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### **REVISION HISTORY**

### 6/2016—Rev. 0 to Rev. A

Changed CP-16-1 to CP-16-21	. Throughout
Changes to Figure 4 and Table 4	5
Deleted Evaluation Boards Section, RF Layout Con-	siderations
Section, Power Supply Section, and Figure 20; Renu	ımbered
Sequentially	9
Deleted Figure 21 and Figure 22	10
Updated Outline Dimensions	11
Changes to Ordering Guide	11

5/2007—Revision 0: Initial Version

# **SPECIFICATIONS**

 $V_{\text{CC}}$  = 5 V, 75  $\Omega$  system,  $T_{\text{A}}$  = 25°C, unless otherwise noted.

Table 1.

Parameter	Conditions	Min	Тур	Max	Unit
DYNAMIC PERFORMANCE					
Bandwidth (–3 dB)			1600		MHz
Specified Frequency Range		54		865	MHz
Gain (S <sub>21</sub> , S <sub>31</sub> )	f = 100 MHz; see Figure 17 and Figure 18		2.8		dB
1 dB Gain Flatness			1000		MHz
NOISE/DISTORTION PERFORMANCE					
Noise Figure <sup>1</sup>	@ 54 MHz		4.0		dB
	@ 550 MHz		4.5		dB
	@ 865 MHz		4.6		dB
Output IP3	$f_1 = 97.25 \text{ MHz}, f_2 = 103.25 \text{ MHz}$		26		dBm
Output IP2	$f_1 = 97.25 \text{ MHz}, f_2 = 103.25 \text{ MHz}$		44.5		dBm
Composite Triple Beat (CTB)	135 channels, 15 dBmV/channel, f = 865 MHz		-72		dBc
Composite Second Order (CSO)	135 channels, 15 dBmV/channel, f = 865 MHz		-62		dBc
Cross Modulation (CXM)	135 channels, 15 dBmV/channel, 100% modulation		-69		dBc
	@ 15.75 kHz, f = 865 MHz				
INPUT CHARACTERISTICS	See Figure 17, Figure 18, and Figure 19				
Input Return Loss (S <sub>11</sub> )	@ 54 MHz		-15	-11	dB
	@ 550 MHz		-35.5	-22	dB
	@ 865 MHz		-13.3	-8	dB
Output-to-Input Isolation $(S_{12}, S_{13})$	Either output, 54 MHz to 865 MHz				
	@ 54 MHz		-32	-30	dB
	@ 550 MHz		-32	-29	dB
	@ 865 MHz		-33	-31	dB
OUTPUT CHARACTERISTICS	See Figure 17, Figure 18, and Figure 19				
Output Return Loss (S <sub>22</sub> , S <sub>33</sub> )	Either output, 54 MHz to 865 MHz				
	@ 54 MHz		-26.7	-21	dB
	@ 550 MHz		-22	-15	dB
	@ 865 MHz		-20	-12	dB
Output-to-Output Isolation (S23, S32)	Either output, 54 MHz to 865 MHz				dB
	@ 54 MHz		-26.7		dB
	@ 550 MHz		-25.1		dB
	@ 865 MHz		-25		dB
1 dB Compression (P <sub>1dB</sub> )	Output referred, f = 100 MHz		8.25		dBm
POWER SUPPLY					
Nominal Supply Voltage		4.75	5.0	5.25	V
Quiescent Supply Current			88	105	mA

 $<sup>^{\</sup>mbox{\tiny 1}}$  Characterized with 50  $\Omega$  noise figure analyzer.

### ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Supply Voltage	5.5 V
Power Dissipation	See Figure 3
Storage Temperature Range	−65°C to +125°C
Operating Temperature Range	-40°C to +85°C
Lead Temperature (Soldering, 10 sec)	300°C
Junction Temperature	150°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

#### THERMAL RESISTANCE

 $\theta_{JA}$  is specified for the device (including exposed pad) soldered to a high thermal conductivity 2s2p circuit board, as described in EIA/JESD 51-7.

**Table 3. Thermal Resistance** 

Package Type	θ <sub>JA</sub>	Unit
16-Lead LFCSP (Exposed Pad)	98	°C/W

#### **Maximum Power Dissipation**

The maximum safe power dissipation in the ADA4304-2 package is limited by the associated rise in junction temperature (T<sub>1</sub>) on the die. At approximately 150°C, which is the glass transition temperature, the plastic changes its properties. Even temporarily exceeding this temperature limit can change the stresses that the package exerts on the die, permanently shifting the parametric performance of the ADA4304-2. Exceeding a junction temperature of 150°C for an extended period can result in changes in the silicon devices, potentially causing failure.

