### **ABSOLUTE MAXIMUM RATINGS**

All voltages referenced to ground (GND). Supply Voltage (V<sub>CC</sub>).....+7V Control Input Voltage (EN, EN, EN\_) .....-0.3V to (V<sub>CC</sub> + 0.3V) Driver Input Voltage (T\_IN)....-0.3V to (V<sub>CC</sub> + 0.3V) Driver Output Voltage (Y\_, Z\_) (Driver Disabled) ....-7.5V to +12.5V Driver Output Voltage (Y\_, Z\_) (Driver Enabled) ....-7.5V to +10V Continuous Power Dissipation (T<sub>A</sub> = +70°C) 16-Pin TSSOP (derate 9.4mW/°C above +70°C) ......755mW

16-Pin Narrow SO (derate 8.70mW/	<sup>o</sup> C above +70°C)696mW
16-Pin Wide SO (derate 9.52mW/°C	c above +70°C)762mW
Operating Temperature Range	
MAX304_C_E	0°C to +70°C
MAX304_E_E	40°C to +85°C
Maximum Junction Temperature	+150°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = +5V \pm 5\%, T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $V_{CC} = +5V$  and  $T_A = +25^{\circ}C$ .) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	МАХ	UNITS
DRIVER						
Driver Differential Output	Vod	Figure 1, R = $50\Omega$ Figure 1, R = $27\Omega$	2.0 1.5			V
Change in Magnitude of Differential Output Voltage	$\Delta V_{OD}$	Figure 1, R = $50\Omega$ or $27\Omega$ (Note 2)			0.2	V
Driver Common-Mode Output Voltage	Voc	Figure 1, R = $50\Omega$ or $27\Omega$		V <sub>CC</sub> / 2	3	V
Change In Magnitude of Common-Mode Voltage	ΔV <sub>OC</sub>	Figure 1, R = $50\Omega$ or $27\Omega$ (Note 2)			0.2	V
Input High Voltage	VIH	T_IN, EN_, EN, EN	2.0			V
Input Low Voltage	VIL	T_IN, EN_, EN, EN			0.8	V
Hot-Swap Driver Input Current	I <sub>HOT</sub> SWAP	EN_, EN, EN (Note 3)			±200	μA
Driver Input Current	l <sub>IN</sub>	T_IN, EN_, EN, EN			±1	μA
Driver Short-Circuit Output Current	ISC	$-7V \le V_{OUT} \le +10V$ (Note 4)	±25		±250	mA
Output Leakage (Y_, Z_)		MAX3040/MAX3041/MAX3042B EN_ = GND			±1	
when Disabled		MAX3043/MAX3044/MAX3045B EN = GND, $\overline{EN} = V_{CC}$		±		μA
ESD Protection (Y_, Z_)		Human Body Model		±10		kV
Electrical Fast Transient/Burst Immunity		IEC 1000-4-4		±4		kV
SUPPLY CURRENT	•		•			
Supply Current	ICC	No load		1	2	mA
Supply Current in Shutdown Mode		MAX3040/MAX3041/MAX3042B EN_ = GND, T <sub>A</sub> = +25°C			10	
	ISHDN	MAX3043/MAX3044/MAX3045B EN = GND, $\overline{EN} = V_{CC}$ , $T_A = +25^{\circ}C$		0.002		μA
2		-	•		////	XI/M

### SWITCHING CHARACTERISTICS—MAX3040/MAX3043

(V<sub>CC</sub> = +5V ±5%,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at V<sub>CC</sub> = +5V and  $T_A = +25^{\circ}$ C.)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS
Maximum Data Rate	fMAX			250			kbps
Driver Propagation Dalay	tPLH	Figures 2 and 3,			0.7	1.5	10
Driver Propagation Delay	t <sub>PHL</sub>	$R_{DIFF} = 54\Omega$ , $C_{DIFF} = 50\mu$	ъF		0.7	1.5	μs
Driver Differential Output	t <sub>F</sub>	Figures 2 and 3,		0.48	0.75	1.33	10
Rise-Time/Fall-Time	t <sub>R</sub>	$R_{DIFF} = 54\Omega$ , $C_{DIFF} = 50\mu$	рF	0.48	0.75	1.33	μs
Change Driver to Driver	<b>t</b> DSKEW	Different chips	Figures 2 and 3,			±350	
Skew Driver to Driver	tsskew	Same chip	$R_{DIFF} = 54\Omega$ , $C_{DIFF} = 50pF$			±100	ns
Driver Differential Output Skew   tPLH - t PHL	<sup>t</sup> SKEW	Figures 2 and 3, RDIFF = $54\Omega$ , CDIFF = $50pF$				±100	ns
Driver Enable to Output High	tzH	MAX3040, Figures 4 and 5, S2 closed, $R_L = 500\Omega$ , $C_L = 100PF$				2.0	μs
Driver Enable from Shutdown to Output High	セト(SHDN)	Figures 4 and 5, S2 closed, $R_L = 500\Omega$ , $C_L = 100pF$				2.0	μs
Driver Enable to Output Low	tzL	MAX3040, Figures 4 and 5, S1 closed, $R_L = 500\Omega$ , $C_L = 100pF$				2.0	μs
Driver Enable from Shutdown to Output Low	tzl(SHDN)	Figures 4 and 5, S1 closed, $R_L = 500\Omega$ , $C_L = 100pF$				2.0	μs
Driver Disable Time from Low	tLZ	Figures 4 and 5, S1 closed, $R_L = 500\Omega$ , $C_L = 15pF$				500	ns
Driver Disable Time from High	t <sub>HZ</sub>	Figures 4 and 5, S2 close $R_L = 500\Omega$ , $C_L = 15pF$	ed,			500	ns

