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1 Block diagram and pin description

Figure 1. Block diagram

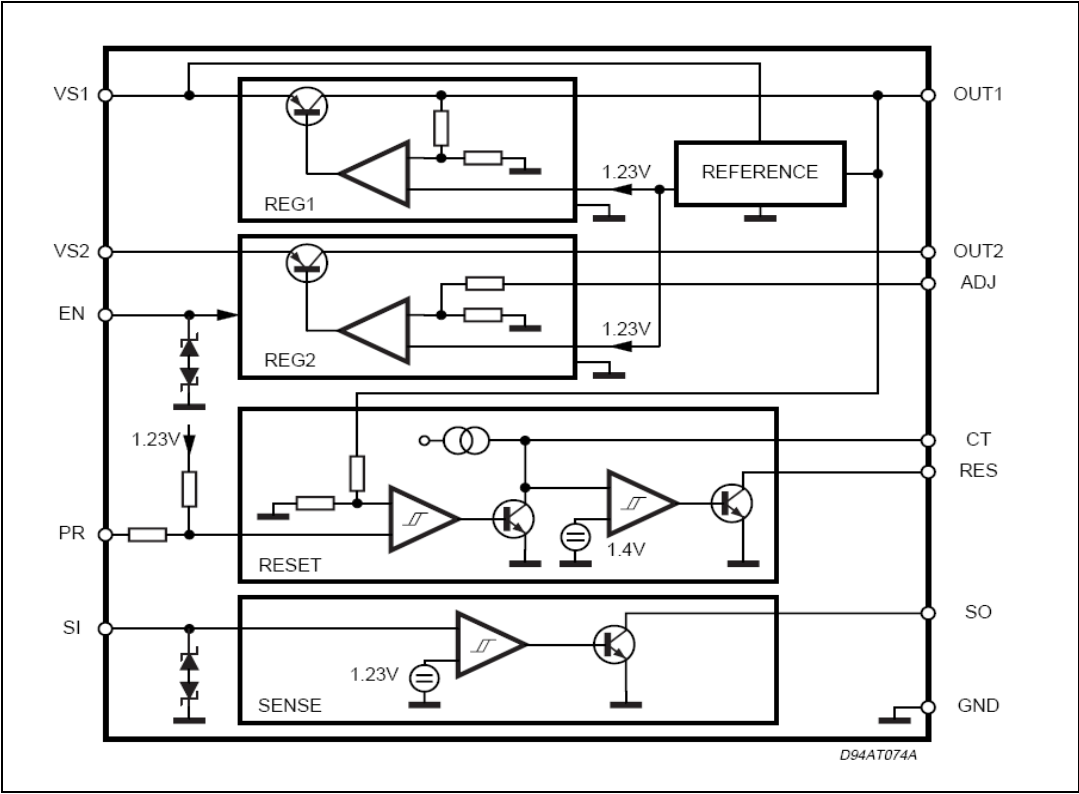


Figure 2. Configuration diagram (top view)

