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REVISION HISTORY

2/09—Rev. 0 to Rev. A

Edits to Features	1
Added Table 5	5
Changes to Table 6	5
Added Table 7	
Changes to Figure 9	13
Added Table 13	16
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5/08—Revision 0: Initial Version

SPECIFICATIONS

Each voltage is relative to its respective ground; $3.0 \text{ V} \le V_{DD1} \le 5.5 \text{ V}$, $3.0 \text{ V} \le V_{DD2} \le 3.6 \text{ V}$. All minimum/maximum specifications apply over the entire recommended operation range, unless otherwise noted. All typical specifications are at $T_A = 25^{\circ}\text{C}$, $V_{DD1} = 5 \text{ V}$, $V_{DD2} = 3.3 \text{ V}$, unless otherwise noted.

Table 1.

Parameter	Symbol	Min	Тур	Max	Unit	Test Conditions
SUPPLY CURRENT						
Power Supply Current, Logic Side	I _{DD1}					
TxD/RxD Data Rate < 500 kbps				3.5	mA	Unloaded output
ADM2487E TxD/RxD Data Rate = 500 kbps				4	mA	Half-duplex configuration, $R_{\text{TERMINATION}} = 120 \Omega$, see Figure 25
ADM2482E TxD/RxD Data Rate = 16 Mbps				6.0	mA	Half-duplex configuration, $R_{\text{TERMINATION}} = 120 \Omega$, see Figure 25
Power Supply Current, Bus Side	I _{DD2}					_
TxD/RxD Data Rate < 500 kbps				17	mA	Unloaded output
ADM2487E TxD/RxD Data Rate = 500 kbps				40	mA	$V_{DD2} = 3.6 \text{ V, half-duplex} \\ \text{configuration, } R_{\text{TERMINATION}} = 120 \ \Omega, \\ \text{see Figure 25} \\$
ADM2482E TxD/RxD Data Rate = 16 Mbps				50	mA	$V_{DD2} = 3.6 \text{ V}$, half-duplex configuration, $R_{TERMINATION} = 120 \Omega$, see Figure 25
DRIVER						
Differential Outputs						
Differential Output Voltage, Loaded	V _{OD2}	2.0		5.0	V	$R_L = 100 \Omega$ (RS-422), see Figure 19
		1.5		5.0	V	$R_L = 54 \Omega$ (RS-485), see Figure 19
	V _{OD3}	1.5		5.0	V	$-7 \text{ V} \leq V_{\text{TEST}} \leq +12 \text{ V}$, see Figure 20
$\Delta V_{OD} $ for Complementary Output States	Δ V _{OD}			0.2	V	$R_L = 54 \Omega$ or 100Ω , see Figure 19
Common-Mode Output Voltage	Voc			3.0	V	$R_L = 54 \Omega$ or 100Ω , see Figure 19
$\Delta V_{oc} $ for Complementary Output States	Δ Voc			0.2	V	$R_L = 54 \Omega$ or 100Ω , see Figure 19
Short-Circuit Output Current	los			250	mA	_
Output Leakage Current (Y, Z)	lo			125	μΑ	DE = 0 V, \overline{RE} = 0 V, V_{CC} = 0 V or 3.6 V, V_{IN} = 12 V
		-100			μΑ	DE = 0 V, \overline{RE} = 0 V, V_{CC} = 0 V or 3.6 V, V_{IN} = -7 V
Logic Inputs						
Input Threshold Low	V _{IL}	$0.25 \times V_{DD1}$			V	DE, RE, TxD
Input Threshold High	V _{IH}			$0.7\times V_{\text{DD1}}$	V	DE, RE, TxD
Input Current	I ₁	-10	+0.01	+10	μΑ	DE, RE, TxD
RECEIVER						
Differential Inputs						
Differential Input Threshold Voltage	V _{TH}	-200	-125	-30	mV	$-7 \text{ V} < \text{V}_{\text{CM}} < +12 \text{ V}$
Input Voltage Hysteresis	V _{HYS}		15		mV	$V_{OC} = 0 V$
Input Current (A, B)	l ₁			125	μΑ	$DE = 0 \text{ V}, V_{DD} = 0 \text{ V or } 3.6 \text{ V}, V_{IN} = 12 \text{ V}$
		-125			μΑ	$DE = 0 \text{ V}, V_{DD} = 0 \text{ V or } 3.6 \text{ V}, V_{IN} = -7 \text{ V}$
Line Input Resistance	R _{IN}	96			kΩ	$-7 \text{ V} < \text{V}_{\text{CM}} < +12 \text{ V}$
Logic Outputs						
Output Voltage Low	V_{OLRxD}		0.2	0.4	V	$I_{ORxD} = 1.5 \text{ mA}, V_A - V_B = -0.2 \text{ V}$
Output Voltage High	V_{OHRxD}	$V_{DD1} - 0.3$	$V_{DD1}-0.2$		V	$I_{ORxD} = -1.5 \text{ mA}, V_A - V_B = 0.2 \text{ V}$
Short-Circuit Current	los			100	mA	
Tristate Output Leakage Current	I _{OZR}			±1	μΑ	$V_{DD1} = 5.0 \text{ V}, 0 \text{ V} < V_{O} < V_{DD1}$

