

UC184xA/284xA/384xA

CURRENT MODE PWM CONTROLLER

PRODUCTION DATA SHEET

ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage (Low Impedance Source) (V_{CC})	30V
Supply Voltage ($I_{CC} < 30mA$)	Self Limiting
Output Current	$\pm 1A$
Output Energy (Capacitive Load)	5 μJ
Analog Inputs (V_{FB} & I_{SENSE})	-0.3V to +6.3V
Error Amp Output Sink Current	10mA
Power Dissipation at $T_A = 25^\circ C$ (M Package)	1W
Storage Temperature Range	-65°C to +150°C
Lead Temperature (Soldering, 10 Seconds)	300°C
Peak Package Solder Reflow Temp. (40 second max. exposure)	260°C (+0, -5)

Note 1. Exceeding these ratings could cause damage to the device. All voltages are with respect to Ground. Currents are positive into, negative out of the specified terminal. Pin numbers refer to DIL packages only.

THERMAL DATA

M PACKAGE:

THERMAL RESISTANCE-JUNCTION TO AMBIENT, θ_{JA}	95°C/W
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DM PACKAGE:

THERMAL RESISTANCE-JUNCTION TO AMBIENT, θ_{JA}	165°C/W
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D PACKAGE:

THERMAL RESISTANCE-JUNCTION TO AMBIENT, θ_{JA}	120°C/W
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Y PACKAGE:

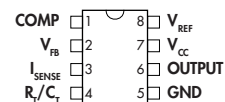
THERMAL RESISTANCE-JUNCTION TO AMBIENT, θ_{JA}	130°C/W
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Junction Temperature Calculation: $T_J = T_A + (P_D \times \theta_{JA})$.

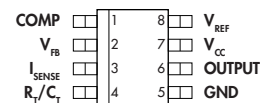
The θ_{JA} numbers are guidelines for the thermal performance of the device/pc-board system.

All of the above assume no ambient airflow

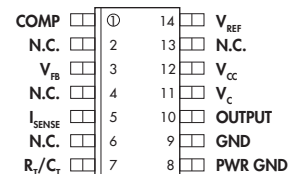
PACKAGE PIN OUTS



M & Y PACKAGE
(Top View)



DM PACKAGE
(Top View)



D PACKAGE
(Top View)

RoHS / Pb-free 100% Matte Tin Lead Finish

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ELECTRICAL CHARACTERISTICS

(Unless otherwise specified, these specifications apply over the operating ambient temperatures for UC384xA with $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$, UC284xA with $-40^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$, UC184xA with $-55^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$; $V_{CC}=15\text{V}$; $R_T=10\text{K}$; $C_T=3.3\text{nF}$. Low duty cycle pulse testing techniques are used which maintains junction and case temperatures equal to the ambient temperature.)

