Low-Voltage, High-Accuracy, Quad Voltage Monitors in µMAX Package

ABSOLUTE MAXIMUM RATINGS

All Pins to GND...-0.3V to +6V
Input/Output Current (all pins) 20mA
Continuous Power Dissipation (T$_A$ = +70°C)
   10-Pin µMAX (derate 5.6mW/°C above +70°C) 444mW

Operating Temperature Range ....-40°C to +85°C
Storage Temperature Range ....-65°C to +150°C
Junction Temperature..+150°C
Lead Temperature (soldering, 10s) ....+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS (MAX6709)

(V$_{CC}$ = 2.0V to 5.5V, T$_A$ = -40°C to +85°C, unless otherwise noted. Typical values are at V$_{CC}$ = 5V and T$_A$ = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Supply Voltage Range	V$_{CC}$			2.0		5.5	V
Supply Current	I$_{CC}$	V$_{CC}$ = 3V			25	50	µA
		V$_{CC}$ = 5V			35	65	
Input Current	I$_{IN\_}$	V$_{IN\_}$ = input threshold voltage			25	40	µA
		V$_{IN\_}$ = 0 to 0.85V (for adjustable threshold)				0.2	
Threshold Voltage	V$_{TH}$	IN_ decreasing	5.0V (-5%)	4.50	4.63	4.75	V
			5.0V (-10%)	4.25	4.38	4.50	
			3.3V (-5%)	3.00	3.08	3.15	
			3.3V (-10%)	2.85	2.93	3.00	
			3.0V (-5%)	2.70	2.78	2.85	
			3.0V (-10%)	2.55	2.63	2.70	
			2.5V (-5%)	2.25	2.32	2.38	
			2.5V (-10%)	2.13	2.19	2.25	
			1.8V (-5%)	1.62	1.67	1.71	
			1.8V (-10%)	1.53	1.58	1.62	
Adjustable Threshold	V$_{TH}$	IN_ decreasing		0.609	0.623	0.635	V
Threshold Voltage Temperature Coefficient	TCV$_{TH}$				60		ppm/°C
Threshold Hysteresis	V$_{HYST}$				0.3 x V$_{TH}$		%
Propagation Delay	t$_{PD}$	V$_{IN\_}$ falling at 10mV/µs from V$_{TH}$ to (V$_{TH}$ - 50mV)			30		µs
		V$_{IN\_}$ rising at 10mV/µs from V$_{TH}$ to (V$_{TH}$ + 50mV)			5		
Output Low Voltage	V$_{OL}$	V$_{CC}$ = 5V, I$_{SINK}$ = 2mA				0.3	V
		V$_{CC}$ = 2.5V, I$_{SINK}$ = 1.2mA				0.3	
		V$_{CC}$ = 1V, I$_{SINK}$ = 50µA (Note 2)				0.3	
Output High Voltage	V$_{OH}$	V$_{CC}$ ≥ 2.0V, I$_{SOURCE}$ = 6µA (min), PWRGD_ unasserted		0.8 x V$_{CC}$			V
Output High Source Current	I$_{OH}$	V$_{CC}$ ≥ 2.0V, PWRGD_ unasserted			10		µA

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Low-Voltage, High-Accuracy Quad, Voltage Monitors in µMAX Package

ELECTRICAL CHARACTERISTICS (MAX6714)

(V_{CC} = 2.0V to 5.5V, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at V_{CC} = 5V and T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Supply Voltage Range	V_{CC}			2.0		5.5	V
Supply Current (Note 3)	I_{CC}	V_{CC} = 3V			60	90	µA
		V_{CC} = 5V			80	105	
Power-Fail Input Current	$I_{PFI\_}$	$V_{PFI\_}$ = 0 to 0.85V				0.2	µA
V_{CC} Reset Threshold	V_{TH}	V_{CC} decreasing	MAX6714B (-5%)	4.50	4.63	4.75	V
			MAX6714A (-10%)	4.25	4.38	4.50	
			MAX6714D (-5%)	3.00	3.08	3.15	
			MAX6714C (-10%)	2.85	2.93	3.00	
Power-Fail Input Threshold	V_{PFI}	$V_{PFI\_}$ decreasing		0.609	0.623	0.635	V
Threshold Hysteresis	V_{HYST}	$V_{PFI\_}$ increasing relative to $V_{PFI\_}$ decreasing			0.3 x V_{TH}		%
Reset Timeout Period	t_{RP}			140	210	280	ms
Reset Delay	t_{RD}	V_{CC} falling at 10mV/µs from (V_{TH} + 100mV) to (V_{TH} - 100mV)			30		µs
Power-Fail Propagation Delay	t_{PFD}	$V_{PFI\_}$ falling at 10mV/µs from V_{TH} to (V_{TH} - 50mV)			30		µs
		V_{CC} falling at 10mV/µs from (V_{TH} + 100mV) to (V_{TH} - 100mV)			5		
$\overline{MR}$ Input Voltage	V_{IL}					0.3 x V_{CC}	V
	V_{IH}			0.7 x V_{CC}			
$\overline{MR}$ Minimum Input Pulse				1			µs
$\overline{MR}$ Glitch Rejection					100		ns
$\overline{MR}$ to $\overline{RESET}$ Delay	t_{MRD}				200		ns
$\overline{MR}$ Pullup Resistance		$\overline{MR}$ to V_{CC}		10	20	50	kΩ
Output Low Voltage	V_{OL}	V_{CC} = 5V, I_{SINK} = 2mA				0.3	V
		V_{CC} = 2.5V, I_{SINK} = 1.2mA				0.3	
		V_{CC} = 1V, I_{SINK} = 50µA (Note 2)				0.3	
Output High Voltage	V_{OH}	V_{CC} ≥ 2.0V, I_{SOURCE} = 6mA (min), $\overline{RESET}$, $\overline{PFO\_}$ unasserted		0.8 x V_{CC}			V
Output High Source Current	I_{OH}	V_{CC} ≥ 2.0V, $\overline{RESET}$ and $\overline{PFO\_}$ unasserted			10		µA