Parameter	Symbol	Min	Тур	Max	Unit	Test Conditions
TRANSFORMER DRIVER						
Oscillator Frequency	fosc	400	500	600	kHz	$V_{DD1} = 5.0 \text{ V}$
		230	330	430	kHz	$V_{DD1} = 3.3 \text{ V}$
Switch-On Resistance	Ron		0.5	1.5	Ω	
Start-Up Voltage	V _{START}		2.2	2.5	V	
COMMON-MODE TRANSIENT IMMUNITY ¹		25			kV/μs	V _{CM} = 1 kV, transient magnitude = 800 V

¹ CM is the maximum common-mode voltage slew rate that can be sustained while maintaining specification-compliant operation. V_{CM} is the common-mode potential difference between the logic and bus sides. The transient magnitude is the range over which the common-mode is slewed. The common-mode voltage slew rates apply to both rising and falling common-mode voltage edges.

TIMING SPECIFICATIONS

 $T_A = -40$ °C to +85°C, unless otherwise noted.

Table 2. ADM2482E

Parameter	Symbol	Min	Тур	Max	Unit	Test Conditions
DRIVER						
Propagation Delay	t _{DPLH} , t _{DPHL}			100	ns	$R_{DIFF} = 54 \Omega$, $C_L = 100 pF$, see Figure 21 and Figure 26
Output Skew	t _{DSKEW}			8	ns	$R_{DIFF} = 54 \Omega$, $C_L = 100 pF$, see Figure 21 and Figure 26
Rise Time/Fall Time	t _{DR} , t _{DF}			15	ns	$R_{DIFF} = 54 \Omega$, $C_L = 100 pF$, see Figure 21 and Figure 26
Enable Time	t _{ZL} , t _{ZH}			120	ns	$R_L = 110 \Omega$, $C_L = 50 pF$, see Figure 22 and Figure 28
Disable Time	t _{LZ} , t _{HZ}			150	ns	$R_L = 110 \Omega$, $C_L = 50 pF$, see Figure 22 and Figure 28
RECEIVER						
Propagation Delay	t _{PLH} , t _{PHL}			110	ns	$C_L = 15 \text{ pF}$, see Figure 23 and Figure 27
Output Skew	t _{SKEW}			8	ns	$C_L = 15$ pF, see Figure 23 and Figure 27
Enable Time	t _{ZL} , t _{ZH}			13	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15 \text{ pF}$, see Figure 24 and Figure 29
Disable Time	t _{LZ} , t _{HZ}			13	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15 \text{ pF}$, see Figure 24 and Figure 29

Table 3. ADM2487E

Parameter	Symbol	Min	Тур	Max	Unit	Test Conditions
DRIVER						
Propagation Delay	t _{DPLH} , t _{DPHL}	250		700	ns	$R_{DIFF} = 54 \Omega$, $C_L = 100 pF$, see Figure 21 and Figure 26
Output Skew	t _{DSKEW}			100	ns	$R_{DIFF} = 54 \Omega$, $C_L = 100 pF$, see Figure 21 and Figure 26
Rise Time/Fall Time	t _{DR} , t _{DF}	200		1100	ns	$R_{DIFF} = 54 \Omega$, $C_L = 100 pF$, see Figure 21 and Figure 26
Enable Time	tzL, tzH			2.5	μs	$R_L = 110 \Omega$, $C_L = 50 pF$, see Figure 22 and Figure 28
Disable Time	t _{LZ} , t _{HZ}			200	ns	$R_L = 110 \Omega$, $C_L = 50 pF$, see Figure 22 and Figure 28
RECEIVER						
Propagation Delay	t _{PLH} , t _{PHL}			200	ns	$C_L = 15$ pF, see Figure 23 and Figure 27
Output Skew	t _{SKEW}			30	ns	$C_L = 15$ pF, see Figure 23 and Figure 27
Enable Time	t _{ZL} , t _{ZH}			13	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15 \text{ pF}$, see Figure 24 and Figure 29
Disable Time	t _{LZ} , t _{HZ}			13	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15 \text{ pF}$, see Figure 24 and Figure 29

PACKAGE CHARACTERISTICS

Table 4.

