ABSOLUTE MAXIMUM RATINGS

| Terminal Voltage (with respect to GND) | Rate of Rise, V _{CC} , V _{BATT} 100V/µs |
|---|--|
| $egin{array}{llll} V_{CC} & & & -0.3 V 	ext{ to } 6.0 V \\ V_{BATT} & & -0.3 V 	ext{ to } 6.0 V \\ \end{array}$ | Continuous Power Dissipation Plastic DIP (derate 9.09mW/°C above +70°C) 727mW |
| All Other Inputs (Note 1)0.3V to (V _{CC} + 0.3V) | SO (derate 5.88mW/°C above +70°C) |
| Input Current | CERDIP (derate 8.00mW/°C above +70°C) 640mW |
| V _{CC} | Operating Temperature Ranges: |
| V _{BATT} 50mA | MAX69_AC, MAX80 C 0°C to +70°C |
| GND 20mA | MAX69_AE, MAX80E40°C to +85°C |
| Output Current | MAX69_AMJA, MAX805LMJA55°C to +125°C |
| V _{OUT} Short-Circuit Protected for up to 10sec All Other Outputs 20mA | Storage Temperature Range65°C to +160°C Lead Temperature (soldering, 10sec) +300°C |

Note 1: The input voltage limits on PFI and WDI may be exceeded if the current into these pins is limited to less than 10mA.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{CC} = 4.75V to 5.5V for MAX690A/MAX802L/MAX805L, V_{CC} = 4.5V to 5.5V for MAX692A/MAX802M, V_{BATT} = 2.8V, T_A = T_{MIN} to T_{MAX}, unless otherwise noted.)

| PARAMETER | SYMBOL | CONI | MIN | TYP | MAX | UNITS | | |
|--|---------------------|---|------------------------------|--|---------------------|-------|----|--|
| Operating Voltage Range, | | MAX69_AC, MAX802_C | | 1.0 | | 5.5 | | |
| V _{CC} , V _{BATT} (Note 2) | | MAX805LC MAX69_AE/M, MAX80E | | 1.1 | | 5.5 | V | |
| CC, 'BATT (Note 2) | | | | 1.2 | | 5.5 | | |
| Supply Current (Excluding I _{OUT}) | 1 | MAX69_AC, MAX8 | 302_C | | 200 | 350 | μΑ | |
| Supply Surrent (Excidently 1001) | I _{SUPPLY} | MAX69_AE/M, MAX | (802_E, MAX805LE/M | | 200 | 500 | | |
| I _{SUPPLY} in Battery-Backup Mode | | $V_{CC} = 0V$, | $T_A = +25^{\circ}C$ | | 0.05 | 1.0 | | |
| (Excluding I _{OUT}) | | $V_{BATT} = 2.8V$ | $T_A = T_{MIN}$ to T_{MAX} | | | 5.0 | μA | |
| V _{BATT} Standby Current (Note 3) | | 5.5V > V _{CC} > | $T_A = +25^{\circ}C$ | -0.1 | | 0.02 | ^ | |
| V _{BATT} Standby Current (Note 3) | | V _{BATT} +0.2V | $T_A = T_{MIN}$ to T_{MAX} | -1.0 | | 0.02 | μΑ | |
| V Output | | I _{OUT} = 5mA | | V _{CC} - 0.05 V _{CC} - 0.025 | | | | |
| V _{OUT} Output | | I _{OUT} = 50mA | | V _{CC} - 0.5 | V _{CC} - C |).25 | V | |
| V _{OUT} in Battery-Backup Mode | | $I_{OUT} = 250\mu A, V_{CC}$ | < V _{BATT} - 0.2V | V _{BATT} - 0.1 | V _{BATT} - | 0.02 | V | |
| Battery Switch Threshold, V _{CC} | | Power-up | | | 20 | | \/ | |
| to V _{BATT} | | $V_{CC} < V_{RT}$ | Power-down | | -20 | | mV | |
| Battery Switchover Hysteresis | | | | | 40 | | mV | |
| | | MAX690A, MAX80 | 2L, MAX805L | 4.50 | 4.65 | 4.75 | | |
| | V | MAX692A, MAX80 | 4.25 | 4.40 | 4.50 | V | | |
| Reset Threshold | V _{RT} | MAX802L, $T_A = +2$ | 4.55 | | 4.70 | | | |
| | | MAX802M, $T_A = +$ | 4.30 | | 4.45 | | | |
| Reset Threshold Hysteresis | | | | | 40 | | mV | |
| Reset Pulse Width | t _{RS} | | | 140 | 200 | 280 | ms | |
| | | I _{SOURCE} = 800μA | | V _{CC} - 1.5 | | | | |
| | | $I_{SINK} = 3.2mA$ | | | | 0.4 | | |
| RESET Output Voltage | | MAX69_AC, MAX802_C, $V_{CC} = 1.0V$ $I_{SINK} = 50\mu A$ | | | | 0.0 | V | |
| | | | | | | 0.3 | | |
| | | MAX69_AE/M, MA | | 0.0 | | | | |
| | | $V_{CC} = 1.2V, I_{SINK} =$ | | 0.3 | | | | |

ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = 4.75V to 5.5V for MAX690A/MAX802L/MAX805L, V_{CC} = 4.5V to 5.5V for MAX692A/MAX802M, V_{BATT} = 2.8V $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted.)

| PARAMETER | SYMBOL | CONE | MIN | TYP | MAX | UNITS | | |
|-------------------------------|-----------------|--|---------------------------|-----------------------|-------|-------|-------|--|
| | | MAX805LC, I _{SOURCE} | $= 4\mu A, V_{CC} = 1.1V$ | 0.8 | | | | |
| RESET Output Voltage | | MAX805LE/M, I _{SOUR} | 0.9 | | |] , | | |
| NESET Output Voltage | | MAX805L, I _{SOURCE} = | V _{CC} - 1.5 | | | | | |
| | | MAX805L, $I_{SINK} = 3.5$ | 2mA | | | 0.4 | 1 | |
| Watchdog Timeout | t _{WD} | | | 1.00 | 1.60 | 2.25 | sec | |
| WDI Pulse Width | t _{WP} | $V_{IL} = 0.4V, V_{IH} = (0.8)$ | B) (V _{CC}) | 50 | | | ns | |
| WDI Input Threshold (Note 4) | | $V_{CC} = 5V$ | Logic low | | | 0.8 | V | |
| WDI Input Tilleshold (Note 4) | | | Logic high | 3.5 | | | | |
| WDI Input Current | | $WDI = V_{CC}$ | | | 50 | 150 | - μΑ | |
| WDI Input Current | | WDI = 0V | -150 | -50 | | | | |
| PFI Input Threshold | | MAX69_A, MAX805L, V _{CC} = 5V | | 1.20 | 1.25 | 1.30 | V | |
| FFI IIIput Tillesiloid | | MAX802_C/E, $V_{CC} = 5V$ | | 1.225 | 1.250 | 1.275 | \ \ \ | |
| PFI Input Current | | | | -25 | 0.01 | 25 | nA | |
| PFO Output Voltage | | I _{SOURCE} = 800μA | | V _{CC} - 1.5 | | | V | |
| 110 Output voitage | | $I_{SINK} = 3.2mA$ | | | | 0.4 |] | |

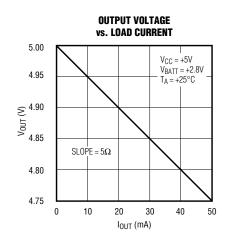
Note 2: Either V_{CC} or V_{BATT} can go to 0V, if the other is greater than 2.0V.