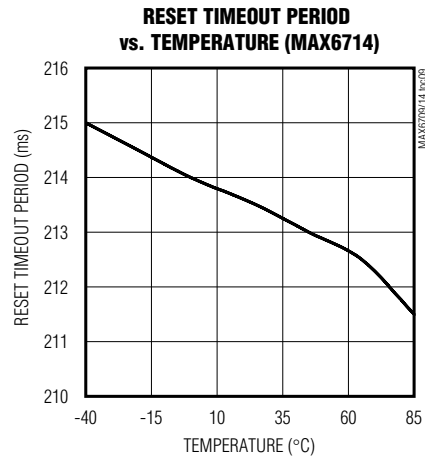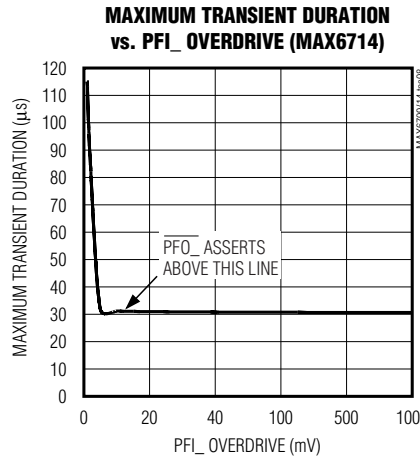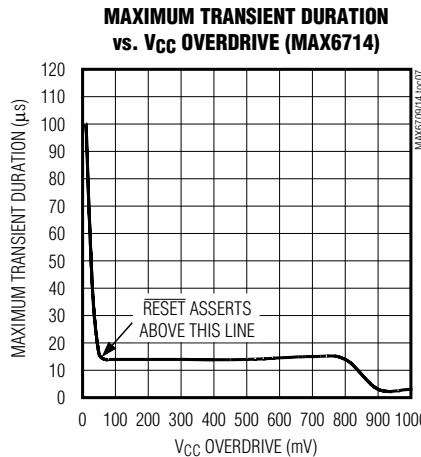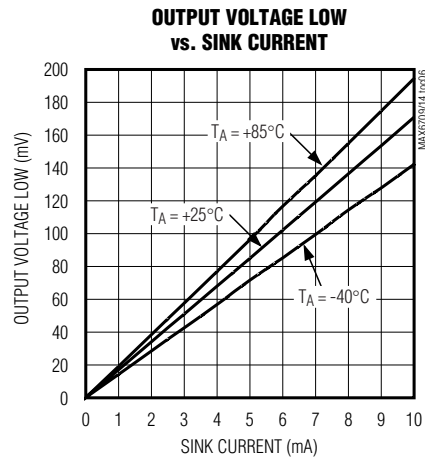**Note 1:** 100% production tested at T_A = +25°C. Overtemperature limits guaranteed by design.
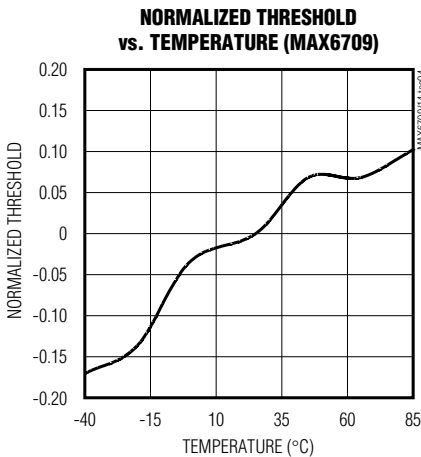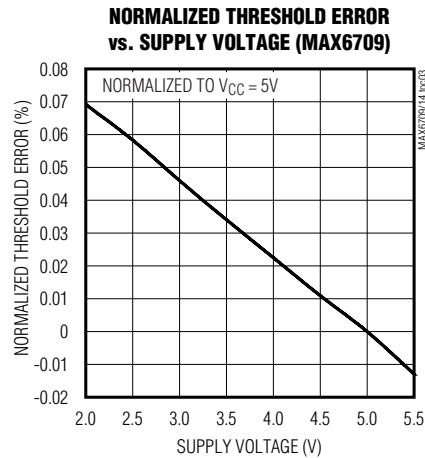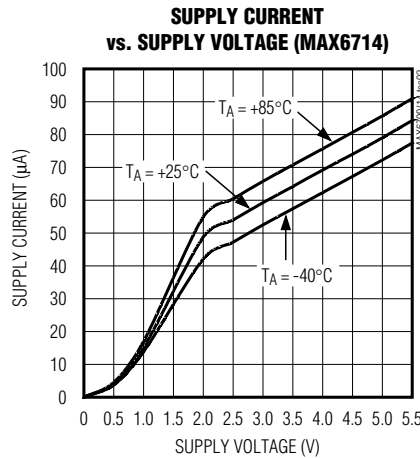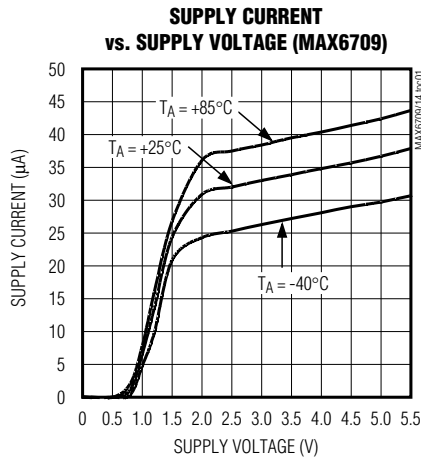**Note 2:** Condition at V_{CC} = 1V is guaranteed only from T_A = 0°C to +70°C.
Note 3: Monitored voltage 5V/3.3V is also the device supply. In the typical condition, supply current splits as follows: 25µA for the resistor-divider, and the rest for other circuitry.

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Low-Voltage, High-Accuracy, Quad Voltage Monitors in µMAX Package

_____*Typical Operating Characteristics*

(V_CC = 5V, T_A = +25°C, unless otherwise noted.)
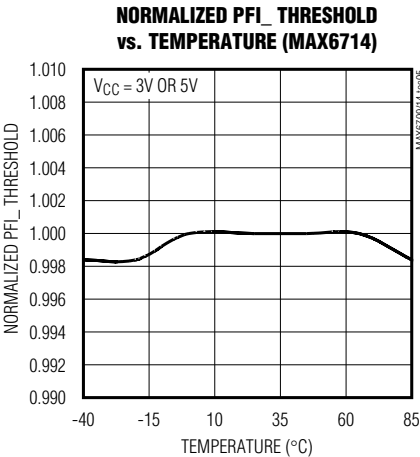
SUPPLY CURRENT vs. SUPPLY VOLTAGE (MAX6709)

SUPPLY CURRENT vs. SUPPLY VOLTAGE (MAX6714)

NORMALIZED THRESHOLD ERROR vs. SUPPLY VOLTAGE (MAX6709)

NORMALIZED THRESHOLD vs. TEMPERATURE (MAX6709)

NORMALIZED PFI_ THRESHOLD vs. TEMPERATURE (MAX6714)

OUTPUT VOLTAGE LOW vs. SINK CURRENT

MAXIMUM TRANSIENT DURATION vs. V_CC OVERDRIVE (MAX6714)

MAXIMUM TRANSIENT DURATION vs. PFI_ OVERDRIVE (MAX6714)

RESET TIMEOUT PERIOD vs. TEMPERATURE (MAX6714)

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Typical Operating Characteristics (continued)

(V$_{CC}$ = 5V, T$_A$ = +25°C, unless otherwise noted.)

