LT1012A/LT1012

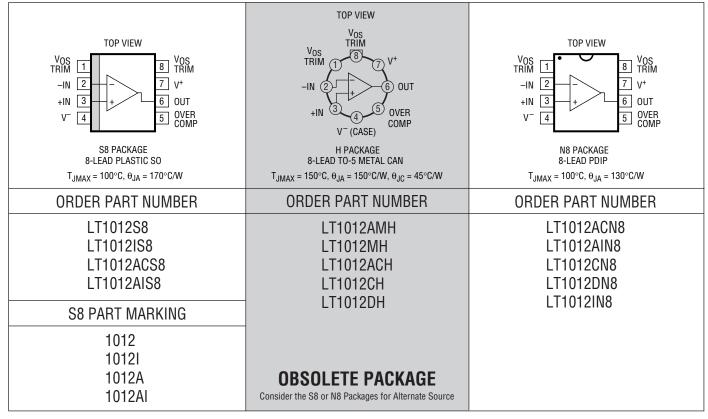
ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage	±20V
Differential Input Current (Note 1)	
Input Voltage	
Output Short Circuit Duration	

Operating Temperature Range

LT1012AM/LT1012M (OBSOLETE) 55°C to	125°C
LT1012I/LT1012AI 40°C to) 85°C
LT1012AC/LT1012C	
LT1012D/LT1012S80°C to) 70°C
Storage Temperature Range – 65°C to	150°C
Lead Temperature (Soldering, 10 sec)	300°C

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.





ELECTRICAL CHARACTERISTICS $V_{S} = \pm 15V$, $V_{CM} = 0V$, $T_{A} = 25^{\circ}C$, unless otherwise noted.

			LT1012AM/AC/AI			Ľ	T1012M	/I	I			
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	ТҮР	MAX	MIN	ТҮР	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 3)		8 20	25 90		8 20	35 90		10 25	50 120	μV μV
	Long Term Input Offset Voltage Stability			0.3			0.3			0.3		μV/month
I _{OS}	Input Offset Current	(Note 3)		15 25	100 150		15 25	100 150		20 30	150 200	pA pA
I _B	Input Bias Current	(Note 3)		±25 ±35	±100 ±150		±25 ±35	±100 ±150		±30 ±40	±150 ±200	pA pA
e _n	Input Noise Voltage	0.1Hz to 10Hz		0.5			0.5			0.5		μV _{P-P}
e _n	Input Noise Voltage Density	f ₀ = 10Hz (Note 4) f ₀ = 1000Hz (Note 5)		17 14	30 22		17 14	30 22		17 14	30 22	nV√Hz nV√Hz
i _n	Input Noise Current Density	f _{0 =} 10Hz		20			20			20		fA/√Hz
A _{VOL}	Large Signal Voltage Gain	$\label{eq:VOUT} \begin{split} V_{OUT} &= \pm 12 V, \ R_L \geq 10 k \Omega \\ V_{OUT} &= \pm 10 V, \ R_L \geq 2 k \Omega \end{split}$	300 300	2000 1000		300 200	2000 1000		200 200	2000 1000		V/mV V/mV
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 13.5V$	114	132		114	132		110	132		dB
PSRR	Power Supply Rejection Ratio	$V_{S} = \pm 1.2V$ to $\pm 20V$	114	132		114	132		110	132		dB
	Input Voltage Range		±13.5	±14		±13.5	±14		±13.5	±14		V
V _{OUT}	Output Voltage Swing	$R_L = 10k\Omega$	±13	±14		±13	±14		±13	±14		V
	Slew Rate		0.1	0.2		0.1	0.2		0.1	0.2		V/µs
I _S	Supply Current	(Note 3)		370 380	500 600		380 380	600		380 380	600	μΑ μΑ