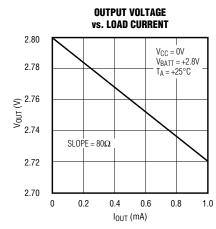
Note 3: "-" = battery-charging current, "+" = battery-discharging current.

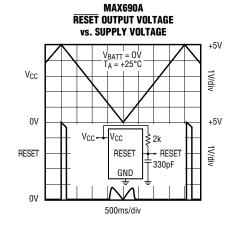
Note 4: WDI is guaranteed to be in an intermediate, non-logic level state if WDI is floating and V_{CC} is in the operating voltage range. WDI is internally biased to 35% of V_{CC} with an input impedance of $50k\Omega$.

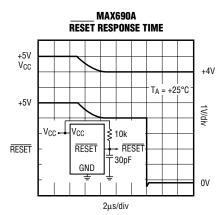
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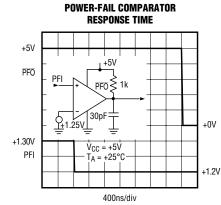
Typical Operating Characteristics

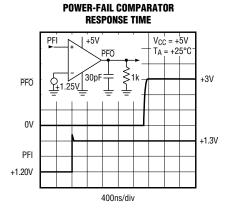


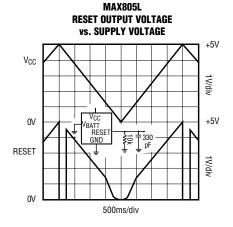


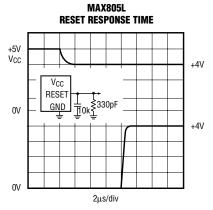












4 /V|/X|/VI

Pin Description

| PIN | | | | | | |
|------------------------------------|---------|-------------------|--|--|--|--|
| MAX690A/MAX692A MAX802L/MAX802M | MAX805L | NAME | FUNCTION | | | |
| 1 | 1 | V _{OUT} | Supply Output for CMOS RAM. When V_{CC} is above the reset threshold, V_{OUT} connects to V_{CC} through a P-channel MOSFET switch. When V_{CC} is below the reset threshold, the higher of V_{CC} or V_{BATT} will be connected to V_{OUT} . | | | |
| 2 | 2 | V _{CC} | +5V Supply Input | | | |
| 3 | 3 | GND | Ground | | | |
| 4 | 4 | PFI | Power-Fail Comparator Input. When PFI is less than 1.25V, PFO goes low. Connect PFI to GND or V _{CC} when not used. | | | |
| 5 | 5 | PFO | Power-Fail Output. When PFI is less than 1.25V, PFO goes low; otherwise PFO stays high. | | | |
| 6 | 6 | WDI | Watchdog Input. If WDI remains high or low for 1.6sec, the internal watchdog timer runs out and reset is triggered. If WDI is left floating or connected to a high-impedance three-state buffer, the watchdog feature is disabled. The internal watchdog timer clears whenever reset is asserted, WDI is three-stated, or WDI sees a rising or falling edge. | | | |
| 7 | | RESET | Reset Output. Whenever $\overline{\text{RESET}}$ is triggered, it pulses low for 200ms. It stays low when V_{CC} is below the reset threshold (4.65V in the MAX690A/MAX802L and 4.4V in the MAX692A/MAX802M) and remains low for 200ms after V_{CC} rises above the reset threshold. A watchdog timeout also triggers $\overline{\text{RESET}}$. | | | |
| - | 7 | RESET | Active-High Reset Output is the inverse of $\overline{\text{RESET}}$. When RESET is asserted, the RESET output voltage = V_{CC} or V_{BATT} , whichever is higher. | | | |
| 8 | 8 | V _{BATT} | Backup-Battery Input. When V_{CC} falls below the reset threshold, V_{BATT} will be switched to V_{OUT} if V_{BATT} is 20mV greater than V_{CC} . When V_{CC} rises to 20mV above V_{BATT} , V_{OUT} will be reconnected to V_{CC} . The 40mV hysteresis prevents repeated switching if V_{CC} falls slowly. | | | |