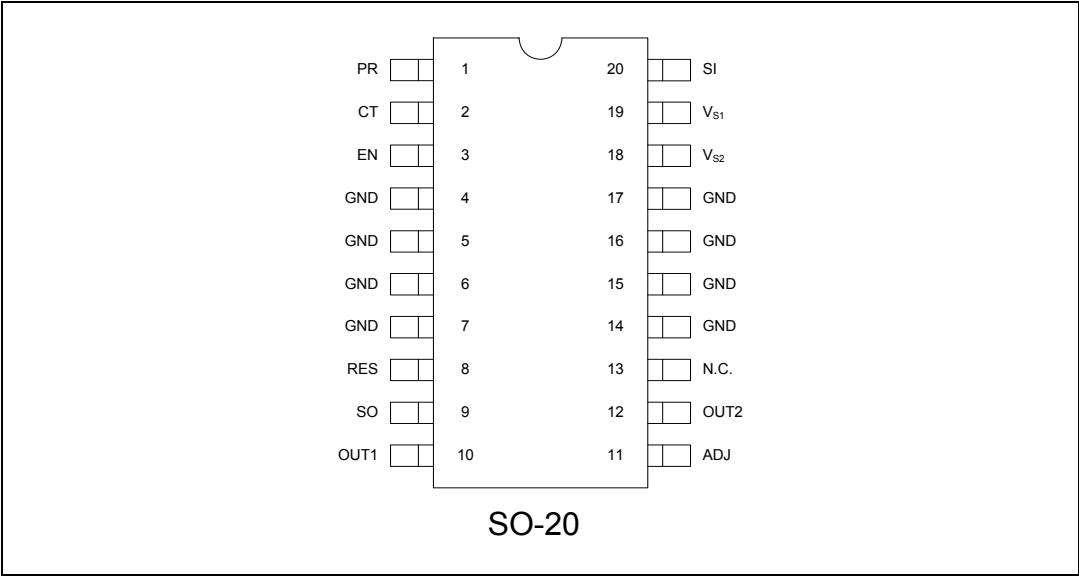


Table 2. Pin definitions and functions

Pin number	Name	Function
SO-20		
18	V _{S2}	Supply voltage (400 mA regulator)
19	V _{S1}	Supply voltage (100 mA regulator, reset, sense)
20	SI	Sense input
1	PR	Reset threshold programming
2	CT	Reset delay capacitor
3	EN	Enable (low activates the 400 mA regulator)
4, 5, 6, 7, 14, 15, 16, 17	GND	Ground
8	RES	Reset output
9	SO	Sense output
10	OUT1	100 mA regulator output
11	ADJ	Feedback of 400 mA regulator
12	OUT2	400 mA regulator output
13	NC	Not connected

2 Electrical specifications

2.1 Absolute maximum ratings

Stressing the device above the ratings listed in [Table 3: Absolute maximum ratings](#) may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to Absolute maximum rating conditions for extended periods may affect device reliability.

Table 3. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{INDC}	DC operating supply voltage	28	V
V_{INTR}	Transient operating supply voltage ($T < 400$ ms)	-14 to 40	V
I_O	Output current	internally limited	
V_{SI}	Sense input voltage (voltage forced) ⁽¹⁾	-20 to 20	V
I_{SI}	Sense input current (current forced) ⁽¹⁾	± 1	mA
V_{EN}	Enable input voltage (voltage forced) ⁽¹⁾	-20 to 20	V
I_{EN}	Sense input current (current forced) ⁽¹⁾	± 1	mA
V_{RES}, V_{SO}	Output voltages	-0.3 to 20	V
I_{RES}, I_{SO}	Output currents (output low)	5	mA
P_O	Power dissipation at $T_{amb} = 80$ °C ⁽²⁾	875	mW
T_{stg}	Storage temperature	-65 to 150	°C
T_J	Operating junction temperature	-40 to 150	°C
T_{JSD}	Thermal shutdown junction temperature output 2 will shutdown typically at T_J 10 K lower than output 1	165	°C

1. Current forced means voltage unlimited but current limited to the specified value voltage forced means voltage limited to the specified values while the current is not limited
2. Typical value soldered on a PC board with 8 cm² copper ground plane (35 mm thick).

Note: The circuit is ESD protected according to MIL-STD-883C.

2.2 Thermal data

Table 4. Thermal data

Symbol	Parameter	SO-20	Unit
$R_{thj-amb}$	Thermal resistance junction to ambient	50	°C/W
$R_{thj-case}$	Thermal resistance junction to case	-	°C/W

Note: Typical value soldered on a PC board with 8 cm² copper ground plane (35 mm thick).

2.3 Electrical characteristics

$V_S = 14$ V; $T_j = -40$ to 150 °C, unless otherwise specified.

Table 5. OUT1

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_{O1}	Supply output voltage	$V_S = 6$ to 28 V; $I_{O1} = 400$ μ A to 100 mA	4.9	5	5.1	V
		$T_J \leq 125$ °C; $I_{O1} = 50$ to 400 μ A	4.8	5	5.2	V
V_{DP1}	Drop output voltage 1	$I_{OUT1} = 10$ mA		0.1	0.2	V
		$I_{OUT1} = 100$ mA; $V_S = 4.8$ V		0.2	0.4	V
V_{OL01}	Load regulation 1	$I_{OUT1} = 1$ to 100 mA (after regulation setting)			25	mV
V_{LIM1}	Current limit 1	$V_{OUT1} = 0.8$ to 4.5 V	100	200	400	mA
I_{QSB}	Quiescent current in standby mode	$I_{EN} \geq 2.4$ V (output 2 disabled) $I_{O1} = 0.1$ mA; $V_{SI} > 1.3$ V		65	90	μ A
		$T_J < 85$ °C; $R_{PR} = 0$		75		μ A

Table 6. OUT2

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_{O2}	Output voltage 2 ADJ connected to OUT2	Enable = low; $V_S = 6$ to 28 V; $I_{O2} = 5$ to 400 mA	4.9		5.1	V
V_{DP2}	Drop output voltage 2	$I_{OUT2} = 100$ mA		0.2	0.3	V
		$I_{OUT2} = 400$ mA; $V_S = 4.8$ V		0.3	0.6	V
V_{OL02}	Load regulation 2	$I_{OUT1} = 5$ to 400 mA (after regulation setting)			50	mV
R_{ADJ}	Adjust input resistance		60	100	150	mA
I_{LIM2}	Current limit 2	$V_{O2} = 0.8$ to 4.5 V	450	650	1300	mA
I_Q	Quiescent current	$I_{OUT1} = 100$ mA; $I_{OUT2} = 400$ mA			20	mA