LT1012A/LT1012

ELECTRICAL CHARACTERISTICS $V_{S} = \pm 15V, V_{CM} = 0V, T_{A}$

 V_S = $\pm\,15V,~V_{CM}$ = 0V, T_A = 25°C, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	LT1012D TYP	MAX	MIN	LT1012S8 TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 3)		12 25	60		15 25	120 180	μV μV
	Long Term Input Offset Voltage Stability			0.3			0.4		μV/month
I _{OS}	Input Offset Current	(Note 3)		20 30	150		50 60	280 380	pA pA
I _B	Input Bias Current	(Note 3)		±30 ±40	±150		±80 ±120	±300 ±400	pA pA
e _n	Input Noise Voltage	0.1Hz to 10Hz		0.5			0.5		μV _{P-P}
e _n	Input Noise Voltage Density	f ₀ = 10Hz (Note 5) f ₀ = 1000Hz (Note 5)		17 14	30 22		17 14	30 22	nV√Hz nV√Hz
i _n	Input Noise Current Density	f ₀ = 10Hz		20			20		fA/√Hz
A _{VOL}	Large-Signal Voltage Gain	$ \begin{array}{l} V_{0UT} = \pm 12 V, R_L \geq 10 k \Omega \\ V_{0UT} = \pm 10 V, R_L \geq 2 k \Omega \end{array} $	200 200	2000 1000		200 120	2000 1000		V/mV V/mV
CMRR	Common Mode Rejection Ratio	V _{CM} = ±13.5V	110	132		110	132		dB
PSRR	Power Supply Rejection Ratio	$V_{S} = \pm 1.2 V$ to $\pm 20 V$	110	132		110	132		dB
	Input Voltage Range		±13.5	±14.0		±13.5	±14.0		V
V _{OUT}	Output Voltage Swing	$R_L = 10k\Omega$	±13	±14		±13	±14		V
	Slew Rate		0.1	0.2		0.1	0.2		V/µs
I _S	Supply Current	(Note 3)		380	600		380	600	μA





ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range of $-55^{\circ}C \le T_A \le 125^{\circ}C$ for LT1012AM and LT1012M, and $-40^{\circ}C \le T_A \le 85^{\circ}C$ for LT1012AI and LT1012I. $V_S = \pm 15V$, $V_{CM} = 0V$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		L' Min	Г1012AM/ Түр	AI Max	MIN	LT1012M/I TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 3)	•		30 40	60 180		30 40	180 250	μV μV
	Average Temperature Coefficient of Input Offset Voltage		•		0.2	0.6		0.2	1.5	μV/°C
I _{OS}	Input Offset Current	(Note 3)	•		30 70	250 350		30 70	250 350	pA pA
	Average Temperature Coefficient of Input Offset Current		•		0.3	2.5		0.3	2.5	pA/°C
IB	Input Bias Current	(Note 3)	•		±80 ±150	±600 ±800		±80 ±150	±600 ±800	pA pA
	Average Temperature Coefficient of Input Bias Current		•		0.6	6.0		0.6	6.0	pA/°C
A _{VOL}	Large-Signal Voltage Gain	$\label{eq:V_OUT} \begin{split} V_{0UT} = \pm 12 V, \ R_L \geq 10 k \Omega \\ V_{0UT} = \pm 10 V, \ R_L \geq 2 k \Omega \end{split}$	•	200 200	1000 600		150 100	1000 600		V/mV V/mV
CMRR	Common Mode Rejection Ratio	V _{CM} = ±13.5V	•	110	128		108	128		dB
PSRR	Power Supply Rejection Ratio	$V_{\rm S}$ = ±1.5V to ±20V	•	110	126		108	126		dB
	Input Voltage Range		•	±13.5			±13.5			V
V _{OUT}	Output Voltage Swing	$R_L = 10k\Omega$	•	±13	±14		±13	±14		V
I _S	Supply Current		•		400	650		400	800	μA



ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range of 0°C \leq T_A \leq 70°C. V_S = \pm 15V, V_{CM} = 0V, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	LT1012AC Typ	МАХ	MIN	LT1012C Typ	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 3)	•		20 30	60 160		20 30	100 200	μV μV
	Average Temperature Coefficient of Input Offset Voltage		•		0.2	0.6		0.2	1.0	μV/°C
I _{OS}	Input Offset Current	(Note 3)	•		25 40	230 300		35 45	230 300	pA pA
	Average Temperature Coefficient of Input Offset Current		•		0.3	2.5		0.3	2.5	pA/°C
I _B	Input Bias Current	(Note 3)	•		±35 ±50	±230 ±300		±35 ±50	±230 ±300	pA pA
	Average Temperature Coefficient of Input Bias Current		•		0.3	2.5		0.3	2.5	pA/°C
A _{VOL}	Large-Signal Voltage Gain	$\label{eq:V_OUT} \begin{split} V_{0UT} = \pm 12 V, \ R_L \geq 10 k \Omega \\ V_{0UT} = \pm 10 V, \ R_L \geq 2 k \Omega \end{split}$	•	200 200	1500 1000		150 150	1500 800		V/mV V/mV
CMRR	Common Mode Rejection Ratio	V _{CM} = 13.5V	•	110	130		108	130		dB
PSRR	Power Supply Rejection Ratio	$V_{\rm S} = \pm 1.3 \text{V} \text{ to } \pm 20 \text{V}$	•	110	128		108	128		dB
	Input Voltage Range		•	±13.5			±13.5			V
V _{OUT}	Output Voltage Swing	$R_L = 10k\Omega$	•	±13	±14		±13	±14		V
I _S	Supply Current		•		400	600		400	800	μA





ELECTRICAL CHARACTERISTICS The • denotes the specifications which apply over the full operating

temperature range of $0^{\circ}C \le T_A \le 70^{\circ}C$. $V_S = \pm 15V$, $V_{CM} = 0V$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	LT1012D Typ	MAX	MIN	LT1012S8 TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 3)	•		25 40	140		30 45	200 270	μV μV
	Average Temperature Coefficient of Input Offset Voltage		•		0.3	1.7		0.3	1.8	μV/°C
I _{OS}	Input Offset Current	(Note 3)	•		35 45	380		60 80	380 500	pA pA
	Average Temperature Coefficient of Input Offset Current		•		0.35	4.0		0.4	4.0	pA/°C
I _B	Input Bias Current	(Note 3)	•		±50 ±65	±420		±100 ±150	±420 ±550	pA pA
	Average Temperature Coefficient of Input Bias Current		•		0.4	5.0		0.5	5.0	pA/°C
A _{VOL}	Large-Signal Voltage Gain	$ \begin{array}{l} V_{0UT}=\pm12V,R_L\geq10k\Omega\\ V_{0UT}=\pm10V,R_L\geq2k\Omega \end{array} $	•	150 150	1500 800		150 100	1500 800		V/mV V/mV
CMRR	Common Mode Rejection Ratio	V _{CM} = ±13.5V	•	108	130		108	130		dB
PSRR	Power Supply Rejection Ratio	$V_{\rm S}$ = ±1.3V to ±20V	•	108	128		108	128		dB
	Input Voltage Range		•	±13.5			±13.5			V
V _{OUT}	Output Voltage Swing	$R_L = 10k\Omega$	•	±13	±14		±13	±14		V
I _S	Supply Current		•		400	800		400	800	μA

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: Differential input voltages greater than 1V will cause excessive current to flow through the input protection diodes unless limiting resistance is used.

Note 3: These specifications apply for V_{MIN} \leq V_S \leq \pm 20V and $-13.5V \leq$ V_{CM} \leq 13.5V (for V_S = \pm 15V). V_{MIN} = \pm 1.2V at 25°C, \pm 1.3V from

 $-13.5V \le V_{CM} \le 13.5V$ (for V_S = ±15V). V_{MIN} = ±1.2V at 25°C, ±1.3V from 0°C to 70°C, ±1.5V from -55°C to 125°C.