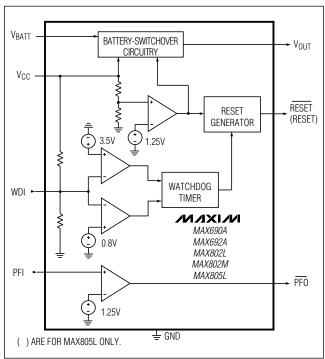


Figure 1. Block Diagram

Detailed Description

Reset Output

A microprocessor's (μ P's) reset input starts the μ P in a known state. When the μ P is in an unknown state, it should be held in reset. The MAX690A/MAX692A/MAX802L/MAX802M assert reset during power-up and prevent code execution errors during power-down or brownout conditions.

On power-up, once V_{CC} reaches 1V, \overline{RESET} is guaranteed to be a logic low. As V_{CC} rises, \overline{RESET} remains low. When V_{CC} exceeds the reset threshold, an internal timer keeps \overline{RESET} low for a time equal to the reset pulse width; after this interval, \overline{RESET} goes high (Figure 2). If a brownout condition occurs (if V_{CC} dips below the reset threshold), \overline{RESET} is triggered. Each time \overline{RESET} is triggered, it stays low for the reset pulse width interval. Any time V_{CC} goes below the reset threshold, the internal timer restarts the pulse. If a brownout condition interrupts a previously initiated reset pulse, the reset pulse continues for another 200ms. On power-down, once V_{CC} goes below the threshold, \overline{RESET} is guaranteed to be logic low until V_{CC} droops below 1V.

RESET is also triggered by a watchdog timeout. If a high or low is continuously applied to the WDI pin for 1.6sec, RESET pulses low. As long as RESET is assert-

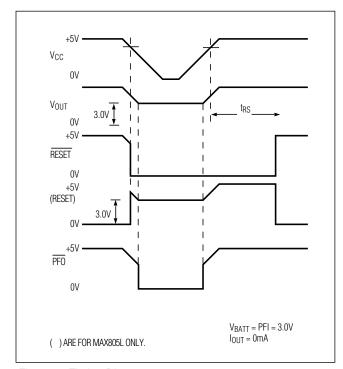


Figure 2. Timing Diagram

ed, the watchdog timer remains clear. When RESET comes high, the watchdog resumes timing and must be serviced within 1.6sec. If WDI is tied high or low, a RESET pulse is triggered every 1.8sec (t_{WD} plus t_{RS}).

The MAX805L active-high RESET output is the inverse of the MAX690A/MAX692A/MAX802L/MAX802M $\overline{\text{RESET}}$ output, and is guaranteed to be valid with V_{CC} down to 1.1V. Some µPs, such as Intel's 80C51, require an active-high reset pulse.

Watchdog Input

The watchdog circuit monitors the μP 's activity. If the μP does not toggle the watchdog input (WDI) within 1.6sec, a reset pulse is triggered. The internal 1.6sec timer is cleared by either a reset pulse or by open circuiting the WDI input. As long as reset is asserted or the WDI input is open circuited, the timer remains cleared and does not count. As soon as reset is released or WDI is driven high or low, the timer starts counting. It can detect pulses as short as 50ns.

Power-Fail Comparator

The PFI input is compared to an internal 1.25V reference. If PFI is less than 1.25V, PFO goes low. The power-fail comparator is intended for use as an undervoltage detector to signal a failing power supply; it need not be dedicated to this function though, as it is

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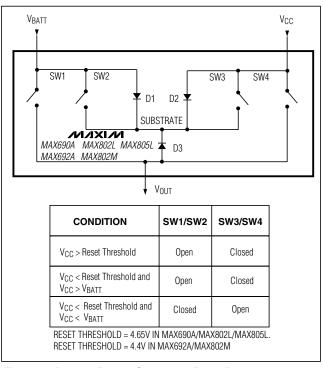


Figure 3. Backup-Battery Switchover Block Diagram

completely separate from the rest of the circuitry. The external voltage divider drives PFI to sense the unregulated DC input to the +5V regulator (see *Typical Operating Circuit*). The voltage-divider ratio can be chosen such that the voltage at PFI falls below 1.25V just before the +5V regulator drops out. $\overline{\text{PFO}}$ then triggers an interrupt which signals the μP to prepare for power-down.