Table 7. OUT1, OUT2

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
$V_{OLi\ 1,2}$	Line regulation	$V_S = 6$ to 28 V; $I_{O1} = 1$ mA, $I_{O2} = 5$ mA, (after regulation setting)			20	mV

Table 8. Enable input

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_{ENL}	Enable input low voltage (output 2 active)		-20		1	V
V_{ENH}	Enable input high voltage		1.4		20	V
V_{ENhyst}	Enable hysteresis		20	30	60	mV
$I_{EN\ LOW}$	Enable input current low	$V_{EN} = 0$	-20	-8	-3	μ A
$I_{EN\ HIGH}$	Enable input current high	$V_{EN} = 1.1$ to 7 V; $T_J < 130$ °C;	-1	0	1	μ A
		$V_{EN} = 1.1$ to 7 V; $T_J = 130$ to 150 °C;	-10	0	10	μ A

Table 9. Reset circuit

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_{RT}	Reset threshold voltage ⁽¹⁾	$R_{PR} = \infty$	4.5	$V_{O1-0.3}$	$V_{O1-0.2}$	V
		$R_{PR} = 0$	3.65	3.8	3.95	V
V_{RTH}	Reset threshold hysteresis	$R_{PR} = \infty$	30	60	120	mV
$t_{RD\ min}$	Reset pulse delay	$C_{RES} = 47$ nF; $t_r \leq 30$ μ s ⁽²⁾	40	60	100	ms
$t_{RD\ nom}$	Reset pulse delay	$C_{RES} = 47$ nF ⁽³⁾	60	100	140	ms
t_{RR}	Reset reaction time	$C_{RES} = 47$ nF	10	50	150	μ s
I_{CT}	Pull down capability of the discharge circuit	$V_{OUT1} < V_{RT}$	3	6	15	mA
I_{CT}	Charge current	$V_{OUT1} > V_{RT}$	-1.3	-1	0.7	μ A
V_{RESL}	Reset output low voltage	$R_{RES} = 10$ K Ω to V_{OUT1} $V_{OUT1} \geq 1.5$ V			0.4	V
V_{RESH}	Reset output high leakage current	$V_{RES} = 5$ V			1	μ A

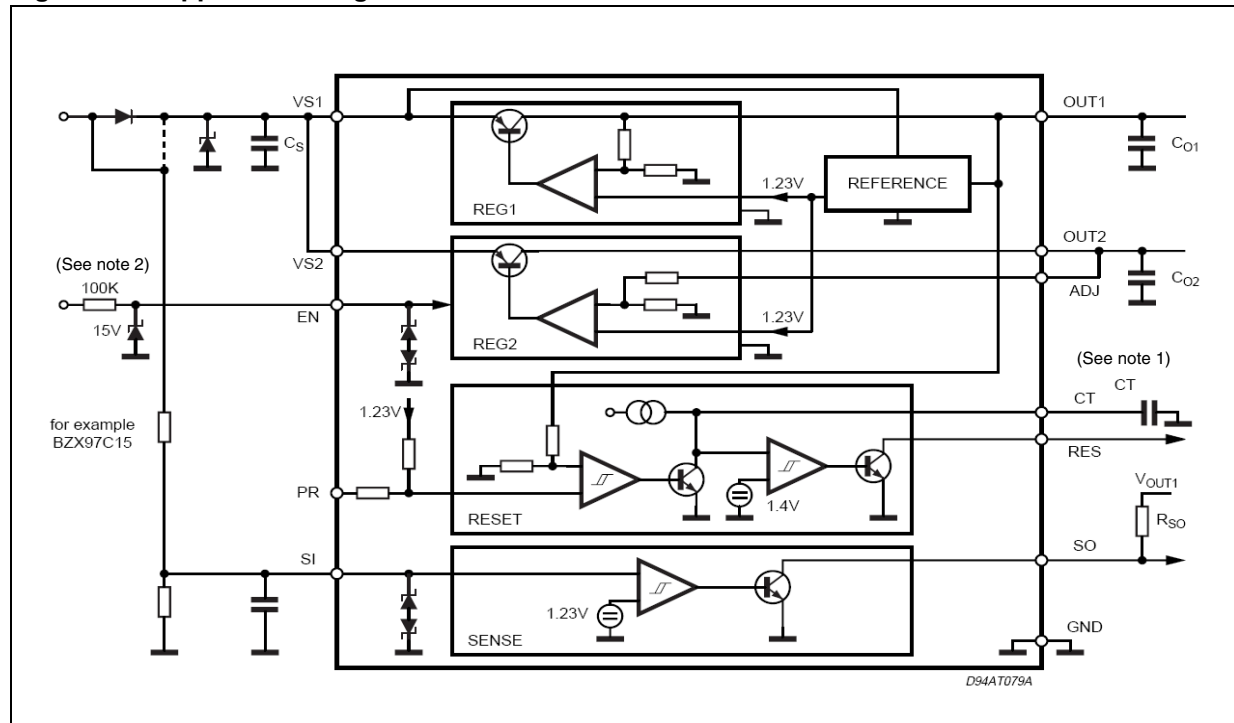
1. The reset threshold can be programmed continuously from typ 3.8 V to 4.7 V by changing a value of an external resistor from pin PR to GND.
2. This is a minimum reset time according to the hysteresis of the comparator. Delay time starts with V_{OUT1} exceeding V_{RT} .
3. This is the nominal reset time depending on the discharging limit of C_T (saturation voltage) and the upper threshold of the timer comparator. Delay time starts with V_{OUT1} exceeding V_{RT} .

