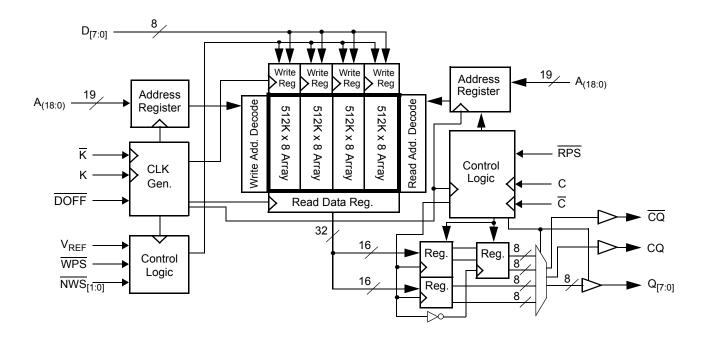
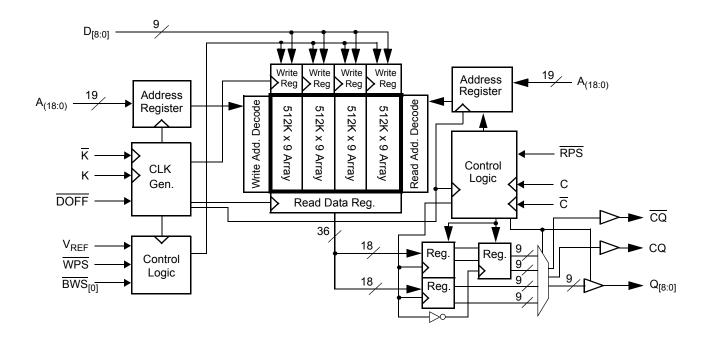


# Logic Block Diagram (CY7C1311BV18)



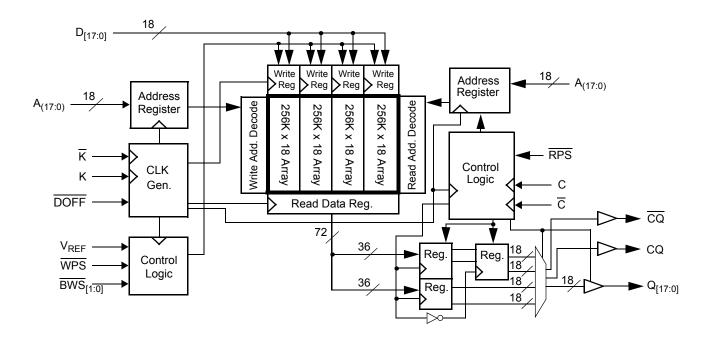
## Logic Block Diagram (CY7C1911BV18)



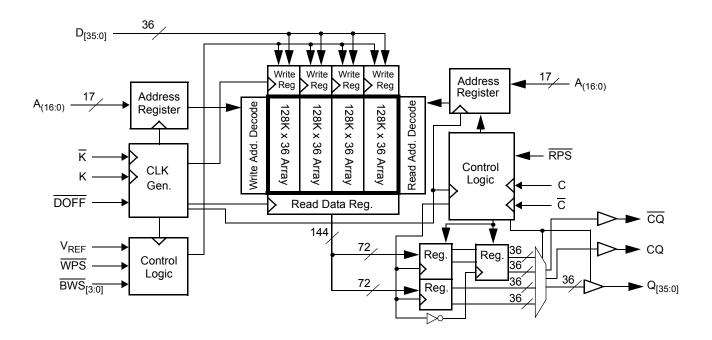
Document Number: 38-05620 Rev. \*F Page 2 of 32



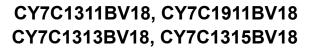
# Logic Block Diagram (CY7C1313BV18)



# Logic Block Diagram (CY7C1315BV18)



Document Number: 38-05620 Rev. \*F Page 3 of 32





### **Contents**

Pin Configuration	5
Pin Definitions	7
Functional Overview	9
Read Operations	9
Write Operations	9
Byte Write Operations	9
Single Clock Mode	
Concurrent Transactions	10
Depth Expansion	10
Programmable Impedance	10
Echo Clocks	10
DLL	10
Application Example	11
Truth Table	11
Write Cycle Descriptions	12
Write Cycle Descriptions	12
Write Cycle Descriptions	13
IEEE 1149.1 Serial Boundary Scan (JTAG)	14
Disabling the JTAG Feature	14
Test Access Port—Test Clock	
Test Mode Select (TMS)	14
Test Data-In (TDI)	
Test Data-Out (TDO)	
Performing a TAP Reset	14
TAP Registers	
TAP Instruction Set	
TAP Controller State Diagram	16
TAP Controller Block Diagram	17

TAP Electrical Characteristics	17
TAP AC Switching Characteristics	
TAP Timing and Test Conditions	
Identification Register Definitions	
Scan Register Sizes	
Instruction Codes	
Boundary Scan Order	20
Power Up Sequence in QDR-II SRAM	
Power Up Sequence	
DLL Constraints	
Maximum Ratings	22
Operating Range	
Electrical Characteristics	
DC Electrical Characteristics	22
AC Electrical Characteristics	23
Capacitance	
Thermal Resistance	24
Switching Characteristics	25
Switching Waveforms	
Ordering Information	28
Ordering Code Definitions	28
Package Diagram	29
Document History Page	30
Sales, Solutions, and Legal Information	32
Worldwide Sales and Design Support	32
Products	32
PSoC Solutions	32

Document Number: 38-05620 Rev. \*F Page 4 of 32

10

NC/36M

NC

NC

NC

D2

NC

NC

 $V_{REF}$ 

Q1

NC

NC

NC

NC

NC

**TMS** 

11

CQ

Q3

D3

NC

Q2

NC

NC

ZQ

D1

NC

Q0

D0

NC

NC

TDI



### Pin Configuration

Н

J

Κ

L

M

N

Р

R

DOFF

NC

NC

NC

NC

NC

NC

TDO

 $V_{REF}$ 

NC

NC

Q6

NC

D7

NC

**TCK** 

 $V_{DDQ}$ 

NC

NC

D6

NC

NC

Q7

Α

The pin configuration for CY7C1311BV18, CY7C1911BV18, CY7C1313BV18, and CY7C1315BV18 follow. [1]

### 165-Ball FBGA (13 x 15 x 1.4 mm) Pinout CY7C1311BV18 (2M x 8)

				C	17013110	V 10 (21VI X	0)			
	1	2	3	4	5	6	7	8	9	Ī
Α	CQ	NC/72M	Α	WPS	NWS <sub>1</sub>	K	NC/144M	RPS	Α	Ī
В	NC	NC	NC	А	NC/288M	K	NWS <sub>0</sub>	Α	NC	Ī
С	NC	NC	NC	$V_{SS}$	Α	NC	Α	$V_{SS}$	NC	Ī
D	NC	D4	NC	$V_{SS}$	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	NC	Ī
E	NC	NC	Q4	$V_{DDQ}$	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{DDQ}$	NC	Ī
F	NC	NC	NC	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>	$V_{DD}$	$V_{DDQ}$	NC	ĺ
G	NC	D5	Q5	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>	$V_{DD}$	$V_{DDQ}$	NC	ĺ

 $V_{DD}$ 

 $V_{DD}$ 

 $V_{DD}$ 

 $V_{SS}$ 

 $V_{SS}$ 

Α

Α

Α

 $V_{SS}$ 

 $V_{SS}$ 

 $V_{SS}$ 

 $V_{SS}$ 

 $V_{SS}$ 

Α

С

С

 $V_{DD}$ 

 $V_{DD}$ 

 $V_{DD}$ 

 $V_{SS}$ 

 $V_{SS}$ 

Α

Α

Α

 $V_{DDQ}$ 

 $V_{DDQ}$ 

 $V_{DDQ}$ 

 $V_{DDQ}$ 

 $V_{SS}$ 

 $V_{SS}$ 

Α

 $V_{DDQ}$ 

NC

NC

NC

NC

NC

NC

Α

 $V_{DDQ}$ 

 $V_{DDQ}$ 

 $V_{DDQ}$ 

 $V_{DDQ}$ 

 $V_{SS}$ 

 $V_{SS}$ 

Α

Α

# CY7C1911BV18 (2M x 9)

	1	2	3	4	5	6	7	8	9	10	11
Α	CQ	NC/72M	Α	WPS	NC	K	NC/144M	RPS	Α	NC/36M	CQ
В	NC	NC	NC	Α	NC/288M	K	BWS <sub>0</sub>	Α	NC	NC	Q4
С	NC	NC	NC	$V_{SS}$	Α	NC	Α	V <sub>SS</sub>	NC	NC	D4
D	NC	D5	NC	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	V <sub>SS</sub>	NC	NC	NC
E	NC	NC	Q5	$V_{DDQ}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	$V_{\mathrm{DDQ}}$	NC	D3	Q3
F	NC	NC	NC	$V_{DDQ}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	NC	NC
G	NC	D6	Q6	$V_{\mathrm{DDQ}}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	NC	NC
Н	DOFF	$V_{REF}$	$V_{\mathrm{DDQ}}$	$V_{DDQ}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{\mathrm{DDQ}}$	$V_{\mathrm{DDQ}}$	$V_{REF}$	ZQ
J	NC	NC	NC	$V_{DDQ}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	Q2	D2
K	NC	NC	NC	$V_{DDQ}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	NC	NC
L	NC	Q7	D7	$V_{DDQ}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	$V_{\mathrm{DDQ}}$	NC	NC	Q1
M	NC	NC	NC	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	$V_{SS}$	NC	NC	D1
N	NC	D8	NC	V <sub>SS</sub>	Α	Α	Α	V <sub>SS</sub>	NC	NC	NC
Р	NC	NC	Q8	А	Α	С	Α	Α	NC	D0	Q0
R	TDO	TCK	Α	А	Α	C	Α	Α	Α	TMS	TDI

Document Number: 38-05620 Rev. \*F Page 5 of 32

<sup>1.</sup> NC/36M, NC/72M, NC/144M, and NC/288M are not connected to the die and can be tied to any voltage level.