To conserve backup-battery power, the power-fail detector comparator is turned off and $\overline{\text{PFO}}$ is forced low when V_{BATT} connects to V_{OUT} .

Backup-Battery Switchover

In the event of a brownout or power failure, it may be necessary to preserve the contents of RAM. With a backup battery installed at $\rm V_{BATT}$, the devices automatically switch RAM to backup power when $\rm V_{CC}$ fails.

As long as V_{CC} exceeds the reset threshold, V_{OUT} connects to V_{CC} through a 5Ω PMOS power switch. Once V_{CC} falls below the reset threshold, V_{CC} or V_{BATT} (whichever is higher) switches to $V_{OUT}.$ Unlike the MAX690/MAX692, the MAX690A/MAX692A/MAX802L/MAX802M/MAX805L don't always connect V_{BATT} to V_{OUT} when V_{BATT} is greater than $V_{CC}.$ V_{BATT} connects to V_{OUT} (through an 80Ω switch) only when V_{CC} is below the reset threshold $% V_{CC}$ and V_{CC} is greater than $V_{CC}.$

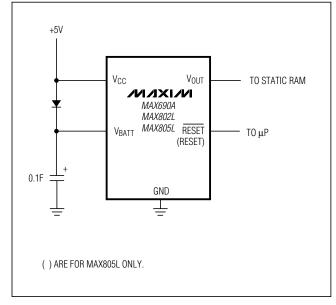


Figure 4. Using a SuperCap as a Backup Power Source with a MAX690A/MAX802L/MAX805L and a +5V ±5% Supply

When V_{CC} exceeds the reset threshold, it is connected to the MAX690A/MAX692A/MAX802L/MAX802M/MAX805L substrate, regardless of the voltage applied to V_{BATT} (Figure 3). During this time, the diode (D1) between V_{BATT} and the substrate will conduct current from V_{BATT} to V_{CC} if V_{BATT} is 0.6V or greater than V_{CC} .

Table 1. Input and Output Status in Battery-Backup Mode

| SIGNAL | STATUS | | | | | |
|-------------------|--|--|--|--|--|--|
| V _{CC} | Disconnected from V _{OUT} | | | | | |
| V _{OUT} | Connected to V_{BATT} through an internal 80Ω PMOS switch | | | | | |
| V _{BATT} | Connected to V_{OUT} . Current drawn from the battery is less than 1 μ A, as long as $V_{CC} < V_{BATT}$ - 1V. | | | | | |
| PFI | Power-fail comparator is disabled. | | | | | |
| PFO | Logic low | | | | | |
| RESET | Logic low | | | | | |
| RESET | Logic high (MAX805L only) | | | | | |
| WDI | Watchdog timer is disabled | | | | | |



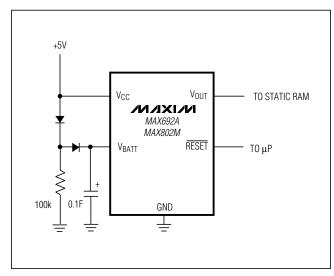


Figure 5. Using a SuperCapTM as a Backup Power Source with the MAX692A/MAX802M and a \pm 5V \pm 10% Supply

When V_{BATT} connects to V_{OUT} , backup mode is activated and the internal circuitry is powered from the battery (Table 1). When V_{CC} is just below V_{BATT} , the current drawn from V_{BATT} is typically 30µA. When V_{CC} drops to more than 1V below V_{BATT} , the internal switchover comparator shuts off and the supply current falls to less than 1µA.