Table 10. Sense comparator

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_{SI}	Functional range		-20		20	V
V_{SIT}	Sense threshold voltage	Falling edge; $T_J < 130\text{ }^{\circ}\text{C}$	1.08	1.16	1.24	V
		Falling edge; $T_J < 130\text{ to }150\text{ }^{\circ}\text{C}$	1.05	1.16	1.29	V
V_{SITH}	Sense threshold hysteresis		10	30	60	mV
V_{SOL}	Sense output low voltage	$V_{SI} \leq 1.05\text{ V}$; $R_{SO} = 10\text{ K}\Omega$ connected to 5 V; $V_S \geq 5\text{ V}$			0.4	V
I_{SOH}	Sense output leakage	$V_{SO} = 5\text{ V}$; $V_{SI} \geq 1.5\text{ V}$			1	μA
$I_{SI\text{ HIGH}}$	Sense input current high	$V_{SI} = 1.1\text{ to }7\text{ V}$; $T_J < 130\text{ }^{\circ}\text{C}$	-1	0	1	μA
		$V_{SI} = 1.1\text{ to }7\text{ V}$; $T_J < 130\text{ to }150\text{ }^{\circ}\text{C}$	-10	0	10	μA
$I_{SI\text{ LOW}}$	Sense input current low	$V_{SI} = 0\text{ V}$	-20	-8	-3	μA

3 Application information

Figure 3. Application diagram



1. The leakage of C_T must be less than 0.5 mA (2 V). If an external resistor between C_T and V_{OUT1} is applied, the leakage current may be increased. The external resistor should have more than 30 K Ω . For stability: $C_s \geq 1 \mu F$, $C_{01} \geq 10 \mu F$, $C_{02} \geq 10 \mu F$, $ESR \leq 5 \Omega$ (designed target).
2. For transients exceeding 20 V or -20 V external protection is required at the pins SI and EN as shown at pin EN. The protection proposed provides proper function for transients in the range of ± 200 V. If the zener diode is omitted the external resistor should be raised to 200 K Ω to limit the current to 1 mA. Without the zener diode, the function 20 V or -20 V can not be guaranteed.

3.1 Functional description

The L4938ED-E is a monolithic integrated dual voltage regulators, based on the STM modulator voltage regulator approach. Several outstanding features and auxiliary functions are implemented to meet the requirements of supplying microprocessor systems in automotive applications. Nevertheless, it is suitable also in other applications where two stabilized voltages are required. The modular approach of this device allows to get easy also other features and functions when required.

3.2 Standby regulator

The standby regulator uses an isolated collector vertical PNP transistor as a regulating element. With this structure very low dropout voltage at currents up to 100 mA is obtained. The dropout operation of the standby regulator is maintained down to 3 V input supply voltage. The output voltage is regulated up to the transient input supply voltage of 40 V. With this feature no functional interruption due to overvoltage pulses is generated.

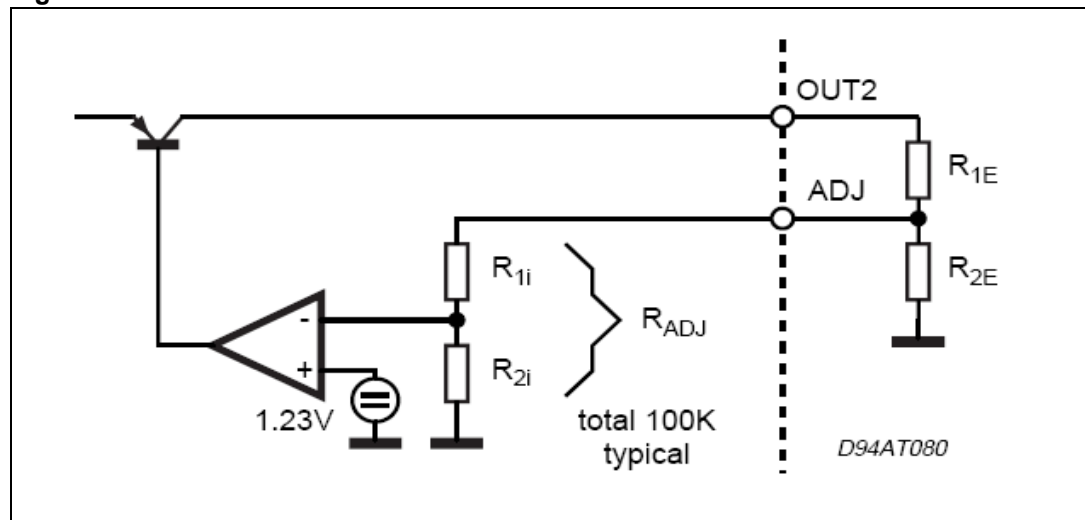
In the standby mode when the output 2 is disabled, the current consumption of the device (quiescent current) is less than 90 μA (14 V supply voltage).