### SWITCHING CHARACTERISTICS—MAX3041/MAX3044

(V<sub>CC</sub> = +5V ±5%, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at V<sub>CC</sub> = +5V and T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL	CONDITI	ONS	MIN	ТҮР	МАХ	UNITS
Maximum Data Rate	fMAX			2.5			Mbps
Driver Propagation Delay	t <sub>PLH</sub>	Figures 2 and 3,			70	150	ns
Divertropagation Delay	tPHL	$R_{DIFF} = 54\Omega, C_{DIFF} = 50$	pF		70	150	115
Driver Differential Output	tF	Figures 2 and 3, RDIFF = $54\Omega$ , C DIFF = $50$ pF		33	70	133	
Rise-Time/Fall-Time	t <sub>R</sub>			33	70	133	ns
	tDSKEW	Different chips	Figures 2 and 3,			±52	
Skew Driver to Driver	tsskew	Same chip	$R_{DIFF} = 54\Omega$ , $C_{DIFF} = 50pF$			±15	ns
Driver Differential Output Skew <sup> </sup> tPLH - t PHL	tskew	Figures 2 and 3, $R_{DIFF} = 54\Omega$ , $C_{DIFF} = 50pF$				±15	ns
Driver Enable to Output High	tzH	MAX3041, Figures 4 and 5, S2 closed, $R_L = 500\Omega$ , $C_L = 100pF$				400	ns

### SWITCHING CHARACTERISTICS—MAX3041/MAX3044 (continued)

(V<sub>CC</sub> = +5V ±5%, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at V<sub>CC</sub> = +5V and T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Driver Enable from Shutdown to Output High	tzh(shdn)	Figures 4 and 5, S2 closed, RL = $500\Omega$ , CL = $100pF$			400	ns
Driver Enable to Output Low	tzL	MAX3041, Figures 4 and 5, S1 closed, $R_L=500\Omega,C_L=100pF$			400	ns
Driver Enable from Shutdown to Output Low	tzl(SHDN)	Figures 4 and 5, S1 closed, $R_L = 500\Omega$ , $C_L = 100pF$			400	ns
Driver Disable Time from Low	tLZ	Figures 4 and 5, S1 closed, $R_L = 500\Omega$ , $C_L = 15pF$			500	ns
Driver Disable Time from High	t <sub>HZ</sub>	Figures 4 and 5, S2 closed, $R_L = 500\Omega$ , $C_L = 15pF$			500	ns

### SWITCHING CHARACTERISTICS—MAX3042B/MAX3045B

(V<sub>CC</sub> = +5V ±5%, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at V<sub>CC</sub> = +5V and T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL	CONDITI	ONS	MIN	ТҮР	MAX	UNITS
Maximum Data Rate	fMAX			20			Mbps
Driver Propagation Delay	t <sub>PLH</sub>	Figures 2 and 3,			23	40	ns
Divertropagation Delay	<b>t</b> PHL	$R_{DIFF} = 54\Omega, C_{DIFF} = 50\mu$	ρF		23	40	115
Driver Differential Output	tF	Figures 2 and 3,				17	ns
Rise-Time/Fall-Time	t <sub>R</sub>	$R_{DIFF} = 54\Omega, C_{DIFF} = 50\mu$	рF			17	115
Skew Driver to Driver	<b>t</b> DSKEW	Different chips	Figures 2 and 3, $R_{DIFF} = 54\Omega$ ,		±8		ns
	<b>İ</b> SSKEW	Same chip	$C_{DIFF} = 50pF$			±8	115
Differential Driver Output Skew   t PLH - tPHL	tSKEW	Figures 2 and 3, RDIFF = 54 $\Omega$ , C DIFF = 50	рF			±8	ns
Driver Enable to Output High	tzh	MAX3042B, Figures 4 and 5, S2 closed, $R_L$ = 500 $\Omega$ , C L = 100pF				300	ns
Driver Enable from Shutdown to Output High	セト(SHDN)	Figures 4 and 5, S2 closed, RL = $500\Omega$ , CL = $100pF$				300	ns
Driver Enable to Output Low	tzL	MAX3042B, Figures 4 and 5, S1 closed, $R_L=500\Omega,C_L=100pF$				300	ns
Driver Enable from Shutdown to Output Low	tzL(SHDN)	Figures 4 and 5, S1 closed, $R_L = 500\Omega$ , $C_L = 100pF$				300	ns
Driver Disable Time from Low	tLZ	Figures 4 and 5, S1 closed, $R_L = 500\Omega$ , $C_L = 15pF$				400	ns
Driver Disable Time from High	t <sub>HZ</sub>	Figures 4 and 5, S2 closed, $R_L = 500\Omega$ , $C_L = 15pF$				400	ns

Note 1: All currents into the device are positive; all currents out of the device are negative. All voltages are referenced to device ground unless otherwise noted.