The power dissipated in the package ( $P_D$ ) is essentially equal to the quiescent power dissipation; the supply voltage ( $V_S$ ) times the quiescent current ( $I_S$ ). In Table 1, the maximum power dissipation of the ADA4304-2 can be calculated as

$$P_{D(MAX)} = 5.25 \text{ V} \times 105 \text{ mA} = 551 \text{ mW}$$

Airflow increases heat dissipation, effectively reducing  $\theta_{JA}$ . In addition, more metal directly in contact with the package leads/exposed pad from metal traces, through-holes, ground, and power planes reduces the  $\theta_{JA}$ .

Figure 3 shows the maximum safe power dissipation in the package vs. the ambient temperature for the 16-lead LFCSP (98°C/W) on a JEDEC standard 4-layer board.

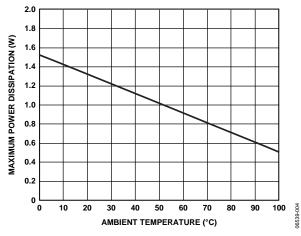


Figure 3. Maximum Power Dissipation vs. Temperature for a 4-Layer Board

### **ESD CAUTION**



**ESD** (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

# PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

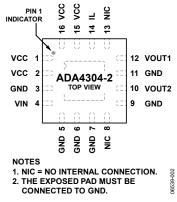


Figure 4. Pin Configuration

**Table 4. Pin Function Descriptions** 

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Pin No.	Mnemonic	Description	
1, 2, 15, 16	VCC	Supply Pin.	
3, 5 to 7, 9, 11	GND	Ground.	
4	VIN	Input.	
8, 13	NIC	No Internal Connection.	
10	VOUT2	Output 2.	
12	VOUT1	Output 1.	
14	IL	Bias Pin.	
	EPAD	Exposed Pad. The exposed pad must be connected to GND.	

### TYPICAL PERFORMANCE CHARACTERISTICS

 $V_{\text{CC}}$  = 5 V, 75  $\Omega$  system,  $T_{\text{A}}$  = 25°C, unless otherwise noted.

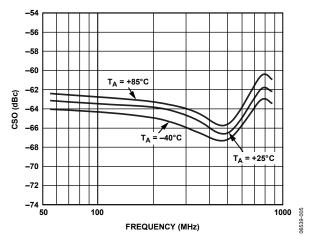


Figure 5. Composite Second Order (CSO) vs. Frequency

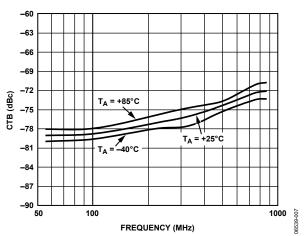


Figure 6. Composite Triple Beat (CTB) vs. Frequency

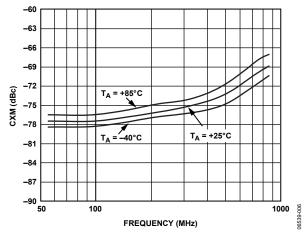


Figure 7. Cross Modulation (CXM) vs. Frequency

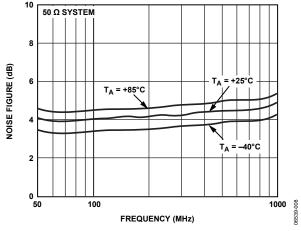


Figure 8. Noise Figure vs. Frequency

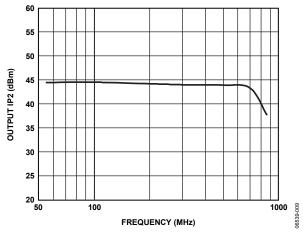


Figure 9. Output IP2 vs. Frequency

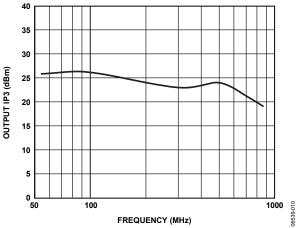


Figure 10. Output IP3 vs. Frequency

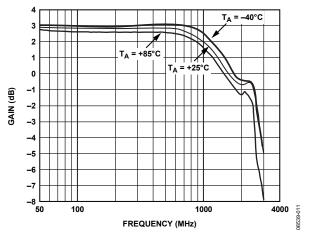


Figure 11. Gain (S<sub>21</sub>, S<sub>31</sub>) vs. Frequency

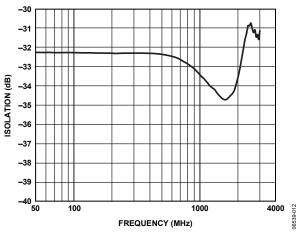
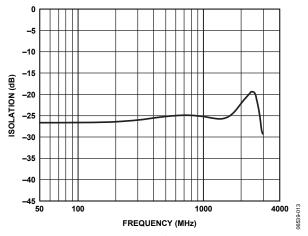


Figure 12. Output-to-Input Isolation (S<sub>12</sub>, S<sub>13</sub>) vs. Frequency