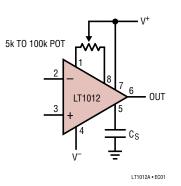
Note 4: 10Hz noise voltage density is sample tested on every lot. Devices 100% tested at 10Hz are available on request.

Note 5: This parameter is tested on a sample basis only.

Optional Offset Nulling and Overcompensation Circuits

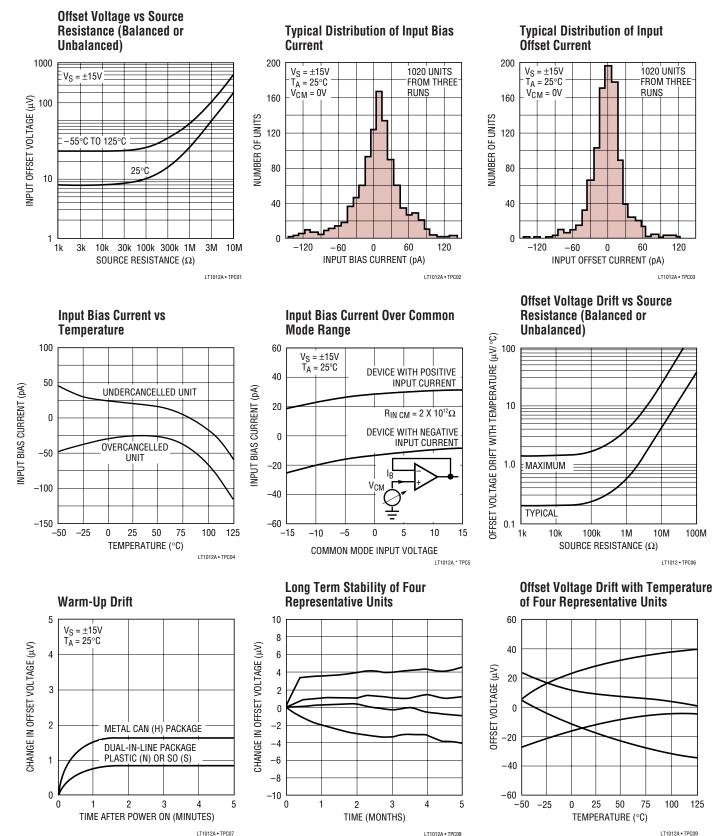
Input offset voltage can be adjusted over a $\pm 800 \mu V$ range with a 5k to 100k potentiometer.

The LT1012 is internally compensated for unity gain stability. The overcompensation capacitor, C_S , can be used to improve capacitive load handling capability, to narrow noise bandwidth, or to stabilize circuits with gain in the feedback loop.

