# **Absolute Maximum Ratings (Note 1)**

Supply Voltage (V <sub>V+</sub> – V <sub>V</sub> _)	
Differentail Input Voltage ( V <sub>IN+</sub> – V <sub>IN-</sub>  )	8V, <b>Note 4</b>
Input Common-Mode Range (V <sub>IN+</sub> , V <sub>IN-</sub> )	$V_{V+}$ to $V_{V-}$
Lead Temperature (soldering, 5 sec.)	260°C
Storage Temperature (T <sub>S</sub> )	150°C
ESD Rating, Note 3	1.5kV

# **Operating Ratings (Note 2)**

Supply Voltage (V <sub>S</sub> )	±2.5V to ±9V
Junction Temperature (T <sub>J</sub> )	40°C to +85°C
Package Thermal Resistance	260°C/W

# Electrical Characteristics (±5V)

 $\underline{V_{V+} = +5V, \, V_{V-} = -5V, \, V_{CM} = 0V, \, V_{OUT} = 0V; \, R_L = 10M\Omega; \, T_J = 25^{\circ}C, \, \text{bold} \, \, \text{values indicate} \, -40^{\circ}C \leq T_J \leq +85^{\circ}C; \, \text{unless noted}.}$ 

Symbol	Parameter	Condition	Min	Тур	Max	Units
V <sub>os</sub>	Input Offset Voltage			1	15	mV
V <sub>OS</sub>	Input Offset Voltage Temperature Coefficient			4		μV/°C
I <sub>B</sub>	Input Bias Current			3.5	5.5 <b>9</b>	μA μA
I <sub>os</sub>	Input Offset Current			0.05	3	μА
V <sub>CM</sub>	Input Common-Mode Range	CMRR > 60dB	-3.25		+3.25	V
CMRR	Common-Mode Rejection Ratio	-2.5V < V <sub>CM</sub> < +2.5V	70 <b>60</b>	90		dB dB
PSRR	Power Supply Rejection Ratio	±5V < V <sub>S</sub> < ±9V	74 <b>70</b>	81		dB dB
A <sub>VOL</sub>	Large-Signal Voltage Gain	$R_L = 2k$ , $V_{OUT} = \pm 2V$	60	71		dB
		$R_L = 200\Omega$ , $V_{OUT} = \pm 2V$	60	71		dB
V <sub>OUT</sub>	Maximum Output Voltage Swing	positive, $R_L = 2k\Omega$	+3.3 +3.0	3.5		V
		negative, $R_L = 2k\Omega$		-3.5	−3.3 <b>−3.0</b>	V
		positive, $R_L = 200\Omega$	+3.0 +2.75	3.2		V
		negative, $R_L = 200\Omega$		-2.8	-2.45 <b>-2.2</b>	V
GBW	Gain-Bandwidth Product	$R_L = 1k\Omega$		125		MHz
BW	-3dB Bandwidth	$A_V = 1, R_L = 100\Omega$		192		MHz
SR	Slew Rate			230		V/µs
I <sub>GND</sub>	Short-Circuit Output Current	source		72		mA
		sink		25		mA
I <sub>GND</sub>	Supply Current			2.4	3.5 <b>4.1</b>	mA mA

## **Electrical Characteristics**

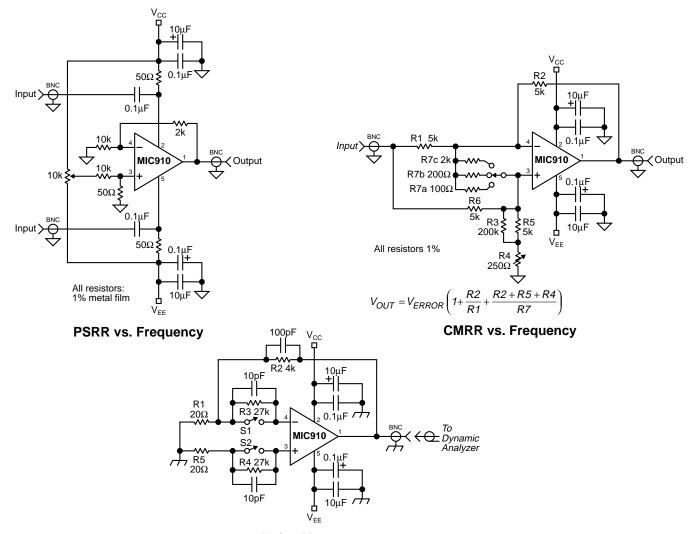
 $V_{V+} = +9V, \ V_{V-} = -9V, \ V_{CM} = 0V, \ V_{OUT} = 0V; \ R_L = 10M\Omega; \ T_J = 25^{\circ}C, \ \textbf{bold} \ values \ indicate} \ -40^{\circ}C \leq T_J \leq +85^{\circ}C; \ unless \ noted$ 