Parameter	Symbol	Min Typ	Max	Unit	Test Conditions
Resistance (Input-to-Output) ¹	R _{I-O}	1012	!	Ω	
Capacitance (Input-to-Output) ¹	C _{I-O}	3		pF	f = 1 MHz
Input Capacitance ²	Cı	4		рF	
Input IC Junction-to-Case Thermal Resistance	θ _{ЈСІ}	33		°C/W	Thermocouple located at center of package underside
Output IC Junction-to-Case Thermal Resistance	θιсο	28		°C/W	Thermocouple located at center of package underside

¹ Device considered a 2-terminal device: Pin 1 to Pin 8 are shorted together, and Pin 9 to Pin 16 are shorted together. ² Input capacitance is from any input data pin to ground.

REGULATORY INFORMATION

Table 5. ADM2482E/ADM2487E Approvals

Organization	Approval Type	Notes
UL	Recognized under the component recognition program of underwriters laboratories, Inc.	In accordance with UL 1577, each ADM2482E/ADM2487E is proof tested by applying an insulation test voltage ≥3000 V rms for 1 second (current leakage detection limit = 5 µA)
VDE	Certified according to DIN V VDE V 0884-10 (VDE V 0884-10):2006-12	In accordance with DIN V VDE V 0884-10, each ADM2482E/ADM2487E is proof tested by applying an insulation test voltage ≥1050 V peak for 1 second (partial discharge detection limit = 5 pC)

INSULATION AND SAFETY-RELATED SPECIFICATIONS

Table 6.

Parameter	Symbol	Value	Unit	Conditions
Rated Dielectric Insulation Voltage		2500	V rms	1-minute duration
Minimum External Air Gap (External Clearance)	L(I01)	5.15 min	mm	Measured from input terminals to output terminals, shortest distance through air
Minimum External Tracking (Creepage)	L(102)	5.5 min	mm	Measured from input terminals to output terminals, shortest distance along body
Minimum Internal Gap (Internal Clearance)		0.017 min	mm	Insulation distance through insulation
Tracking Resistance (Comparative Tracking Index)	CTI	>175	V	DIN IEC 112/VDE 0303-1
Isolation Group		Illa		Material group (DIN VDE 0110: 1989-01, Table 1)

VDE 0884-2 INSULATION CHARACTERISTICS

This isolator is suitable for basic electrical isolation only within the safety limit data. Maintenance of the safety data must be ensured by means of protective circuits. An asterisk (*) on packages denotes DIN V VDE V 0884-10 approval.

Table 7.

Description	Conditions	Symbol	Characteristic	Unit
CLASSIFICATIONS				
Installation Classification per DIN VDE 0110 for Rated				
Mains Voltage				
≤150 V rms			I to IV	
≤300 V rms			l to III	
≤400 V rms			l to II	
Climatic Classification			40/85/21	
Pollution Degree	(DIN VDE 0110: 1989-01, see Table 1)		2	
VOLTAGE				
Maximum Working Insulation Voltage		V_{IORM}	560	V peak
Input-to-Output Test Voltage		V_{PR}		
	$V_{IORM} \times 1.875 = V_{PR}$, 100% production tested, $t_m =$			
Method b1	1 sec, partial discharge < 5 pC		1050	V peak
Method a:			896	V peak
After Environmental Tests, Subgroup 1	$V_{IORM} \times 1.6 = V_{PR}$, $t_m = 60$ sec, partial discharge <5 pC			
Method a			672	V peak
After Input and/or Safety Test, Subgroup 2/3):	$V_{IORM} \times 1.2 = V_{PR}$, $t_m = 60$ sec, partial discharge <5 pC			
Highest Allowable Overvoltage ¹		V_{TR}	4000	V peak
SAFETY-LIMITING VALUES ²				
Case Temperature		Ts	150	°C
Input Current		I _{S, INPUT}	265	mA
Output Current		Is, output	335	mA
Insulation Resistance at T _S ³		R_{S}	>109	Ω

¹ Transient overvoltage, t_{TR} = 10 sec.
² The safety-limiting value is the maximum value allowed in the event of a failure. See Figure 3 for the thermal derating curve.

 $^{^{3}}$ V_{IO} = 500 V.

ABSOLUTE MAXIMUM RATINGS

Each voltage is relative to its respective ground; $T_A = 25$ °C, unless otherwise noted.

Table 8.

D	D-4:
Parameter	Rating
V_{DD1}	−0.5 V to +6 V
V_{DD2}	−0.5 V to +6 V
Digital Input Voltages (DE, RE, TxD)	$-0.5 \mathrm{V}$ to $\mathrm{V}_{\mathrm{DD1}} + 0.5 \mathrm{V}$
Digital Output Voltages	
RxD	$-0.5 \mathrm{V}$ to $\mathrm{V}_{\mathrm{DD1}} + 0.5 \mathrm{V}$
D1, D2	13 V
Driver Output/Receiver Input Voltage Range	−9 V to +14 V
Average Output Current per Pin	−35 mA to +35 mA
ESD (Human Body Model) on A, B, Y, and Z pins	±15 kV
Operating Temperature Range	−40°C to +85°C
Storage Temperature Range	−55°C to +150°C
Lead Temperature	
Soldering (10 sec)	300°C
Vapor Phase (60 sec)	215°C
Infrared (15 sec)	220°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

