

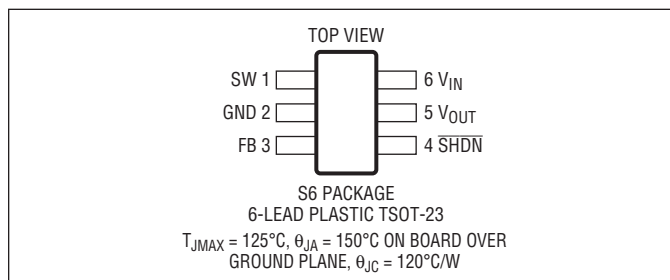
# LT3461/LT3461A

## ABSOLUTE MAXIMUM RATINGS

(Note 1)

Input Voltage ( $V_{IN}$ ).....	16V
$V_{OUT}$ , SW Voltage.....	40V
FB Voltage.....	5V
SHDN Voltage .....	16V
Operating Ambient Temperature Range (Note 2) .....	–40°C to 85°C
Maximum Junction Temperature .....	125°C
Storage Temperature Range .....	–65°C to 150°C
Lead Temperature (Soldering, 10 sec).....	300°C

## PIN CONFIGURATION



## ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3461AES6#PBF	LT3461AES6#TRPBF	LTAHG	8-Lead Plastic TSOT-23	–40°C to 85°C
LT3461ES6#PBF	LT3461ES6#TRPBF	LTAEB	8-Lead Plastic TSOT-23	–40°C to 85°C

Consult LTC Marketing for parts specified with wider operating temperature ranges.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreel/>. Some packages are available in 500 unit reels through designated sales channels with #TRMPBF suffix.

# ELECTRICAL CHARACTERISTICS

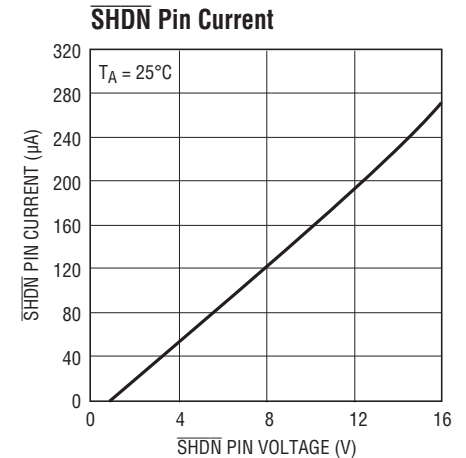
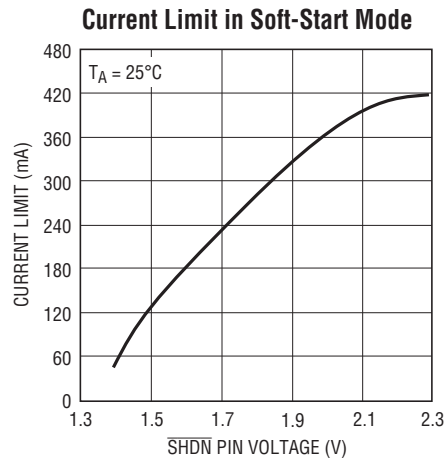
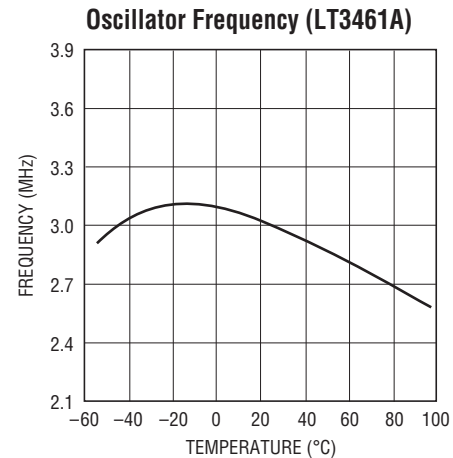
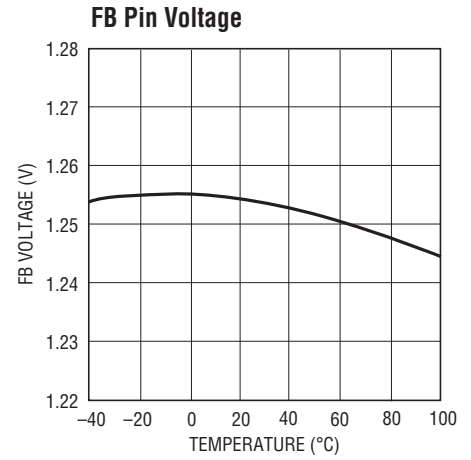
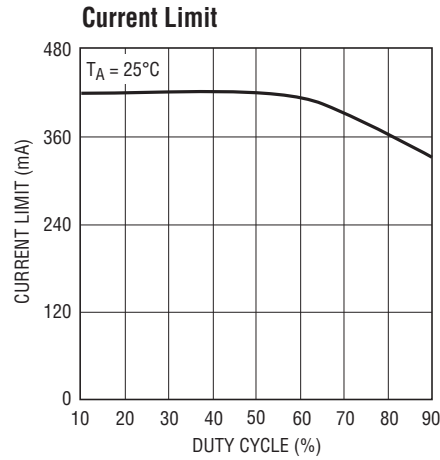
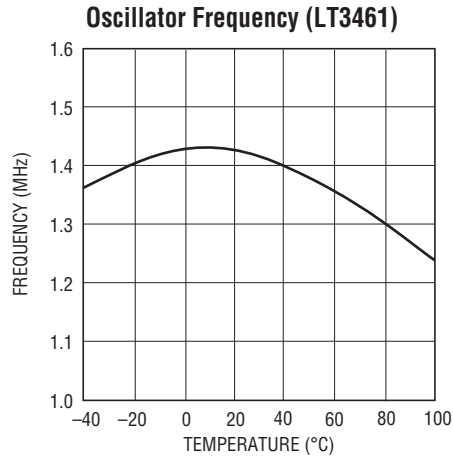
The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ ,  $V_{IN} = 3\text{V}$ ,  $V_{SHDN} = 3\text{V}$ , unless otherwise noted.

PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
Minimum Operating Voltage			2.5			V
Maximum Operating Voltage			16			V
Feedback Voltage		●	1.235 1.225	1.255	1.275 1.280	V V
Feedback Line Regulation			0.005		%/V	
FB Pin Bias Current		●	40		100	nA
Supply Current	FB = 1.3V, Not Switching SHDN = 0V		2.8 0.1		3.6 0.5	mA μA
Switching Frequency (LT3461A)		●	2.1	3.0	3.9	MHz
Switching Frequency (LT3461)		●	1.0	1.3	1.7	MHz
Maximum Duty Cycle (LT3461A)		●	82			%
Maximum Duty Cycle (LT3461)		●	92			%
Switch Current Limit			300	420	600	mA
Switch V <sub>CESAT</sub>	I <sub>SW</sub> = 250mA		260		350	mV
Switch Leakage Current	V <sub>SW</sub> = 5V		0.01		1	μA
Schottky Forward Voltage	I <sub>SCHOTTKY</sub> = 250mA		800		1100	mV
Schottky Reverse Leakage	V <sub>OUT</sub> – SW = 40V		0.03		4	μA
SHDN Voltage High			1.5			V
SHDN Voltage Low			0.4			V
SHDN Pin Bias Current			35		50	μA

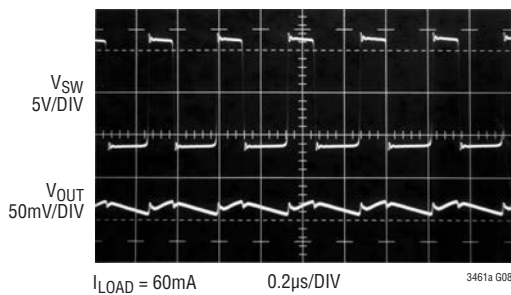
**Note 1:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

**Note 2:** The LT3461E/LT3461AE is guaranteed to meet specifications from  $0^\circ\text{C}$  to  $70^\circ\text{C}$ . Specifications over the  $-40^\circ\text{C}$  to  $85^\circ\text{C}$  operating temperature range are assured by design, characterization and correlation with statistical process controls.