# Pin Configuration (continued)

The pin configuration for CY7C1311BV18, CY7C1911BV18, CY7C1313BV18, and CY7C1315BV18 follow. [1]

# 165-Ball FBGA (13 x 15 x 1.4 mm) Pinout CY7C1313BV18 (1M x 18)

	1	2	3	4	5	6	7	8	9	10	11
Α	CQ	NC/144M	NC/36M	WPS	BWS <sub>1</sub>	K	NC/288M	RPS	Α	NC/72M	CQ
В	NC	Q9	D9	Α	NC	K	BWS <sub>0</sub>	Α	NC	NC	Q8
С	NC	NC	D10	$V_{SS}$	Α	NC	Α	$V_{SS}$	NC	Q7	D8
D	NC	D11	Q10	$V_{SS}$	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	NC	NC	D7
E	NC	NC	Q11	$V_{\mathrm{DDQ}}$	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{\mathrm{DDQ}}$	NC	D6	Q6
F	NC	Q12	D12	$V_{\mathrm{DDQ}}$	$V_{DD}$	V <sub>SS</sub>	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	NC	Q5
G	NC	D13	Q13	$V_{\mathrm{DDQ}}$	$V_{DD}$	V <sub>SS</sub>	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	NC	D5
Н	DOFF	$V_{REF}$	$V_{DDQ}$	$V_{\mathrm{DDQ}}$	$V_{DD}$	V <sub>SS</sub>	$V_{DD}$	$V_{\mathrm{DDQ}}$	$V_{\mathrm{DDQ}}$	$V_{REF}$	ZQ
J	NC	NC	D14	$V_{\mathrm{DDQ}}$	$V_{DD}$	V <sub>SS</sub>	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	Q4	D4
K	NC	NC	Q14	$V_{\mathrm{DDQ}}$	$V_{DD}$	V <sub>SS</sub>	$V_{DD}$	$V_{\mathrm{DDQ}}$	NC	D3	Q3
L	NC	Q15	D15	$V_{\mathrm{DDQ}}$	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{\mathrm{DDQ}}$	NC	NC	Q2
М	NC	NC	D16	$V_{SS}$	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	$V_{SS}$	NC	Q1	D2
N	NC	D17	Q16	$V_{SS}$	Α	Α	Α	$V_{SS}$	NC	NC	D1
Р	NC	NC	Q17	Α	Α	С	Α	Α	NC	D0	Q0
R	TDO	TCK	Α	Α	Α	C	А	Α	Α	TMS	TDI

#### CY7C1315BV18 (512K x 36)

	1	2	3	4	5	6	7	8	9	10	11
Α	CQ	NC/288M	NC/72M	WPS	BWS <sub>2</sub>	K	BWS <sub>1</sub>	RPS	NC/36M	NC/144M	CQ
В	Q27	Q18	D18	Α	BWS <sub>3</sub>	K	BWS <sub>0</sub>	Α	D17	Q17	Q8
С	D27	Q28	D19	$V_{SS}$	Α	NC	Α	V <sub>SS</sub>	D16	Q7	D8
D	D28	D20	Q19	$V_{SS}$	$V_{SS}$	$V_{SS}$	$V_{SS}$	V <sub>SS</sub>	Q16	D15	D7
E	Q29	D29	Q20	$V_{\mathrm{DDQ}}$	$V_{SS}$	$V_{SS}$	$V_{SS}$	$V_{DDQ}$	Q15	D6	Q6
F	Q30	Q21	D21	$V_{\mathrm{DDQ}}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	D14	Q14	Q5
G	D30	D22	Q22	$V_{\mathrm{DDQ}}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	Q13	D13	D5
Н	DOFF	$V_{REF}$	$V_{DDQ}$	$V_{\mathrm{DDQ}}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	$V_{DDQ}$	$V_{REF}$	ZQ
J	D31	Q31	D23	$V_{\mathrm{DDQ}}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	D12	Q4	D4
K	Q32	D32	Q23	$V_{\mathrm{DDQ}}$	$V_{DD}$	$V_{SS}$	$V_{DD}$	$V_{DDQ}$	Q12	D3	Q3
L	Q33	Q24	D24	$V_{\mathrm{DDQ}}$	$V_{SS}$	$V_{SS}$	$V_{SS}$	$V_{DDQ}$	D11	Q11	Q2
М	D33	Q34	D25	$V_{SS}$	$V_{SS}$	$V_{SS}$	$V_{SS}$	V <sub>SS</sub>	D10	Q1	D2
N	D34	D26	Q25	$V_{SS}$	Α	Α	Α	V <sub>SS</sub>	Q10	D9	D1
Р	Q35	D35	Q26	Α	Α	С	Α	Α	Q9	D0	Q0
R	TDO	TCK	Α	Α	Α	С	Α	Α	Α	TMS	TDI

Document Number: 38-05620 Rev. \*F Page 6 of 32



# **Pin Definitions**

Pin Name	Ю	Pin Description
D <sub>[x:0]</sub>	Input- Synchronous	Data Input Signals. Sampled on the rising edge of K and $\overline{K}$ clocks during valid write operations. CY7C1311BV18 – $D_{[7:0]}$ CY7C1911BV18 – $D_{[8:0]}$ CY7C1313BV18 – $D_{[17:0]}$ CY7C1315BV18 – $D_{[35:0]}$
WPS	Input- Synchronous	Write Port Select – Active LOW. Sampled on the rising edge of the K clock. When asserted active, a write operation is initiated. Deasserting deselects the write port. Deselecting the write port ignores $D_{[x:0]}$ .
NWS <sub>0</sub> , NWS <sub>1</sub>	Input- Synchronous	Nibble Write Select 0, 1 – Active LOW (CY7C1311BV18 Only). Sampled on the rising edge of the K and $\overline{K}$ clocks during write operations. Used to select which nibble is written into the device during the $\overline{K}$ current portion of the write $\overline{K}$ operations. Nibbles not written remain unaltered. $\overline{K}$ Controls $\overline{K}$ controls $\overline{K}$ controls $\overline{K}$ controls $\overline{K}$ controls $\overline{K}$ and $\overline{K}$ write Selects are sampled on the same edge as the data. Deselecting a Nibble Write Select ignores the corresponding nibble of data and it is not written into the device.
BWS <sub>0</sub> , BWS <sub>1</sub> , BWS <sub>2</sub> , BWS <sub>3</sub>	Input- Synchronous	Byte Write Select 0, 1, 2, and 3 – Active LOW. Sampled on the rising edge of the K and $\overline{K}$ clocks during write operations. Used to select which byte is written into the device during the current portion of the write operations. Bytes not written remain unaltered. $CY7C1911BV18 - \underline{BWS}_0 \text{ controls } D_{[8:0]} \underline{\qquad} CY7C1313BV18 - \underline{BWS}_0 \text{ controls } D_{[8:0]} \underline{\qquad} \underline{\qquad} \underline{\qquad} \underline{\qquad} \underline{\qquad} \underline{\qquad} \underline{\qquad} \qquad$
Α	Input- Synchronous	<b>Address Inputs</b> . Sampled on the rising edge of the K clock during active read and write operations. These address inputs are multiplexed for both read and write operations. Internally, the device is organized as 2M x 8 (4 arrays each of 512K x 8) for CY7C1311BV18, 2M x 9 (4 arrays each of 512K x 9) for CY7C1911BV18,1M x 18 (4 arrays each of 256K x 18) for CY7C1313BV18 and 512K x 36 (4 arrays each of 128K x 36) for CY7C1315BV18. Therefore, only 19 address inputs are needed to access the entire memory array of CY7C1311BV18 and CY7C1911BV18, 18 address inputs for CY7C1313BV18 and 17 address inputs for CY7C1315BV18. These inputs are ignored when the appropriate port is deselected.
Q <sub>[x:0]</sub>	Output- Synchronous	<b>Data Output Signals</b> . These pins drive out the requested data during a read operation. Valid data is driven out on the rising edge of both the C and $\overline{C}$ clocks during read operations, or K and K when in single clock mode. When the read port is deselected, $Q_{[x:0]}$ are automatically tri-stated. CY7C1311BV18 – $Q_{[7:0]}$ CY7C1911BV18 – $Q_{[8:0]}$ CY7C1313BV18 – $Q_{[17:0]}$ CY7C1315BV18 – $Q_{[35:0]}$
RPS	Input- Synchronous	<b>Read Port Select</b> – <b>Active LOW</b> . Sampled on the rising edge of positive input clock (K). When active, a read operation is initiated. Deasserting deselects the read port. When deselected, the pending access is allowed to complete and the output drivers are automatically tri-stated following the next rising edge of the C clock. Each read access consists of a burst of four sequential transfers.
С	Input Clock	<b>Positive Input Clock for Output Data</b> . C is used in conjunction with $\overline{C}$ to clock out the read data from the device. C and $\overline{C}$ can be used together to deskew the flight times of various devices on the board back to the controller. See Application Example on page 11 for further details.
C	Input Clock	<b>Negative Input Clock for Output Data</b> . $\overline{C}$ is used in conjunction with C to clock out the read data from the device. C and $\overline{C}$ can be used together to deskew the flight times of various devices on the board back to the controller. See Application Example on page 11 for further details.
K	Input Clock	<b>Positive Input Clock Input</b> . The rising edge of K is used to capture synchronous inputs to the device and to drive out data through $Q_{[x:0]}$ when in single clock mode. All accesses are initiated on the rising edge of K.
ĸ	Input Clock	<b>Negative Input Clock Input.</b> $\overline{K}$ is used to capture synchronous inputs being presented to the device and to drive out data through $Q_{[x:0]}$ when in single clock mode.