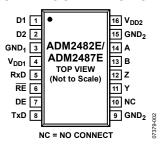


Figure 2. Pin Configuration

Table 9. Pin Function Descriptions

Pin No.	Mnemonic	Description
1	D1	Transformer Driver Terminal 1.
2	D2	Transformer Driver Terminal 2.
3	GND ₁	Ground, Logic Side.
4	V_{DD1}	Power Supply, Logic Side (3.3 V or 5 V). Decoupling capacitor to GND ₁ required; capacitor value should be between 0.01 μ F and 0.1 μ F.
5	RxD	Receiver Output Data. This output is high when $(A - B) > +200 \text{ mV}$ and low when $(A - B) < -200 \text{ mV}$. The output is tristated when the receiver is disabled, that is, when $\overline{\text{RE}}$ is driven high.
6	RE	Receiver Enable Input. This is an active low input. Driving this input low enables the receiver; driving it high disables the receiver.
7	DE	Driver Enable Input. Driving this input high enables the driver; driving it low disables the driver.
8	TxD	Transmit Data.
9	GND ₂	Ground, Bus Side.
10	NC	No Connect. This pin must be left floating.
11	Υ	Driver Noninverting Output.
12	Z	Driver Inverting Output.
13	В	Receiver Inverting Input.
14	Α	Receiver Noninverting Input.
15	GND ₂	Ground, Bus Side.
16	V_{DD2}	Power Supply, Bus Side (Isolated 3.3 V Supply). Decoupling capacitor to GND ₂ required; capacitor value should be between 0.01 μF and 0.1 μF.

TYPICAL PERFORMANCE CHARACTERISTICS

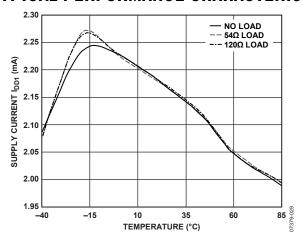


Figure 3. ADM2487E I_{DD1} Supply Current vs. Temperature (Data Rate = 500 kbps, V_{DD1} = 5 V, V_{DD2} = 3.3 V, DE = 1 V, \overline{RE} = 0 V)

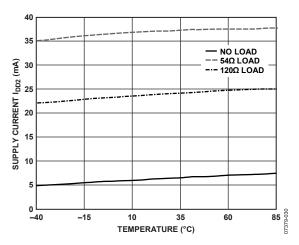


Figure 4. ADM2487E I_{DD2} Supply Current vs. Temperature (See Figure 25) (Data Rate = 500 kbps, V_{DD1} = 5 V, V_{DD2} = 3.3 V, DE = 1 V, \overline{RE} = 0 V)

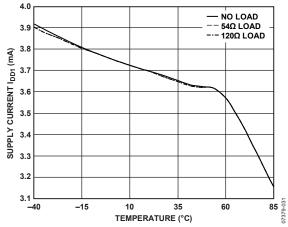


Figure 5. ADM2482E I_{DD1} Supply Current vs. Temperature (Data Rate = 16 Mbps, $V_{DD1} = 5$ V, $V_{DD2} = 3.3$ V, DE = 1 V, $\overline{RE} = 0$ V)

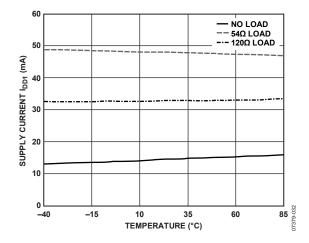


Figure 6. ADM2482E Supply Current vs. Temperature (See Figure 25) (Data Rate = 16 Mbps, V_{DD1} = 5 V, V_{DD2} = 3.3 V, DE = 1, \overline{RE} = 0 V)