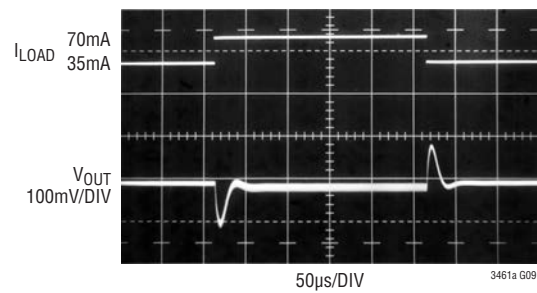
## TYPICAL PERFORMANCE CHARACTERISTICS



**Switching Waveform  
Circuit of Figure 4**



**Load Transient Response  
Circuit of Figure 4**



## PIN FUNCTIONS

**SW (Pin 1):** Switch Pin. Connect inductor here. Minimize trace at this pin to reduce EMI.

**GND (Pin 2):** Ground Pin. Tie directly to local ground plane.

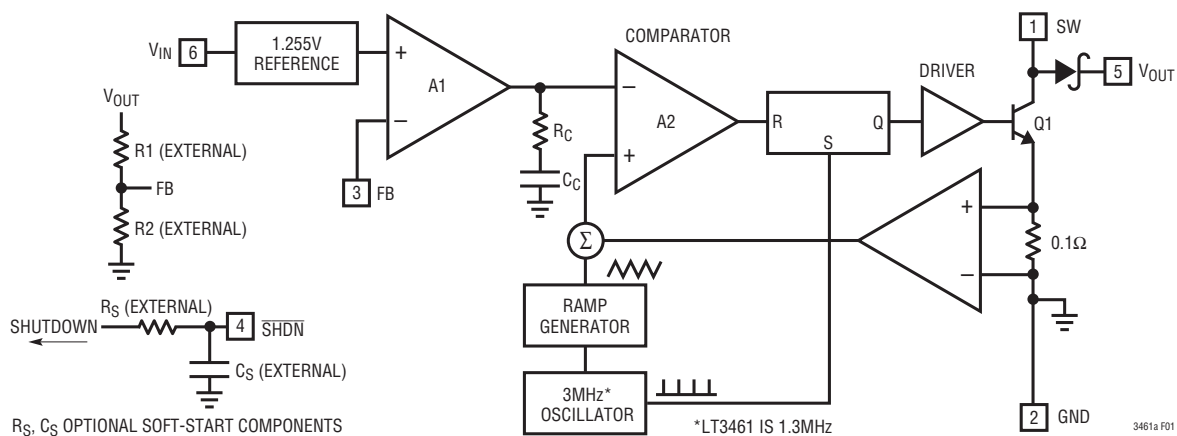
**FB (Pin 3):** Feedback Pin. Reference voltage is 1.255V. Connect resistor divider tap here. Minimize trace area at FB. Set  $V_{OUT}$  according to  $V_{OUT} = 1.255V (1 + R1/R2)$ .

**SHDN (Pin 4):** Shutdown Pin. Tie to 1.5V or higher to enable device; 0.4V or less to disable device. Also functions as soft-start. Use RC filter (47k, 47nF typ) as shown in Figure 1.

**V<sub>OUT</sub> (Pin 5):** Output Pin. Connect to resistor divider. Put capacitor close to pin and close to GND plane.

**V<sub>IN</sub> (Pin 6):** Input Supply Pin. Must be locally bypassed.

## BLOCK DIAGRAM



## APPLICATIONS INFORMATION

### Inrush Current

The LT3461 has a built-in Schottky diode. When supply voltage is applied to the  $V_{IN}$  pin, the voltage difference between  $V_{IN}$  and  $V_{OUT}$  generates inrush current flowing from input through the inductor and the Schottky diode to charge the output capacitor. The maximum nonrepetitive surge current the Schottky diode in the LT3461 can sustain is 1.5A. The selection of inductor and capacitor value should ensure the peak of the inrush current to be below 1.5A. In addition, turn-on of the LT3461 should be delayed until the inrush current is less than the maximum current limit. The peak inrush current can be calculated as follows:

$$I_P = \frac{V_{IN} - 0.6}{\sqrt{\frac{L}{C} - 1\Omega^2}} \exp\left(-\frac{\pi \cdot \Omega}{2\sqrt{\frac{L}{C} - 1\Omega^2}}\right)$$

where L is the inductance, r is the resistance of the inductor and C is the output capacitance.

