the reference frequency oscillator
- $R$  : Preset divide ratio of binary 14-bit programmable reference counter (5 to 16,383)
- $P$  : Preset divide ratio of modules prescaler (32 or 64)

### Serial Data Input

Serial data is processed using the Data, Clock, and LE pins. Serial data controls the programmable reference divider and the programmable divider separately.

Binary serial data is entered through the Data pin.

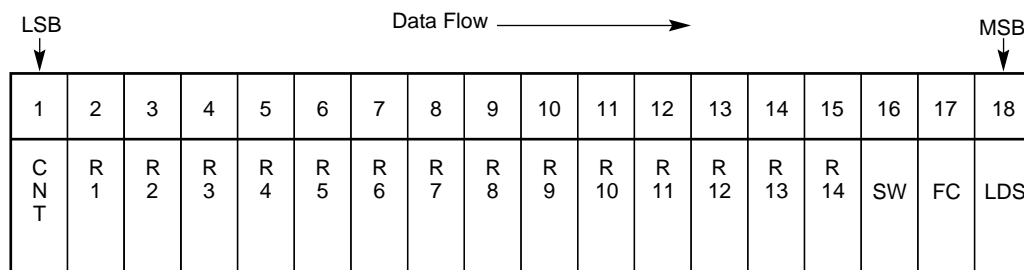
One bit of data is shifted into the shift register on the rising edge of the clock. When the load enable pin is high, stored data is latched according to the control bit data as follows:

**Table.1 Control Bit**

Control bit (CNT)	Destination of serial data
H	17 bit latch (for the programmable reference divider)
L	18 bit latch (for the programmable divider)

### Shift Register Configuration

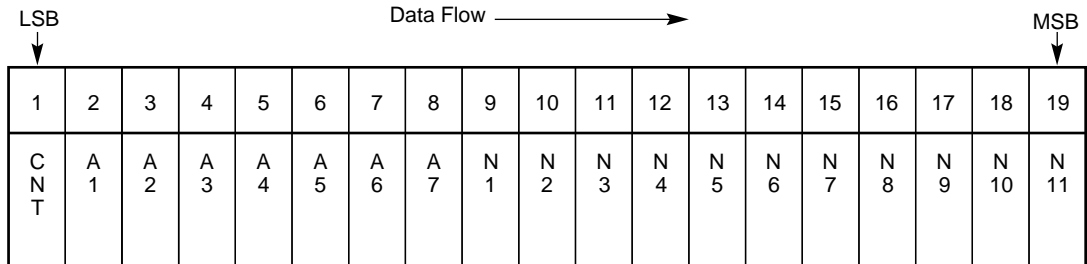
#### Programmable Reference Counter



- CNT : Control bit [Table. 1]
- R1 to R14 : Divide ratio setting bit for the programmable reference counter (5 to 16,383) [Table. 2]
- SW : Divide ratio setting bit for the prescaler (32/33 or 64/65) [Table. 5]
- FC : Phase control bit for the phase comparator [Table. 7]
- LDS : LD/fout signal select bit [Table. 6]

Note : Start data input with MSB first

## Programmable Reference Counter



CNT : Control bit

N1 to N11 : Divide ratio setting bits for the programmable counter (5 to 2,047)

A1 to A7 : Divide ratio setting bits for the swallow counter (0 to 127)

[Table. 1]

[Table. 2]

[Table. 4]

Note : Start data input with MSB first

**Table2. Binary 14-bit Programmable Reference Counter Data Setting**

Divide ratio (R)	R14	R13	R12	R11	R10	R9	R8	R7	R6	R5	R4	R3	R2	R1
5	0	0	0	0	0	0	0	0	0	0	0	1	0	1
6	0	0	0	0	0	0	0	0	0	0	0	1	1	0
•	•	•	•	•	•	•	•	•	•	•	•	•	•	•
16383	1	1	1	1	1	1	1	1	1	1	1	1	1	1

Note: • Divide ratio less than 5 is prohibited.

**Table.3 Binary 11-bit Programmable Counter Data Setting**

Divide ratio (N)	N11	N10	N9	N8	N7	N6	N5	N4	N3	N2	N1
5	0	0	0	0	0	0	0	0	1	0	1
6	0	0	0	0	0	0	0	0	1	1	0
•	•	•	•	•	•	•	•	•	•	•	•
2047	1	1	1	1	1	1	1	1	1	1	1

Note: • Divide ratio less than 5 is prohibited.

• Divide ratio (N) range = 5 to 2,047

**Table.4 Binary 7-bit Swallow Counter Data Setting**

Divide ratio (A)	A 7	A 6	A 5	A 4	A 3	A 2	A 1
0	0	0	0	0	0	0	0
1	0	0	0	0	0	0	1
•	•	•	•	•	•	•	•
127	1	1	1	1	1	1	1

Note: • Divide ratio (A) range = 0 to 127

**Table. 5 Prescaler Data Setting**

SW	Prescaler Divide ratio
H	32/33
L	64/65

**Table. 6 LD/fout Output Select Data Setting**

LDS	LD/fout output signal
H	fout signal
L	LD signal

## Relation between the FC input and phase characteristics

The FC bit changes the phase characteristics of the phase comparator. Both the internal charge pump output level ( $D_o$ ) and the phase comparator output ( $\phi_R$ ,  $\phi_P$ ) are reversed according to the FC bit. Also, the monitor pin ( $f_{OUT}$ ) output is controlled by the FC bit. The relationship between the FC bit and each of  $D_o$ ,  $\phi_R$ , and  $\phi_P$  is shown below.