To reduce the quiescent current peak in the undervoltage region and to improve the transient response in this region, the dropout voltage is controlled. A second regulation path keeps the output voltage without load below 5.5 V even at high temperatures.

3.3 Output 2 voltage

The output 2 regulator uses the same output structure as the standby regulator but rated for the output current of 400 mA. The output voltage is internally fixed to 5 V if ADJ is connected to $V_{\text{OUT}2}$. The output 2 regulator can be switches OFF via the enable input.

Figure 4. OUT2



Connecting a resistor divider R_{1E} , R_{2E} to the ADJ, OUT2 pin the output voltage 2 can be programmed to the value of

Equation 1

$$V_{\text{OUT}2} = V_{\text{OUT}1} \left(1 + \frac{R_{1E}(R_{2E} + R_{\text{ADJ}})}{R_{2E} \cdot R_{\text{ADJ}}} \right)$$

with $R_{\text{ADJ}} = 60 \text{ K}$ to 150 K and $V_{\text{OUT}1} = 4.95$ to 5.05 V . For an exact calculation the temperature coefficient ($T_C - 2000 \text{ ppm}$) of the internal resistor (R_{ADJ}) must be taken into account. Pin ADJ in this mode should not have a capacitive burden because this would reduce the phase margin of the regulator loop.

3.4 Reset circuit

The reset circuit supervises the standby output voltage. The reset output (RES) is defined from $V_{\text{OUT}} \geq 1 \text{ V}$. Even if V_S is lacking, the reset generator is supplied by the output voltage $V_{\text{OUT}1}$.

The reset threshold of 4.7 V is defined with the internal reference voltage⁽¹⁾ and standby output divider, when pin PR is left open. The reset threshold voltage can be programmed in the range from 3.8 V to 4.7 V by connecting an external resistor from pin PR to GND.

1. The reference is alternatively supplied from V_S or V_{OUT1} . If one supply is present, the reference is operating.

The value of the programming resistor R_{PR} can be calculated with:

Equation 2

$$R_{PR} = \frac{22K}{\frac{4.7K}{V_{RT}} - 1} - 92.9K$$

Equation 3

$$3.8V \leq V_{RT} \leq 4.7V$$

The reset pulse delay time t_{RD} , is defined with the charge time of an external capacitor C_T :

Equation 4

$$t_{RDmin} = \frac{C_T \cdot 0.6V}{1\mu A}$$

Equation 5

$$t_{RDnom} = \frac{C_T \cdot 1.4V}{1\mu A}$$

The reaction time of the reset circuit originates from the noise immunity. Standby output voltage drops below the reset threshold only a bit longer than the reaction time results in a shorter reset delay time. The nominal reset delay time is generated for standby output voltage drops longer than approximately 50 μs . The minimum reset time is generated if reset condition only occurs for a short time triggering a reset pulse but not completely discharging C_T . The reset can be related to output2 on request. If higher charge currents for the reset capacitor are required a resistors from pin C_T to OUT1, may be used to increase the current. We recommended the use of 10 K Ω to 5 V as an output pull up.

3.5 Sense comparator

The sense comparator compares an input signal with an internal voltage reference of typical 1.23 V. The use of an external voltage divider makes this comparator very flexible in the application. It can be used to supervise the input voltage either before or after the protection diode and to give additional information to the microprocessor like low voltage warnings. We recommended the use of 10 K Ω to 5 V as an output pull up.

3.6 Thermal protection

Both outputs are provided with an overtemperature shutdown regulation power dissipation down to uncritical values. Output 2 shuts down approximately 10 K before output 1. Under normal conditions shutdown of output 2 allows the chip to cool down again. Thus output 1 is unaffected. The thermal shutdown reduces the output voltages until power dissipation and the flow of thermal energy out of the chip balance.