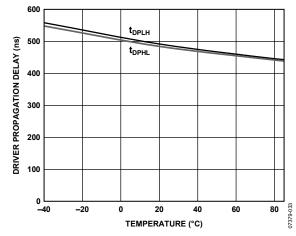


Figure 7. ADM2487E Driver Propagation Delay vs. Temperature

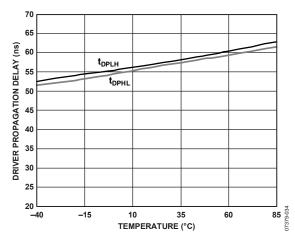


Figure 8. ADM2482E Driver Propagation Delay vs. Temperature

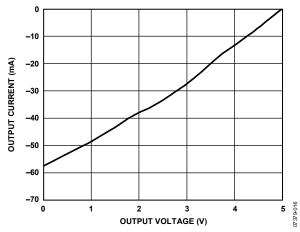


Figure 9. Output Current vs. Receiver Output High Voltage

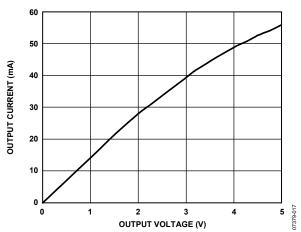


Figure 10. Output Current vs. Receiver Output Low Voltage

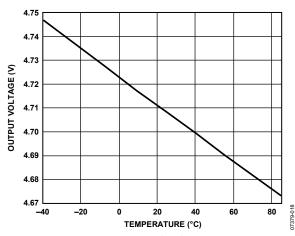


Figure 11. Receiver Output High Voltage vs. Temperature $(I_{DD2} = -4 \text{ mA})$

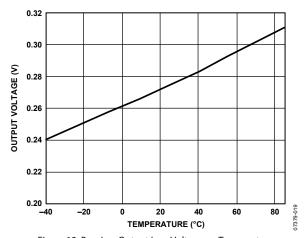


Figure 12. Receiver Output Low Voltage vs. Temperature $(I_{DD2} = 4 \text{ mA})$

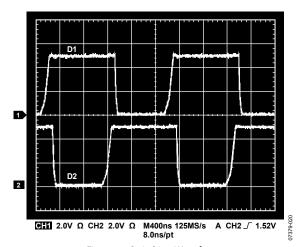


Figure 13. Switching Waveforms (50 Ω Pull-Up to V_{DD1} on D1 and D2)

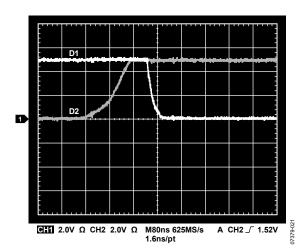


Figure 14. Switching Waveforms (Break-Before-Make, 50 Ω Pull-Up to V_{DD1} on D1 and D2)

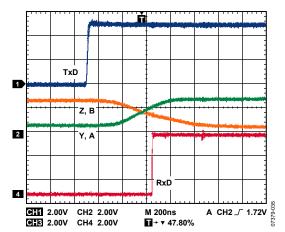


Figure 15. ADM2487E Driver/Receiver Propagation Delay, Low to High $(R_{DIFF} = 54 \Omega, C_{L1} = C_{L2} = 100 \text{ pF})$

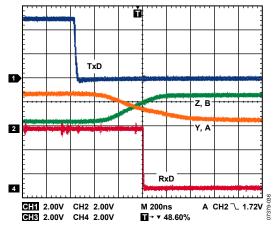


Figure 16. ADM2487E Driver/Receiver Propagation Delay, High to Low $(R_{DIFF}=54~\Omega,~C_{L1}=C_{L2}=100~pF)$

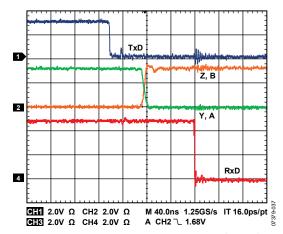


Figure 17. ADM2482E Driver/Receiver Propagation Delay, High to Low $(R_{DIFF}=54~\Omega,~C_{L1}=C_{L2}=100~pF)$

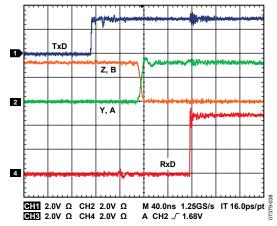


Figure 18. ADM2482E Driver/Receiver Propagation Delay, Low to High $(R_{DIFF}=54~\Omega,~C_{L1}=C_{L2}=100~pF)$

TEST CIRCUITS

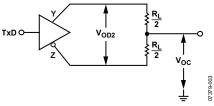


Figure 19. Driver Voltage Measurement

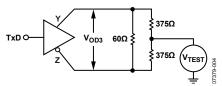


Figure 20. Driver Voltage Measurement

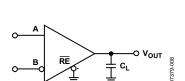


Figure 22. Driver Enable/Disable

Figure 23. Receiver Propagation Delay

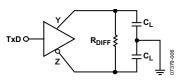


Figure 21. Driver Propagation Delay

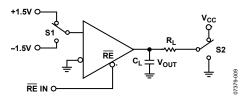


Figure 24. Receiver Enable/Disable

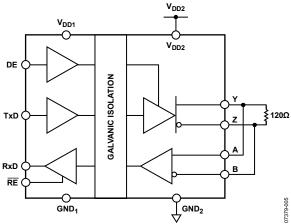


Figure 25. Supply Current Measurement Test Circuit

SWITCHING CHARACTERISTICS

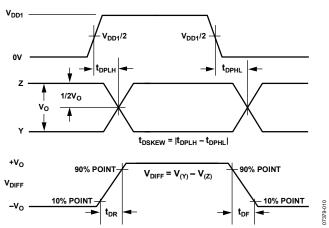
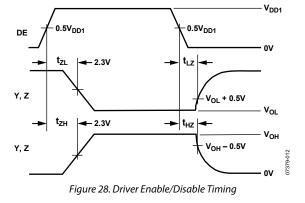