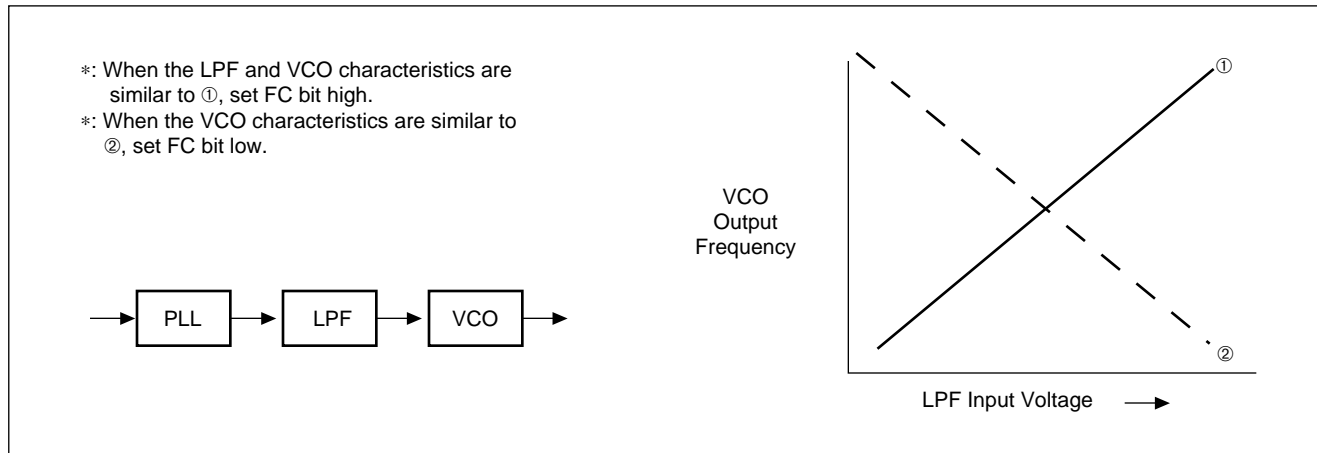
**Table. 7 FC Bit Data Setting (LDS = “H”)**

	FC = High				FC = Low			
	$D_o$	$\phi_R$	$\phi_P$	LD/fout	$D_o$	$\phi_R$	$\phi_P$	LD/fout
$f_r > f_p$	H	L	L	(fr)	L	H	Z*	(fp)
$f_r < f_p$	L	H	Z*	(fr)	H	L	L	(fp)
$f_r = f_p$	Z*	L	Z*	(fr)	Z*	L	Z*	(fp)

\* : High impedance



When designing a synthesizer, the FC pin setting depends on the VCO and LPF characteristics.



## Power Saving Mode (Intermittent Mode Control Circuit)

Setting a PS pin to Low, the IC enters into power saving mode resultatly current sonsumption can be limited to 10 $\mu$ A (max.). Setting PS pin to High, power saving mode is released so that the IC works normally.

In addition, the intermittent operation control circuit is included which helps smooth start up from the power saving mode. In general, the power consumption can be saved by the intermittent operation that powering down or waking up the synthesizer. Such case, if the PLL is powered up uncontrolled, the resulting phase comparator output signal is unpredictable due to an undefined phase relation between reference frequency ( $f_r$ ) and comparison frequency ( $f_p$ ) and may in the worst case take longer time for lock up of the loop.

To prevent this, the intermittent operation control circuit enforces a limited error signal output of the phase detector during power up, thus keeping the loop locked.

During the power saving mode, the corresponding section except for indispensable circuit for the power saving function stops working, then current consumption is reduced to 10  $\mu$ A (max.).

At that time, the Do and LD become the same state as when a loop is locking. That is, the Do becomes high impedance.

A VCO control voltage is naturally kept at the locking voltage which defined by a LPF's time constant. As a result of this, VCO's frequency is kept at the locking frequency.

- Note:
- While the power saving mode is executed, ZC pin should be set at "H" or open. If ZC is set at "L" during power saving mode, approximately 10  $\mu$ A current flows.
  - PS pin must be set "L" at Power-ON.
  - The power saving mode can be released (PS : L  $\rightarrow$  H) 1 $\mu$ s later after power supply remains stable.
  - During the power saving mode, it is possible to input the serial data.