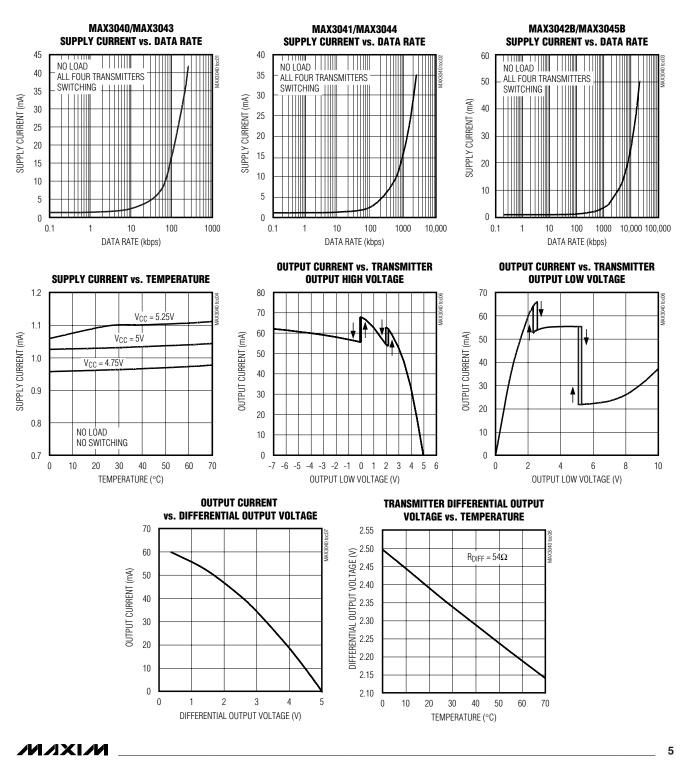
Note 2:  $\Delta V_{OD}$  and  $\Delta V_{OC}$  are the changes in  $V_{OD}$  and  $V_{OC}$ , respectively, when the transmitter input changes state.

**Note 3:** This input current level is for the hot-swap enable (EN\_, EN, EN) inputs and is present until the first transition only. After the first transition the input reverts to a standard high-impedance CMOS input with input current I<sub>IN</sub>. For the first 20µs the input current may be as high as 1mA. During this period the input is disabled.

Note 4: Maximum current level applies to peak current just prior to foldback-current limiting. Minimum current level applies during current limiting.

### **Typical Operating Characteristics**

 $(V_{CC} = +5V, T_A = +25^{\circ}C, unless otherwise noted.)$ 



**Pin Description** 

Р	IN		
MAX3040/MAX3041/ MAX3042B	MAX3043/MAX3044/ MAX3045B	NAME	FUNCTION
1	1	T1IN	Transmitter 1 Input
2	2	Y1	Noninverting Transmitter 1 Output
3	3	Z1	Inverting Transmitter 1 Output
_	4	EN	Transmitter Enable High Input. Drive EN high to enable all four transmitters. When EN is low and $\overline{\text{EN}}$ is high, all transmitters are disabled and the part enters a low-power shutdown state. The transmitter outputs are high impedance when disabled.
4	_	EN12	Transmitter Enable Input to Control Transmitters 1 and 2. Drive EN12 high to enable transmitters 1 and 2. Drive EN12 low to disable transmitters 1 and 2. The transmitter outputs are high impedance when disabled. The part enters a low-power shutdown state when both EN12 and EN34 are low.
5	5	Z2	Inverting Transmitter 2 Output
6	6	Y2	Noninverting Transmitter 2 Output
7	7	T2IN	Transmitter 2 Input
8	8	GND	Ground
9	9	T3IN	Transmitter 3 Input
10	10	Y3	Noninverting Transmitter 3 Output
11	11	Z3	Inverting Transmitter 3 Output
	12	ĒN	Transmitter Enable Low Input. Drive $\overline{\text{EN}}$ low to enable all four transmitters. When EN is low and $\overline{\text{EN}}$ is high, all transmitters are disabled and the part enters a low-power shutdown state. The transmitter outputs are high impedance when disabled.
12		EN34	Transmitter Enable Input to Control Transmitters 3 and 4. Drive EN34 high to enable transmitters 3 and 4. Drive EN34 low to disable transmitters 3 and 4. The transmitter outputs are high impedance when disabled. The part enters a low-power shutdown state when both EN12 and EN34 are low.
13	13	Z4	Inverting Transmitter 4 Output
14	14	Y4	Noninverting Transmitter 4 Output
15	15	T4IN	Transmitter 4 Input
16	16	Vcc	Positive Supply. Bypass with a 0.1µF capacitor to GND.