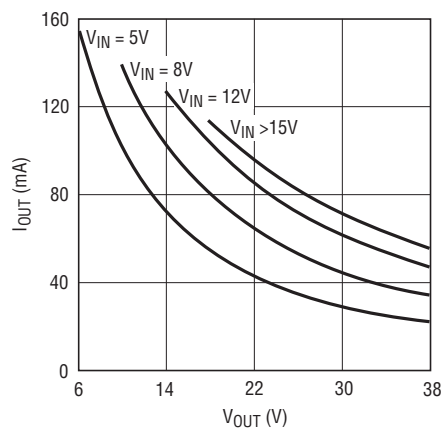
Table 3 gives inrush peak currents for some component selections.

**Table 3. Inrush Peak Current**

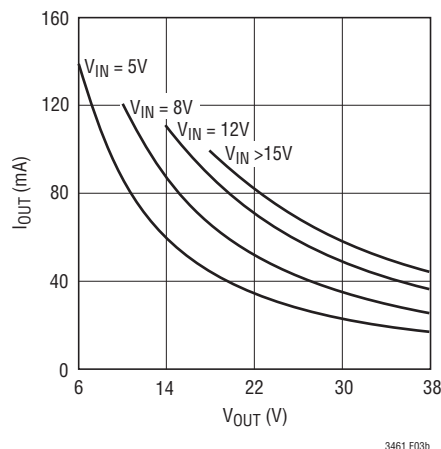
$V_{IN}$ (V)	L (μH)	C (μF)	$I_P$ (A)
5	4.7	1	1.1
5	10	1	0.9

### Thermal Considerations

Significant power dissipation can occur on the LT3461 and LT3461A, particularly at high input voltage. Device load, voltage drops in the power path components, and switching losses are the major contributors. It is important to measure device power dissipation in an application to ensure that the LT3461 does not exceed the absolute maximum operating junction temperature of 125°C over the operating ambient temperature range. Generally, for supply voltages below 5V the integrated current limit function provides adequate protection for nonfault conditions. For supply voltages above 5V, Figures 3a and 3b show the recommended operating region of the LT3461 and LT3461A, respectively. These graphs are based on 250mW on-chip dissipation. Improvement of these numbers can be expected if the LT3461 is supplied from a separate low voltage rail.



**Figure 3a. LT3461 Operating Region**



**Figure 3b. LT3461A Operating Region**

### Switching Frequency

The key difference between the LT3461 and LT3461A is the faster switching frequency of the LT3461A. At 3MHz, the LT3461A switches at twice the rate of the LT3461. The higher switching frequency of the LT3461A allows physically smaller inductors and capacitors to be used in a given application, but with a slight decrease in efficiency and maximum output current when compared to the LT3461. Generally if efficiency and maximum output current are crucial, or a high output voltage is being generated, the LT3461 should be used. If application size and cost are more important, the LT3461A will be the better choice.

## APPLICATIONS INFORMATION

### Inductor Selection

The inductors used with the LT3461/LT3461A should have a saturation current rating of 0.3A or greater. If the device is used in an application where the input supply will be hot-plugged, then the saturation current rating should be equal to or greater than the peak inrush current. For the LT3461, an inductor value between 10μH and 47μH, depending upon output voltage, will usually be the best choice for most designs. For the LT3461A, inductor values between 4.7μH and 15μH inductor will suffice for most applications. For best loop stability results, the inductor value selected should provide a ripple current of 70mA or more. For a given  $V_{IN}$  and  $V_{OUT}$  the inductor value to use with LT3461A is estimated by the formula:

$$L \text{ (in microhenries)} = \frac{D \cdot V_{IN} \cdot \sqrt{V_{OUT}} \cdot 1\text{sec}}{1A \cdot \sqrt{1V}}$$

$$\text{where } D = \frac{V_{OUT} + 1V - V_{IN}}{V_{OUT} + 1V}$$

Use twice this value for the LT3461.