Symbol	Parameter	Condition	Min	Тур	Max	Units
V <sub>OS</sub>	Input Offset Voltage			1	15	mV
V <sub>OS</sub>	Input Offset Voltage Temperature Coefficient			4		μV/°C
I <sub>B</sub>	Input Bias Current			3.5	5.5 <b>9</b>	μA μA
I <sub>os</sub>	Input Offset Current			0.05	3	μΑ

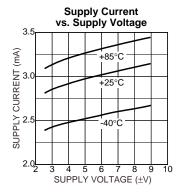
Symbol	Parameter	Condition	Min	Тур	Max	Units
$V_{CM}$	Input Common-Mode Range	CMRR > 60dB	-7.25		+7.25	V
CMRR	Common-Mode Rejection Ratio	-6.5V < V <sub>CM</sub> < 6.5V	70 <b>60</b>	98		dB dB
$\overline{A_{VOL}}$	Large-Signal Voltage Gain	$R_L = 2k\Omega, V_{OUT} = \pm 6V$	60	73		dB
V <sub>OUT</sub>	Maximum Output Voltage Swing	positive, $R_L = 2k\Omega$	+7.2 +6.8	+7.4		V V
		negative, $R_L = 2k\Omega$		-7.4	-7.2 - <b>6.8</b>	V V
GBW	Gain-Bandwidth Product	$R_L = 1k\Omega$		135		MHz
SR	Slew Rate			270		V/μs
I <sub>GND</sub>	Short-Circuit Output Current	source		90		mA
		sink		32		mA
I <sub>GND</sub>	Supply Current			2.5	3.7 <b>4.3</b>	mA mA

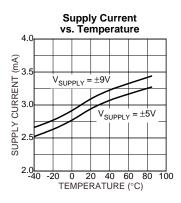
- Note 1. Exceeding the absolute maximum rating may damage the device.
- Note 2. The device is not guaranteed to function outside its operating rating.
- Note 3. Devices are ESD sensitive. Handling precautions recommended. Human body model, 1.5k in series with 100pF.
- Note 4. Exceeding the maximum differential input voltage will damage the input stage and degrade performance (in particular, input bias current is likely to increase.