Parameter	Symbol	Test Conditions	UC184xA/284xA			UC384xA			Units
			Min.	Typ.	Max.	Min.	Typ.	Max.	
Reference Section									
Output Voltage	V _{REF}	T _J = 25°C, I _L = 1mA	4.95	5.00	5.05	4.90	5.00	5.10	V
Line Regulation		12 ≤ V _N ≤ 25V		6	20		6	20	mV
Load Regulation		1 ≤ I _O ≤ 20mA		6	25		6	25	mV
Temperature Stability (Note 2 & 7)				0.2	0.4		0.2	0.4	mV/°C
Total Output Variation		Over Line, Load, and Temperature	4.9		5.1	4.82		5.18	V
Output Noise Voltage (Note 2)	V _N	10Hz ≤ f ≤ 10kHz, T _J = 25°C		50			50		μV
Long Term Stability (Note 2)		T _A = 125°C, t = 1000hrs		5	25		5	25	mV
Output Short Circuit Current	I _{SC}		-30	-100	-180	-30	-100	-180	mA
Oscillator Section									
Initial Accuracy (Note 6)		T _J = 25°C	47	52	57	47	52	57	kHz
Voltage Stability		12 ≤ V _{CC} ≤ 25V		0.2	1		0.2	1	%
Temperature Stability (Note 2)		T _{MIN} ≤ T _A ≤ T _{MAX}		5			5		%
Amplitude (Note 2)				1.7			1.7		V
Discharge Current		T _J = 25°C, V _{PIN 4} = 2V	7.8	8.3	8.8	7.8	8.3	8.8	mA
		V _{PIN 4} = 2V, T _{MIN} ≤ T _A ≤ T _{MAX}	7.5		8.8	7.6		8.8	mA
Error Amp Section									
Input Voltage		V _{PIN 1} = 2.5V	2.45	2.50	2.55	2.42	2.50	2.58	V
Input Bias Current	I _B			-0.3	-1		-0.3	-2	μA
Open Loop Gain	A _{VOL}	2 ≤ V _O ≤ 4V	65	90		65	90		dB
Unity Gain Bandwidth (Note 2)	UGBW	T _J = 25°C	0.7	1		0.7	1		MHz
Power Supply Rejection Ratio (Note 3)	PSRR	12 ≤ V _{CC} ≤ 25V	60	70		60	70		dB
Output Sink Current	I _{OL}	V _{PIN 2} = 2.7V, V _{PIN 1} = 1.1V	2	6		2	6		mA
Output Source Current	I _{OH}	V _{PIN 2} = 2.3V, V _{PIN 1} = 5V	-0.5	-0.8		-0.5	-0.8		mA
Output Voltage High Level	V _{OH}	V _{PIN 2} = 2.3V, R _L = 15K to ground	5	6		5	6		V
Output Voltage Low Level	V _{OL}	V _{PIN 2} = 2.7V, R _L = 15K to V _{REF}		0.7	1.1		0.7	1.1	V
Current Sense Section									
Gain (Note 3 & 4)	A _{VOL}		2.85	3	3.15	2.85	3	3.15	V/V
Maximum Input Signal (Note 3)		V _{PIN 1} = 5V	0.9	1	1.1	0.9	1	1.1	V
Power Supply Rejection Ratio (Note 3)	PSRR	12 ≤ V _{CC} ≤ 25V		70			70		dB
Input Bias Current	I _B			-2	-10		-2	-10	μA
Delay to Output (Note 2)	T _{pd}	V _{PIN 3} = 0 to 2V		150	300		150	300	ns
Output Section									
Output Low Level	V _{OL}	I _{SINK} = 20mA		0.1	0.4		0.1	0.4	V
		I _{SINK} = 200mA		1.5	2.2		1.5	2.2	V
Output High Level	V _{OH}	I _{SOURCE} = 20mA	13	13.5		13	13.5		V
		I _{SOURCE} = 200mA	12	13.5		12	13.5		V
Rise Time (Note 2)	T _R	T _J = 25°C, C _L = 1nF		50	150		50	150	ns
Fall Time (Note 2)	T _F	T _J = 25°C, C _L = 1nF		50	150		50	150	ns
UVLO Saturation	V _{SAT}	V _{CC} = 5V, I _{SINK} = 10mA		0.7	1.2		0.7	1.2	V

(Electrical Characteristics continue next page.)

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ELECTRICAL CHARACTERISTICS (Con't.)

Parameter	Symbol	Test Conditions	UC184xA/284xA			UC384xA			Units
			Min.	Typ.	Max.	Min.	Typ.	Max.	
► Under-Voltage Lockout Section									
Start Threshold		x842A/4A	15	16	17	14.5	16	17.5	V
		x843A/5A	7.8	8.4	9.0	7.8	8.4	9.0	V
Min. Operation Voltage After Turn-On		x842A/4A	9	10	11	8.5	10	11.5	V
		x843A/5A	7.0	7.6	8.2	7.0	7.6	8.2	V
► PWM Section									
Maximum Duty Cycle		x842A/3A	94	96	100	94	96	100	%
		x844A/5A	47	48	50	47	48	50	%
Minimum Duty Cycle					0			0	%
► Total Standby Section									
Start-Up Current				0.3	0.5		0.3	0.5	mA
Operating Supply Current	I _{CC}			11	17		11	17	mA
Zener Voltage	V _Z	I _{CC} = 25mA	30	35		30	35		V

Notes: 2. These parameters, although guaranteed, are not 100% tested in production.