### **Detailed Description**

The MAX3040–MAX3045 are quad RS-485/RS-422 transmitters. They operate from a single +5V power supply and are designed to give optimum performance when used with the MAX3093E/MAX3095 5V quad RS-485/RS-422 receivers or MAX3094E/MAX3096 3V quad RS-485/RS-422 receivers. The MAX3040–MAX3045 only need 1mA of operating supply current and consume 2nA when they enter a low-power shutdown mode. The MAX3040–MAX3045 also feature a hot-swap capability allowing line insertion without erroneous data transfer. The MAX3042B/MAX3045B are capable of transferring data up to 20Mbps, the MAX3041/MAX3044 for data rates up to 2.50Mbps, and the MAX3040/MAX3043 for data rates up to 250Mbps. All transmitter outputs are protected to  $\pm$ 10kV using the Human Body Model.

#### ±10kV ESD Protection

As with all Maxim devices, ESD-protection structures are incorporated on all pins to protect against electrostatic discharges (ESD) encountered during handling and assembly. The MAX3040–MAX3045 transmitter outputs have extra protection against electrostatic discharges found in normal operation. Maxim's engineers have developed state-of-the-art structures to protect these pins against the application of ±10kV ESD (Human Body Model), without damage.

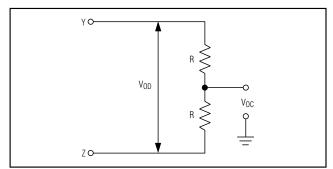


Figure 1. Driver DC Test Circuit

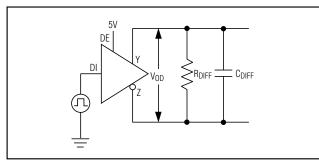


Figure 2. Driver Timing Test Circuit

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#### ESD Test Conditions

ESD performance depends on a number of conditions. Contact Maxim for a reliability report that documents test setup, methodology, and results.

#### Human Body Model

Figure 6a shows the Human Body Model, and Figure 6b shows the current waveform it generates when discharged into low impedance. This model consists of a 100pF capacitor charged to the ESD voltage of interest, which is then discharged into the device through a  $1.5k\Omega$  resistor.

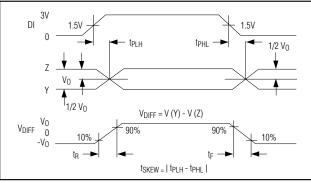


Figure 3. Driver Propagation Delays

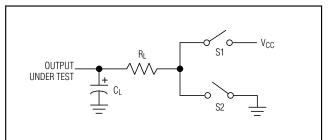


Figure 4. Driver Enable/Disable Timing Test Load

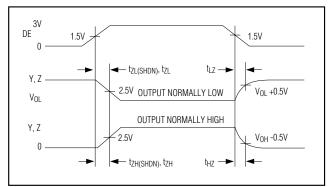


Figure 5. Driver Enable and Disable Times

#### Machine Model

The Machine Model for ESD testing uses a 200pF storage capacitor and zero-discharge resistance. It mimics the stress caused by handling during manufacturing and assembly. Of course, all pins (not just RS-485 inputs) require this protection during manufacturing. Therefore, the Machine Model is less relevant to the I/O ports than are the Human Body Model.

#### ±4kV Electrical Fast Transient/Burst Testing (IEC 1000-4-4)

IEC 1000-4-4 Electrical Fast Transient/Burst (EFT/B) is an immunity test for the evaluation of electrical and electronic systems during operating conditions. The test was adapted for evaluation of integrated circuits with power applied. Repetitive fast transients with severe pulsed EMI were applied to signal and control ports. Over 15,000 distinct discharges per minute are sent to each interface port of the IC or equipment under test (EUT) simultaneously with a minimum test duration time of one minute. This simulates stress due to displacement current from electrical transients on AC mains, or other telecommunication lines in close proximity. Short rise times and very specific repetition rates are essential to the validity of the test.

Stress placed on the EUT is severe. In addition to the controlled individual discharges placed on the EUT, extraneous noise and ringing on the transmission line can multiply the number of discharges as well as increase the magnitude of each discharge. All cabling was left unterminated to simulate worst-case reflections.

The MAX3040–MAX3045 were setup as specified in IEC 1000-4-4 and the *Typical Operating Circuit* of this data sheet. The amplitude, pulse rise time, pulse duration, pulse repetition period, burst duration, and burst period (Figure 8) of the burst generator were all verified with a digital oscilloscope according to the specifications in IEC 1000-4-4 sections 6.1.1 and 6.1.2. A simplified diagram of the EFT/B generator is shown in Figure 7. The burst stresses were applied to Y1–Y4 and Z1–Z4 simultaneously.

IEC 1000-4-4 provides several levels of test severity (see Table 1). The MAX3040–MAX3045 pass the 4000V stress, a special category "X" beyond the highest level for severe (transient) industrial environments for telecommunication lines.