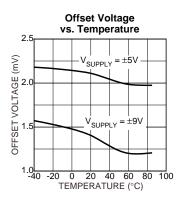
## **Test Circuits**

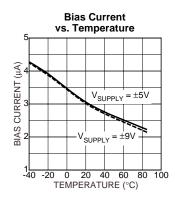


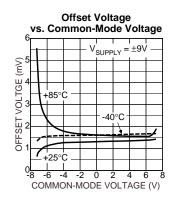
### **Electrical Characteristics**

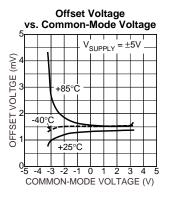


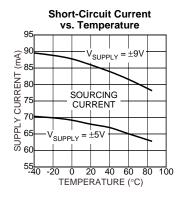


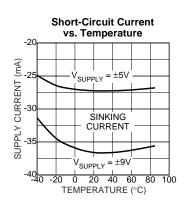


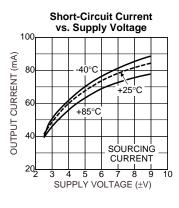


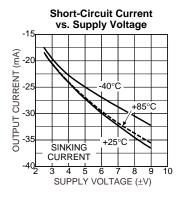


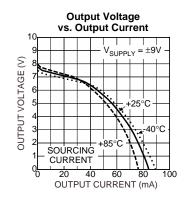


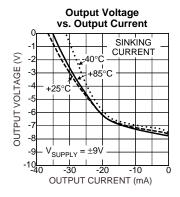


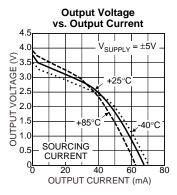


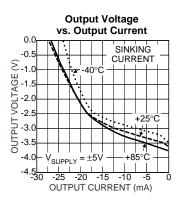


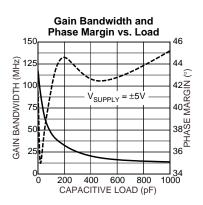


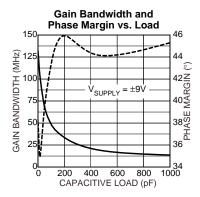


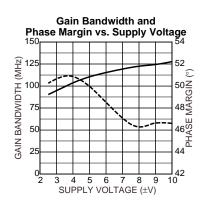


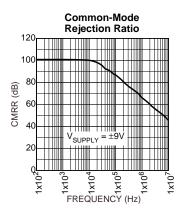




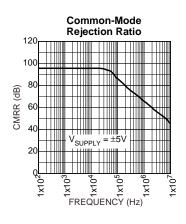


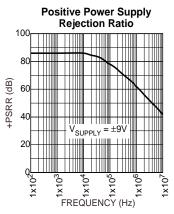


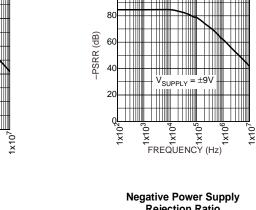




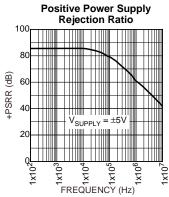
Negative Power Supply Rejection Ratio

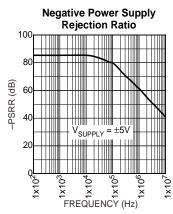


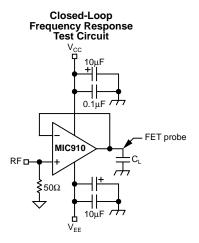


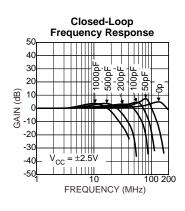


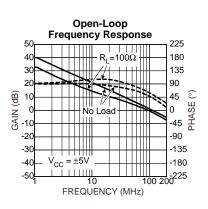
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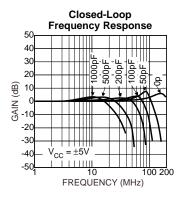