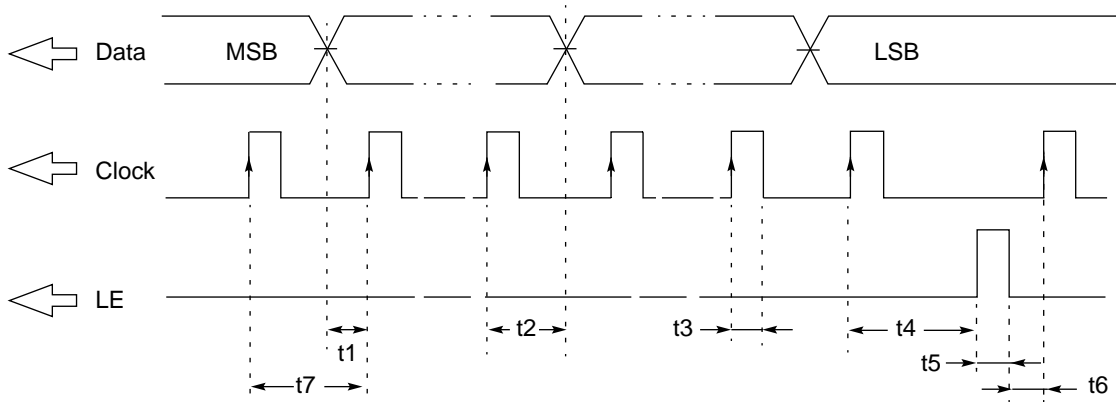
**Table.8 PS Pin Setting**

PS pin	Status
H	Normal mode
L	Power saving mode

**Table.9 ZC Pin Setting**

ZC pin	Do output
H	Normal output
L	High impedance

## ■ SERIAL DATA INPUT TIMING

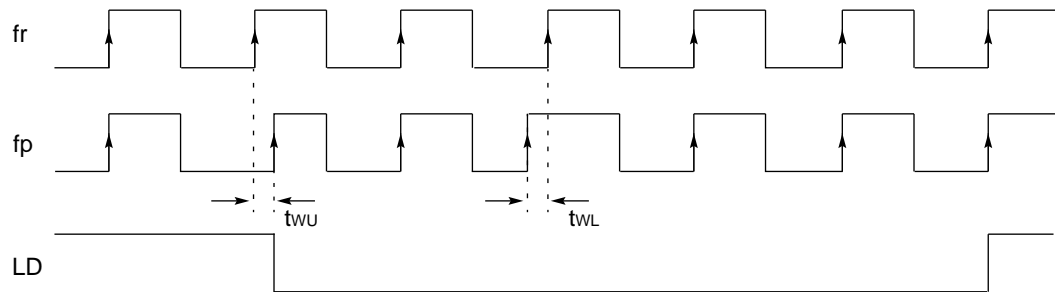


On rising edge of the clock, one bit of the data is transferred into the shift register.

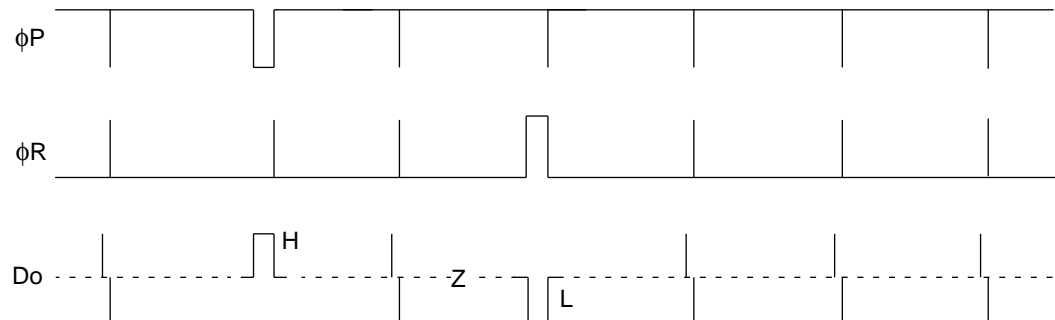
Parameter	Min.	Typ.	Max.	Unit
t1	20	–	–	ns
t2	20	–	–	ns
t3	30	–	–	ns
t4	30	–	–	ns

Parameter	Min.	Typ.	Max.	Unit
t5	100	–	–	ns
t6	20	–	–	ns
t7	100	–	–	ns

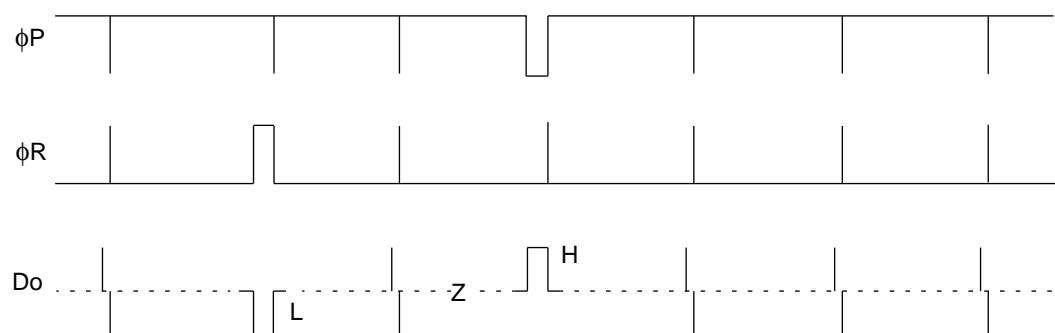
## ■ PHASE COMPARATOR OUTPUT WAVEFORM



[ FC = "H" ]