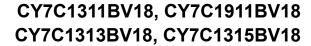
Document Number: 38-05620 Rev. \*F Page 7 of 32



# Pin Definitions (continued)

Pin Name	Ю	Pin Description
CQ	Echo Clock	<b>CQ Referenced with Respect to C</b> . This is a free-running clock and is synchronized to the input clock for output data (C) of the QDR-II. In single clock mode, CQ is generated with respect to K. The timing for the echo clocks is shown in Switching Characteristics on page 25.
CQ	Echo Clock	CQ Referenced with Respect to C. This is a free-running clock and is synchronized to the input clock for output data (C) of the QDR-II. In single clock mode, CQ is generated with respect to K. The timing for the echo clocks is shown in Switching Characteristics on page 25.
ZQ	Input	Output Impedance Matching input. This input is used to tune the device outputs to the system data bus impedance. CQ, CQ, and $Q_{[x:0]}$ output impedance are set to $0.2 \times RQ$ , where RQ is a resistor connected between ZQ and ground. Alternatively, this pin can be connected directly to $V_{DDQ}$ , which enables the minimum impedance mode. This pin cannot be connected directly to GND or left unconnected.
DOFF	Input	DLL Turn Off – Active LOW. Connecting this pin to ground turns off the DLL inside the device. The timings in the DLL turned off differs from those listed in this data sheet.
TDO	Output	TDO for JTAG.
TCK	Input	TCK Pin for JTAG.
TDI	Input	TDI Pin for JTAG.
TMS	Input	TMS Pin for JTAG.
NC	N/A	Not Connected to the Die. Can be tied to any voltage level.
NC/36M	N/A	Not Connected to the Die. Can be tied to any voltage level.
NC/72M	N/A	Not Connected to the Die. Can be tied to any voltage level.
NC/144M	N/A	Not Connected to the Die. Can be tied to any voltage level.
NC/288M	N/A	Not Connected to the Die. Can be tied to any voltage level.
V <sub>REF</sub>	Input- Reference	Reference Voltage Input. Static input used to set the reference level for HSTL inputs, outputs, and AC measurement points.
$V_{DD}$	Supply	Power Supply Inputs to the Core of the Device.
V <sub>SS</sub>	Ground	Ground for the Device.
$V_{\mathrm{DDQ}}$	Power Supply	Power Supply Inputs for the Outputs of the Device.

Document Number: 38-05620 Rev. \*F Page 8 of 32





#### **Functional Overview**

The CY7C1311BV18, CY7C1911BV18, CY7C1313BV18, CY7C1315BV18 are synchronous pipelined Burst SRAMs with a read port and a write port. The read port is dedicated to read operations and the write port is dedicated to write operations. Data flows into the SRAM through the write port and flows out through the read port. These devices multiplex the address inputs to minimize the number of address pins required. By having separate read and write ports, the QDR-II completely eliminates the need to "turn-around" the data bus and avoids any possible data contention, thereby simplifying system design. Each access consists of four 8-bit data transfers in the case of CY7C1311BV18, four 9-bit data transfers in the case of CY7C1313BV18, and four 36-bit data transfers in the case of CY7C1315BV18 in two clock cycles.

Accesses for both ports are initiated on the positive input clock (K). All synchronous input timing is referenced from the rising edge of the input clocks (K and  $\overline{K}$ ) and all output timing is referenced to the output clocks (C and  $\overline{C}$ , or K and  $\overline{K}$  when in single clock mode).

All synchronous data inputs  $(D_{[x:0]})$  pass through input registers controlled by the input clocks (K and  $\overline{K}$ ). All synchronous data outputs  $(Q_{[x:0]})$  pass through output registers controlled by the rising edge of the output clocks (C and  $\overline{C}$ , or K and  $\overline{K}$  when in single clock mode).

All synchronous control ( $\overline{RPS}$ ,  $\overline{WPS}$ ,  $\overline{BWS}_{[x:0]}$ ) inputs pass through input registers controlled by the rising edge of the input clocks (K and  $\overline{K}$ ).

CY7C1313BV18 is described in the following sections. The same basic descriptions apply to CY7C1311BV18, CY7C1911BV18, and CY7C1315BV18.

#### **Read Operations**

The CY7C1313BV18 is organized internally as four arrays of 256K x 18. Accesses are completed in a burst of four sequential 18-bit data words. Read operations are initiated by asserting RPS active at the rising edge of the positive input clock (K). The address presented to address inputs is stored in the read address register. Following the next K clock rise, the corresponding lowest order 18-bit word of data is driven onto the Q<sub>[17:0]</sub> using C as the output timing reference. On the subsequent rising edge of C the next 18-bit data word is driven onto the Q<sub>[17:0]</sub>. This process continues until all four 18-bit data words have been driven out onto Q<sub>[17:0]</sub>. The requested data is valid 0.45 ns from the rising edge of the output clock (C or C, or K or K when in single clock mode). To maintain the internal logic, each read access must be allowed to complete. Each read access consists of four 18-bit data words and takes two clock cycles to complete. Therefore, read accesses to the device cannot be initiated on two consecutive K clock rises. The internal logic of the device ignores the second read request. Read accesses can be initiated on every other K clock rise. Doing so pipelines the data flow such that data is transferred out of the device on every rising edge of the output clocks (C and  $\overline{C}$ , or K and  $\overline{K}$  when in single clock mode).

When the read port is deselected, the CY7C1311BV18 first completes the pending read transactions. Synchronous internal circuitry automatically tri-states the outputs following the next rising edge of the positive output clock (C). This enables for a seamless transition between devices without the insertion of wait states in a depth expanded memory.

#### **Write Operations**

Write operations are initiated by asserting WPS active at the rising edge of the positive input clock (K). On the following K clock rise the data presented to D<sub>[17:0]</sub> is latched and stored into the lower 18-bit write data register, provided BWS<sub>[1:0]</sub> are both asserted active. On the subsequent rising edge of the negative input clock  $(\overline{K})$  the information presented to  $D_{[17:0]}$  is also stored in the write data register, provided BWS<sub>[1:0]</sub> are both asserted active. This process continues for one more cycle until four 18-bit words (a total of 72 bits) of data are stored in the SRAM. The 72 bits of data are then written into the memory array at the specified location. Therefore, write accesses to the device cannot be initiated on two consecutive K clock rises. The internal logic of the device ignores the second write request. Write accesses can be initiated on every other rising edge of the positive input clock (K). Doing so pipelines the data flow such that 18 bits of data can be transferred into the device on every rising edge of the input clocks (K and K).

When deselected, the write port ignores all inputs after the pending write operations have been completed.

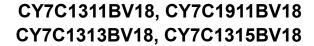
#### **Byte Write Operations**

Byte write operations are supported by the CY7C1311BV18. A write operation is initiated as described in the Write Operations section. The bytes that are written are determined by  $BWS_0$  and  $\overline{BWS}_1$ , which are sampled with each set of 18-bit data words. Asserting the appropriate Byte Write Select input during the data portion of a write latches the data being presented and writes it into the device. Deasserting the Byte Write Select input during the data portion of a write enables the data stored in the device for that byte to remain unaltered. This feature can be used to simplify read, modify, or write operations to a byte write operation.

#### Single Clock Mode

The CY7C1311BV18 can be used with a single clock that controls both the input and output registers. In this mode the device recognizes only a single pair of input clocks (K and  $\overline{K}$ ) that control both the input and output registers. This operation is identical to the operation if the device had zero skew between the K/K and C/ $\overline{C}$  clocks. All timing parameters remain the same in this mode. To use this mode of operation, the user must tie C and  $\overline{C}$  HIGH at power on. This function is a strap option and not alterable during device operation.