**PROPAGATION DELAY
(WITH 100mV OVERDRIVE)**

MAX6709/14 toc10

IN_ (PFI_)
100mV/div
AC-COUPLED

PWRGD_ ($\overline{PFO\_}$)
2V/div

10µs/div

**$\overline{PFO\_}$ PULLUP AND PULLDOWN RESPONSE
(C$_{\overline{PFO\_}}$ = 47pF)**

MAX6709/14 toc11

PFI_
50mV/div
AC-COUPLED

$\overline{PFO\_}$
2V/div

10µs/div

RESET TIMEOUT DELAY

MAX6709/14 toc12

$\overline{MR}$
2V/div

$\overline{RESET}$
2V/div

40ms/div

MAX6709/MAX6714

PIN		NAME	FUNCTION
MAX6709	MAX6714		
1	—	IN1	Input Voltage 1. See *Selector Guide* for monitored voltages.
2	—	IN2	Input Voltage 2. See *Selector Guide* for monitored voltages.
3	—	IN3	Input Voltage 3. See *Selector Guide* for monitored voltages.
4	—	IN4	Input Voltage 4. See *Selector Guide* for monitored voltages.
5	5	GND	Ground
6	—	PWRGD4	Output 4. PWRGD4 asserts low when IN4 falls below its threshold voltage. PWRGD4 is open drain with a 10µA internal pullup current source to V_{CC}.
7	—	PWRGD3	Output 3. PWRGD3 asserts low when IN3 falls below its threshold voltage. PWRGD3 is open drain with a 10µA internal pullup current source to V_{CC}.
8	—	PWRGD2	Output 2. PWRGD2 asserts low when IN2 falls below its threshold voltage. PWRGD2 is open drain with a 10µA internal pullup current source to V_{CC}.
9	—	PWRGD1	Output 1. PWRGD1 asserts low when IN1 falls below its threshold voltage. PWRGD1 is open drain with a 10µA internal pullup current source to V_{CC}.
10	10	V_{CC}	Power-Supply Input. Connect V_{CC} to a 2.0V to 5.5V supply. An undervoltage lockout circuit forces all PWRGD_ outputs low when V_{CC} drops below the minimum operating voltage. V_{CC} is not a monitored voltage for the MAX6709. For the MAX6714, $\overline{RESET}$ asserts low when V_{CC} drops below its threshold.
—	1	$\overline{MR}$	Manual Reset Input. Force $\overline{MR}$ low to assert the $\overline{RESET}$ output. $\overline{RESET}$ remains asserted for the reset timeout period after $\overline{MR}$ goes high. $\overline{MR}$ is internally pulled up to V_{CC}.
—	2	PFI1	Power-Fail Input 1. Input to noninverting input of the power-fail comparator. PFI1 is compared to an internal 0.62V reference. Use an external resistor-divider network to adjust the monitor threshold.
—	3	PFI2	Power-Fail Input 2. Input to noninverting input of the power-fail comparator. PFI2 is compared to an internal 0.62V reference. Use an external resistor-divider network to adjust the monitor threshold.
—	4	PFI3	Power-Fail Input 3. Input to noninverting input of the power-fail comparator. PFI3 is compared to an internal 0.62V reference. Use an external resistor-divider network to adjust the monitor threshold.
—	6	$\overline{PFO3}$	Power-Fail Output 3. $\overline{PFO3}$ is an active-low, open-drain output with a 10µA internal pullup to V_{CC}. $\overline{PFO3}$ asserts low when PFI3 is below the selected threshold.
—	7	$\overline{PFO2}$	Power-Fail Output 2. $\overline{PFO2}$ is an active-low, open-drain output with a 10µA internal pullup to V_{CC}. $\overline{PFO2}$ asserts low when PFI2 is below the selected threshold.
—	8	$\overline{PFO1}$	Power-Fail Output 1. $\overline{PFO1}$ is an active-low, open-drain output with a 10µA internal pullup to V_{CC}. $\overline{PFO1}$ asserts low when PFI1 is below the selected threshold.
—	9	$\overline{RESET}$	Reset Output. $\overline{RESET}$ is an active-low, open-drain output that asserts low when V_{CC} drops below its preset threshold voltage or when a manual reset is initiated. $\overline{RESET}$ remains low for the reset timeout period after V_{CC} exceeds the selected reset threshold or $\overline{MR}$ is released.

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_____Detailed Description

The MAX6709/MAX6714 are low-power, quad voltage monitors designed for multivoltage systems. Preset voltage options for 5.0V, 3.3V, 3.0V, 2.5V, and 1.8V make these quad monitors ideal for applications such as telecommunications, desktop and notebook computers, high-end printers, data storage equipment, and networking equipment.

The MAX6709/MAX6714 have an internally trimmed threshold that minimizes or eliminates the need for external components. The four open-drain outputs have weak (10µA) internal pullups to V_{CC}, allowing them to interface easily with other logic devices. The weak internal pullups can be overdriven by external pullups to any voltage from 0 to 5.5V. Internal circuitry prevents current flow from the external pullup voltage to V_{CC}. The outputs can be wire-ORed for a single power-good signal.