### Capacitor Selection

Low ESR capacitors should be used at the output to minimize the output voltage ripple. Multilayer ceramic capacitors using X5R/X7R dielectrics are preferred as

they have a low ESR and maintain capacitance over wide voltage and temperature range. A 2.2μF output capacitor is sufficient for most applications using the LT3461, while a 1μF capacitor is sufficient for most applications using the LT3461A. High output voltages typically require less capacitance for loop stability. Always use a capacitor with sufficient voltage rating.

Either ceramic or solid tantalum capacitors may be used for the input decoupling capacitor, which should be placed as close as possible to the LT3461/LT3461A. A 1μF capacitor is sufficient for most applications.

### Phase Lead Capacitor

A small value capacitor can be added across resistor R1 between the output and the FB pin to reduce output perturbation due to a load step and to improve transient response. This phase lead capacitor introduces a pole-zero pair to the feedback that boosts phase margin near the cross-over frequency. The following formula is useful to estimate the capacitor value needed:

$$C_{PL} = \frac{500k\Omega}{R2} \cdot 1pF$$

For an application running 50μA in the feedback divider, capacitor values from 10pF to 22pF work well.

## TYPICAL APPLICATIONS

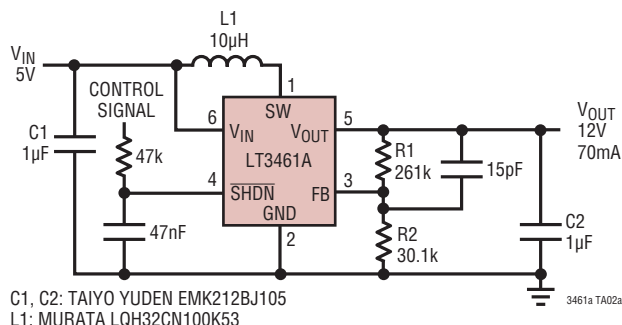
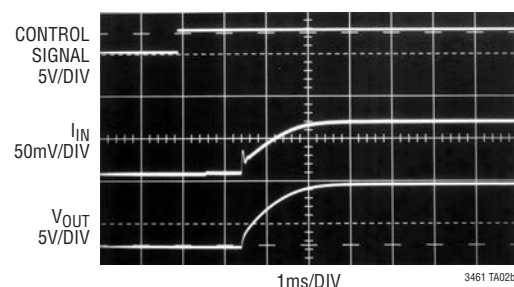


Figure 4. 5V to 12V with Soft-Start Circuit (LT3461A)

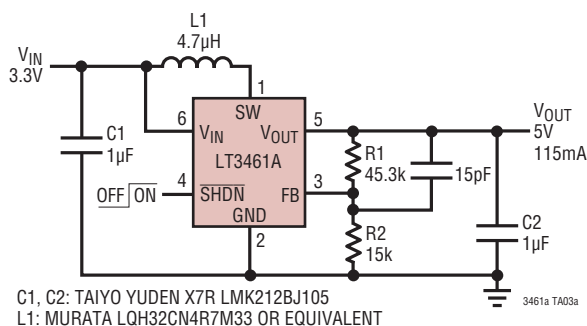
### Input Current and Output Voltage



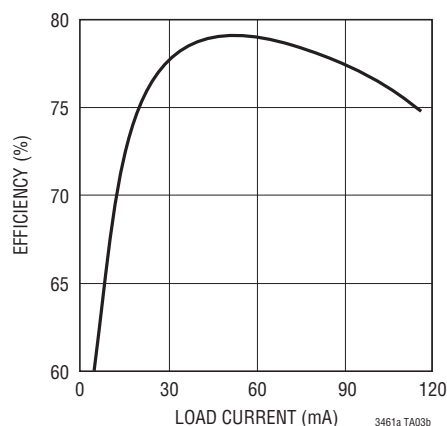
# LT3461/LT3461A

## TYPICAL APPLICATIONS

3.3V to 5V Step-Up Converter (LT3461A)