Document Number: 38-05620 Rev. \*F Page 9 of 32





#### **Concurrent Transactions**

The read and write ports on the CY7C1311BV18 operate completely independently of one another. As each port latches the address inputs on different clock edges, the user can read or write to any location, regardless of the transaction on the other port. If the ports access the same location when a read follows a write in successive clock cycles, the SRAM delivers the most recent information associated with the specified address location. This includes forwarding data from a write cycle that was initiated on the previous K clock rise.

Read accesses and write access must be scheduled such that one transaction is initiated on any clock cycle. If both ports are selected on the same K clock rise, the arbitration depends on the previous state of the SRAM. If both ports were deselected, the read port takes priority. If a read was initiated on the previous cycle, the write port takes priority (as read operations cannot be initiated on consecutive cycles). If a write was initiated on the previous cycle, the read port takes priority (as write operations cannot be initiated on consecutive cycles). Therefore, asserting both port selects active from a deselected state results in alternating read or write operations being initiated, with the first access being a read.

#### **Depth Expansion**

The CY7C1311BV18 has a port select input for each port. This enables for easy depth expansion. Both port selects are sampled on the rising edge of the positive input clock only (K). Each port select input can deselect the specified port. Deselecting a port does not affect the other port. All pending transactions (read and write) are completed before the device is deselected.

#### **Programmable Impedance**

An external resistor, RQ, must be connected between the ZQ pin on the SRAM and  $V_{SS}$  to allow the SRAM to adjust its output driver impedance. The value of RQ must be 5X the value of the intended line impedance driven by the SRAM. The allowable range of RQ to guarantee impedance matching with a tolerance of  $\pm 15\%$  is between  $175\Omega$  and  $350\Omega$ , with  $V_{DDQ}$  = 1.5V. The output impedance is adjusted every 1024 cycles upon power up to account for drifts in supply voltage and temperature.

#### **Echo Clocks**

Echo clocks are provided on the QDR-II to simplify data capture on high-speed systems. Two echo clocks are generated by the QDR-II. CQ is referenced with respect to C and  $\overline{\text{CQ}}$  is referenced with respect to  $\overline{\text{C}}$ . These are free-running clocks and are synchronized to the output clock of the QDR-II. In single clock mode, CQ is generated with respect to K and  $\overline{\text{CQ}}$  is generated with respect to  $\overline{\text{K}}$ . The timing for the echo clocks is shown in the Switching Characteristics on page 25.

#### **DLL**

These chips use a DLL that is designed to function between 120 MHz and the specified maximum clock frequency. During power up, when the  $\overline{\text{DOFF}}$  is tied HIGH, the DLL is locked after 1024 cycles of stable clock. The DLL can also be reset by slowing or stopping the input clock K and K for a minimum of 30 ns. However, it is not necessary to reset the DLL to lock to the desired frequency. The DLL automatically locks 1024 clock cycles after a stable clock is presented. The DLL may be disabled by applying ground to the  $\overline{\text{DOFF}}$  pin. For information refer to the application note AN5062, DLL Considerations in QDRII/DDRII/QDRII+/DDRII+.

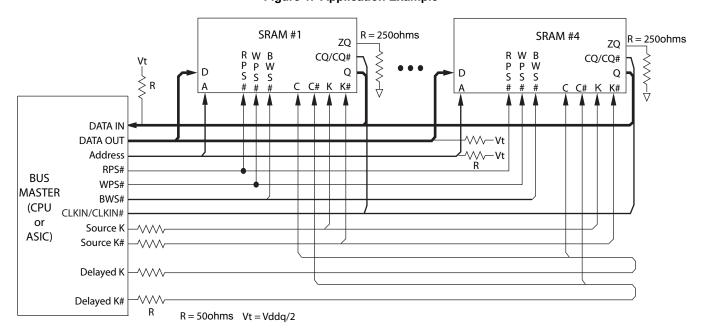
Document Number: 38-05620 Rev. \*F Page 10 of 32



#### Application Example

Figure 1 shows four QDR-II used in an application.

Figure 1. Application Example



#### Truth Table

The truth table for CY7C1311BV18, CY7C1911BV18, CY7C1313BV18, and CY7C1315BV18 follows. [2, 3, 4, 5, 6, 7]

Operation	K	RPS	WPS	DQ	DQ	DQ	DQ
Write Cycle: Load address on the rising edge of K; write data on_ two consecutive K and K rising edges.	L-H	H <sup>[8]</sup>	L [9]	D(A) at K(t + 1)↑	D(A + 1) at K(t + 1)↑	D(A + 2) at K(t + 2)↑	D(A + 3) at $\overline{K}$ (t + 2) $\uparrow$
Read Cycle: Load address on the rising edge of K; wait one and a half cycle; read data on two consecutive C and C rising edges.	다	L [9]	X	Q(A) at <del>C</del> (t + 1)↑	Q(A + 1) at C(t + 2)1	Q(A + 2) at $\overline{C}$ (t + 2) $\uparrow$	Q(A + 3) at C(t + 3)↑
NOP: No Operation	L-H	Ι	Н	D = X Q = High-Z	D = X Q = High-Z	D = X Q = High-Z	D = X Q = High-Z
Standby: Clock Stopped	Stopped	Х	Х	Previous State	Previous State	Previous State	Previous State

- 2. X = "Don't Care," H = Logic HIGH, L = Logic LOW, ↑represents rising edge.
- Device powers up deselected with the outputs in a tri-state condition.
- "A" represents address location latched by the devices when transaction was initiated. A + 1, A + 2, and A +3 represents the address sequence in the burst.
- "t" represents the cycle at which a read/write operation is started. t + 1, t + 2, and t + 3 are the first, second and third clock cycles respectively succeeding the "t" clock cycle.
- Data inputs are registered at K and K rising edges. Data outputs are delivered on C and  $\overline{C}$  rising edges, except when in single clock mode. It is recommended that  $K = \overline{K}$  and  $C = \overline{C} = HIGH$  when clock is stopped. This is not essential, but permits most rapid restart by overcoming transmission line charging
- If this signal was LOW to initiate the previous cycle, this signal becomes a "Don't Care" for this operation.
- This signal was HIGH on previous K clock rise. Initiating consecutive read or write operations on consecutive K clock rises is not permitted. The device ignores the second read or write request.

Document Number: 38-05620 Rev. \*F Page 11 of 32



# **Write Cycle Descriptions**

The write cycle description table for CY7C1311BV18 and CY7C1313BV18 follows. [2, 10]

$\frac{\overline{\text{BWS}}_0}{\text{NWS}_0}$	BWS <sub>1</sub> / NWS <sub>1</sub>	К	ĸ	Comments
L	L	L–H	_	During the data portion of a write sequence: CY7C1311BV18 – both nibbles ( $D_{[7:0]}$ ) are written into the device. CY7C1313BV18 – both bytes ( $D_{[17:0]}$ ) are written into the device.
L	L	-		During the data portion of a write sequence: CY7C1311BV18 – both nibbles ( $D_{[7:0]}$ ) are written into the device. CY7C1313BV18 – both bytes ( $D_{[17:0]}$ ) are written into the device.
L	Н	L–H		During the data portion of a write sequence: CY7C1311BV18 – only the lower nibble ( $D_{[3:0]}$ ) is written into the device, $D_{[7:4]}$ remains unaltered. CY7C1313BV18 – only the lower byte ( $D_{[8:0]}$ ) is written into the device, $D_{[17:9]}$ remains unaltered.
L	Н	-	L–H	During the data portion of a write sequence: CY7C1311BV18 – only the lower nibble $(D_{[3:0]})$ is written into the device, $D_{[7:4]}$ remains unaltered. CY7C1313BV18 – only the lower byte $(D_{[8:0]})$ is written into the device, $D_{[17:9]}$ remains unaltered.
Н	L	L–H	_	During the data portion of a write sequence: CY7C1311BV18 – only the upper nibble ( $D_{[7:4]}$ ) is written into the device, $D_{[3:0]}$ remains unaltered. CY7C1313BV18 – only the upper byte ( $D_{[17:9]}$ ) is written into the device, $D_{[8:0]}$ remains unaltered.
Н	L	-	L–H	During the data portion of a write sequence: CY7C1311BV18 – only the upper nibble ( $D_{[7:4]}$ ) is written into the device, $D_{[3:0]}$ remains unaltered. CY7C1313BV18 – only the upper byte ( $D_{[17:9]}$ ) is written into the device, $D_{[8:0]}$ remains unaltered.
Н	Н	L–H	1	No data is written into the devices during this portion of a write operation.
Н	Н	-	L–H	No data is written into the devices during this portion of a write operation.

**Write Cycle Descriptions**The write cycle description table for CY7C1911BV18 follows. <sup>[2, 10]</sup>

BWS <sub>0</sub>	K	K	Comments					
L	L–H	-	During the data portion of a write sequence, the single byte (D <sub>[8:0]</sub> ) is written into the device.					
L	-	L–H	During the data portion of a write sequence, the single byte (D <sub>[8:0]</sub> ) is written into the device.					
Н	L–H	-	No data is written into the device during this portion of a write operation.					
Н	1	L–H	No data is written into the device during this portion of a write operation.					