3. Parameter measured at trip point of latch with $V_{VFB} = 0$.

4. Gain defined as: $A_{VOL} = \frac{\Delta V_{COMP}}{\Delta V_{ISENSE}}$; $0 \leq V_{ISENSE} \leq 0.8V$.

5. Adjust V_{CC} above the start threshold before setting at 15V.

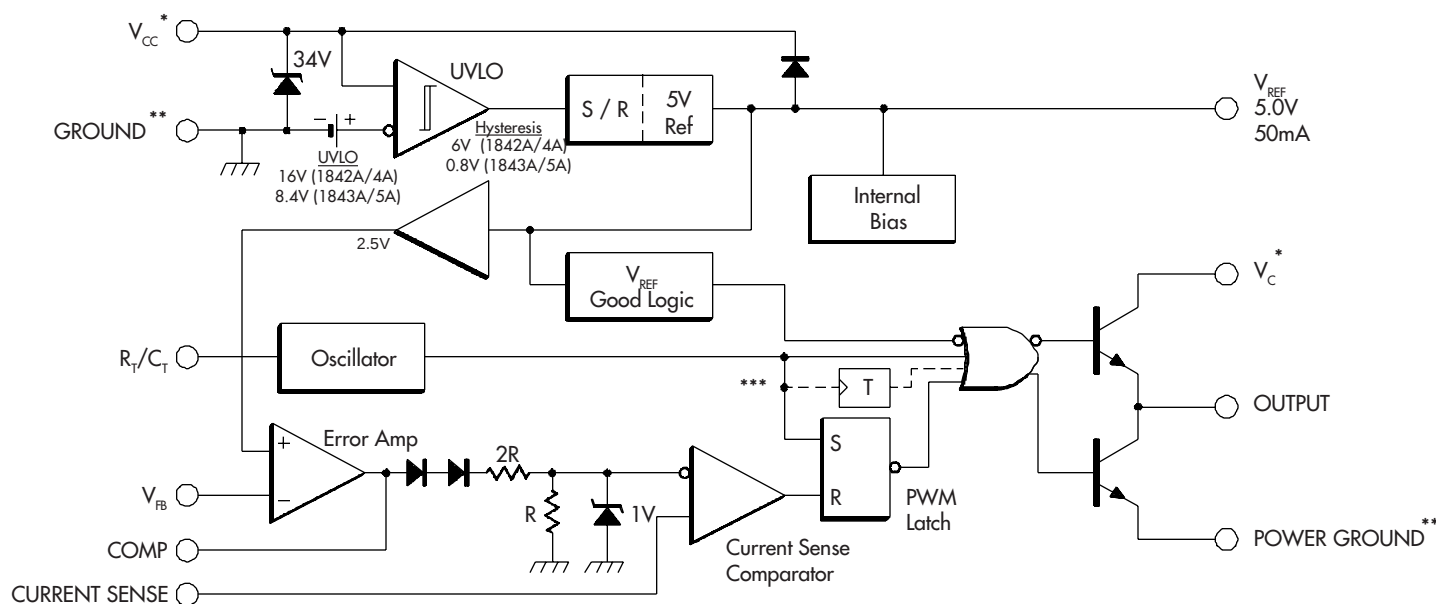
6. Output frequency equals oscillator frequency for the UC1842A and UC1843A. Output frequency is one half oscillator frequency for the UC1844A and UC1845A.

7. "Temperature stability, sometimes referred to as average temperature coefficient, is described by the equation:

$$\text{Temp Stability} = \frac{V_{REF}(\text{max.}) - V_{REF}(\text{min.})}{T_j(\text{max.}) - T_j(\text{min.})}$$

$V_{REF}(\text{max.})$ & $V_{REF}(\text{min.})$ are the maximum & minimum reference voltage measured over the appropriate temperature range. Note that the extremes in voltage do not necessarily occur at the extremes in temperature."