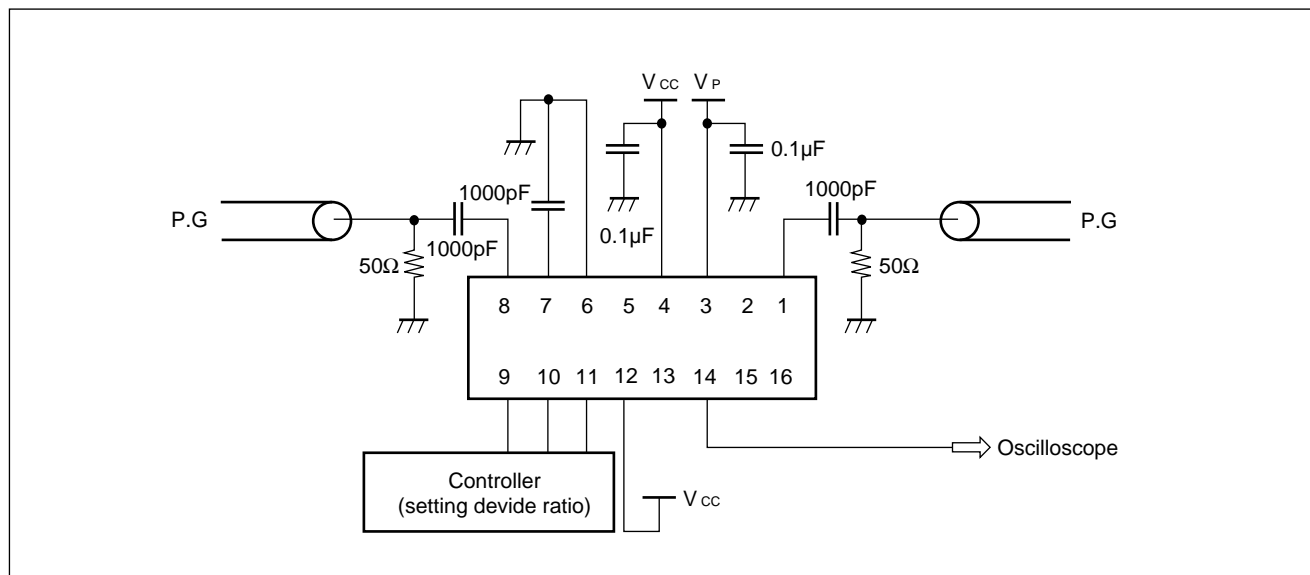


[ FC = "L" ]



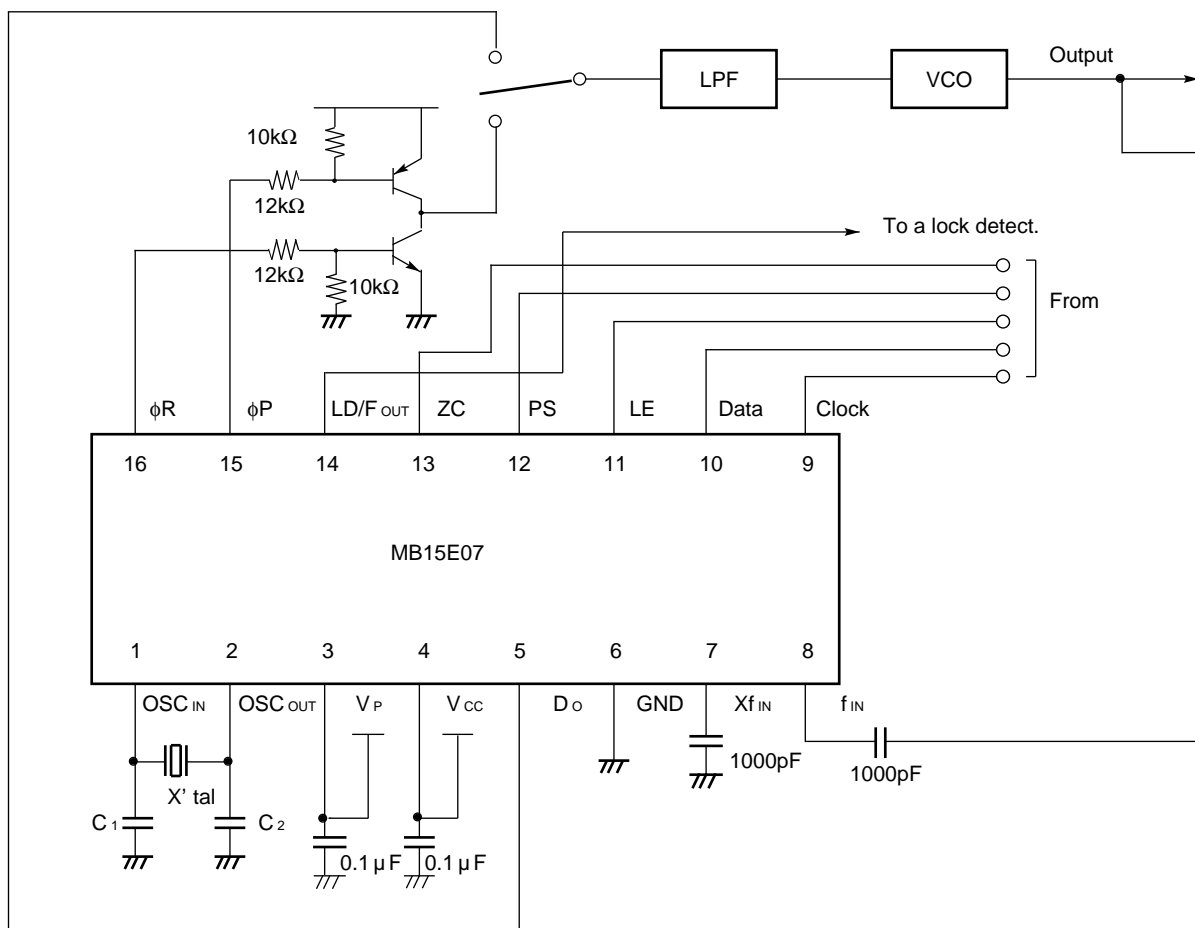
- Notes:
1. Phase error detection range:  $-2\pi$  to  $+2\pi$
  2. Pulses on Do output signal during locked state are output to prevent dead zone.
  3. LD output becomes low when phase is  $twu$  or more. LD output becomes high when phase error is  $twl$  or less and continues to be so for three cycles or more.
  4.  $twu$  and  $twl$  depend on OSCin input frequency.  
 $twu \geq 8/fosc$  (e.g.  $twu \geq 625ns$ ,  $fosc = 12.8 MHz$ )  
 $twl \leq 16/fosc$  (e.g.  $twl \leq 1250ns$ ,  $fosc = 12.8 MHz$ )
  5. LD becomes high during the power saving mode (PS = "L".)