Figure 26. Driver Propagation Delay, Rise/Fall Timing



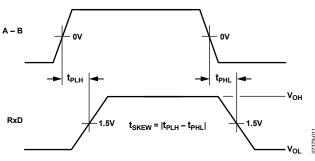


Figure 27. Receiver Propagation Delay

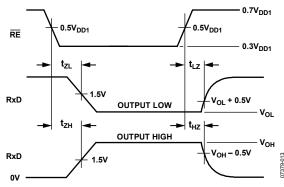


Figure 29. Receiver Enable/Disable Timing

CIRCUIT DESCRIPTION ELECTRICAL ISOLATION

In the ADM2482E/ADM2487E, electrical isolation is implemented on the logic side of the interface. Therefore, the part has two main sections: a digital isolation section and a transceiver section (see Figure 30). Driver input and data enable applied to the TxD and DE pins, respectively, and referenced to logic ground (GND₁) are coupled across an isolation barrier to appear at the transceiver section referenced to isolated ground (GND₂). Similarly, the receiver output, referenced to isolated ground in the transceiver section, is coupled across the isolation barrier to appear at the RxD pin referenced to logic ground.

iCoupler Technology

The digital signals transmit across the isolation barrier using iCoupler technology. This technique uses chip scale transformer windings to couple the digital signals magnetically from one side of the barrier to the other. Digital inputs are encoded into waveforms that are capable of exciting the primary transformer winding. At the secondary winding, the induced waveforms are decoded into the binary value that was originally transmitted.

Positive and negative logic transitions at the input cause narrow pulses (~1 ns) to be sent to the decoder, via the transformer. The decoder is bistable and is, therefore, either set or reset by the pulses, indicating input logic transitions. In the absence of logic transitions at the input for more than ~1 μs , a periodic set of refresh pulses indicative of the correct input state are sent to ensure dc correctness at the output. If the decoder receives no internal pulses for more than about 5 μs , then the input side is assumed to be unpowered or nonfunctional, in which case the output is forced to a default state (see Table 10).

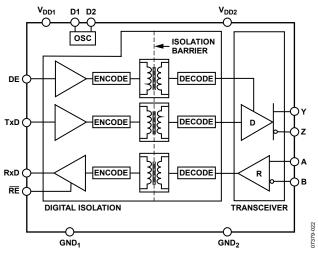


Figure 30. ADM2482E/ADM2487E Digital Isolation and Transceiver Sections

TRUTH TABLES

The truth tables in this section use the abbreviations found in Table 10.

Table 10. Truth Table Abbreviations

Letter	Description
Н	High level
1	Indeterminate
L	Low level
Χ	Irrelevant
Z	High impedance (off)
NC	Disconnected

Table 11. Transmitting

Supply	Status	Inp	uts	Outputs		
V _{DD1}	V_{DD2}	DE	TxD	Υ	Z	
On	On	Н	Н	Н	L	
On	On	Н	L	L	Н	
On	On	L	Χ	Z	Z	
On	Off	Χ	Χ	Z	Z	
Off	On	L	Χ	Z	Z	
Off	Off	Χ	Χ	Z	Z	

Table 12. Receiving

Suppl	y Status	Inputs	Outputs	
V_{DD1}	V_{DD2}	A – B	RE	RxD
On	On	>-0.03 V	L or NC	Н
On	On	<-0.2 V	L or NC	L
On	On	-0.2 V < A - B < -0.03 V	L or NC	1
On	On	Inputs open	L or NC	Н
On	On	X	Н	Z
On	Off	X	L or NC	Н
Off	Off	X	L or NC	L

THERMAL SHUTDOWN

The ADM2482E/ADM2487E contain thermal shutdown circuitry that protects the parts from excessive power dissipation during fault conditions. Shorting the driver outputs to a low impedance source can result in high driver currents. The thermal sensing circuitry detects the increase in die temperature under this condition and disables the driver outputs. This circuitry is designed to disable the driver outputs when a die temperature of 150°C is reached. As the device cools, the drivers are re-enabled at a temperature of 140°C.

TRUE FAIL-SAFE RECEIVER INPUTS

The receiver inputs have a true fail-safe feature that ensures that the receiver output is high when the inputs are open or shorted. During line idle conditions, when no driver on the bus is enabled, the voltage across a terminating resistance at the receiver input decays to 0 V. With traditional transceivers, receiver input thresholds specified between -200 mV and +200 mV mean that external bias resistors are required on the A and B pins to ensure that the receiver outputs are in a known state. The true fail-safe receiver input feature eliminates the need for bias resistors by specifying the receiver input threshold between -30 mV and -200 mV. The guaranteed negative threshold means that when the voltage between A and B decays to 0 V, the receiver output is guaranteed to be high.