BLOCK DIAGRAM



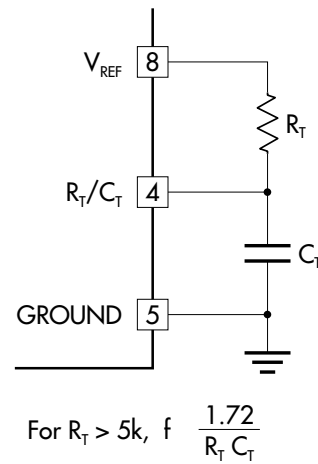
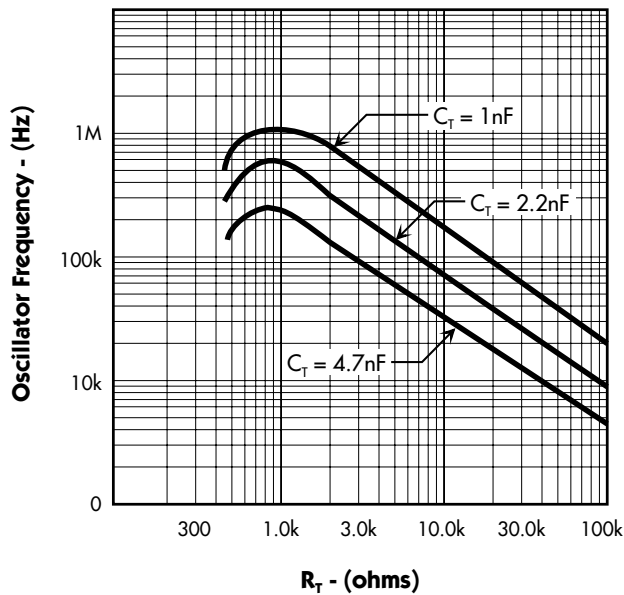
* - V_{CC} and V_c are internally connected for 8 pin packages.

** - POWER GROUND and GROUND are internally connected for 8 pin packages.

*** - Toggle flip flop used only in x844A and x845A series.

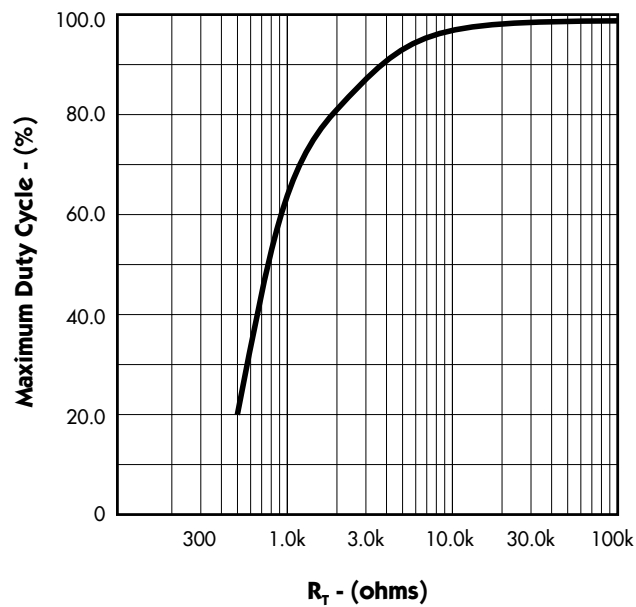
CHARACTERISTIC CURVES

FIGURE 1. — OSCILLATOR FREQUENCY vs. TIMING RESISTOR



Note: Output drive frequency is half the oscillator frequency for the UCx844A/5A devices.

FIGURE 2. — MAXIMUM DUTY CYCLE vs. TIMING RESISTOR



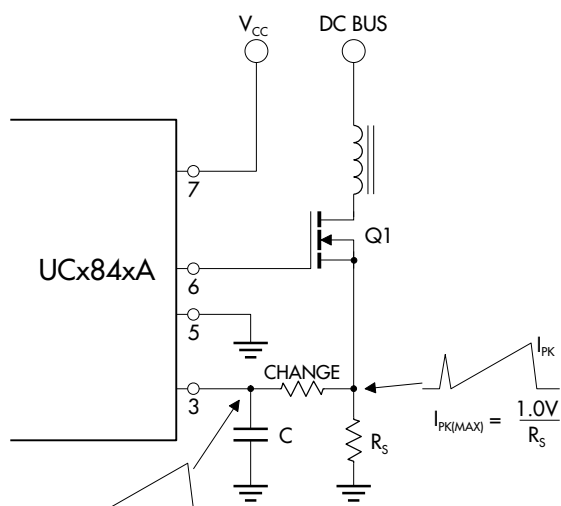
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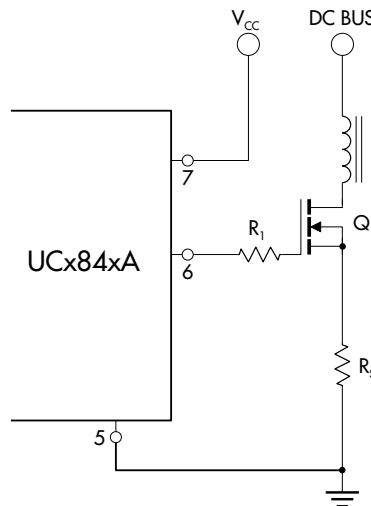
TYPICAL APPLICATION CIRCUITS