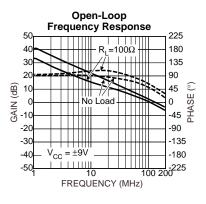


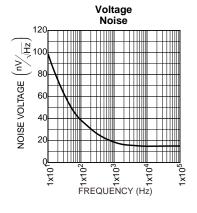


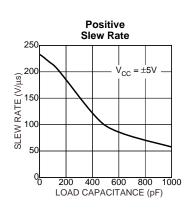


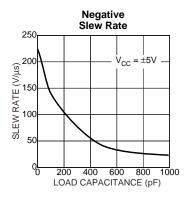


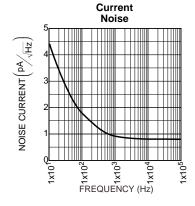


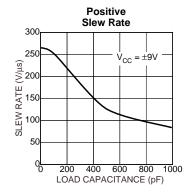


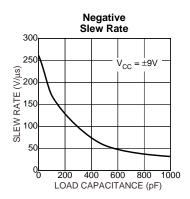


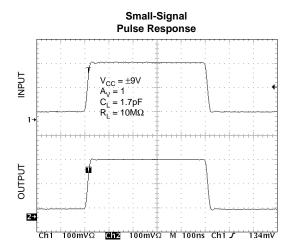


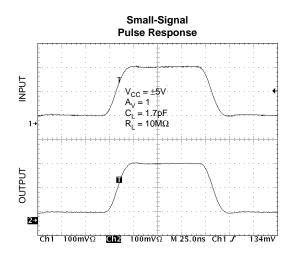


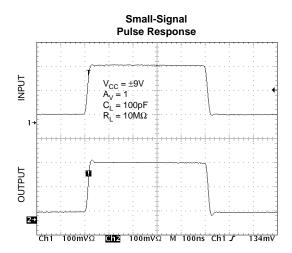


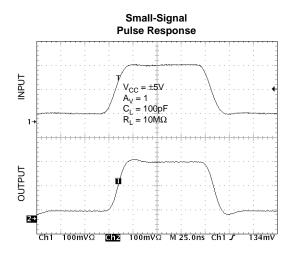


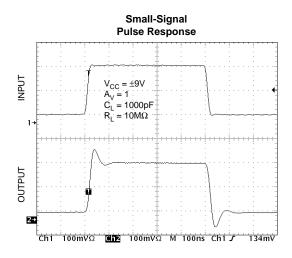


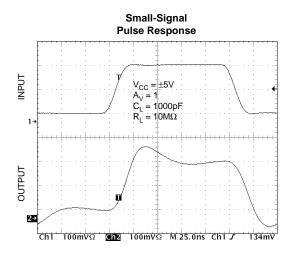




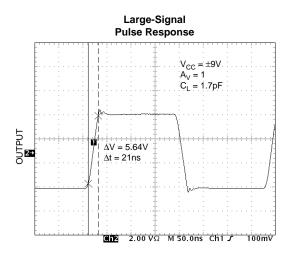


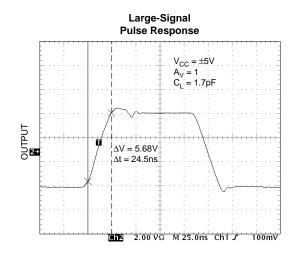


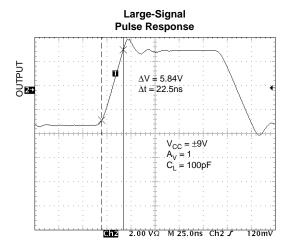


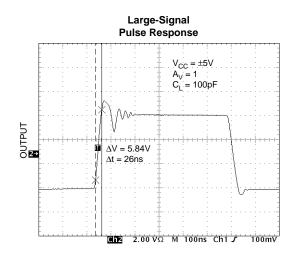


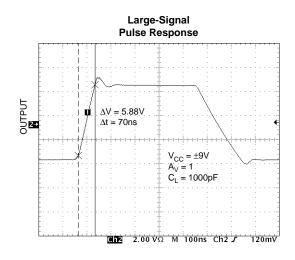
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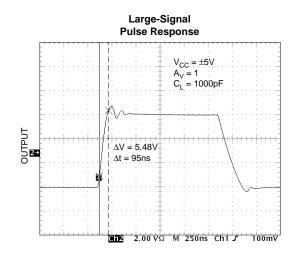












## **Applications Information**

The MIC910 is a high-speed, voltage-feedback operational amplifier featuring very low supply current and excellent stability. This device is unity gain stable and capable of driving high capacitance loads.

### **Driving High Capacitance**

The MIC910 is stable when driving any capacitance (see "Typical Characteristics: Gain Bandwidth and Phase Margin vs. Load Capacitance") making it ideal for driving long coaxial cables or other high-capacitance loads.

Phase margin remains constant as load capacitance is increased. Most high-speed op amps are only able to drive limited capacitance.

Note: increasing load capacitance does reduce the speed of the device (see "Typical Characteristics: Gain Bandwidth and Phase Margin vs. Load"). In applications where the load capacitance reduces the speed of the op amp to an unacceptable level, the effect of the load capacitance can be reduced by adding a small resistor  $(<100\Omega)$  in series with the output.

#### **Feedback Resistor Selection**

Conventional op amp gain configurations and resistor selection apply, the MIC910 is NOT a current feedback device. Resistor values in the range of 1k to 10k are recommended.

### **Layout Considerations**

All high speed devices require careful PCB layout. The high stability and high PSRR of the MIC910 make this op amp easier to use than most, but the following guidelines should be observed: Capacitance, particularly on the two inputs pins will degrade performance; avoid large copper traces to the inputs. Keep the output signal away from the inputs and use a ground plane.

It is important to ensure adequate supply bypassing capacitors are located close to the device.

#### **Power Supply Bypassing**

Regular supply bypassing techniques are recommended. A  $10\mu F$  capacitor in parallel with a  $0.1\mu F$  capacitor on both the positive and negative supplies are ideal. For best performance all bypassing capacitors should be located as close to the op amp as possible and all capacitors should be low ESL (equivalent series inductance), ESR (equivalent series resistance). Surface-mount ceramic capacitors are ideal.

#### **Thermal Considerations**

The SOT-23-5 package, like all small packages, has a high thermal resistance. It is important to ensure the IC does not exceed the maximum operating junction (die) temperature of 85°C. The part can be operated up to the absolute maximum temperature rating of 125°C, but between 85°C and 125°C performance will degrade, in particular CMRR will reduce.

A MIC910 with no load, dissipates power equal to the quiescent supply current \* supply voltage

$$P_{D(noload)} = (V_{V+} - V_{V-})I_{S}$$

When a load is added, the additional power is dissipated in the output stage of the op amp. The power dissipated in the device is a function of supply voltage, output voltage and output current.

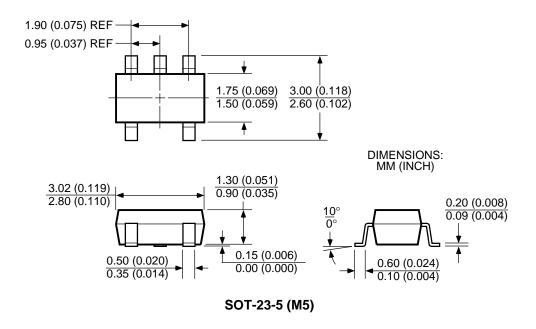
$$P_{D(output \, stage)} = (V_{V+} - V_{OUT})I_{OUT}$$

$$Total\ Power\ Dissipation = P_{D(no\ load)} + P_{D(output\ stage)}$$

Ensure the total power dissipated in the device is no greater than the thermal capacity of the package. The SOT23-5 package has a thermal resistance of 260°C/W.

Max. Allowable Power Dissipation = 
$$\frac{T_{J(max)} - T_{A(max)}}{260W}$$

# **Package Information**



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