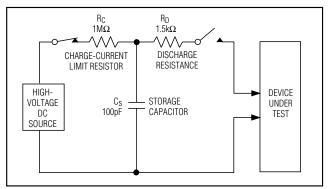


Figure 6a. Human Body ESD Test Model

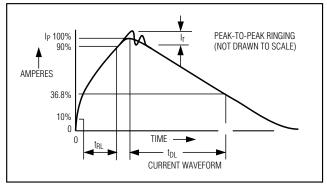


Figure 6b. Human Body Model Current Waveform

# Table 1. Test Severity Levels forCommunication Lines

LEVEL	ON I/O, SIGNAL, DATA AND CONTROL PORTS	EFT	INDUSTRIAL ELECTRO- MAGNETIC
	PEAK VOLTAGE	REPETITION RATE (kHz)	ENVIROMENT
1	250	5	Well protected
2	500	5	Protected
3	1000	5	Typical
4	2000	5	Severe
Х	4000	5	MAX3040-MAX3045



#### IEC 1000-4-4 Burst/Electrical Fast Transient Test Levels (For Communication Lines)

The stresses are applied while the MAX3040–MAX3045 are powered up. Test results are reported as:

- 1) Normal performance within the specification limits.
- 2) Temporary degradation or loss of function or performance which is self-recoverable.
- Temporary degradation, loss of function or performance requiring operator intervention, such as system reset.
- Degradation or loss of function not recoverable due to damage.

The MAX3040–MAX3045 meets classification 2 listed above. Additionally, the MAX3040–MAX3045 will not latchup during the IEC burst stress events.

#### **Hot-Swap Capability**

#### Hot-Swap Inputs

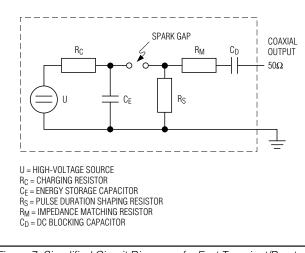
When circuit boards are plugged into a "hot" backplane, there can be disturbances to the differential signal levels that could be detected by receivers connected to the transmission line. This erroneous data could cause data errors to an RS-485/RS-422 system. To avoid this, the MAX3040–MAX3045 have hot-swap capable inputs.

When a circuit board is plugged into a "hot" backplane there is an interval during which the processor is going through its power-up sequence. During this time, the processor's output drivers are high impedance and will be unable to drive the enable inputs of the MAX3040-MAX3045 (EN, EN, EN\_) to defined logic levels. Leakage currents from these high impedance drivers, of as much as 10µA, could cause the enable inputs of the MAX3040-MAX3045 to drift high or low. Additionally, parasitic capacitance of the circuit board could cause capacitive coupling of the enable inputs to either GND or V<sub>CC</sub>. These factors could cause the enable inputs of the MAX3040-MAX3045 to drift to levels that may enable the transmitter outputs (Y\_ and Z\_). To avoid this problem, the hot-swap input provides a method of holding the enable inputs of the MAX3040-MAX3045 in the disabled state as VCC ramps up. This hot-swap input is able to overcome the leakage currents and parasitic capacitances that may pull the enable inputs to the enabled state.

#### Hot-Swap Input Circuitry

In the MAX3040–MAX3045 the enable inputs feature hot-swap capability. At the input there are two NMOS

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MAX3040-MAX3045

Figure 7. Simplified Circuit Diagram of a Fast Transient/Burst Generator

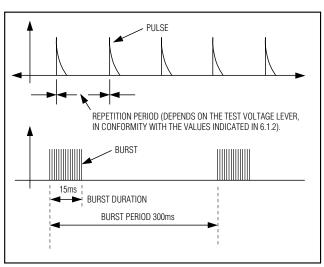


Figure 8. General Graph of a Fast Transient Burst

devices, Q1 and Q2 (Figure 9). When V<sub>CC</sub> is ramping up from 0, an internal 10µs timer turns on Q2 and sets the SR latch, which also turns on Q1. Transistors Q2, a 700µA current sink, and Q1, an 85µA current sink, pull EN to GND through a 5.6k $\Omega$  resistor. Q2 is designed to pull the EN input to the disabled state against an external parasitic capacitance of up to 100pF that is trying to enable the EN input. After 10µs, the timer turns Q2 off and Q1 remains on, holding the EN input low against three-state output leakages that might enable EN. Q1 remains on until an external source overcomes the

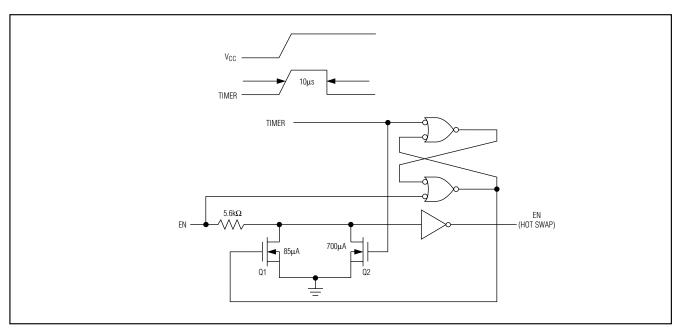


Figure 9. Simplified Structure of the Driver Enable Pin (EN)

required input current. At this time the SR latch resets and Q1 turns off. When Q1 turns off, EN reverts to a standard, high-impedance CMOS input. Whenever  $V_{CC}$  drops below 1V, the hot-swap input is reset.