FIGURE 3. — CURRENT SENSE SPIKE SUPPRESSION



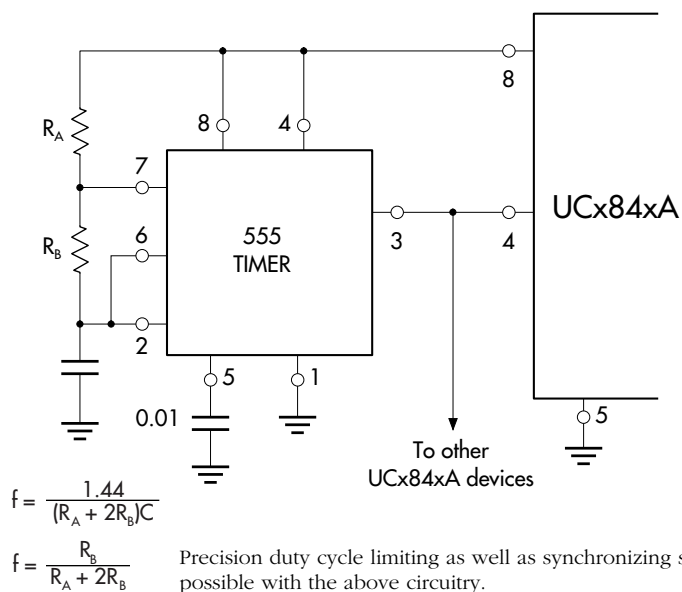
The RC low pass filter will eliminate the leading edge current spike caused by parasitics of Power MOSFET.

FIGURE 4. — MOSFET PARASITIC OSCILLATIONS



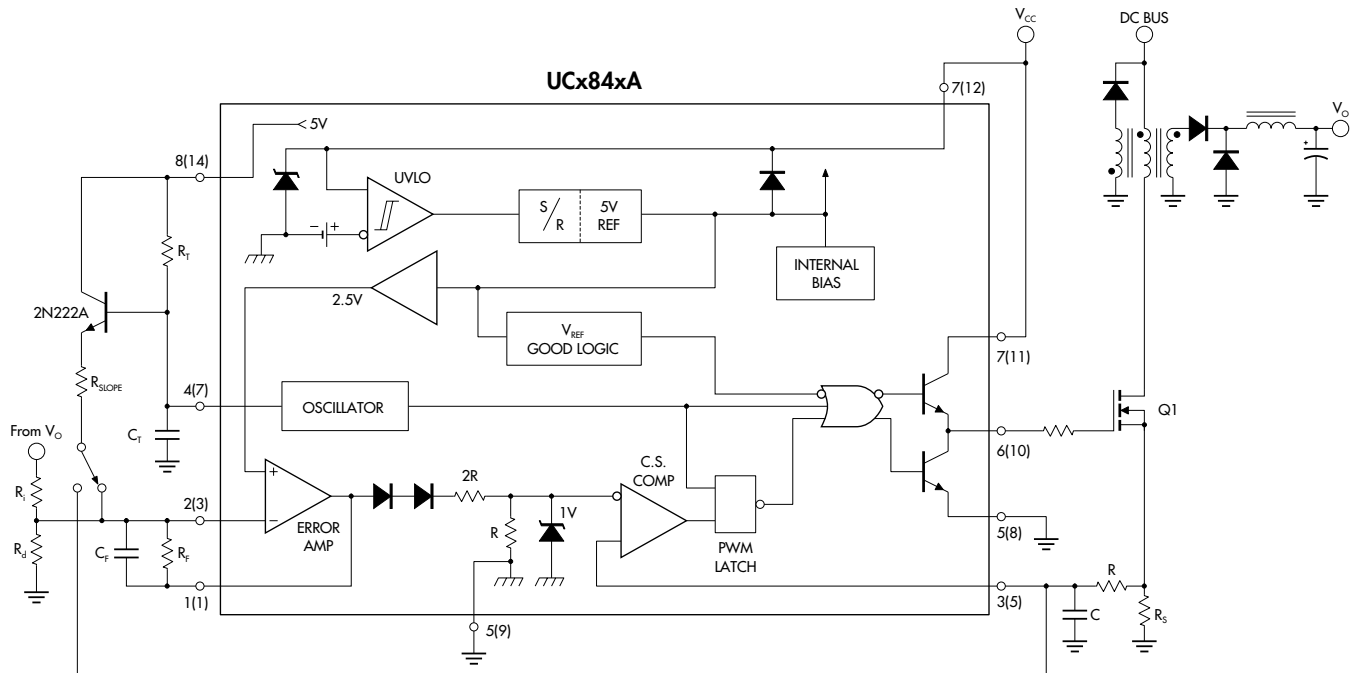
A resistor (R_1) in series with the MOSFET gate will reduce overshoot & ringing caused by the MOSFET input capacitance and any inductance in series with the gate drive. (Note: It is very important to have a low inductance ground path to insure correct operation of the I.C. This can be done by making the ground paths as short and as wide as possible.)