Document Number: 38-05620 Rev. \*F Page 12 of 32

<sup>10.</sup> Is based on a write cycle that was initiated in accordance with the Write Cycle Descriptions table. NWS<sub>0</sub>, NWS<sub>1</sub>, BWS<sub>0</sub>, BWS<sub>1</sub>, BWS<sub>2</sub>, and BWS<sub>3</sub> can be altered on different portions of a write cycle, as long as the setup and hold requirements are achieved.



Write Cycle Descriptions The write cycle description table for CY7C1315BV18 follows.  $^{[2,\ 10]}$ 

BWS <sub>0</sub>	BWS <sub>1</sub>	BWS <sub>2</sub>	BWS <sub>3</sub>	K	K	Comments
L	L	L	L	L–H	1	During the data portion of a write sequence, all four bytes ( $D_{[35:0]}$ ) are written into the device.
L	L	L	L	_	L–H During the data portion of a write sequence, all four bytes (D <sub>[35:0]</sub> ) are writted the device.	
L	Η	Н	Н	L–H	ı	During the data portion of a write sequence, only the lower byte $(D_{[8:0]})$ is written into the device. $D_{[35:9]}$ remains unaltered.
L	Η	Н	Н	-	L–H	During the data portion of a write sequence, only the lower byte $(D_{[8:0]})$ is written into the device. $D_{[35:9]}$ remains unaltered.
Н	L	Н	Н	L–H	ı	During the data portion of a write sequence, only the byte $(D_{[17:9]})$ is written into the device. $D_{[8:0]}$ and $D_{[35:18]}$ remains unaltered.
Н	L	Н	Н	_	L–H	During the data portion of a write sequence, only the byte $(D_{[17:9]})$ is written into the device. $D_{[8:0]}$ and $D_{[35:18]}$ remains unaltered.
Н	Н	L	Н	L–H	-	During the data portion of a write sequence, only the byte $(D_{[26:18]})$ is written into the device. $D_{[17:0]}$ and $D_{[35:27]}$ remains unaltered.
Н	Н	L	Н	_	L–H	During the data portion of a write sequence, only the byte $(D_{[26:18]})$ is written into the device. $D_{[17:0]}$ and $D_{[35:27]}$ remains unaltered.
Н	Н	Н	L	L–H	-	During the data portion of a write sequence, only the byte $(D_{[35:27]})$ is written into the device. $D_{[26:0]}$ remains unaltered.
Н	Н	Н	L	-	L–H	During the data portion of a write sequence, only the byte $(D_{[35:27]})$ is written into the device. $D_{[26:0]}$ remains unaltered.
Н	Н	Н	Н	L–H	_	No data is written into the device during this portion of a write operation.
Н	Н	Н	Н	_	L–H	No data is written into the device during this portion of a write operation.

Document Number: 38-05620 Rev. \*F Page 13 of 32



### IEEE 1149.1 Serial Boundary Scan (JTAG)

These SRAMs incorporate a serial boundary scan Test Access Port (TAP) in the FBGA package. This part is fully compliant with IEEE Standard #1149.1-1900. The TAP operates using JEDEC standard 1.8V IO logic levels.

#### Disabling the JTAG Feature

It is possible to operate the SRAM without using the JTAG feature. To disable the TAP controller, TCK must be tied LOW ( $V_{SS}$ ) to prevent clocking of the device. TDI and TMS are internally pulled up and may be unconnected. They may alternatively be connected to  $V_{DD}$  through a pull up resistor. TDO must be left unconnected. Upon power up, the device comes up in a reset state, which does not interfere with the operation of the device.

#### **Test Access Port—Test Clock**

The test clock is used only with the TAP controller. All inputs are captured on the rising edge of TCK. All outputs are driven from the falling edge of TCK.

#### Test Mode Select (TMS)

The TMS input is used to give commands to the TAP controller and is sampled on the rising edge of TCK. This pin may be left unconnected if the TAP is not used. The pin is pulled up internally, resulting in a logic HIGH level.

#### Test Data-In (TDI)

The TDI pin is used to serially input information into the registers and can be connected to the input of any of the registers. The register between TDI and TDO is chosen by the instruction that is loaded into the TAP instruction register. For information on loading the instruction register, see the TAP Controller State Diagram on page 16. TDI is internally pulled up and can be unconnected if the TAP is unused in an application. TDI is connected to the most significant bit (MSB) on any register.

#### Test Data-Out (TDO)

The TDO output pin is used to serially clock data out from the registers. The output is active, depending upon the current state of the TAP state machine (see Instruction Codes on page 19). The output changes on the falling edge of TCK. TDO is connected to the least significant bit (LSB) of any register.

#### Performing a TAP Reset

A Reset is performed by forcing TMS HIGH ( $V_{DD}$ ) for five rising edges of TCK. This Reset does not affect the operation of the SRAM and can be performed while the SRAM is operating. At power up, the TAP is reset internally to ensure that TDO comes up in a high-Z state.

#### **TAP Registers**

Registers are connected between the TDI and TDO pins to scan the data in and out of the SRAM test circuitry. Only one register can be selected at a time through the instruction registers. Data is serially loaded into the TDI pin on the rising edge of TCK. Data is output on the TDO pin on the falling edge of TCK.

#### **IDCODE**

The IDCODE instruction loads a vendor-specific, 32-bit code into

#### Instruction Register

Three-bit instructions can be serially loaded into the instruction register. This register is loaded when it is placed between the TDI and TDO pins, as shown in TAP Controller Block Diagram on page 17. Upon power up, the instruction register is loaded with the IDCODE instruction. It is also loaded with the IDCODE instruction if the controller is placed in a reset state, as described in the previous section.

When the TAP controller is in the Capture-IR state, the two least significant bits are loaded with a binary "01" pattern to allow for fault isolation of the board level serial test path.

#### Bypass Register

To save time when serially shifting data through registers, it is sometimes advantageous to skip certain chips. The bypass register is a single-bit register that can be placed between TDI and TDO pins. This enables shifting of data through the SRAM with minimal delay. The bypass register is set LOW ( $V_{SS}$ ) when the BYPASS instruction is executed.

#### Boundary Scan Register

The boundary scan register is connected to all of the input and output pins on the SRAM. Several No Connect (NC) pins are also included in the scan register to reserve pins for higher density devices.

The boundary scan register is loaded with the contents of the RAM input and output ring when the TAP controller is in the Capture-DR state and is then placed between the TDI and TDO pins when the controller is moved to the Shift-DR state. The EXTEST, SAMPLE/PRELOAD, and SAMPLE Z instructions can be used to capture the contents of the input and output ring.

The Boundary Scan Order on page 20 shows the order in which the bits are connected. Each bit corresponds to one of the bumps on the SRAM package. The MSB of the register is connected to TDI, and the LSB is connected to TDO.

#### Identification (ID) Register

The ID register is loaded with a vendor-specific, 32-bit code during the Capture-DR state when the IDCODE command is loaded in the instruction register. The IDCODE is hardwired into the SRAM and can be shifted out when the TAP controller is in the Shift-DR state. The ID register has a vendor code and other information described in Identification Register Definitions on page 19.

#### **TAP Instruction Set**

Eight different instructions are possible with the three-bit instruction register. All combinations are listed in Instruction Codes on page 19. Three of these instructions are listed as RESERVED and must not be used. The other five instructions are described in detail below.

Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO pins. To execute the instruction after it is shifted in, the TAP controller must be moved into the Update-IR state.

the instruction register. It also places the instruction register between the TDI and TDO pins and shifts the IDCODE out of the

Document Number: 38-05620 Rev. \*F Page 14 of 32



# CY7C1311BV18, CY7C1911BV18 CY7C1313BV18, CY7C1315BV18

device when the TAP controller enters the Shift-DR state. The IDCODE instruction is loaded into the instruction register at power up or whenever the TAP controller is supplied a Test-Logic-Reset state.

#### SAMPLE Z

The SAMPLE Z instruction connects the boundary scan register between the TDI and TDO pins when the TAP controller is in a Shift-DR state. The SAMPLE Z command puts the output bus into a High-Z state until the next command is supplied during the Update IR state.

#### SAMPLE/PRELOAD

SAMPLE/PRELOAD is a 1149.1 mandatory instruction. When the SAMPLE/PRELOAD instructions are loaded into the instruction register and the TAP controller is in the Capture-DR state, a snapshot of data on the input and output pins is captured in the boundary scan register.

The user must be aware that the TAP controller clock can only operate at a frequency up to 20 MHz, while the SRAM clock operates more than an order of magnitude faster. Because there is a large difference in the clock frequencies, it is possible that during the Capture-DR state, an input or output undergoes a transition. The TAP may then try to capture a signal while in transition (metastable state). This does not harm the device, but there is no guarantee as to the value that is captured. Repeatable results may not be possible.

To guarantee that the boundary scan register captures the correct value of a signal, the SRAM signal must be stabilized long enough to meet the TAP controller's capture setup plus hold times ( $t_{CS}$  and  $t_{CH}$ ). The SRAM clock input might not be captured correctly if there is no way in a design to stop (or slow) the clock during a SAMPLE/PRELOAD instruction. If this is an issue, it is still possible to capture all other signals and simply ignore the value of the CK and  $\overline{CK}$  captured in the boundary scan register.

After the data is captured, it is possible to shift out the data by putting the TAP into the Shift-DR state. This places the boundary scan register between the TDI and TDO pins.