The MAX6709 quad voltage monitor includes an accurate reference, four precision comparators, and a series of internally trimmed resistor-divider networks to set the factory-fixed threshold options. The resistor networks scale the specified IN_ reset voltages to match the internal reference/comparator voltage. Adjustable threshold options bypass the internal resistor networks and connect directly to one of the comparator inputs (an external resistor-divider network is required for threshold matching). The MAX6709 monitors power supplies with either 5% or 10% tolerance specifications, depending on the selected version. Additional high-input-impedance comparator options can be used
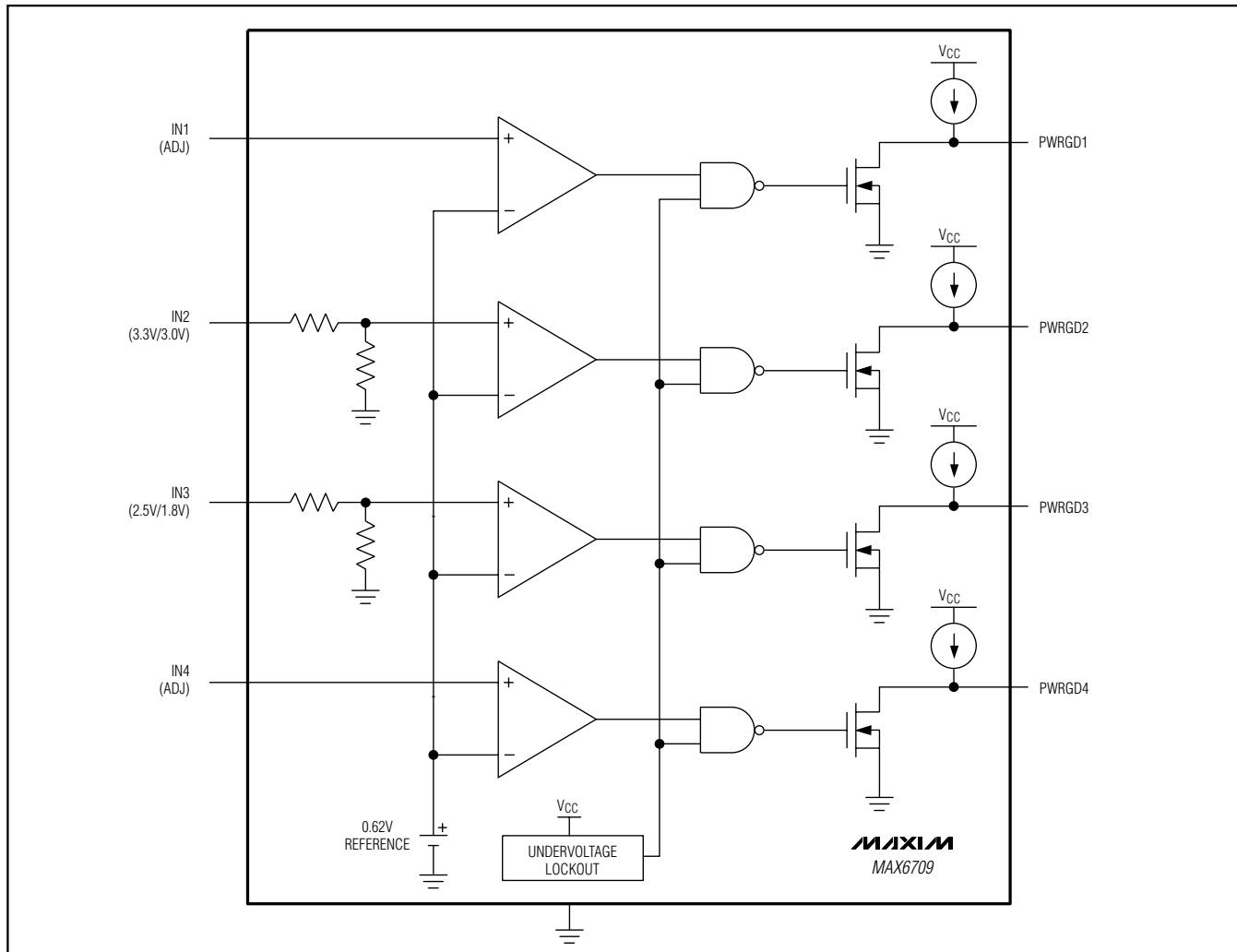

Figure 1. MAX6709 Functional Diagram
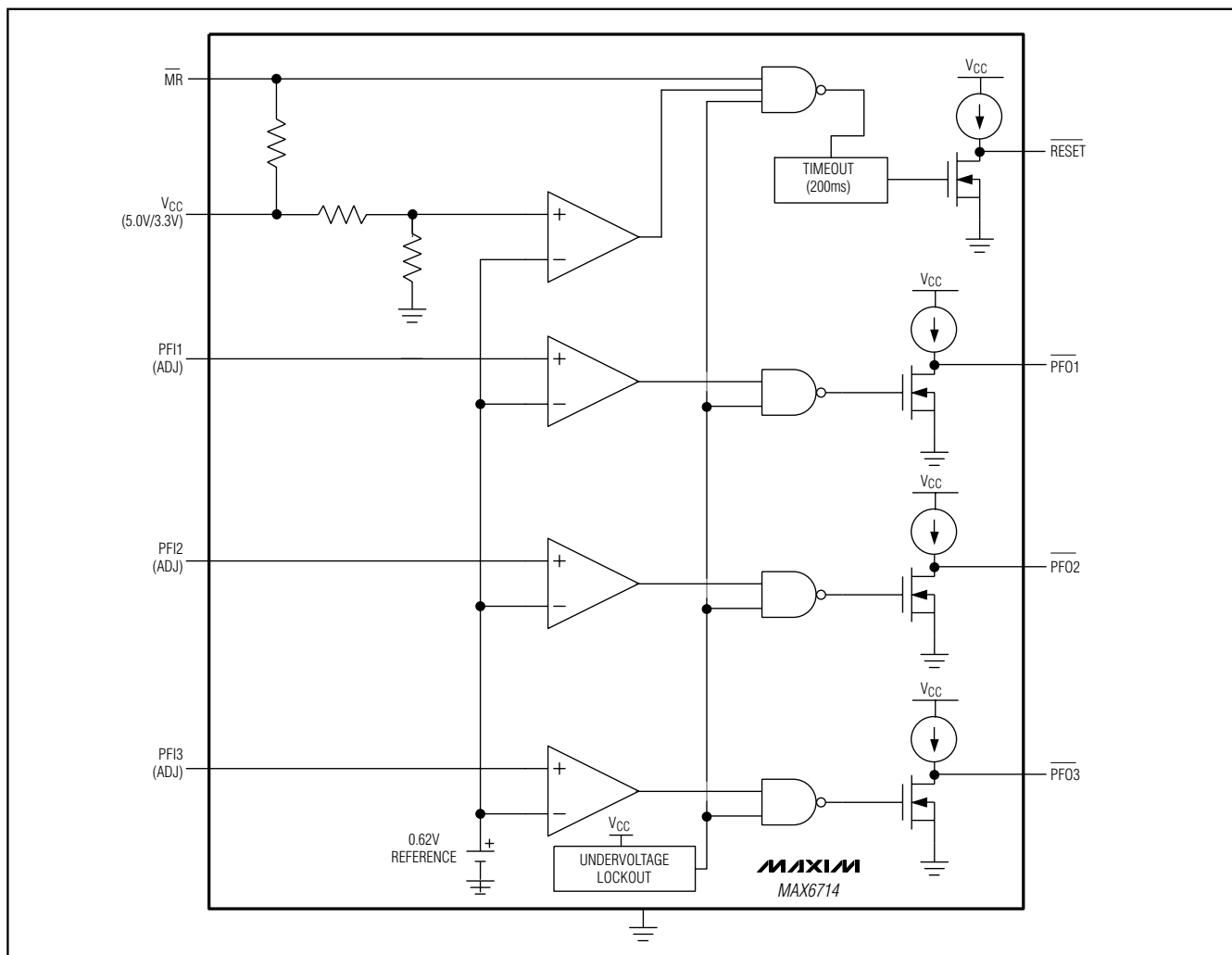
MAX6709/MAX6714

Figure 2. MAX6714 Functional Diagram

as an adjustable voltage monitor, general-purpose comparator, or digital-level translator.