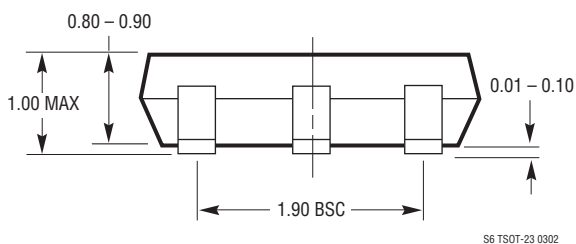
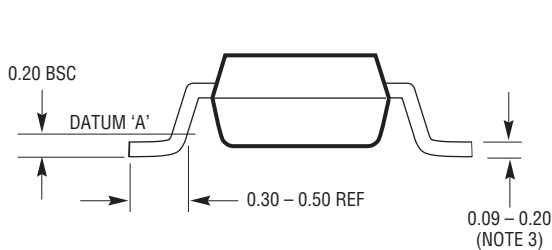
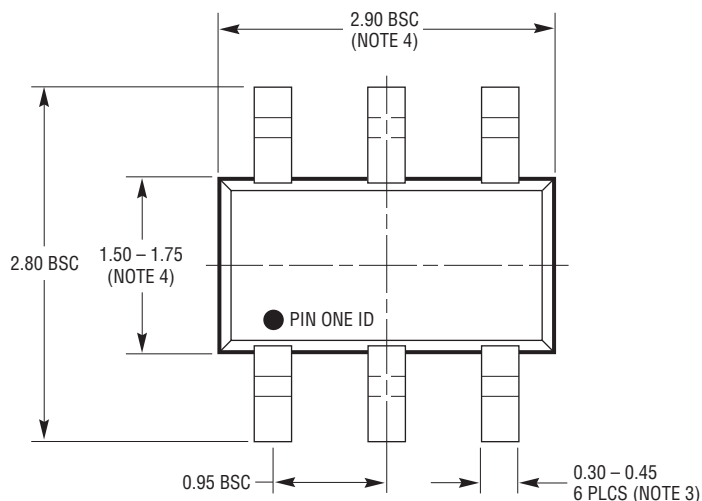
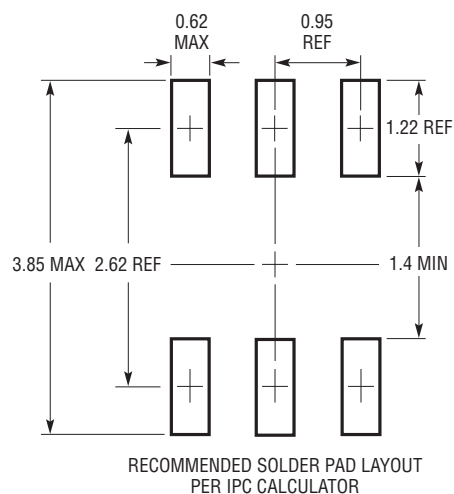
3.3V to 5V Step-Up Converter Efficiency



## PACKAGE DESCRIPTION

Please refer to <http://www.linear.com/product/LT3461#packaging> for the most recent package drawings.

**S6 Package**  
**6-Lead Plastic TSOT-23**  
(Reference LTC DWG # 05-08-1636)



**NOTE:**

1. DIMENSIONS ARE IN MILLIMETERS
2. DRAWING NOT TO SCALE
3. DIMENSIONS ARE INCLUSIVE OF PLATING
4. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR
5. MOLD FLASH SHALL NOT EXCEED 0.254mm
6. JEDEC PACKAGE REFERENCE IS MO-193

3461afa

**REVISION HISTORY** (Revision history begins at Rev B)

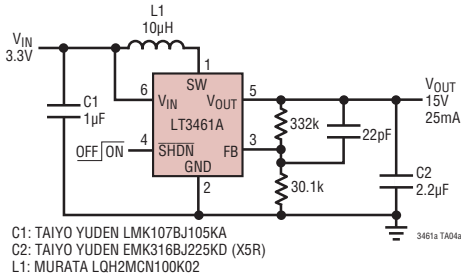
REV	DATE	DESCRIPTION	PAGE NUMBER
A	01/16	Modified inrush current $I_P$ equation.	6