PRELOAD places an initial data pattern at the latched parallel outputs of the boundary scan register cells before the selection of another boundary scan test operation.

The shifting of data for the SAMPLE and PRELOAD phases can occur concurrently when required, that is, while the data captured is shifted out, the preloaded data can be shifted in.

#### **BYPASS**

When the BYPASS instruction is loaded in the instruction register and the TAP is placed in a Shift-DR state, the bypass register is placed between the TDI and TDO pins. The advantage of the BYPASS instruction is that it shortens the boundary scan path when multiple devices are connected together on a board.

#### **EXTEST**

The EXTEST instruction drives the preloaded data out through the system output pins. This instruction also connects the boundary scan register for serial access between the TDI and TDO in the Shift-DR controller state.

#### EXTEST OUTPUT BUS TRI-STATE

IEEE Standard 1149.1 mandates that the TAP controller be able to put the output bus into a tri-state mode.

The boundary scan register has a special bit located at bit #47. When this scan cell, called the "extest output bus tri-state," is latched into the preload register during the Update-DR state in the TAP controller, it directly controls the state of the output (Q-bus) pins, when the EXTEST is entered as the current instruction. When HIGH, it enables the output buffers to drive the output bus. When LOW, this bit places the output bus into a High-Z condition.

This bit can be set by entering the SAMPLE/PRELOAD or EXTEST command, and then shifting the desired bit into that cell, during the Shift-DR state. During Update-DR, the value loaded into that shift-register cell latches into the preload register. When the EXTEST instruction is entered, this bit directly controls the output Q-bus pins. Note that this bit is pre-set HIGH to enable the output when the device is powered up, and also when the TAP controller is in the Test-Logic-Reset state.

#### Reserved

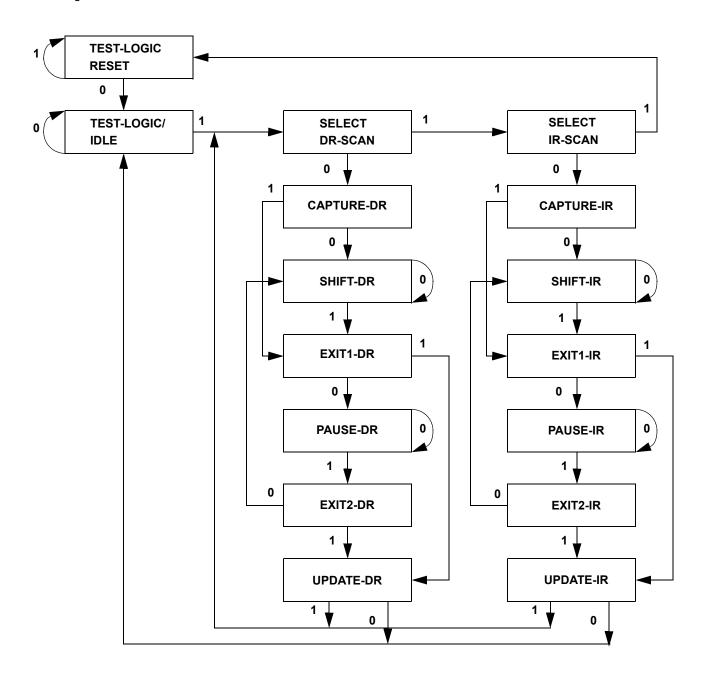
These instructions are not implemented but are reserved for future use. Do not use these instructions.

Document Number: 38-05620 Rev. \*F Page 15 of 32



# **TAP Controller State Diagram**

The state diagram for the TAP controller follows. [11]



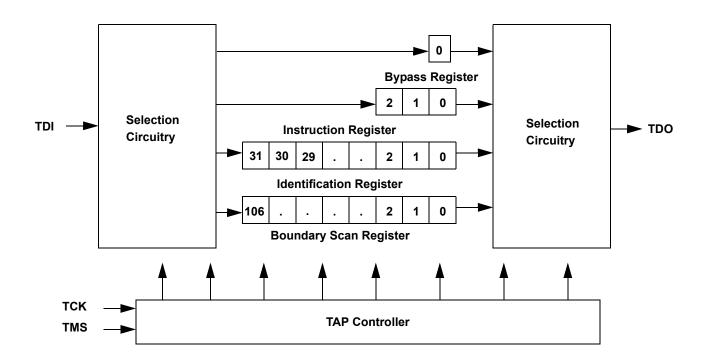
#### Note

11. The 0/1 next to each state represents the value at TMS at the rising edge of TCK.

Document Number: 38-05620 Rev. \*F Page 16 of 32



# **TAP Controller Block Diagram**



# **TAP Electrical Characteristics**

Over the Operating Range [12, 13, 14]

Parameter	Description	Test Conditions	Min	Max	Unit
V <sub>OH1</sub>	Output HIGH Voltage	I <sub>OH</sub> = -2.0 mA	1.4		V
V <sub>OH2</sub>	Output HIGH Voltage	I <sub>OH</sub> = -100 μA	1.6		V
V <sub>OL1</sub>	Output LOW Voltage	I <sub>OL</sub> = 2.0 mA		0.4	V
V <sub>OL2</sub>	Output LOW Voltage	I <sub>OL</sub> = 100 μA		0.2	V
V <sub>IH</sub>	Input HIGH Voltage		0.65V <sub>DD</sub>	V <sub>DD</sub> + 0.3	V
V <sub>IL</sub>	Input LOW Voltage		-0.3	0.35V <sub>DD</sub>	V
I <sub>X</sub>	Input and Output Load Current	$GND \leq V_I \leq V_{DD}$	<b>–</b> 5	5	μА

#### Notes

Document Number: 38-05620 Rev. \*F Page 17 of 32

<sup>12.</sup> These characteristics pertain to the TAP inputs (TMS, TCK, TDI and TDO). Parallel load levels are specified in the Electrical Characteristics Table.

<sup>13.</sup> Overshoot:  $V_{IH}(AC) < V_{DDQ} + 0.85V$  (Pulse width less than  $t_{CYC}/2$ ), Undershoot:  $V_{IL}(AC) > -1.5V$  (Pulse width less than  $t_{CYC}/2$ ). 14. All Voltage referenced to Ground.



# **TAP AC Switching Characteristics**

Over the Operating Range [15, 16]

Parameter	Description	Min	Max	Unit
t <sub>TCYC</sub>	TCK Clock Cycle Time	50		ns
t <sub>TF</sub>	TCK Clock Frequency		20	MHz
t <sub>TH</sub>	TCK Clock HIGH	20		ns
t <sub>TL</sub>	TCK Clock LOW	20		ns
Setup Times				
t <sub>TMSS</sub>	TMS Setup to TCK Clock Rise	5		ns
t <sub>TDIS</sub>	TDI Setup to TCK Clock Rise	5		ns
t <sub>CS</sub>	Capture Setup to TCK Rise	5		ns
Hold Times				
t <sub>TMSH</sub>	TMS Hold after TCK Clock Rise	5		ns
t <sub>TDIH</sub>	TDI Hold after Clock Rise	5		ns
t <sub>CH</sub>	Capture Hold after Clock Rise	5		ns
<b>Output Times</b>		•	•	
t <sub>TDOV</sub>	TCK Clock LOW to TDO Valid		10	ns
t <sub>TDOX</sub>	TCK Clock LOW to TDO Invalid	0		ns

# **TAP Timing and Test Conditions**

Figure 2 shows the TAP timing and test conditions. [16]

0.9V ALL INPUT PULSES 50Ω TDO -0V  $Z_0 = 50\Omega$  $C_{L} = 20 pF$ GND (a) **Test Clock** TCK  $t_{\mathsf{TMSS}}$ Test Mode Select **TMS** t<sub>TDIS</sub> t<sub>TDIH</sub> Test Data In TDI Test Data Out TDO

Figure 2. TAP Timing and Test Conditions

15.  $t_{CS}$  and  $t_{CH}$  refer to the setup and hold time requirements of latching data from the boundary scan register. 16. Test conditions are specified using the load in TAP AC Test Conditions.  $t_{R}/t_{F}$  = 1 ns.

Document Number: 38-05620 Rev. \*F Page 18 of 32



# **Identification Register Definitions**

Instruction Field		Description				
msu ucuon i leiu	CY7C1311BV18	CY7C1911BV18	CY7C1313BV18	CY7C1315BV18	Description	
Revision Number (31:29)	000	000	000	000	Version number.	
Cypress Device ID (28:12)	11010011011000101	11010011011001101	11010011011010101	11010011011100101	Defines the type of SRAM.	
Cypress JEDEC ID (11:1)	00000110100	00000110100	00000110100	00000110100	Allows unique identification of SRAM vendor.	
ID Register Presence (0)	1	1	1	1	Indicates the presence of an ID register.	