The MAX6714 quad voltage monitor/reset offers one fixed input with internal timing for µP reset, three power-fail comparators, and a manual reset input ($\overline{MR}$). $\overline{RESET}$ asserts low when V_{CC} drops below its threshold or $\overline{MR}$ is driven low. Each of the three power-fail inputs connects directly to one of the comparator inputs.

When any input is higher than the threshold level, the output is high. The output goes low as the input drops below the threshold voltage. The undervoltage lockout circuitry remains active and all outputs remain low with V_{CC} down to 1V (Figures 1 and 2).

Applications Information

Hysteresis

When the voltage on one comparator input is at or near the voltage on another input, ambient noise generally causes the comparator output to oscillate. The most common way to eliminate this problem is through hysteresis. When the two comparator input voltages are equal, hysteresis causes one comparator input voltage to move quickly past the other, thus taking the input out of the region where oscillation occurs. Standard comparators require hysteresis to be added through the use of external resistors. The external resistive network usually provides a positive feedback to the input in order to cause a jump in the threshold voltage when output toggles in one direction or the other. These
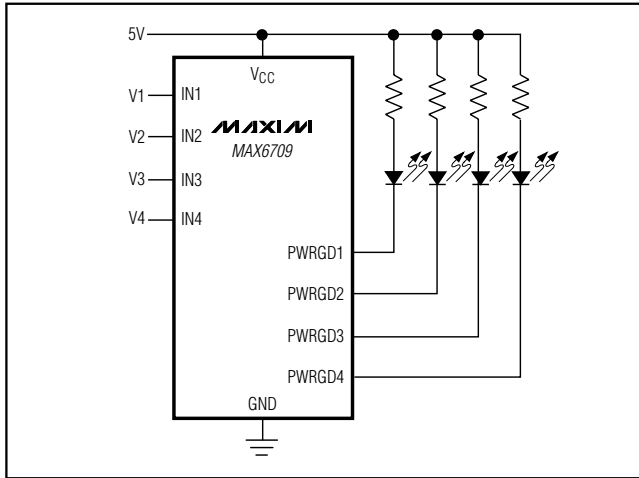
8 _____

MAXIM

Figure 3. Quad Undervoltage Detector with LED Indicators
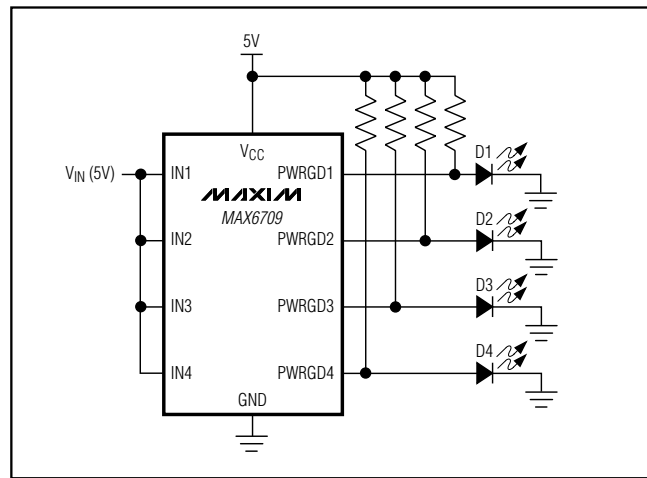
Figure 4. V_{CC} Bar Graph Monitoring

Figure 5. Window Detection
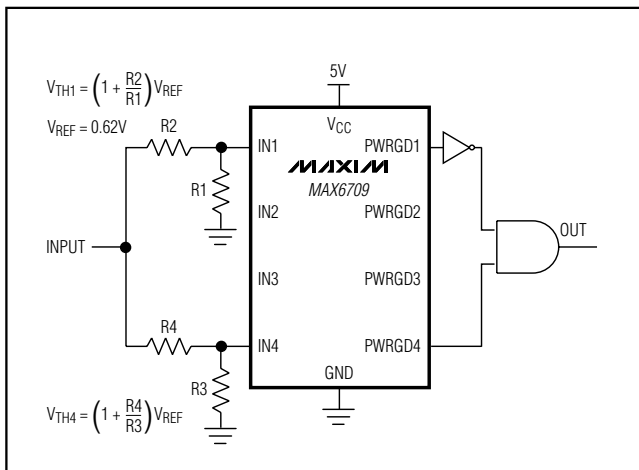

Figure 6. Output Response of Window Detector Circuit

resistors are not required when using the MAX6709/MAX6714 because hysteresis is built into the device. MAX6709/MAX6714 hysteresis is typically 0.3% of the threshold voltage.

Undervoltage Detection Circuit

The open-drain outputs of the MAX6709/MAX6714 can be configured to detect an undervoltage condition. Figure 3 shows a configuration where an LED turns on when the comparator output is low, indicating an undervoltage condition.

The MAX6709/MAX6714 can also be used in applications such as system supervisory monitoring, multivoltage level detection, and V_{CC} bar graph monitoring (Figure 4).
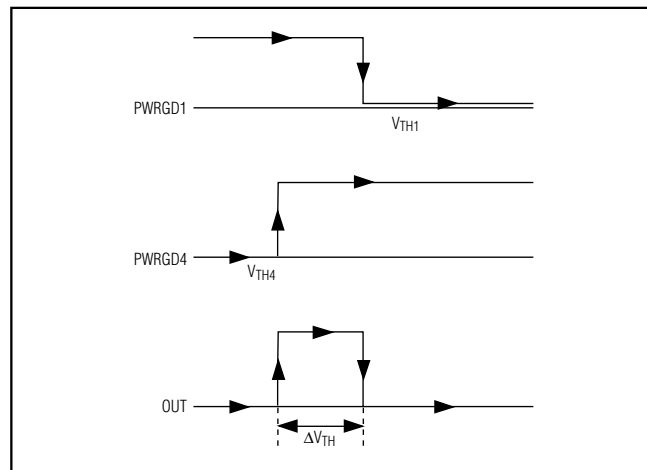
Window Detection

A window detector circuit uses two auxiliary inputs in a configuration such as the one shown in Figure 5. External resistors R1–R4 set the two threshold voltages (V_{TH1} and V_{TH4}) of the window detector circuit. Window width (ΔV_{TH}) is the difference between the threshold voltages (Figure 6).