Applications Information

Using a SuperCap[™] as a Backup Power Source

SuperCaps are capacitors with extremely high capacitance values, on the order of 0.1F. Figure 4 shows a SuperCap used as a backup power source. Do not allow the SuperCap's voltage to exceed the maximum reset threshold by more than 0.6V. In Figure 4's circuit, the SuperCap rapidly charges to within a diode drop of V_{CC} . However, after a long time, the diode leakage current will pull the SuperCap voltage up to V_{CC} . When using a SuperCap with the MAX690A/MAX802L/MAX805L, V_{CC} may not exceed 4.75V + 0.6V = 5.35V.

Use the SuperCap circuit of Figure 5 with a MAX692A or MAX802M and a $\pm 10\%$ supply. This circuit ensures that the SuperCap only charges to V_{CC} -0.5V. At the maximum V_{CC} of 5.5V, the SuperCap charges up to 5.0V, only 0.5V above the maximum reset threshold—well within the requisite 0.6V.

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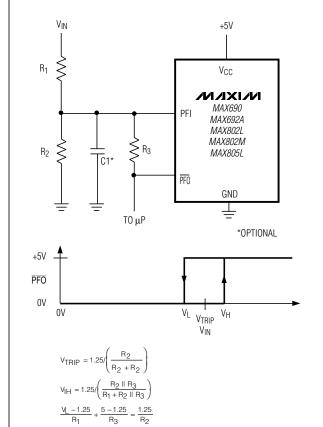


Figure 6. Adding Hysteresis to the Power-Fail Comparator

Allowable Backup Power-Source Batteries

Lithium batteries work very well as backup batteries due to very low self-discharge rates and high energy density. Single lithium batteries with open-circuit voltages of 3.0V to 3.6V are ideal. Any battery with an open-circuit voltage less than the minimum reset threshold plus 0.3V can be connected directly to the $V_{\rm BATT}$ input of the MAX690A/MAX692A/MAX802L/MAX802M/MAX805L with

Table 2. Allowable Backup-Battery Voltages
(see Using a SuperCap as a Backup Power Source section for use with a SuperCap)

| PART NO. | MAXIMUM BACKUP-BATTERY VOLTAGE (V) |
|-----------------------------|------------------------------------|
| MAX690A/ MAX802L/MAX805L | 4.80 |
| MAX692A/ MAX802M | 4.55 |

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no additional circuitry (see the *Typical Operating Circuit*). However, batteries with open-circuit voltages that are greater **cannot** be used for backup, as current is sourced into the substrate through the diode (D1 in Figure 3) when $V_{\rm CC}$ is close to the reset threshold.

Operation Without a Backup Power Source

If a backup power source is not used, ground V_{BATT} and connect V_{OUT} to V_{CC} . Since there is no need to switch over to any backup power source, V_{OUT} does not need to be switched. A direct connection to V_{CC} eliminates any voltage drops across the switch which may push V_{OUT} below V_{CC} .

Replacing the Backup Battery

The backup battery can be removed while V_{CC} remains valid, without danger of triggering RESET/RESET. As long as V_{CC} stays above the reset threshold, battery-backup mode cannot be entered. In other switchover ICs where battery-backup mode is entered whenever V_{BATT} gets close to V_{CC} , an unconnected V_{BATT} pin

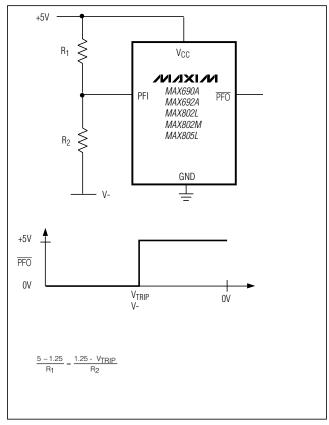


Figure 7. Monitoring a Negative Voltage

accumulates leakage charge and triggers RESET/RESET in error.