FIGURE 5. — EXTERNAL DUTY CYCLE CLAMP AND MULTI-UNIT SYNCHRONIZATION



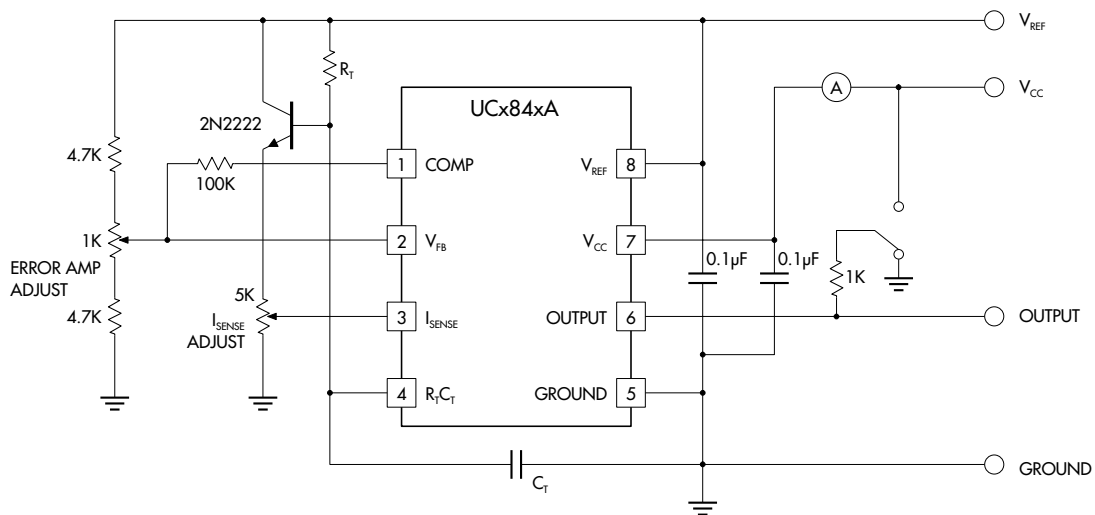
TYPICAL APPLICATION CIRCUITS (continued)

FIGURE 6. — SLOPE COMPENSATION



Due to inherent instability of current mode converters running above 50% duty cycle, slope compensation should be added to either the current sense pin or the error amplifier. Figure 6 shows a typical slope compensation technique.

FIGURE 7. — OPEN LOOP LABORATORY FIXTURE



High peak currents associated with capacitive loads necessitate careful grounding techniques. Timing and bypass capacitors should be connected to pin 5 in a single point ground. The transistor and 5k potentiometer are used to sample the oscillator waveform and apply an adjustable ramp to pin 3.