## ■ TEST CIRCUIT (for Measuring Input Sensitivity fin/OSCin)



# MB15E07

## ■ APPLICATION EXAMPLE

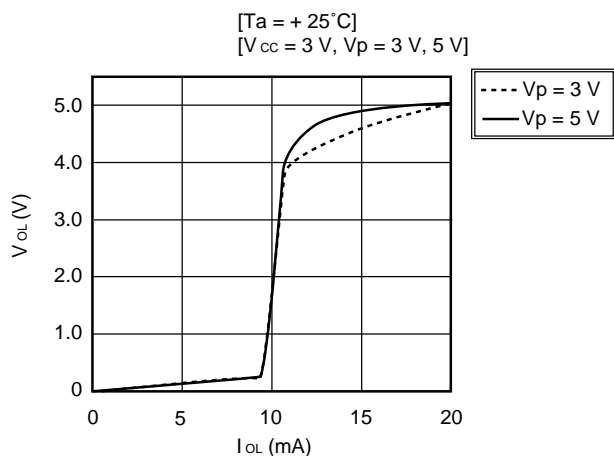
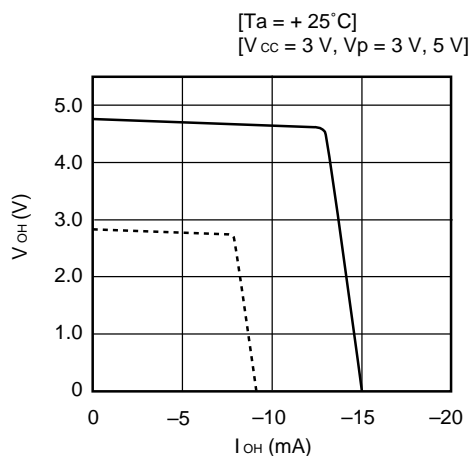


$C_1, C_2$  : Depend on the crystal parameters

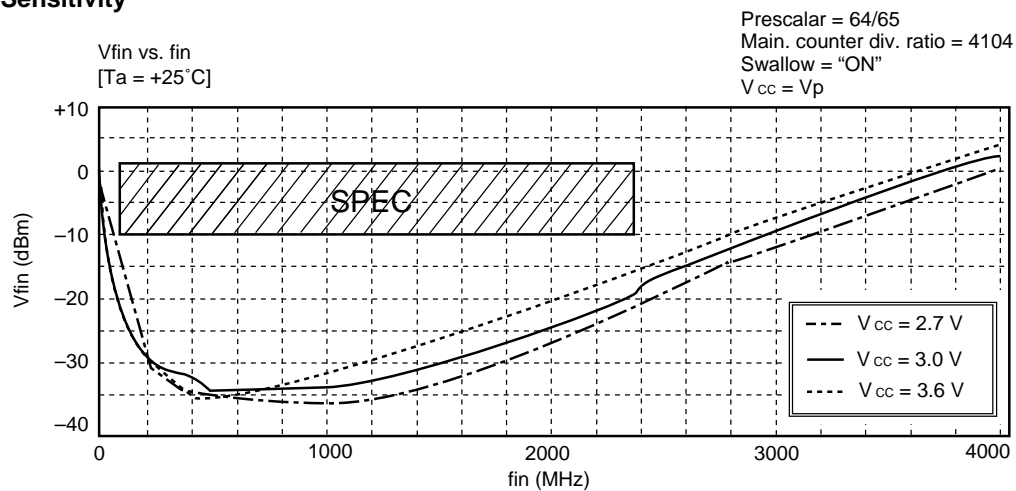
(Continued)