MAGNETIC FIELD IMMUNITY

The limitation on the magnetic field immunity of the *i*Coupler is set by the condition in which an induced voltage in the receiving coil of the transformer is large enough to either falsely set or reset the decoder. The following analysis defines the conditions under which this may occur. The 3 V operating condition of the ADM2482E/ADM2487E is examined because it represents the most susceptible mode of operation.

The pulses at the transformer output have an amplitude greater than 1 V. The decoder has a sensing threshold of about 0.5 V, thus establishing a 0.5 V margin in which induced voltages can be tolerated.

The voltage induced across the receiving coil is given by

$$V = \left(\frac{-d\beta}{dt}\right) \sum \pi r_n^2; n = 1, 2, ..., N$$

where:

 β is the magnetic flux density (gauss). N is the number of turns in the receiving coil. r_n is the radius of the nth turn in the receiving coil (cm). Given the geometry of the receiving coil and an imposed requirement that the induced voltage is, at most, 50% of the 0.5 V margin at the decoder, a maximum allowable magnetic field can be determined using Figure 31.

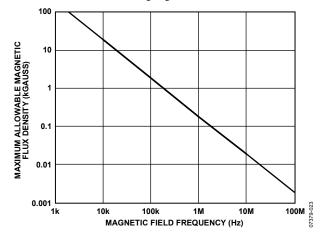


Figure 31. Maximum Allowable External Magnetic Flux Density

For example, at a magnetic field frequency of 1 MHz, the maximum allowable magnetic field of 0.2 kgauss induces a voltage of 0.25 V at the receiving coil. This is about 50% of the sensing threshold and does not cause a faulty output transition. Similarly, if such an event occurs during a transmitted pulse and is the worst-case polarity, it reduces the received pulse from $>1.0~\rm V$ to 0.75 V, still well above the 0.5 V sensing threshold of the decoder.

Figure 32 shows the magnetic flux density values in terms of more familiar quantities, such as maximum allowable current flow at given distances away from the ADM2482E/ADM2487E transformers.

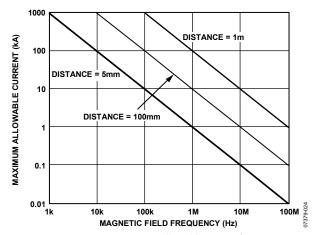


Figure 32. Maximum Allowable Current for Various Current-to-ADM2482E/ADM2487E Spacings

With combinations of strong magnetic field and high frequency, any loops formed by PCB traces can induce error voltages large enough to trigger the thresholds of succeeding circuitry.

Care should be taken in the layout of such traces to avoid this possibility.

APPLICATIONS INFORMATION PRINTED CIRCUIT BOARD LAYOUT

The isolated RS-485 transceiver of the ADM2482E/ADM2487E requires no external interface circuitry for the logic interfaces. Power supply bypassing is required at the input and output supply pins (see Figure 33).

Bypass capacitors are most conveniently connected between Pin 3 and Pin 4 for $V_{\rm DD1}$ and between Pin 15 and Pin 16 for $V_{\rm DD2}.$ The capacitor value must be between 0.01 μF and 0.1 μF . The total lead length between both ends of the capacitor and the input power supply pin must not exceed 20 mm.

Bypassing Pin 9 and Pin 16 is also recommended unless the ground pair on each package side is connected close to the package.

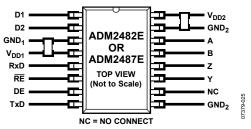


Figure 33. Recommended Printed Circuit Board Layout

In applications involving high common-mode transients, care must be taken to ensure that board coupling across the isolation barrier is minimized. Furthermore, the board layout must be designed such that any coupling that does occur equally affects all pins on a given component side.

Failure to ensure this can cause voltage differentials between pins exceeding the absolute maximum ratings of the device, thereby leading to latch-up or permanent damage.

TRANSFORMER SUPPLIERS

The transformer primarily used with the ADM2482E/ADM2487E must be a center-tapped transformer winding. The turns ratio of the transformer must be set to provide the minimum required output voltage at the maximum anticipated load with the minimum input voltage. Table 13 shows ADM2482E/ADM2487E transformer suppliers.

Table 13. Transformer Supplies

Manufacturer	Primary Voltage 3.3 V	Primary Voltage 5 V			
Coilcraft	DA2303-AL	GA3157			
Murata	782482/33VC	782482/53VC			

ISOLATED POWER SUPPLY CIRCUIT

The ADM2482E/ADM2487E integrate a transformer driver that, when used with an external transformer and linear voltage regulator (LDO), generates an isolated 3.3 V power supply to be supplied between $V_{\rm DD2}$ and GND_2 , as shown in Figure 34.