Adding Hysteresis to the Power-Fail Comparator

Hysteresis adds a noise margin to the power-fail comparator and prevents repeated triggering of PFO when V_{IN} is close to its trip point. Figure 6 shows how to add hysteresis to the power-fail comparator. Select the ratio of R_1 and R_2 such that PFI sees 1.25V when V_{IN} falls to its trip point (V_{TRIP}). R_3 adds the hysteresis. It will typically be an order of magnitude greater than R_1 or R_2 (about 10 times either R_1 or R_2). The current through R_1 and R_2 should be at least 1µA to ensure that the 25nA (max) PFI input current does not shift the trip point. R_3 should be larger than 10k Ω so it does not load down the \overline{PFO} pin. Capacitor C1 adds additional noise rejection.

Monitoring a Negative Voltage

The power-fail comparator can be used to monitor a negative supply rail using the circuit of Figure 7. When the negative rail is good (a negative voltage of large magnitude), $\overline{\text{PFO}}$ is low. When the negative rail is degraded (a negative voltage of lesser magnitude), $\overline{\text{PFO}}$ goes high. This circuit's accuracy is affected by the PFI threshold tolerance, the V_{CC} line, and the resistors.

Interfacing to µPs with Bidirectional Reset Pins

μPs with bidirectional reset pins, such as the Motorola 68HC11 series, can contend with the MAX690A/MAX692A/MAX802L/MAX802M RESET output. If, for example, the RESET output is driven high and the μP wants to pull it low, indeterminate logic levels may result. To correct this, connect a 4.7kΩ resistor between the RESET output and the μP reset I/O, as in Figure 8. Buffer the RESET output to other system components



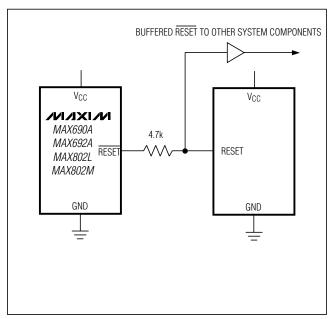


Figure 8. Interfacing to μPs with Bidirectional Reset I/O

$_{f \mu}$ P Supervisory Circuits

| Part Number | Nominal Reset Threshold (V) | Minimum Reset Pulse Width (ms) | Nominal Watchdog Timeout Period (sec) | Backup- Battery Switch | CE - Write Protect | Power- Fail Com- parator | Manual- Reset Input | Watch- dog Output | Low- Line Output | Active- High Reset | Battery- On Output |
|---------------------|--------------------------------------|--|---|------------------------------|--------------------------|-----------------------------------|---------------------------|-------------------------|------------------------|--------------------------|--------------------------|
| MAX690A/692A | 4.65/4.40 | 140 | 1.6 | ✓ | | ' | | | | | |
| MAX691A/693A | 4.65/4.40 | 140/adj. | 1.6/adj. | ✓ | ✓ /10ns | ' | | ' | ' | ~ | ' |
| MAX696 | Adj. | 35/adj. | 1.6/adj. | ' | | ' | | ~ | ' | ' | ' |
| MAX697 | Adj. | 35/adj. | 1.6/adj. | | ' | ' | | / | / | v | |
| MAX700 | 4.65/adj. | 200 | - | | | | ~ | | | ✓ | |
| MAX703/704 | 4.65/4.40 | 140 | - | ✓ | | ✓ | ✓ | | | | |
| MAX705/706 | 4.65/4.40 | 140 | 1.6 | | | ✓ | ✓ | ✓ | | | |
| MAX706P | 2.63 | 140 | 1.6 | | | ~ | / | ~ | | ~ | |
| MAX706R/S/T | 2.63/2.93/ 3.08 | 140 | 1.6 | | | • | • | • | | | |
| MAX707/708 | 4.65/4.40 | 140 | - | | | ✓ | ✓ | | | ✓ | |
| MAX708R/S/T | 2.63/2.93/ 3.08 | 140 | - | | | • | V | | | ~ | |
| MAX709L/M/ R/S/T | 4.65/4.40/ 2.63/2.93/3.08 | 140 | - | | | | | | | | |
| MAX791 | 4.65 | 140 | 1 | • | ✓ /10ns | ✓ | ✓ | • | ✓ | ~ | ✓ |
| MAX792L/M/ R/S/T | 4.65/4.40/ 2.63/2.93/3.08 | 140 | 1 | | ✓ /10ns | • | ~ | • | • | ~ | |
| MAX800L/M | 4.60/4.40 | 140 | 1.6/adj. | ~ | ✓ /10ns | ✓ /±2% | | ~ | ✓ | ~ | / |
| MAX802L/M | 4.60/4.40 | 140 | 1.6 | ~ | | ✓ /±2% | | | | | |
| MAX805L | 4.65 | 140 | 1.6 | ~ | | ~ | | | | ~ | |
| MAX813L | 4.65 | 140 | 1.6 | | | / | / | ~ | | ~ | |
| MAX820L/M/ R/S/T | 4.65/4.40/ 2.63/2.93/3.08 | 140 | 1 | | ✓ /10ns | ✓ /±2% | ~ | ~ | • | ~ | |
| MAX1232 | 4.37/4.62 | 250 | 0.15/0.60/1.2 | ! | | | ~ | | | ~ | |
| MAX1259 | - | - | - | ~ | | V | | | | | |