3.7 Transient sensitivity

In proper operation ($V_{OUT} > 4.5\text{ V}$) the reference is supplied by V_{OUT1} thus reducing sensitivity to input transients.

Figure 5. Reset generator

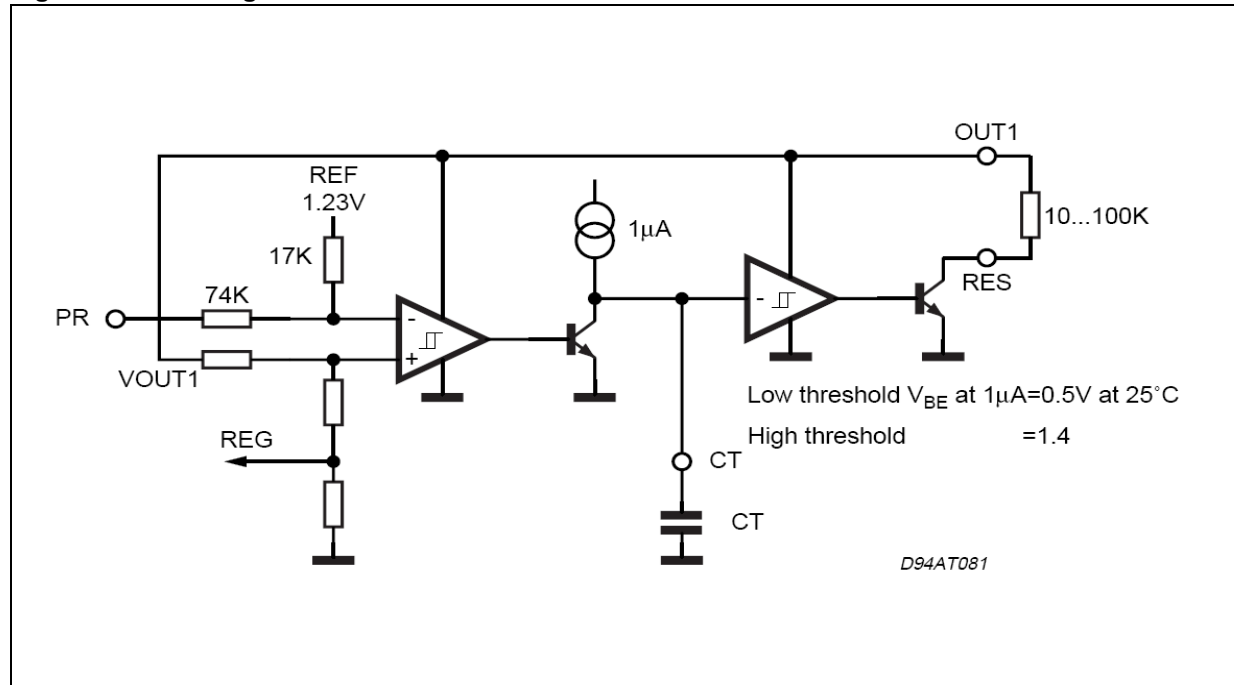
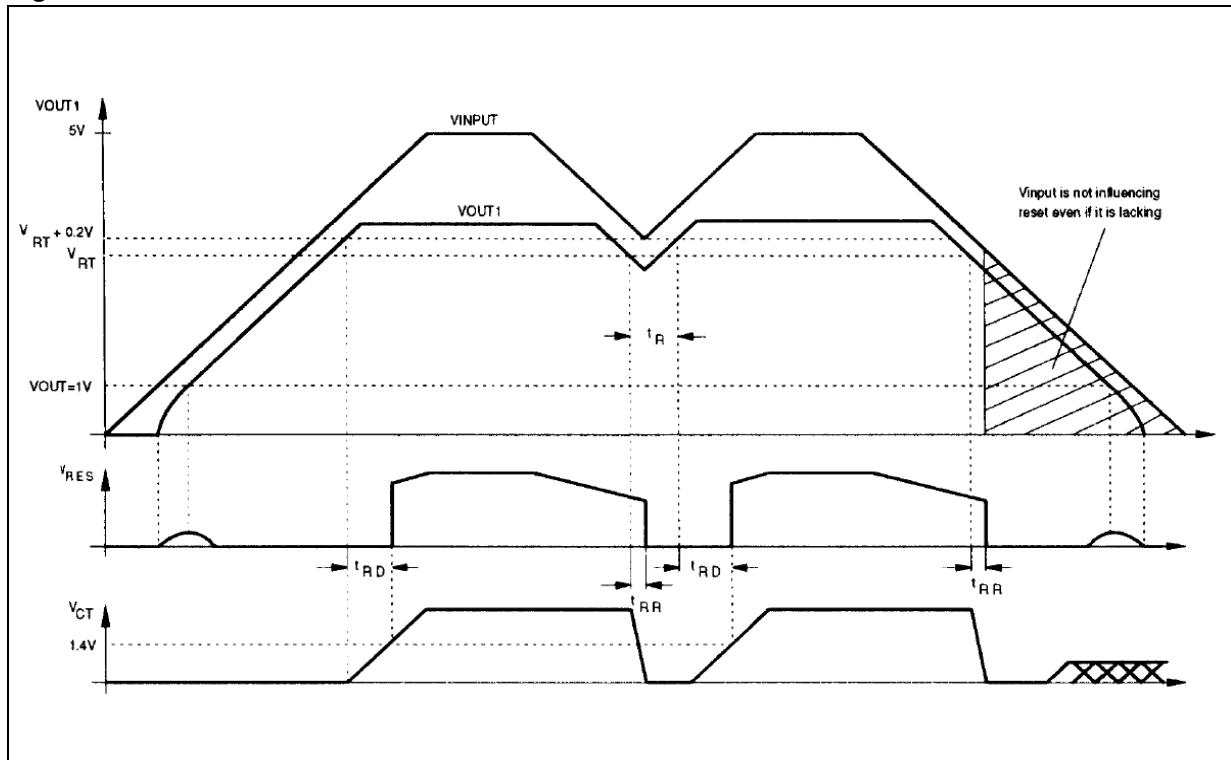