The EN12 and EN34 input structures are identical to the EN input. For the  $\overline{\text{EN}}$  input, there is a complimentary circuit employing two PMOS devices pulling the  $\overline{\text{EN}}$  input to V<sub>CC</sub>.

#### Hot-Swap Line Transient

The circuit of Figure 10 shows a typical offset termination used to guarantee a greater than 200mV offset when a line is not driven. The 50pF represents the minimum parasitic capacitance which would exist in a typical application. In most cases, more capacitance exists in the system and will reduce the magnitude of the glitch. During a "hot-swap" event when the driver is connected to the line and is powered up, the driver must not cause the differential signal to drop below 200mV. Figures 11 and 12 show the results of the MAX3040-MAX3045 during power-up for two different  $V_{CC}$  ramp rates (0.1V/µs and 1V/µs). The photos show the VCC ramp, the single-ended signal on each side of the  $100\Omega$  termination, the differential signal across the termination, and shows the hot-swap line transient stays above the 200mV RS-485 specification.

#### **Operation of Enable Pins**

The MAX3040–MAX3045 family has two enable-functional versions:

The MAX3040/MAX3041/MAX3042B have two transmitter enable inputs EN12 and EN34. EN12 controls the transmitters 1 and 2, and EN34 controls transmitters 3 and 4. EN12 and EN34 are active-high and the part will enter the low-power shutdown mode when both are pulled low. The transmitter outputs are high impedance when disabled (Table 2).

The MAX3043/MAX3044/MAX3045B have two transmitter enable inputs EN and  $\overline{\text{EN}}$ , which are active-high and active-low, respectively. When EN is logic high or  $\overline{\text{EN}}$  is logic low all transmitters are active. When EN is pulled low and  $\overline{\text{EN}}$  is driven high, all transmitters are disabled and the part enters the low-power shutdown mode. The transmitter outputs are high impedance when disabled (Table 3).

### **Applications Information**

#### **Typical Applications**

The MAX3040–MAX3045 offer optimum performance when used with the MAX3093E/MAX3095 5V quad receivers or MAX3094E/MAX3096 3V quad differential line receivers. Figure 13 shows a typical RS-485 connection for transmitting and receiving data and Figure 14 shows a typical multi-point connection.

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# Table 2. Function Table for MAX3040/MAX3041/MAX3042B

### (Each Pair of Transmitters)

INDUT		OUTP	UTS		
INPUT	EN_	Y_	<b>Z</b> _		
Н	Н	Н	L		
L	Н	L	Н		
Х	L	High-Z	High-Z		
H = Logic High X = Don't Care L = Logic Low High-Z = High Impedance					

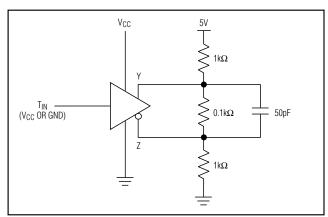


Figure 10. Differential Power-Up Glitch (Hot Swap)

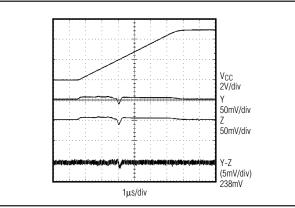


Figure 12. Differential Power-Up Glitch (1V/µs)

### Table 3. Function Table for MAX3043/ MAX3044/MAX3045B

### (All Transmitters)

H H X H L   H H X H L   L H X L H   H X L H L   H X L H L   H X L H L	INPUT	EN	ĒN	OUTPL	JTS
LHXLHHXLHLLXLLHXLHHigh-ZHigh-ZH = Logic High $X = Don't Care$ X	INPUT	EN	EN	Y	Z
HXLHLLXLLHXLHHigh-ZHigh-ZH = Logic High $X = Don't Care$	Н	Н	Х	Н	L
LXLLHXLHHigh-ZHigh-ZH = Logic High $X = Don't Care$	L	Н	Х	L	Н
XLHHigh-ZHigh-Z $H = Logic High$ $X = Don't Care$	Н	Х	L	Н	L
H = Logic High $X = Don't Care$	L	Х	L	L	Н
	Х	L	Н	High-Z	High-Z

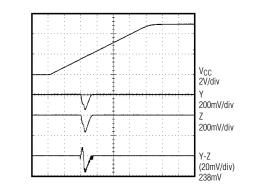


Figure 11. Differential Power-Up Glitch (0.1V/µs)

Downloaded from Arrow.com.