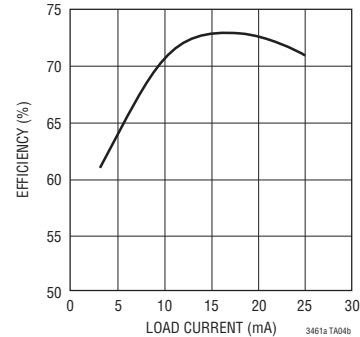
# LT3461/LT3461A

## TYPICAL APPLICATION

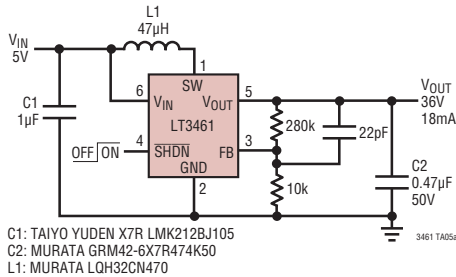
### Low Profile (1mm) 3.3V to 15V Step-Up Converter



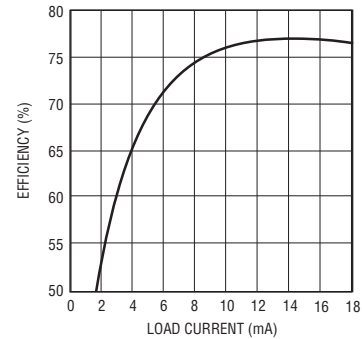
### 3.3V to 15V Efficiency



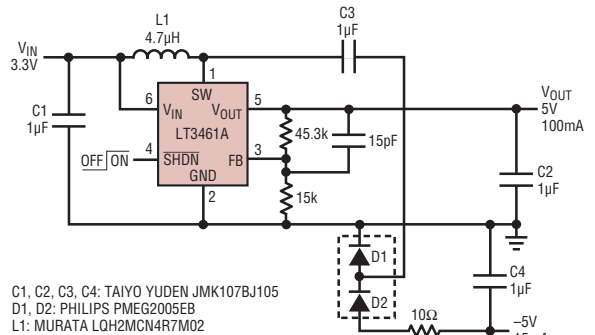
### 5V to 36V Step-Up Converter (LT3461)



### 5V to 36V Efficiency



### 3.3V to ±5V Dual Output Converter



## RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
<a href="#">LT1615/LT1615-1</a>	300mA/80mA ( $I_{SW}$ ) Constant Off-Time, High Efficiency Step-Up DC/DC Converter	$V_{IN}$ : 1.2V to 15V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 20µA, $I_{SD}$ < 1µA, ThinSOT Package
<a href="#">LT1944/LT1944-1</a>	Dual Output 350mA/100mA ( $I_{SW}$ ), Constant Off-Time, High Efficiency Step-Up DC/DC Converter	$V_{IN}$ : 1.2V to 15V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 20µA, $I_{SD}$ < 1µA, MS Package
<a href="#">LTC3400/LTC3400B</a>	600mA ( $I_{SW}$ ), 1.2MHz, Synchronous Step-Up DC/DC Converter	$V_{IN}$ : 0.85V to 5V, $V_{OUT(MAX)}$ = 5V, $I_Q$ = 19µA/300µA, $I_{SD}$ < 1µA, ThinSOT Package
<a href="#">LT3460</a>	0.32A ( $I_{SW}$ ), 1.3MHz, High Efficiency Step-Up DC/DC Converter	$V_{IN}$ : 2.5V to 16V, $V_{OUT(MAX)}$ = 36V, $I_Q$ = 2mA, $I_{SD}$ < 1µA, SC70, ThinSOT Packages
<a href="#">LT3465/LT3465A</a>	Constant Current, 1.2MHz/2.7MHz, High Efficiency White LED Boost Regulator with Integrated Schottky Diode	$V_{IN}$ : 2.7V to 16V, $V_{OUT(MAX)}$ = 34V, $I_Q$ = 1.9mA, $I_{SD}$ < 1µA, ThinSOT Package

3461afa