Figure 6. Waveforms



3.8 Input protection

The Inputs Enable (EN) and Sense In (SI) are protected against negative transients. [Figure 7](#) is showing the simplified schematic

Figure 7. Input protection

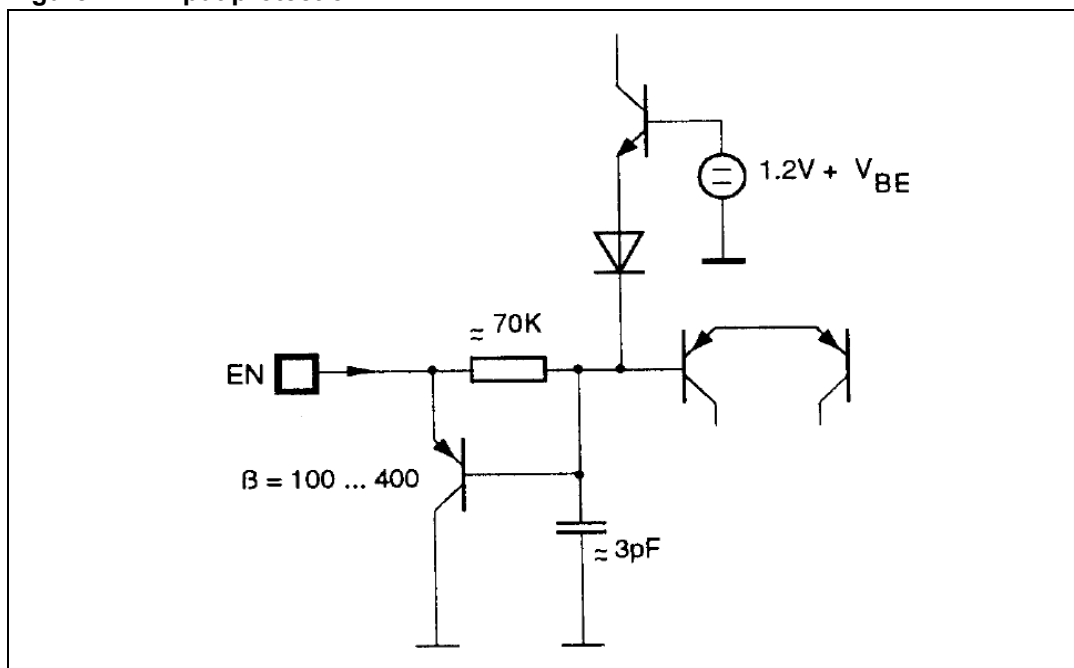
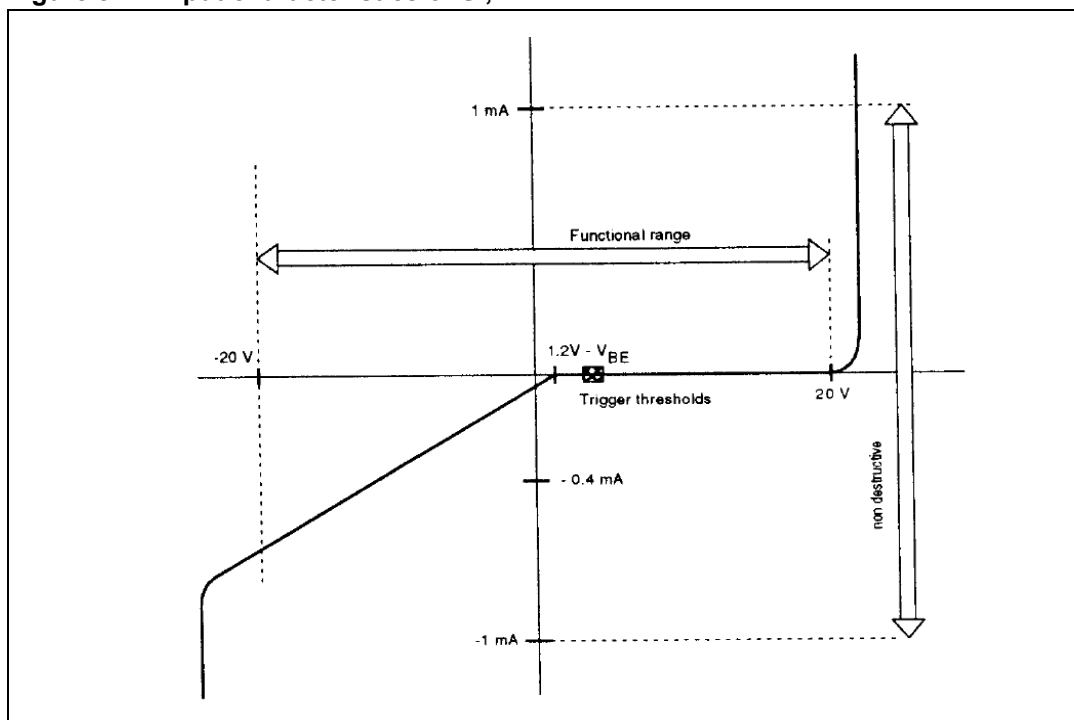