MAX3040-MAX3045

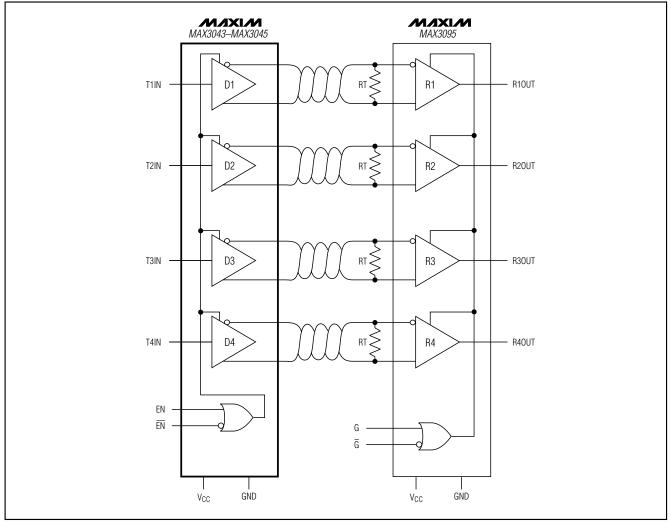


Figure 13. Typical Connection of a Quad Transmitter and a Quad Receiver as a Pair

#### **Typical Multiple-Point Connection**

Figure 14 shows a typical multiple-point connection for the MAX3040–MAX3045 with the MAX3095. Because of the high frequencies and the distances involved, high attention must be paid to transmission-line effects while using termination resistors. A terminating resistor (RT) is simply a resistor that should be placed at the extreme ends of the cable to match the characteristic impedance of the cable. When the termination resistance is not the same value as the characteristic impedance of the cable, reflections will occur as the signal is traveling down the cable. Although some reflections are inevitable due to the cable and resistor tolerances, large mismatches can cause significant reflections resulting in errors in the data. With this in mind, it is very important to match the terminating resistance and the characteristic impedance as closely as possible. As a general rule in a multi-drop system, termination resistors should always be placed at both ends of the cable.

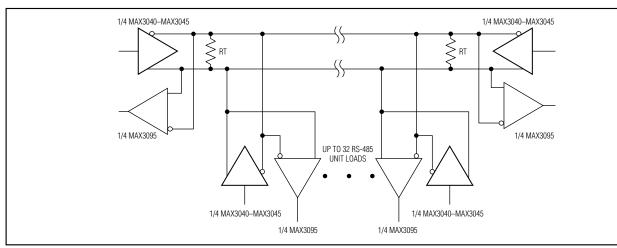


Figure 12. Typical Connection for Multiple-Point RS-485 Bus

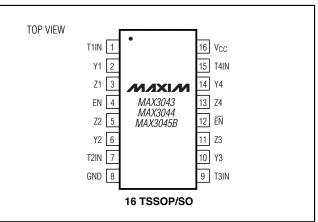
PART	TEMP RANGE	PIN-PACKAGE	DATA RATE
MAX3041CUE	0°C to +70°C	16 TSSOP	2.5Mbps
MAX3041CSE	0°C to +70°C	16 Narrow SO	2.5Mbps
MAX3041CWE	0°C to +70°C	16 Wide SO	2.5Mbps
MAX3041EUE	-40°C to +85°C	16 TSSOP	2.5Mbps
MAX3041ESE	-40°C to +85°C	16 Narrow SO	2.5Mbps
MAX3041EWE	-40°C to +85°C	16 Wide SO	2.5Mbps
MAX3042BCUE	0℃ to +70°C	16 TSSOP	20Mbps
MAX3042BCSE	0℃ to +70°C	16 Narrow SO	20Mbps
MAX3042BCWE	0℃ to +70°C	16 Wide SO	20Mbps
MAX3042BEUE	-40°C to +85°C	16 TSSOP	20Mbps
MAX3042BESE	-40°C to +85°C	16 Narrow SO	20Mbps
MAX3042BEWE	-40°C to +85°C	16 Wide SO	20Mbps
MAX3043CUE	0°C to +70°C	16 TSSOP	250kbps
MAX3043CSE	0°C to +70°C	16 Narrow SO	250kbps
MAX3043EWE	0°C to +70°C	16 Wide SO	250kbps
MAX3043EUE	-40°C to +85°C	16 TSSOP	250kbps
MAX3043ESE	-40°C to +85°C	16 Narrow SO	250kbps
MAX3043EWE	-40°C to +85°C	16 Wide SO	250kbps
MAX3044CUE	0°C to +70°C	16 TSSOP	2.5Mbps
MAX3044CSE	0°C to +70°C	16 Narrow SO	2.5Mbps
MAX3044CWE	0°C to +70°C	16 Wide SO	2.5Mbps
MAX3044EUE	-40°C to +85°C	16 TSSOP	2.5Mbps
MAX3044ESE	-40°C to +85°C	16 Narrow SO	2.5Mbps
MAX3044EWE	-40°C to +85°C	16 Wide SO	2.5Mbps