Adjustable Input

The MAX6709 offers several monitor options with adjustable reset thresholds. The MAX6714 has three monitored inputs with adjustable thresholds. The threshold voltage at each adjustable IN_ (PFI_) input is typically 0.62V. To monitor a voltage >0.62V, connect a resistor-divider network to the circuit as shown in Figure 7.

$$V_{INTH} = 0.62V \times (R1 + R2) / R2$$

MAX6709/MAX6714

Or, solved in terms of R1:

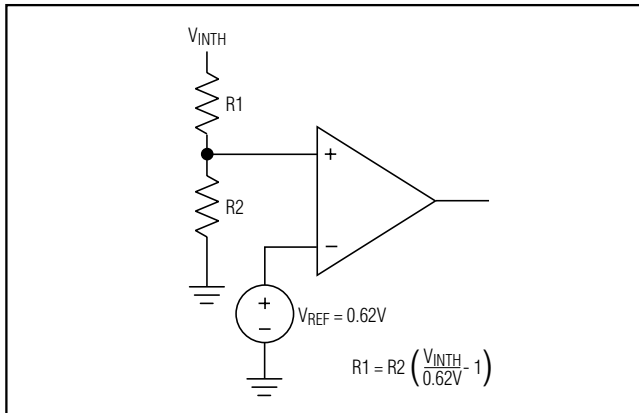$$R1 = R2 \left((V_{INTH} / 0.62V) - 1 \right)$$

Figure 7. Setting the Auxiliary Monitor

Unused Inputs
The unused inputs (except the adjustable) are internally connected to ground through the lower resistors of the threshold-setting resistor pairs. The adjustable input, however, must be connected to ground if unused.

Reset Output
The MAX6714 $\overline{RESET}$ output asserts low when V_{CC} drops below its specified threshold or $\overline{MR}$ asserts low and remains low for the reset timeout period (140ms min) after V_{CC} exceeds its threshold and $\overline{MR}$ deasserts (Figure 8). The output is open drain with a weak (10µA) internal pullup to V_{CC}. For many applications, no external pullup resistor is required to interface with other logic devices. An external pullup resistor to any voltage from 0 to 5.5V overdrives the internal pullup if interfacing to different logic supply voltages (Figure 9). Internal circuitry prevents reverse current flow from the external pullup voltage to V_{CC}.

Manual Reset Input
Many µP-based products require manual reset capability, allowing the operator, a test technician, or external logic circuitry to initiate a reset. A logic low on $\overline{MR}$ asserts $\overline{RESET}$ low. $\overline{RESET}$ remains asserted while $\overline{MR}$ is low, and during the reset timeout period (140ms min) after $\overline{MR}$ returns high. The $\overline{MR}$ input has an internal 20kΩ pullup resistor to V_{CC}, so it can be left open if unused. Drive $\overline{MR}$ with TTL or CMOS-logic levels, or with open-drain/collector outputs. Connect a normally open momentary switch from $\overline{MR}$ to GND to create a manual reset function; external debounce circuitry is not required. If $\overline{MR}$ is driven from long cables or if the device is used in a noisy environment, connecting a 0.1µF capacitor from $\overline{MR}$ to GND provides additional noise immunity.

Reseting the µP from a 2nd Voltage (MAX6714)
The MAX6714 can be configured to assert a reset from a second voltage by connecting the power-fail output to manual reset. As the $V_{PFI\_}$ falls below its threshold, $\overline{PFO}$ goes low and asserts the reset output for the reset timeout period after the manual reset input is deasserted. (See *Typical Operating Circuit*.)

Power-Supply Bypassing and Grounding
The MAX6709/MAX6714 operate from a single 2.0V to 5.5V supply. In noisy applications, bypass V_{CC} with a 0.1µF capacitor as close to V_{CC} as possible.
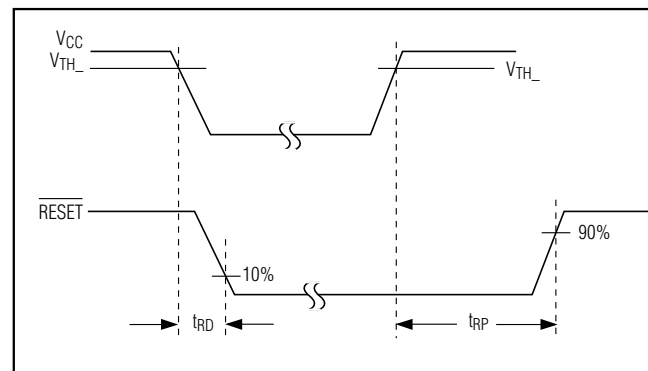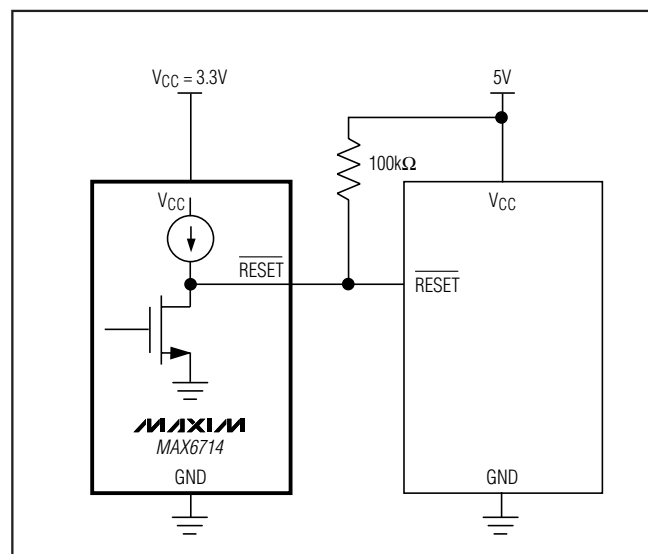

Figure 8. $\overline{RESET}$ Output Timing Diagram

Figure 9. Interfacing to Different Logic Supply Voltage

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Low-Voltage, High-Accuracy Quad, Voltage Monitors in µMAX Package