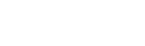




TYPICAL PERFORMANCE CHARACTERISTICS



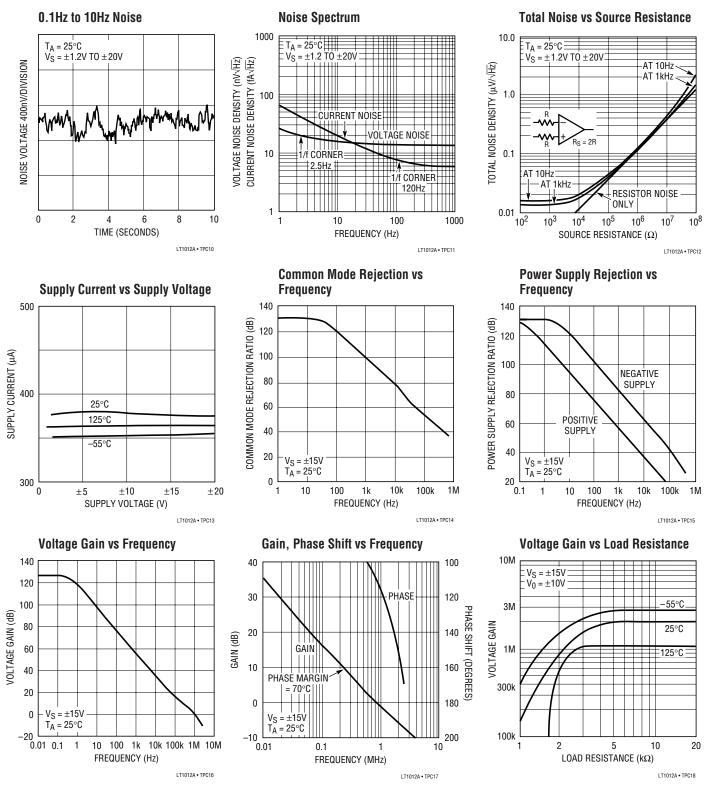
LT1012A • TPC08





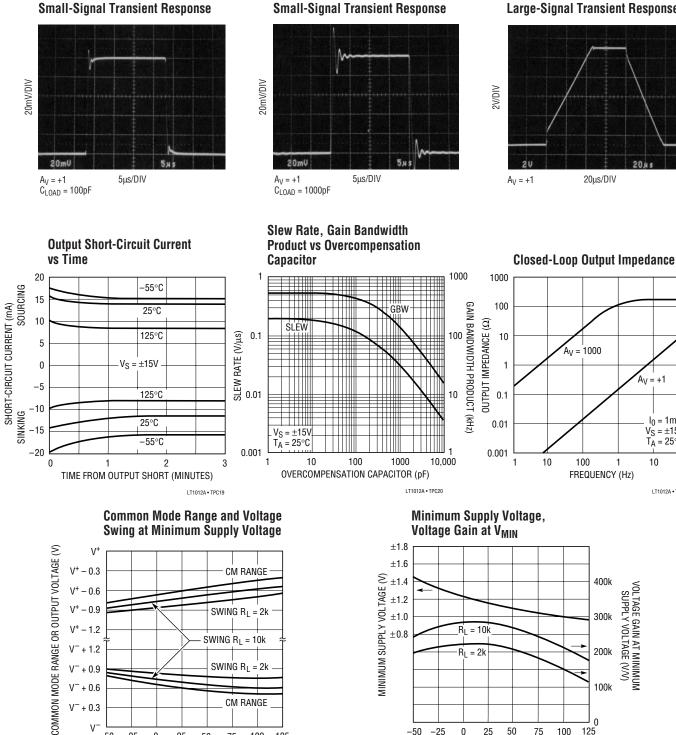
LT1012A • TPC09

TYPICAL PERFORMANCE CHARACTERISTICS





TYPICAL PERFORMANCE CHARACTERISTICS



100 125

LT1012A • TPC22

50 75

TEMPERATURE (°C)

Large-Signal Transient Response

T_A = 25°C 1 10 FREQUENCY (Hz) LT1012A • TPC21 400k

 $A_V = +1$

 $I_0 = 1mA$

 $V_{\rm S} = \pm 15 V$

100

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V.

-50 -25 0 25



sn1012 1012afbs

VOLTAGE GAIN AT MINIMUM SUPPLY VOLTAGE (V/V)

100k

125

LT1012A • TPC23

75 100

25 50

TEMPERATURE (°C)

-50 -25 0

APPLICATIONS INFORMATION

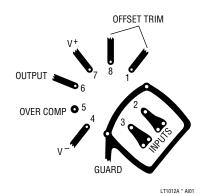
The LT1012 may be inserted directly into OP-07, LM11, 108A or 101A sockets with or without removal of external frequency compensation or nulling components. The LT1012 can also be used in 741, LF411, LF156 or OP-15 applications provided that the nulling circuitry is removed.

Although the OP-97 is a copy of the LT1012, the LT1012 directly replaces and upgrades OP-97 applications. The LT1012C and D have lower offset voltage and drift than the OP-97F. The LT1012A has lower supply current than the OP-97A/E. In addition, all LT1012 grades guarantee operation at \pm 1.2V supplies.