## ■ TYPICAL CHARACTERISTICS

### Do Output Current



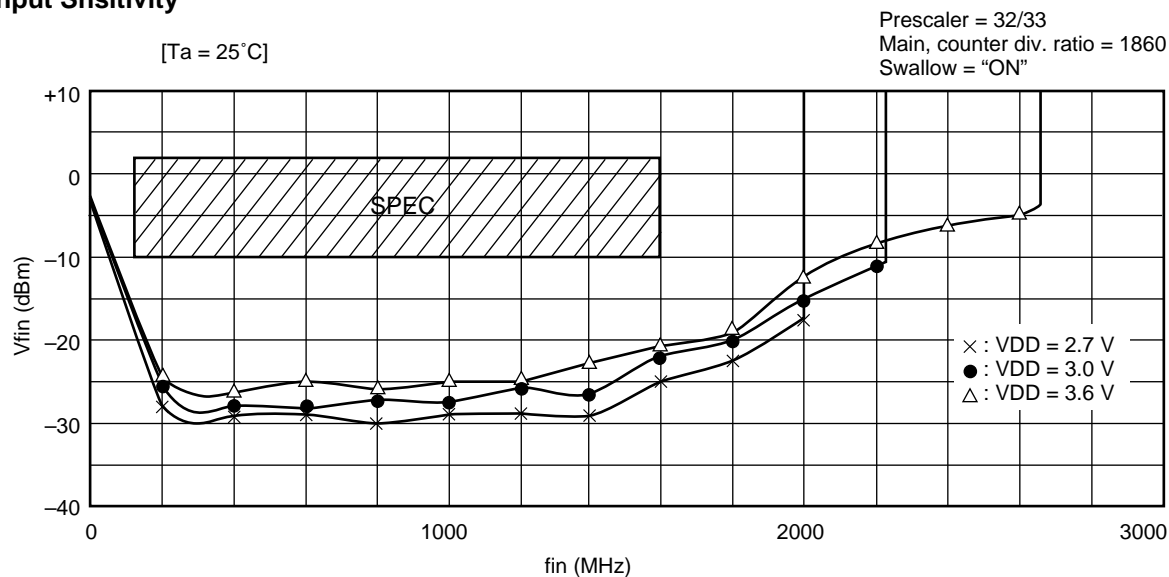
### fin Input Sensitivity



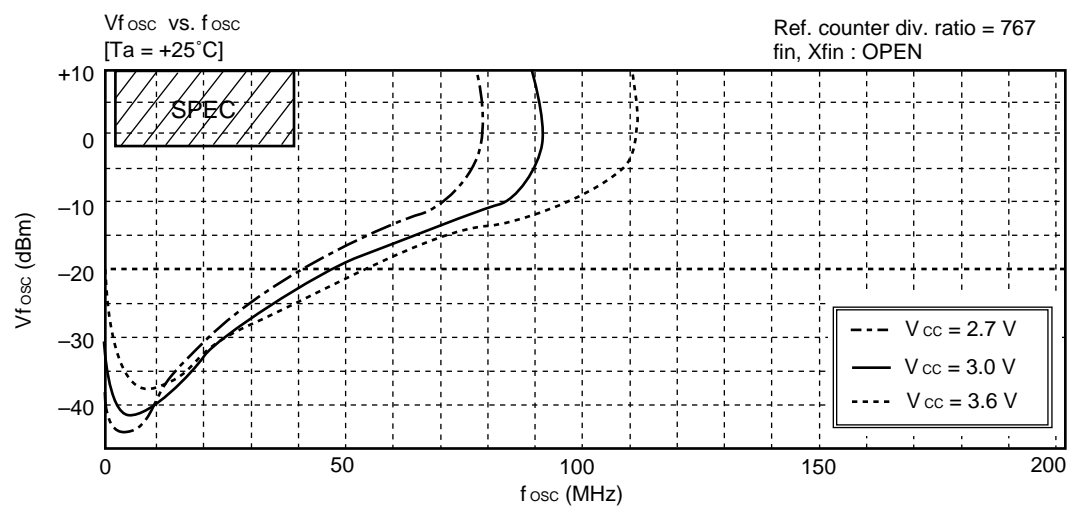
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## fin Input Snsitivity



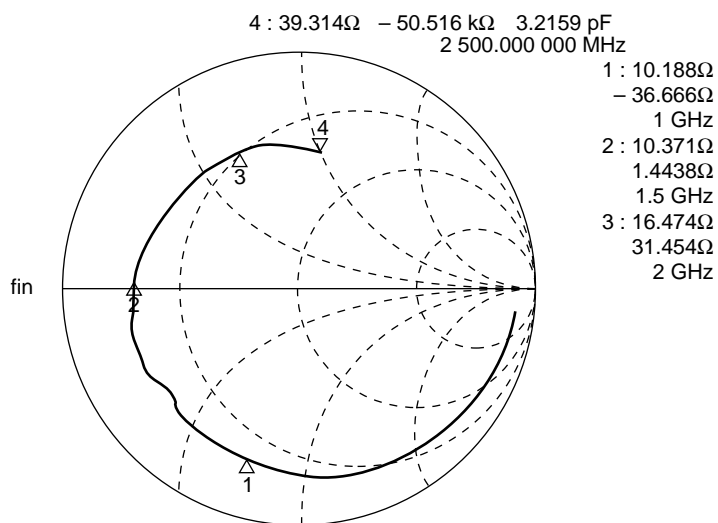
## OSCin Input Characteristics



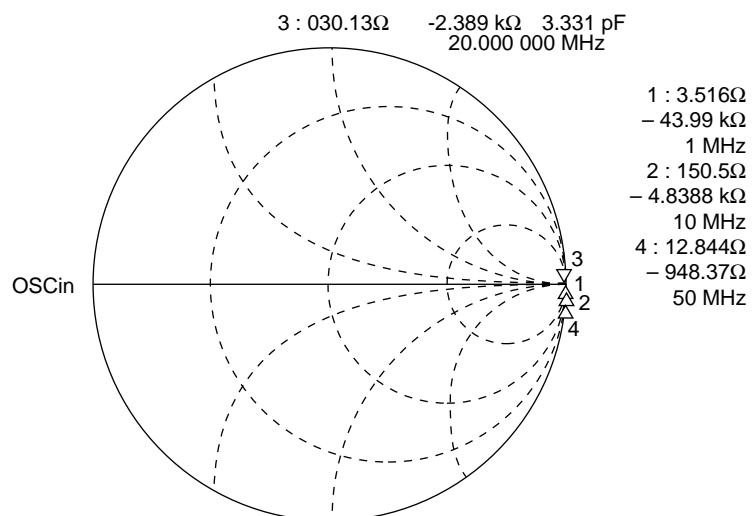
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## fin Input Impedance