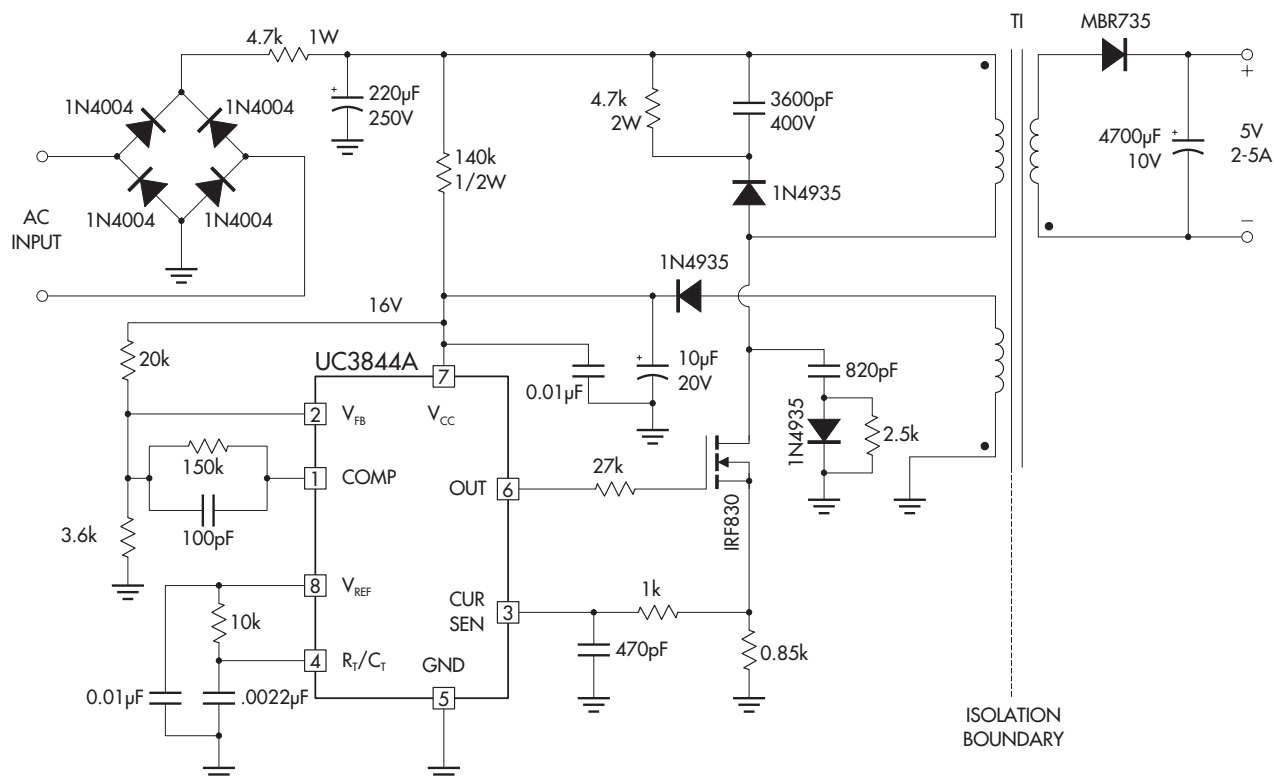
UC184xA/284xA/384xA

CURRENT MODE PWM CONTROLLER

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TYPICAL APPLICATION CIRCUITS (continued)

FIGURE 8. — OFF-LINE FLYBACK REGULATOR



SPECIFICATIONS

Input line voltage:	90VAC to 130VAC
Input frequency:	50 or 60Hz
Switching frequency:	40KHz $\pm 10\%$
Output power:	25W maximum
Output voltage:	5V $\pm 5\%$
Output current:	2 to 5A
Line regulation:	0.01%/V
Load regulation:	8%/A*
Efficiency @ 25 Watts,	
$V_{IN} = 90VAC$:	70%
$V_{IN} = 130VAC$:	65%
Output short-circuit current:	2.5Amp average

* This circuit uses a low-cost feedback scheme in which the DC voltage developed from the primary-side control winding is sensed by the UC3844A error amplifier. Load regulation is therefore dependent on the coupling between secondary and control windings, and on transformer leakage inductance.