Achieving Picoampere/Microvolt Performance

In order to realize the picoampere/microvolt level accuracy of the LT1012, proper care must be exercised. For example, leakage currents in circuitry external to the op amp can significantly degrade performance. High quality insulation should be used (e.g. Teflon, KeI-F); cleaning of all insulating surfaces to remove fluxes and other residues will probably be required. Surface coating may be necessary to provide a moisture barrier in high humidity environments.

Board leakage can be minimized by encircling the input circuitry with a guard ring operated at a potential close to that of the inputs: in inverting configurations the guard ring should be tied to ground, in non-inverting connections to the inverting input at Pin 2. Guarding both sides of the printed circuit board is required. Bulk leakage reduction depends on the guard ring width. Nanoampere level leakage into the offset trim terminals can affect offset voltage and drift with temperature.



Microvolt level error voltages can also be generated in the external circuitry. Thermocouple effects caused by temperature gradients across dissimilar metals at the contacts to the input terminals can exceed the inherent drift of the amplifier. Air currents over device leads should be minimized, package leads should be short, and the two input leads should be as close together as possible and maintained at the same temperature.

Noise Testing

For application information on noise testing and calculations, please see the LT1008 data sheet.

Frequency Compensation

The LT1012 can be overcompensated to improve capacitive load handling capability or to narrow noise bandwidth. In many applications, the feedback loop around the amplifier has gain (e.g. logarithmic amplifiers); overcompensation can stabilize these circuits with a single capacitor.

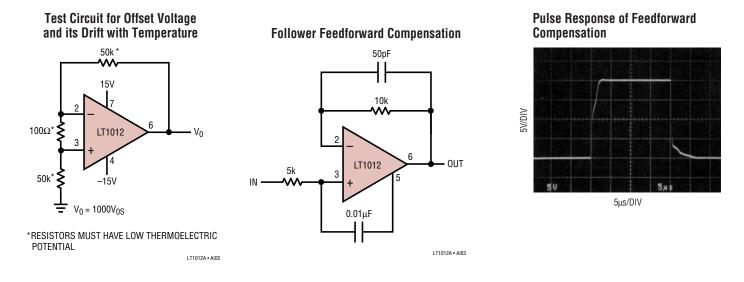
The availability of the compensation terminal permits the use of feedforward frequency compensation to enhance slew rate. The voltage follower feedforward scheme bypasses the amplifier's gain stages and slews at nearly $10V/\mu s$.

The inputs of the LT1012 are protected with back-to-back diodes. Current limiting resistors are not used, because the leakage of these resistors would prevent the realization of picoampere level bias currents at elevated temperatures. In the voltage follower configuration, when the input is driven by a fast, large signal pulse (>1V), the input protection diodes effectively short the output to the input during slewing, and a current, limited only by the output short-circuit protection will flow through the diodes.

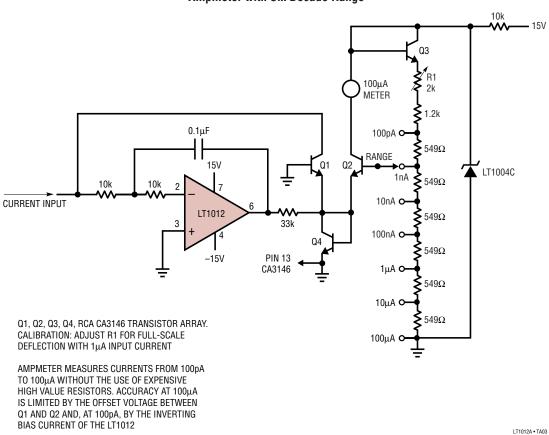
The use of a feedback resistor, as shown in the voltage follower feedforward diagram, is recommended because this resistor keeps the current below the short-circuit limit, resulting in faster recovery and settling of the output.