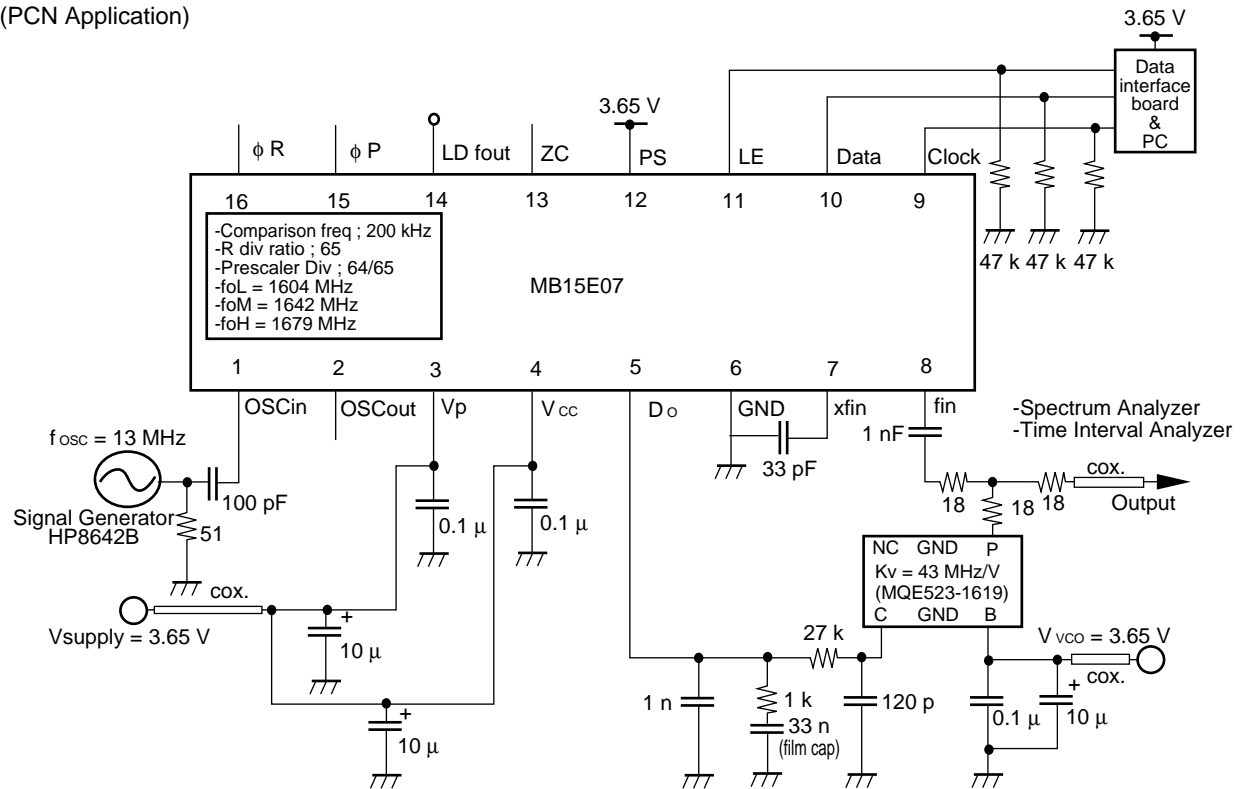
## OSCin Input Impedance





## REFERENCE INFORMATION

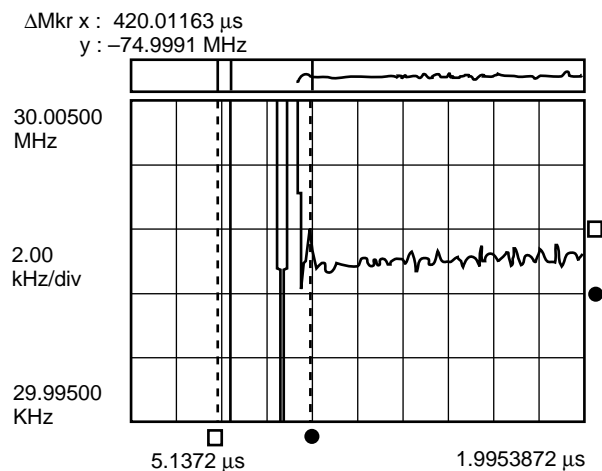
Test Circuit Diagram  
(PCN Application)