Figure 8. Input characteristics of SI, EN



4 Package and packing information

4.1 ECOPACK® packages

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com.

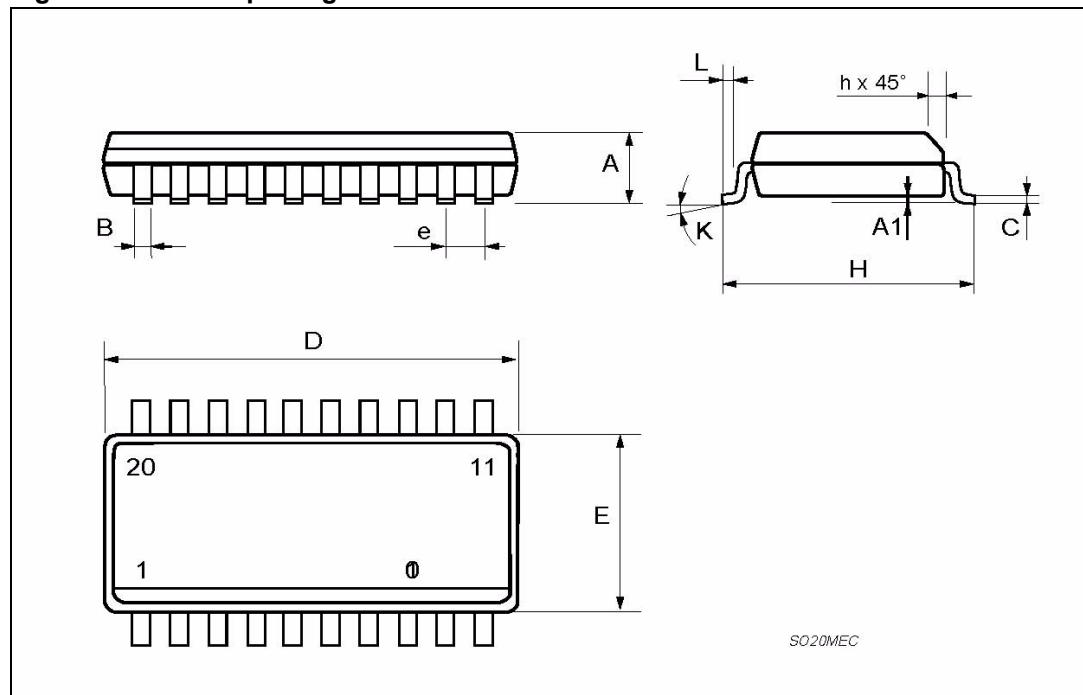
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4.2 SO-20 package information

Table 11. SO-20 mechanical data

Dim.	mm		
	Min.	Typ.	Max.
A	2.35		2.65
A1	0.1		0.3
B	0.33		0.51
C	0.23		0.32
D	12.6		13
E	7.4		7.6
e		1.27	
H	10		10.65
h	0.25		0.75
L	0.4		1.27
K	0°		8°

Figure 9. SO-20 package dimensions



5 Revision history

Table 12. Document revision history

Date	Revision	Changes
14-Dec-2011	1	Initial release.



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