Pin D1 and Pin D2 of the ADM2482E/ADM2487E drive a center-tapped Transformer T1. A pair of Schottky diodes and a smoothing capacitor are used to create a rectified signal from the secondary winding. The ADP3330 LDO provides a regulated 3.3 V power supply to the ADM2482E/ADM2487E bus side circuitry ($V_{\rm DD2}$).

When the ADM2482E/ADM2487E are powered by 3.3 V on the logic side, a step-up transformer is required to compensate for the forward voltage drop of the Schottky diodes and the voltage drop across the regulator. The transformer turns ratio should be chosen to ensure just enough headroom for the ADP3330 LDO to output a regulated 3.3 V output under all operating conditions.

If the ADM2482E/ADM2487E are powered by 5 V on the logic side, then a step-down transformer should be used. For optimum efficiency, the transformer turns ratio should be chosen to ensure just enough headroom for the ADP3330 LDO to output a regulated 3.3 V output under all operating conditions.

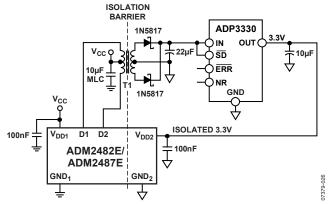


Figure 34. Applications Diagram

TYPICAL APPLICATIONS

Figure 35 and Figure 36 show typical applications of the ADM2482E/ADM2487E in half-duplex and full-duplex RS-485 network configurations. Up to 256 transceivers can be connected to the RS-485 bus. To minimize reflections, the

line must be terminated at the receiving end in its characteristic impedance, and stub lengths off the main line must be kept as short as possible. For a half-duplex operation, this means that both ends of the line must be terminated, because either end can be the receiving end.

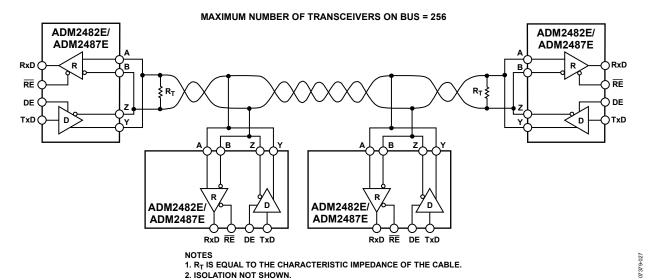


Figure 35. ADM2482E/ADM2487E Typical Half-Duplex RS-485 Network

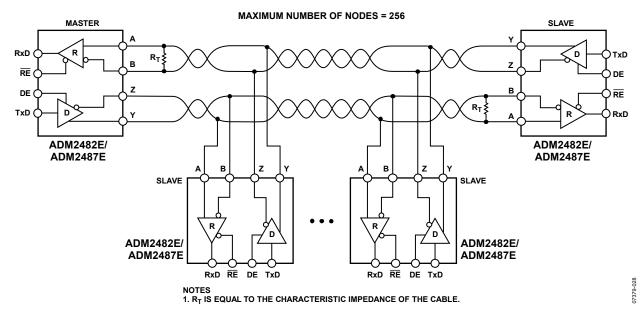
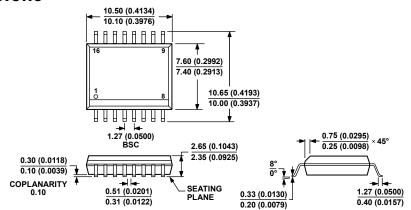


Figure 36. ADM2482E/ADM2487E Typical Full-Duplex RS-485 Network

OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-013-AA
CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS
(IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR
REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

_W]

Figure 37. 16-Lead Standard Small Outline Package [SOIC_W]
Wide Body
(RW-16)
Dimensions shown in millimeters and (inches)

ORDERING GUIDE

Model	Data Rate (Mbps)	Temperature Range	Package Description	Package Option
ADM2482EBRWZ ¹	16	-40°C to +85°C	16-Lead SOIC_W	RW-16
ADM2482EBRWZ-REEL7 ¹	16	-40°C to +85°C	16-Lead SOIC_W	RW-16
ADM2487EBRWZ ¹	0.5	-40°C to +85°C	16-Lead SOIC_W	RW-16
ADM2487EBRWZ-REEL7 ¹	0.5	-40°C to +85°C	16-Lead SOIC_W	RW-16
EVAL-ADM2482EEB3Z			Evaluation Board, 3.3 V Supply	
EVAL2482EEB5Z			Evaluation Board, 5 V Supply	
EVAL-ADM2487EEB3Z			Evaluation Board, 3.3 V Supply	
EVAL2487EEB5Z			Evaluation Board, 5 V Supply	

 $^{^{1}}$ Z = RoHS Compliant Part.

NOTES

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