APPLICATIONS INFORMATION



TYPICAL APPLICATIONS

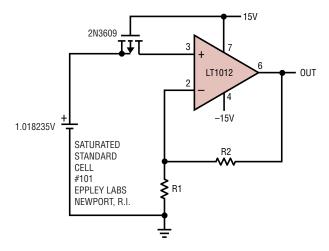


Ampmeter with Six Decade Range





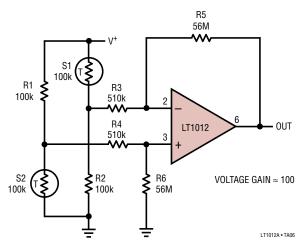
TYPICAL APPLICATIONS



Saturated Standard Cell Amplifier

THE TYPICAL 30pA BIAS CURRENT OF THE LT1012 WILL DEGRADE THE STANDARD CELL BY ONLY 1ppm/YEAR. NOISE IS A FRACTION OF A ppm. UNPROTECTED GATE MOSFET ISOLATES STANDARD CELL ON POWER DOWN

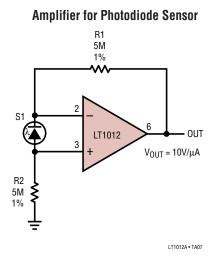
LT1012A • TA05



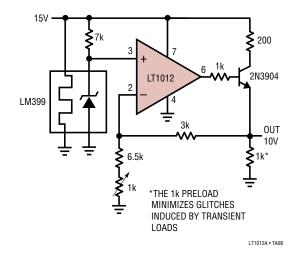
Amplifier for Bridge Transducers



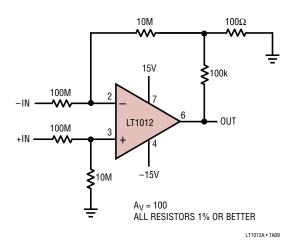
TYPICAL APPLICATIONS



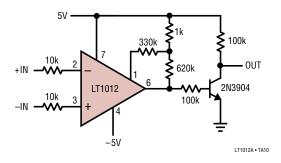
Buffered Reference for A-to-D Converters



Instrumentation Amplifier with $\pm 100V$ Common Mode Range



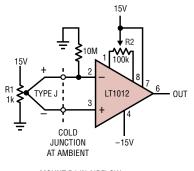
Low Power Comparator with <10µV Hysteresis



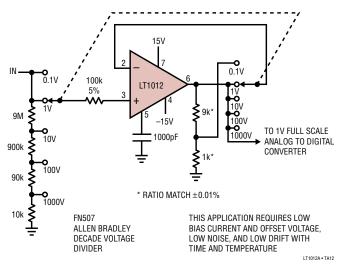


TYPICAL APPLICATIONS

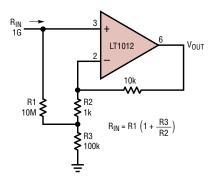




MOUNT R1 IN AIRFLOW. ADJUST R2 SO OUTPUT GOES HIGH WHEN AIRFLOW STOPS LT1012A • TA11 Input Amplifier for 4.5 Digit Voltmeter

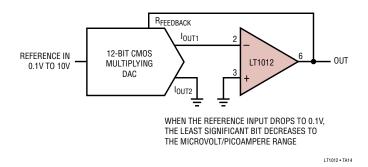


Resistor Multiplier



LT1012 • TA13

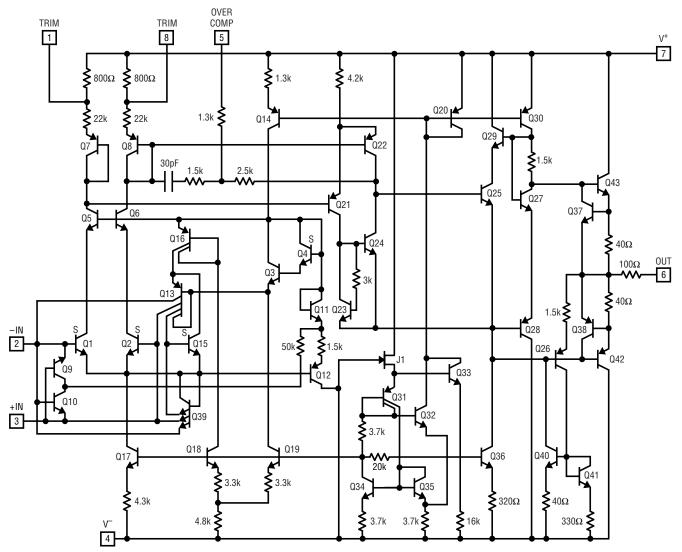
"No Trims" 12-Bit Multiplying DAC Output Amplifier