`Figure 13. Output-to-Output Isolation (S23, S32) vs. Frequency

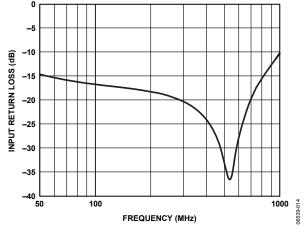


Figure 14. Input Return Loss (S<sub>11</sub>) vs. Frequency

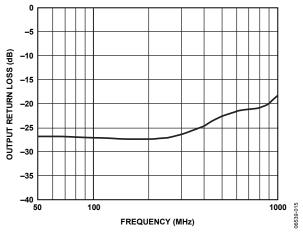


Figure 15. Output Return Loss (S<sub>22</sub>, S<sub>33</sub>) vs. Frequency

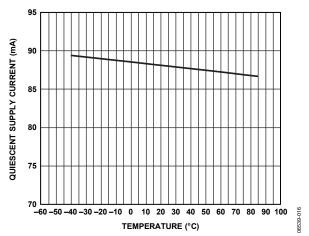


Figure 16. Quiescent Supply Current vs. Temperature

# **TEST CIRCUITS**

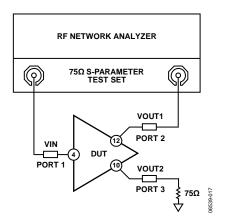


Figure 17. Test Circuit for  $S_{11}$ ,  $S_{12}$ ,  $S_{21}$ ,  $S_{22}$  Measurements

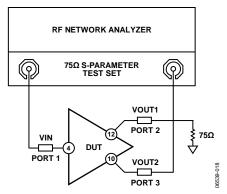


Figure 18. Test Circuit for  $S_{13}$ ,  $S_{31}$ ,  $S_{33}$  Measurements

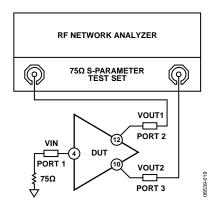


Figure 19. Test Circuit for S<sub>23</sub>, S<sub>32</sub> Measurements

### APPLICATIONS INFORMATION

The ADA4304-2 active splitter is primarily intended for use in the downstream path of television set-top boxes (STBs) that contain multiple tuners. It is typically located directly after the diplexer in a bidirectional CATV customer premise unit. The ADA4304-2 provides a single-ended input and two single-ended outputs that allow the delivery of the RF signal to two different signal paths. These paths can include, but are not limited to, a main picture tuner, the picture-in-picture (PIP) tuner, an out-of-band (OOB) tuner, a digital video recorder (DVR), and a cable modem (CM).

The ADA4304-2 exhibits composite second order (CSO) and composite triple beat (CTB) products that are -62 dBc and -72 dBc, respectively. The use of the SiGe bipolar process also allows the ADA4304-2 to achieve a noise figure (NF) of 4 dB.

#### **CIRCUIT DESCRIPTION**

The ADA4304-2 consists of a low noise buffer amplifier followed by a resistive power divider. This arrangement provides 2.8 dB of gain relative to the RF signal present at the input of the device. The input and each output must be properly matched to a 75  $\Omega$  environment for distortion and noise performance to match the data sheet specifications. AC coupling capacitors of 0.01  $\mu F$  are recommended for the input and outputs.

A 1  $\mu$ H RF choke (Coilcraft chip inductor 0805LS-102X) is required to correctly bias internal nodes of the ADA4304-2. It should be connected between the 5 V supply and the IL pin (Pin 14). The choke should be placed as close as possible to the ADA4304-2 to minimize parasitic capacitance on the IL pin, which is critical for achieving the specified bandwidth and flatness.

### **OUTLINE DIMENSIONS**

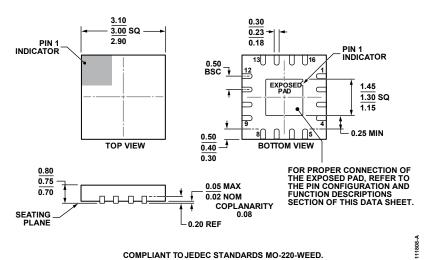


Figure 20. 16-Lead Lead Frame Chip Scale Package [LFCSP] 3 mm × 3 mm Body and 0.75 mm Package Height (CP-16-21) Dimensions shown in millimeters

#### **ORDERING GUIDE**

Model <sup>1</sup>	Temperature Range	Package Description	Package Option	Ordering Quantity	Branding
ADA4304-2ACPZ-RL	−40°C to +85°C	16-Lead LFCSP	CP-16-21	5,000	H0Z
ADA4304-2ACPZ-R7	-40°C to +85°C	16-Lead LFCSP	CP-16-21	1,500	H0Z
ADA4304-2ACPZ-R2	-40°C to +85°C	16-Lead LFCSP	CP-16-21	250	H0Z

 $<sup>^{1}</sup>$  Z = RoHS Compliant Part.



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