_____Selector Guide (MAX6709)

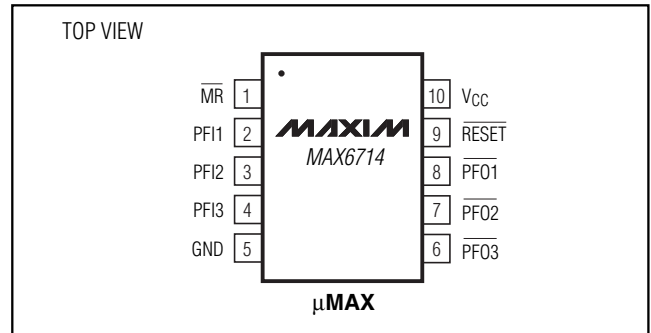| PART	NOMINAL INPUT VOLTAGE				
	IN1 (V)	IN2 (V)	IN3 (V)	IN4 (V)	SUPPLY TOLERANCE (%)
MAX6709AUB	5	3.3	2.5	Adj*	10
MAX6709BUB	5	3.3	2.5	Adj*	5
MAX6709CUB	5	3.3	1.8	Adj*	10
MAX6709DUB	5	3.3	1.8	Adj*	5
MAX6709EUB	Adj*	3.3	2.5	1.8	10
MAX6709FUB	Adj*	3.3	2.5	1.8	5
MAX6709GUB	5	3.3	Adj*	Adj*	10
MAX6709HUB	5	3.3	Adj*	Adj*	5
MAX6709IUB	Adj*	3.3	2.5	Adj*	10
MAX6709JUB	Adj*	3.3	2.5	Adj*	5
MAX6709KUB	Adj*	3.3	1.8	Adj*	10
MAX6709LUB	Adj*	3.3	1.8	Adj*	5
MAX6709MUB	Adj*	3	Adj*	Adj*	10
MAX6709NUB	Adj*	3	Adj*	Adj*	5
MAX6709OUB	Adj*	Adj*	Adj*	Adj*	N/A

*Adjustable voltage based on 0.62V internal threshold. External threshold voltage can be set using an external resistor-divider.

_____Selector Guide (MAX6714)

PART	NOMINAL INPUT VOLTAGE				
	V_{CC} (V)	PFI1 (V)	PFI2 (V)	PFI3 (V)	SUPPLY TOLERANCE (%)
MAX6714AUB	5	Adj*	Adj*	Adj*	10
MAX6714BUB	5	Adj*	Adj*	Adj*	5
MAX6714CUB	3.3	Adj*	Adj*	Adj*	10
MAX6714DUB	3.3	Adj*	Adj*	Adj*	5

*Adjustable voltage based on 0.62V internal threshold. External threshold voltage can be set using an external resistor-divider.

_____Pin Configurations (continued)

TOP VIEW

```
        ___
        MR   ┌1      10┐ V_CC
       PFI1  │2   ╱╲╱╲╱  9│ RESET
       PFI2  │3   MAXIM   8│ PFO1
       PFI3  │4  MAX6714  7│ PFO2
        GND  │5       6│ PFO3
             └─────────┘
              µMAX
```
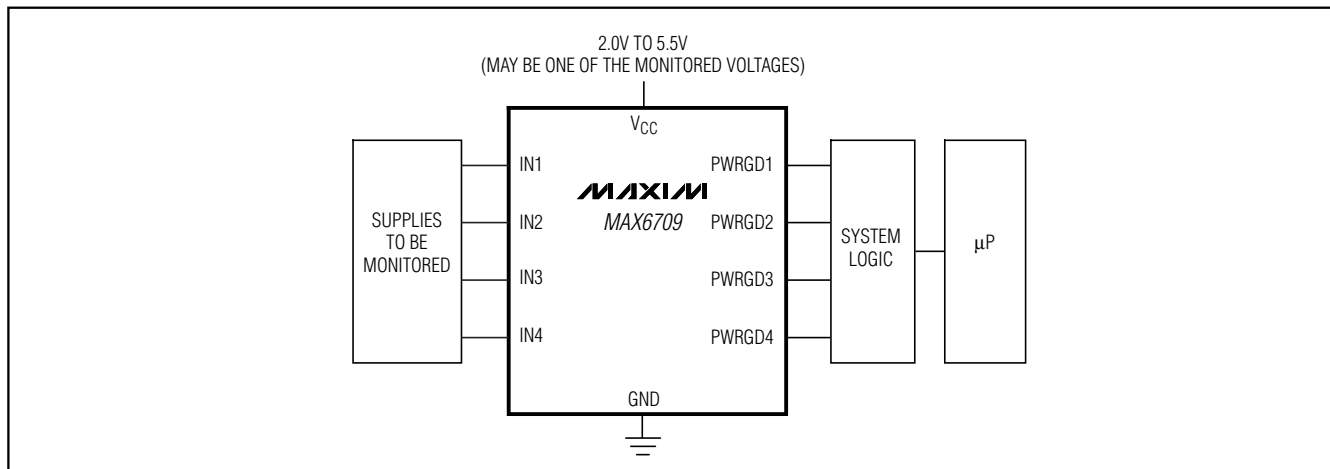
_____Chip Information
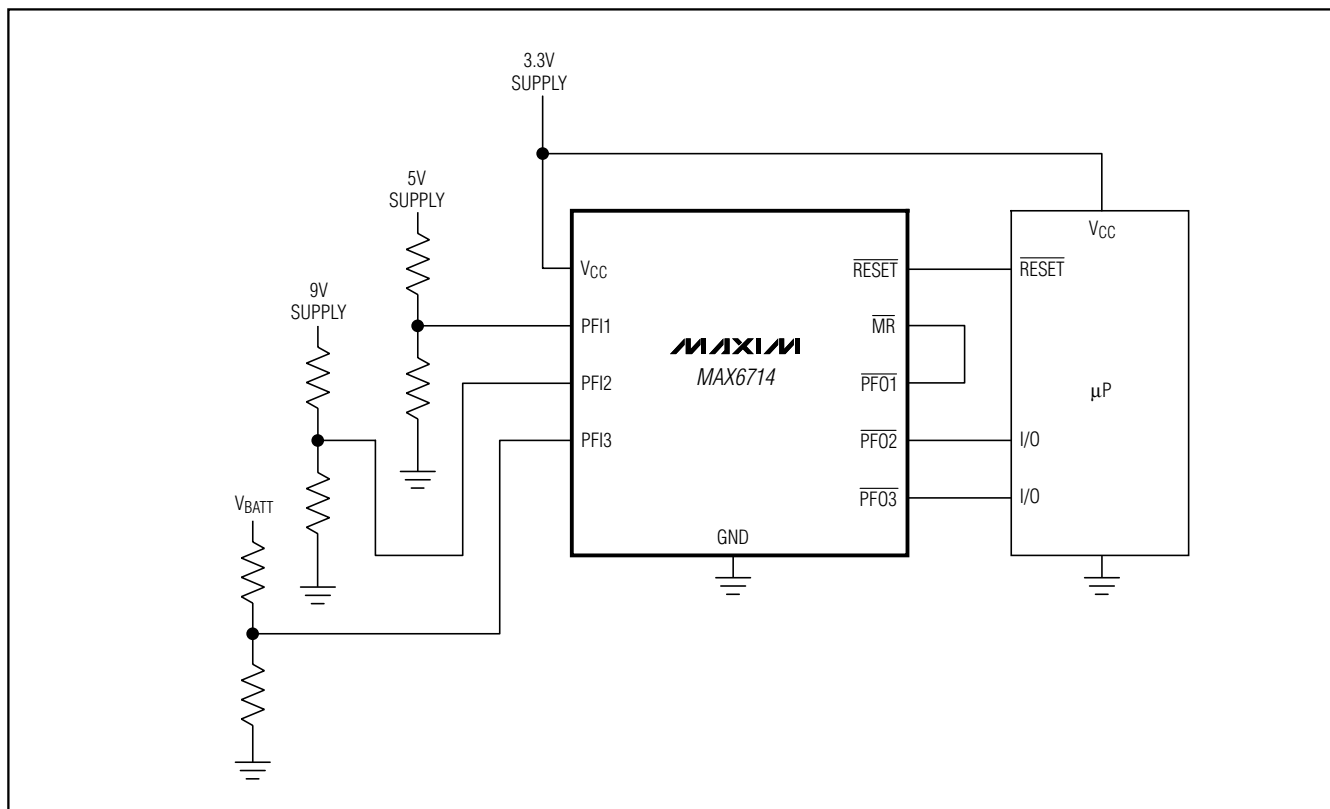
TRANSISTOR COUNT: 1029
PROCESS: BiCMOS