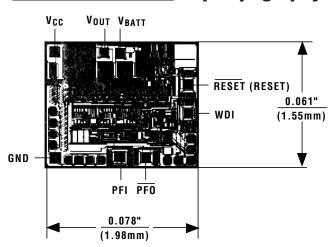
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Ordering Information (continued)

| | | (|
|------------|-----------------|---------------|
| PART | TEMP. RANGE | PIN-PACKAGE |
| MAX692ACPA | 0°C to +70°C | 8 Plastic DIP |
| MAX692ACSA | 0°C to +70°C | 8 SO |
| MAX692AC/D | 0°C to +70°C | Dice* |
| MAX692AEPA | -40°C to +85°C | 8 Plastic DIP |
| MAX692AESA | -40°C to +85°C | 8 SO |
| MAX692AMJA | -55°C to +125°C | 8 CERDIP** |
| MAX802LCPA | 0°C to +70°C | 8 Plastic DIP |
| MAX802LCSA | 0°C to +70°C | 8 SO |
| MAX802LEPA | -40°C to +85°C | 8 Plastic DIP |
| MAX802LESA | -40°C to +85°C | 8 SO |
| MAX802MCPA | 0°C to +70°C | 8 Plastic DIP |
| MAX802MCSA | 0°C to +70°C | 8 SO |
| MAX802MEPA | -40°C to +85°C | 8 Plastic DIP |
| MAX802MESA | -40°C to +85°C | 8 SO |
| MAX805LCPA | 0°C to +70°C | 8 Plastic DIP |
| MAX805LCSA | 0°C to +70°C | 8 SO |
| MAX805LC/D | 0°C to +70°C | Dice* |
| MAX805LEPA | -40°C to +85°C | 8 Plastic DIP |
| MAX805LESA | -40°C to +85°C | 8 SO |
| MAX805LMJA | -55°C to +125°C | 8 CERDIP** |
| | | |

^{*} Dice are specified at $T_A = +25$ °C.

Chip Topography



() ARE FOR MAX805L ONLY. TRANSISTOR COUNT: 573; SUBSTRATE MUST BE LEFT UNCONNECTED.

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^{**}Contact factory for availability and processing to MIL-STD-883. Devices in PDIP and SO packages are available in both leaded and lead-free packaging. Specify lead free by adding the + symbol at the end of the part number when ordering. Lead free not available for CERDIP package.