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LT1012A/LT1012

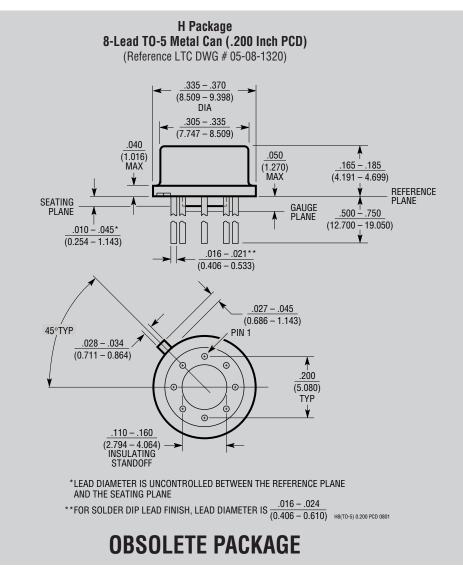
SCHEMATIC DIAGRAM



LT1012A • SD01

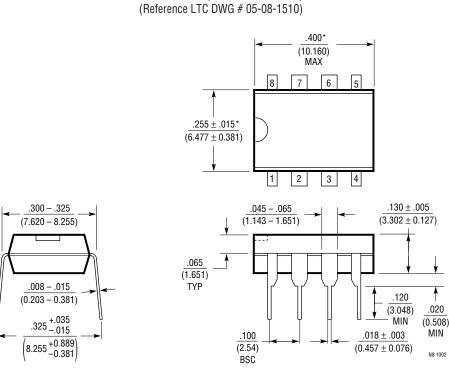


PACKAGE DESCRIPTION





PACKAGE DESCRIPTION



N8 Package 8-Lead PDIP (Narrow .300 Inch)

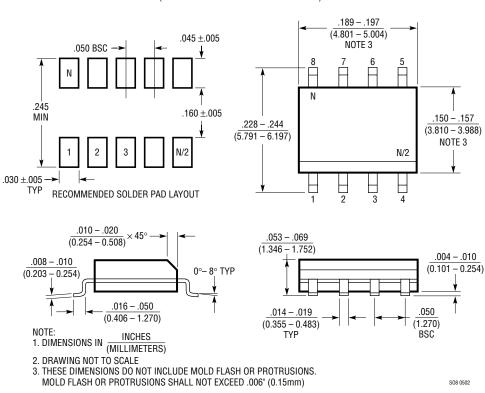
NOTE:

NOTE: 1. DIMENSIONS ARE <u>INCHES</u> *THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .010 INCH (0.254mm)





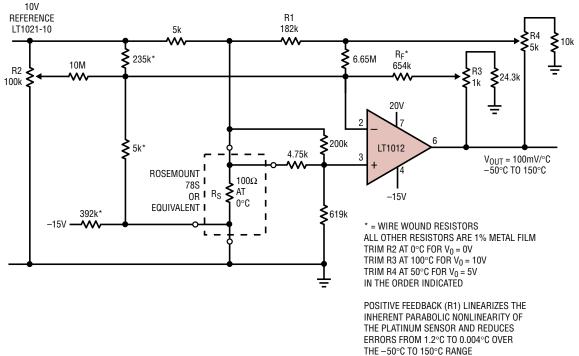
PACKAGE DESCRIPTION



S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)



TYPICAL APPLICATION



Kelvin-Sensed Platinum Temperature Sensor Amplifier

LT1012A • TA04