# **Scan Register Sizes**

Register Name	Bit Size
Instruction	3
Bypass	1
ID	32
Boundary Scan	107

# **Instruction Codes**

Instruction	Code	Description
EXTEST	000	Captures the input and output ring contents.
IDCODE	001	Loads the ID register with the vendor ID code and places the register between TDI and TDO. This operation does not affect SRAM operation.
SAMPLE Z	010	Captures the input and output contents. Places the boundary scan register between TDI and TDO. Forces all SRAM output drivers to a High-Z state.
RESERVED	011	Do not use: This instruction is reserved for future use.
SAMPLE/PRELOAD	100	Captures the input and output ring contents. Places the boundary scan register between TDI and TDO. Does not affect the SRAM operation.
RESERVED	101	Do not use: This instruction is reserved for future use.
RESERVED	110	Do not use: This instruction is reserved for future use.
BYPASS	111	Places the bypass register between TDI and TDO. This operation does not affect SRAM operation.

Document Number: 38-05620 Rev. \*F Page 19 of 32



# **Boundary Scan Order**

Bit#	Bump ID
0	6R
1	6P
2	6N
3	7P
4	7N
5	7R
6	8R
7	8P
8	9R
9	11P
10	10P
11	10N
12	9P
13	10M
14	11N
15	9M
16	9N
17	11L
18	11M
19	9L
20	10L
21	11K
22	10K
23	9J
24	9K
25	10J
26	11J

Bit#	Bump ID
27	11H
28	10G
29	9G
30	11F
31	11G
32	9F
33	10F
34	11E
35	10E
36	10D
37	9E
38	10C
39	11D
40	9C
41	9D
42	11B
43	11C
44	9B
45	10B
46	11A
47	Internal
48	9A
49	8B
50	7C
51	6C
52	8A
	7A

Bit #	Bump ID
54	7B
55	6B
56	6A
57	5B
58	5A
59	4A
60	5C
61	4B
62	3A
63	1H
64	1A
65	2B
66	3B
67	1C
68	1B
69	3D
70	3C
71	1D
72	2C
73	3E
74	2D
75	2E
76	1E
77	2F
78	3F
79	1G
80	1F

Bit #	Bump ID
81	3G
82	2G
83	1J
84	2J
85	3K
86	3J
87	2K
88	1K
89	2L
90	3L
91	1M
92	1L
93	3N
94	3M
95	1N
96	2M
97	3P
98	2N
99	2P
100	1P
101	3R
102	4R
103	4P
104	5P
105	5N
106	5R

Document Number: 38-05620 Rev. \*F Page 20 of 32



### Power Up Sequence in QDR-II SRAM

QDR-II SRAMs must be powered up and initialized in a predefined manner to prevent undefined operations.

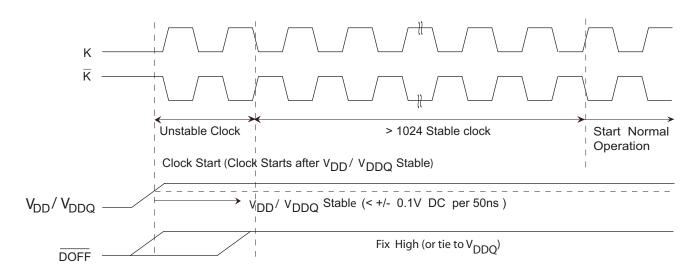
### **Power Up Sequence**

- Apply power and drive DOFF either HIGH or LOW (all other inputs can be HIGH or LOW).
  - $\square$  Apply  $V_{DD}$  before  $V_{DDQ}$ .
  - $\square$  Apply  $\underline{V_{DDQ}}$  before  $V_{REF}$  or at the same time as  $V_{REF}$ .
  - □ Drive DOFF HIGH.
- Provide stable DOFF (HIGH), power and clock (K, K) for 1024 cycles to lock the DLL.

#### **DLL Constraints**

- DLL uses K clock as its synchronizing input. The input must have low phase jitter, which is specified as t<sub>KC Var</sub>.
- The DLL functions at frequencies down to 120 MHz.
- If the input clock is unstable and the DLL is enabled, then the DLL may lock onto an incorrect frequency, causing unstable SRAM behavior. To avoid this, provide1024 cycles stable clock to relock to the desired clock frequency.

Figure 3. Power Up Waveforms



Document Number: 38-05620 Rev. \*F Page 21 of 32



# CY7C1311BV18, CY7C1911BV18 CY7C1313BV18, CY7C1315BV18

## **Maximum Ratings**

Exceeding maximum ratings may impair the useful life of the device. These user guidelines are not tested. Storage Temperature ......-65 C to +150 C Ambient Temperature with Power Applied -55 C to +125 C Supply Voltage on V<sub>DD</sub> Relative to GND ......–0.5V to +2.9V Supply Voltage on  $V_{DDQ}$  Relative to GND......–0.5V to + $V_{DD}$ DC Applied to Outputs in High-Z .......-0.5V to V<sub>DDQ</sub> + 0.3V DC Input Voltage [13]......-0.5V to V<sub>DD</sub> + 0.3V

Current into Outputs (LOW)	20 mA
Static Discharge Voltage (MIL-STD-883, M. 3015) >	> 2001V
Latch up Current >	200 mA

### **Operating Range**

Range	Ambient Temperature (T <sub>A</sub> )	<b>V</b> <sub>DD</sub> <sup>[17]</sup>	<b>V</b> <sub>DDQ</sub> [17]
Commercial	0 C to +70 C	1.8 ± 0.1V	1.4V to V <sub>DD</sub>
Industrial	–40°C to +85°C		

#### **Electrical Characteristics**

#### **DC Electrical Characteristics**

Over the Operating Range [14]

Parameter	Description	Test Cond	litions		Min	Тур	Max	Unit
$V_{DD}$	Power Supply Voltage				1.7	1.8	1.9	V
$V_{\mathrm{DDQ}}$	IO Supply Voltage				1.4	1.5	$V_{DD}$	V
V <sub>OH</sub>	Output HIGH Voltage	Note 18			$V_{DDQ}/2 - 0.12$		V <sub>DDQ</sub> /2 + 0.12	V
V <sub>OL</sub>	Output LOW Voltage	Note 19			$V_{DDQ}/2 - 0.12$		V <sub>DDQ</sub> /2 + 0.12	V
V <sub>OH(LOW)</sub>	Output HIGH Voltage	I <sub>OH</sub> = -0.1 mA, Nomina	I Impedance		V <sub>DDQ</sub> – 0.2		$V_{\mathrm{DDQ}}$	V
V <sub>OL(LOW)</sub>	Output LOW Voltage	I <sub>OL</sub> = 0.1 mA, Nominal	Impedance		V <sub>SS</sub>		0.2	V
V <sub>IH</sub>	Input HIGH Voltage				V <sub>REF</sub> + 0.1		V <sub>DDQ</sub> + 0.3	V
V <sub>IL</sub>	Input LOW Voltage				-0.3		V <sub>REF</sub> – 0.1	V
I <sub>X</sub>	Input Leakage Current	$GND \le V_I \le V_{DDQ}$			<b>-5</b>		5	μΑ
I <sub>OZ</sub>	Output Leakage Current	$GND \le V_I \le V_{DDQ}$ , Output Disabled			<b>-</b> 5		5	μΑ
V <sub>REF</sub>	Input Reference Voltage [20]	Typical Value = 0.75V			0.68	0.75	0.95	V
I <sub>DD</sub> <sup>[21]</sup>	$V_{DD}$ Operating Supply $V_{DD} = Max$ ,	V <sub>DD</sub> = Max,	300 MHz	(x8)			765	mA
		I <sub>OUT</sub> = 0 mA,		(x9)			800	
		$f = f_{MAX} = 1/t_{CYC}$		(x18)			840	
				(x36)			985	
			278 MHz	(x8)			720	mA
				(x9)			730	
				(x18)			760	
				(x36)			910	
			250 MHz	(x8)			665	mA
				(x9)			675	
				(x18)			705	
				(x36)			830	

#### Notes

- 17. Power up: Assumes a linear ramp from 0V to  $V_{DD}$ (min) within 200 ms. During this time  $V_{IH} < V_{DD}$  and  $V_{DDQ} \le V_{DD}$ .

  18. Output are impedance controlled.  $I_{OH} = -(V_{DDQ}/2)/(RQ/5)$  for values of 175 ohms <= RQ <= 350 ohms.

  19. Output are impedance controlled.  $I_{OL} = (V_{DDQ}/2)/(RQ/5)$  for values of 175 ohms <= RQ <= 350 ohms.

- $20. V_{REF}$  (min) = 0.68V or 0.46V<sub>DDQ</sub>, whichever is larger,  $V_{REF}$  (max) = 0.95V or 0.54V<sub>DDQ</sub>, whichever is smaller. 21. The operation current is calculated with 50% read cycle and 50% write cycle.