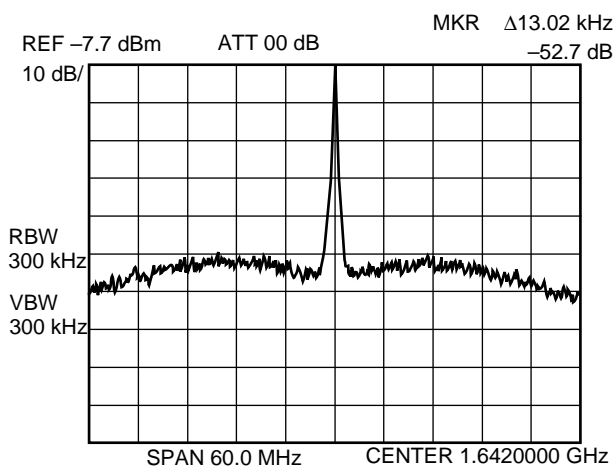
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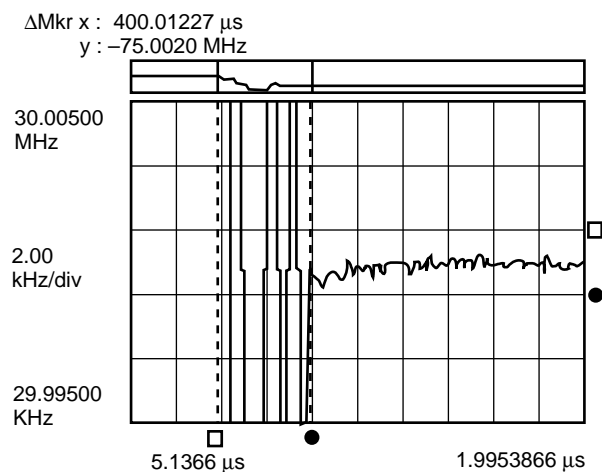
PLL Lock Up toem = 420  $\mu$ s  
(1604 MHz  $\rightarrow$  1679 MHz, within  $\pm$  1 kHz)



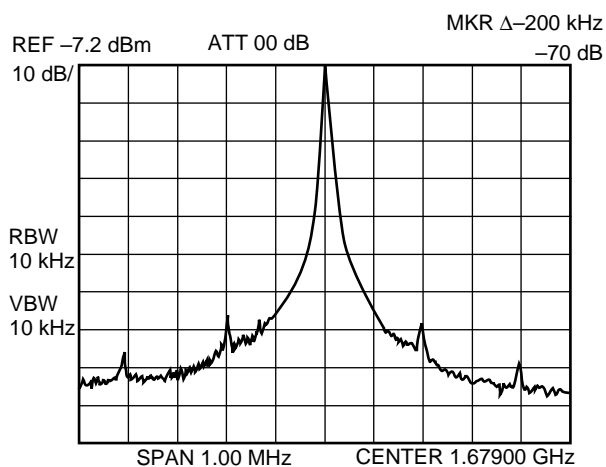
PLL Phase Noise  
@within loop band = -77.5 dBc/Hz



PLL Lock Up toem = 400  $\mu$ s  
(1679 MHz  $\rightarrow$  1604 MHz, within  $\pm$  1 kHz)



PLL Reference Leakage  
200 MHz offset = -70.0 dBc



# MB15E07

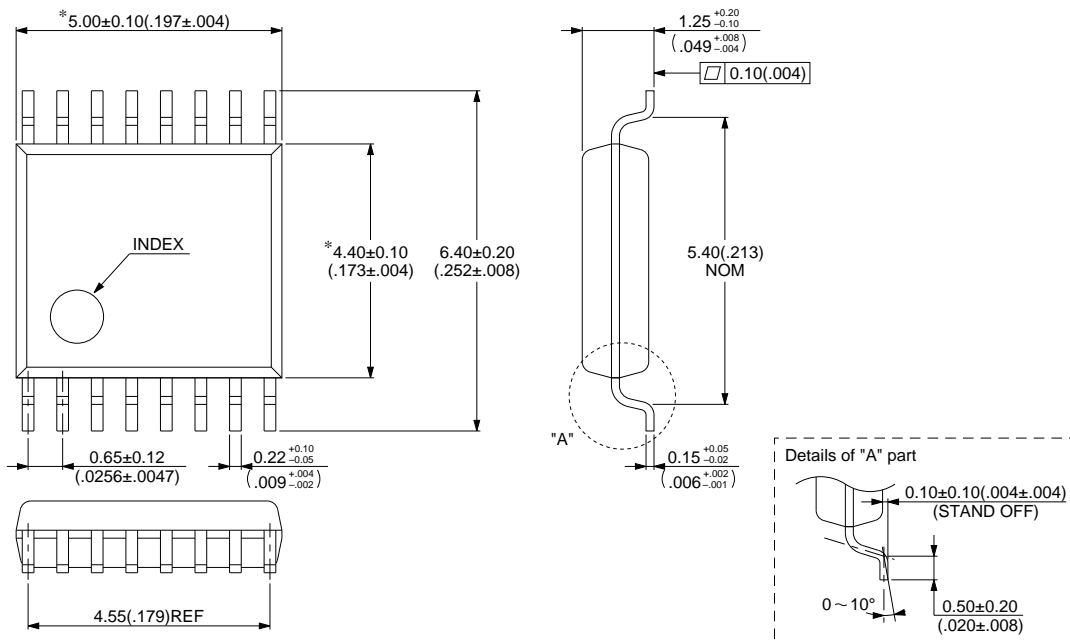
## ■ ORDERING INFORMATION

Part number	Package	Remarks
MB15E07PFV1	16-pin Plastic SSOP (FPT-16P-M05)	

## ■ PACKAGE DIMENSION

16 pins, Plastic SSOP  
(FPT-16P-M05)

\* : These dimensions do not include resin protrusion.



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Dimensions in mm (inches)

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