M X X M

## Ordering Information (continued)

PART	TEMP RANGE	PIN-PACKAGE	DATA RATE
MAX3045BCUE	0°C to +70°C	16 TSSOP	20Mbps
MAX3045BCSE	0°C to +70°C	16 Narrow SO	20Mbps
MAX3045BCWE	0°C to +70°C	16 Wide SO	20Mbps
MAX3045BEUE	-40°C to +85°C	16 TSSOP	20Mbps
MAX3045BESE	-40°C to +85°C	16 Narrow SO	20Mbps
MAX3045BEWE	-40°C to +85°C	16 Wide SO	20Mbps

# Pin Configurations (continued)



## \_Chip Information

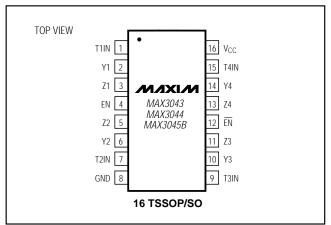
TRANSISTOR COUNT: 545 PROCESS: CMOS

PART	TEMP. RANGE	PIN-PACKAGE	DATA RATE
MAX3041CUE	0°C to +70°C	16 TSSOP	2.5Mbps
MAX3041CSE	0°C to +70°C	16 Narrow SO	2.5Mbps
MAX3041CWE	0°C to +70°C	16 Wide SO	2.5Mbps
MAX3041EUE	-40°C to +85°C	16 TSSOP	2.5Mbps
MAX3041ESE	-40°C to +85°C	16 Narrow SO	2.5Mbps
MAX3041EWE	-40°C to +85°C	16 Wide SO	2.5Mbps
MAX3042BCUE	0°C to +70°C	16 TSSOP	20Mbps
MAX3042BCSE	0°C to +70°C	16 Narrow SO	20Mbps
MAX3042BCWE	0°C to +70°C	16 Wide SO	20Mbps
MAX3042BEUE	-40°C to +85°C	16 TSSOP	20Mbps
MAX3042BESE	-40°C to +85°C	16 Narrow SO	20Mbps
MAX3042BEWE	-40°C to +85°C	16 Wide SO	20Mbps
MAX3043CUE	0°C to +70°C	16 TSSOP	250kbps
MAX3043CSE	0°C to +70°C	16 Narrow SO	250kbps
MAX3043EWE	0°C to +70°C	16 Wide SO	250kbps
MAX3043EUE	-40°C to +85°C	16 TSSOP	250kbps
MAX3043ESE	-40°C to +85°C	16 Narrow SO	250kbps
MAX3043EWE	-40°C to +85°C	16 Wide SO	250kbps

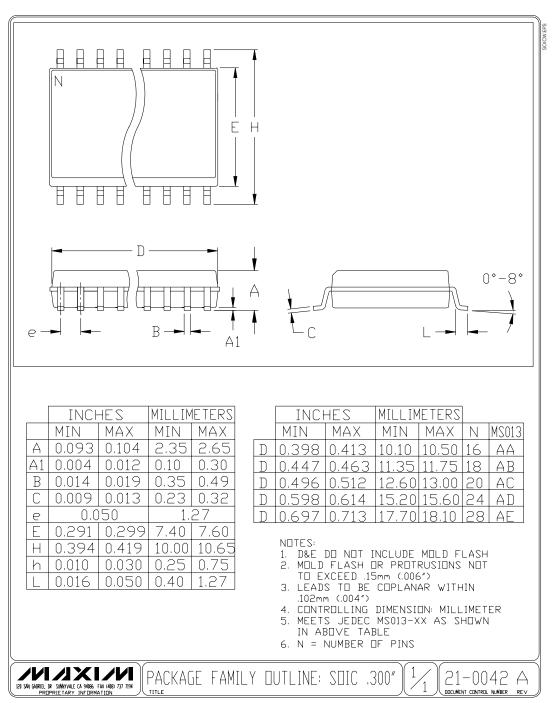
### Ordering Information (continued)

PART	TEMP. RANGE	PIN-PACKAGE	DATA RATE
MAX3044CUE	$0^{\circ}C$ to $+70^{\circ}C$	16 TSSOP	2.5Mbps
MAX3044CSE	0°C to +70°C	16 Narrow SO	2.5Mbps
MAX3044CWE	0°C to +70°C	16 Wide SO	2.5Mbps
MAX3044EUE	-40°C to +85°C	16 TSSOP	2.5Mbps
MAX3044ESE	-40°C to +85°C	16 Narrow SO	2.5Mbps
MAX3044EWE	-40°C to +85°C	16 Wide SO	2.5Mbps
MAX3045BCUE	0°C to +70°C	16 TSSOP	20Mbps
MAX3045BCSE	0°C to +70°C	16 Narrow SO	20Mbps
MAX3045BCWE	0°C to +70°C	16 Wide SO	20Mbps
MAX3045BEUE	-40°C to +85°C	16 TSSOP	20Mbps
MAX3045BESE	-40°C to +85°C	16 Narrow SO	20Mbps
MAX3045BEWE	-40°C to +85°C	16 Wide SO	20Mbps

## \_Pin Configurations (continued)



### **Package Information (continued)**



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