Low-Voltage, High-Accuracy, Quad Voltage Monitors in µMAX Package

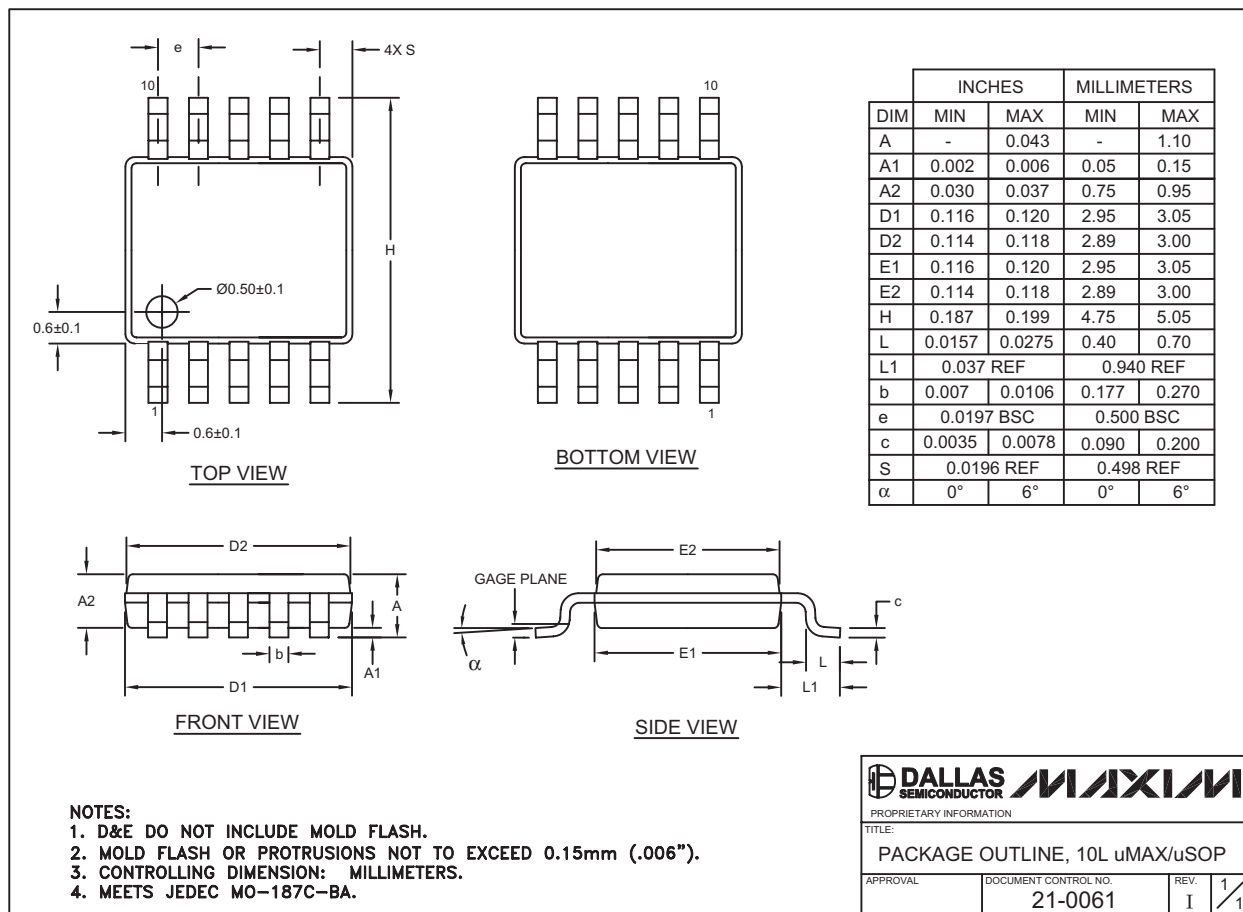_____*Typical Operating Circuit (MAX6709)*

_____*Typical Operating Circuit (MAX6714)*

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Low-Voltage, High-Accuracy, Quad Voltage Monitors in µMAX Package

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to **www.maxim-ic.com/packages**.)

DIM	INCHES MIN	INCHES MAX	MILLIMETERS MIN	MILLIMETERS MAX
A	-	0.043	-	1.10
A1	0.002	0.006	0.05	0.15
A2	0.030	0.037	0.75	0.95
D1	0.116	0.120	2.95	3.05
D2	0.114	0.118	2.89	3.00
E1	0.116	0.120	2.95	3.05
E2	0.114	0.118	2.89	3.00
H	0.187	0.199	4.75	5.05
L	0.0157	0.0275	0.40	0.70
L1	0.037 REF		0.940 REF	
b	0.007	0.0106	0.177	0.270
e	0.0197 BSC		0.500 BSC	
c	0.0035	0.0078	0.090	0.200
S	0.0196 REF		0.498 REF	
α	0°	6°	0°	6°

TOP VIEW

BOTTOM VIEW

FRONT VIEW

SIDE VIEW

10LUMAX.EPS

NOTES:
1. D&E DO NOT INCLUDE MOLD FLASH.
2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED 0.15mm (.006").
3. CONTROLLING DIMENSION: MILLIMETERS.
4. MEETS JEDEC MO−187C−BA.

DALLAS SEMICONDUCTOR /V/IXI/VI

PROPRIETARY INFORMATION

TITLE:
PACKAGE OUTLINE, 10L uMAX/uSOP

APPROVAL	DOCUMENT CONTROL NO. 21-0061	REV. I	1/1

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