Document Number: 38-05620 Rev. \*F Page 22 of 32



# **Electrical Characteristics** (continued)

# **DC Electrical Characteristics** Over the Operating Range <sup>[14]</sup>

Parameter	Description	Test Condi	tions		Min	Тур	Max	Unit
I <sub>DD</sub> <sup>[21]</sup>	V <sub>DD</sub> Operating Supply	V <sub>DD</sub> = Max,	200 MHz	(x8)			560	mA
		$I_{OUT} = 0 \text{ mA},$ $f = f_{MAX} = 1/t_{CYC}$		(x9)			570	
		I - IMAX - I/ICYC		(x18)			590	
				(x36)			675	
			167 MHz	(x8)			495	mA
				(x9)			490	
				(x18)			505	
				(x36)			570	
I <sub>SB1</sub>	Automatic Power Down	Max V <sub>DD</sub> ,	300 MHz	(x8)			290	mA
	Current	Both Ports Deselected, $V_{IN} \ge V_{IH}$ or $V_{IN} \le V_{IL}$		(x9)			300	
		$f = f_{MAX} = 1/t_{CYC}$		(x18)			325	
		f = f <sub>MAX</sub> = 1/t <sub>CYC</sub> , Inputs Static		(x36)			385	7
			278 MHz	(x8)			260	mA
				(x9)			265	
				(x18)			300	
				(x36)			355	
			250 MHz	(x8)			250	mA
				(x9)			250	
				(x18)			290	
				(x36)			325	
			200 MHz	(x8)			230	mA
				(x9)			230	
				(x18)			250	
				(x36)			270	
			167 MHz	(x8)			220	mA
				(x9)			220	
				(x18)			230	
				(x36)			250	

#### **AC Electrical Characteristics**

Over the Operating Range [13]

Parameter	Description	Test Conditions	Min	Тур	Max	Unit
$V_{IH}$	Input HIGH Voltage		V <sub>REF</sub> + 0.2	_	_	V
V <sub>IL</sub>	Input LOW Voltage		-	-	V <sub>REF</sub> – 0.2	V

Document Number: 38-05620 Rev. \*F Page 23 of 32



#### Capacitance

Tested initially and after any design or process change that may affect these parameters.

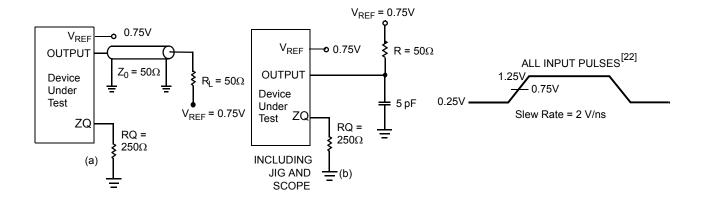
Parameter	Description	Test Conditions	Max	Unit
C <sub>IN</sub>	Input Capacitance	$T_A = 25$ °C, $f = 1$ MHz, $V_{DD} = 1.8$ V, $V_{DDQ} = 1.5$ V	5	pF
C <sub>CLK</sub>	Clock Input Capacitance		6	pF
Co	Output Capacitance		7	pF

#### **Thermal Resistance**

Tested initially and after any design or process change that may affect these parameters.

Parameter	Description	Test Conditions	165 FBGA Package	Unit
$\Theta_{JA}$		Test conditions follow standard test methods and procedures for measuring thermal impedance, in	18.7	°C/W
Θ <sub>JC</sub>	Thermal Resistance (Junction to Case)	accordance with EIA/JESD51.	4.5	°C/W

Figure 4. AC Test Loads and Waveforms



#### Note

Document Number: 38-05620 Rev. \*F Page 24 of 32

<sup>22.</sup> Unless otherwise noted, test conditions are based on signal transition time of 2V/ns, timing reference levels of 0.75V, Vref = 0.75V, RQ = 250Ω, V<sub>DDQ</sub> = 1.5V, input pulse levels of 0.25V to 1.25V, and output loading of the specified I<sub>OL</sub>/I<sub>OH</sub> and load capacitance shown in (a) of AC Test Loads and Waveforms.



# **Switching Characteristics**

Over the Operating Range [22, 23]

Cypress	Consortium	Description		300 MHz		278 MHz		250 MHz		MHz	167 MHz		Unit
Parameter	Parameter	Description	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
t <sub>POWER</sub>		V <sub>DD</sub> (Typical) to the First Access [24]	1		1		1		1		1		ms
t <sub>CYC</sub>	t <sub>KHKH</sub>	K Clock and C Clock Cycle Time	3.3	8.4	3.6	8.4	4.0	8.4	5.0	8.4	6.0	8.4	ns
t <sub>KH</sub>	t <sub>KHKL</sub>	Input Clock (K/K; C/C) HIGH	1.32	_	1.4	_	1.6	-	2.0	_	2.4	_	ns
t <sub>KL</sub>	t <sub>KLKH</sub>	Input Clock (K/K; C/C) LOW	1.32	_	1.4	_	1.6	_	2.0	_	2.4	_	ns
t <sub>KHK</sub> H	t <sub>KHK</sub> H	K Clock Rise to K Clock Rise and C to C Rise (rising edge to rising edge)	1.49	-	1.6	-	1.8	-	2.2	-	2.7	-	ns
t <sub>KHCH</sub>	<sup>t</sup> кнсн	K/K Clock Rise to C/C Clock Rise (rising edge to rising edge)	0	1.45	0	1.55	0	1.8	0	2.2	0	2.7	ns
Setup Time	es		•			•	•		•			•	
t <sub>SA</sub>	t <sub>AVKH</sub>	Address Setup to K Clock Rise	0.4	_	0.4	_	0.5	_	0.6	_	0.7	_	ns
t <sub>SC</sub>	t <sub>IVKH</sub>	Control Setup to K Clock Rise (RPS, WPS)	0.4	-	0.4	_	0.5	-	0.6	_	0.7	_	ns
t <sub>SCDDR</sub>	t <sub>IVKH</sub>	Double Data Rate Control Setup to Clock (K/K) Rise (BWS <sub>0</sub> , BWS <sub>1</sub> , BWS <sub>2</sub> , BWS <sub>3</sub> )	0.3	-	0.3	-	0.35	-	0.4	-	0.5	_	ns
t <sub>SD</sub> <sup>[25]</sup>	t <sub>DVKH</sub>	$D_{[X:0]}$ Setup to Clock (K/ $\overline{K}$ ) Rise	0.3	_	0.3	_	0.35	_	0.4	_	0.5	_	ns
Hold Time:	S		•			•	•		•			•	
t <sub>HA</sub>	t <sub>KHAX</sub>	Address Hold after K Clock Rise	0.4	_	0.4	_	0.5	_	0.6	_	0.7	_	ns
t <sub>HC</sub>	t <sub>KHIX</sub>	Control Hold after K Clock Rise (RPS, WPS)	0.4	-	0.4	_	0.5	-	0.6	-	0.7	-	ns
t <sub>HCDDR</sub>	t <sub>KHIX</sub>	Double Data Rate Control Hold after Clock (K/K) Rise (BWS <sub>0</sub> , BWS <sub>1</sub> , BWS <sub>2</sub> , BWS <sub>3</sub> )	0.3	_	0.3	_	0.35	-	0.4	_	0.5	_	ns
t <sub>HD</sub>	t <sub>KHDX</sub>	$D_{[X:0]}$ Hold after Clock (K/ $\overline{K}$ ) Rise	0.3	_	0.3	_	0.35	-	0.4	_	0.5	_	ns

Document Number: 38-05620 Rev. \*F Page 25 of 32

<sup>23.</sup> When a part with a maximum frequency above 167 MHz is operating at a lower clock frequency, it requires the input timings of the frequency range in which it is being operated and outputs data with the output timings of that frequency range.

24. This part has a voltage regulator internally; tpower is the time that the power must be supplied above VDD minimum initially before a read or write operation can be initiated.

 $<sup>25.\,</sup> For\, D2\,\, data\,\, signal\,\, on\,\, CY7C1911BV18\,\, device,\, t_{SD}\, is\,\, 0.5\,\, ns\,\, for\,\, 200\,\, MHz,\, 250\,\, MHz,\, 278\,\, MHz,\, and\,\, 300\,\, MHz\,\, frequencies.$ 



# Switching Characteristics (continued) Over the Operating Range $^{[22, 23]}$

Cypress	Consortium	2	300 MHz		278 MHz		250 MHz		200 MHz		167 MHz		I I mit
Parameter	Parameter Description		Min	Max	Unit								
Output Tin	nes												
t <sub>CO</sub>		C/C Clock Rise (or K/K in single clock mode) to Data Valid	-	0.45	-	0.45	-	0.45	-	0.45	-	0.50	ns
t <sub>DOH</sub>		Data Output Hold after Output C/C Clock Rise (Active to Active)	-0.45	-	-0.45	-	-0.45	_	-0.45	-	-0.50	-	ns
t <sub>ccqo</sub>	t <sub>CHCQV</sub>	C/C Clock Rise to Echo Clock Valid	_	0.45	_	0.45	_	0.45	_	0.45	_	0.50	ns
t <sub>CQOH</sub>	000.	Echo Clock Hold after C/C Clock Rise	-0.45	-	-0.45	-	-0.45	_	-0.45	-	-0.50	-	ns
t <sub>CQD</sub>	t <sub>CQHQV</sub>	Echo Clock High to Data Valid		0.27		0.27		0.30		0.35		0.40	ns
t <sub>CQDOH</sub>	t <sub>CQHQX</sub>	Echo Clock High to Data Invalid	-0.27	-	-0.27	-	-0.30	_	-0.35	_	-0.40	_	ns
t <sub>CHZ</sub>	t <sub>CHQZ</sub>	Clock (C/C) Rise to High-Z (Active to High-Z) [26, 27]	-	0.45	-	0.45	-	0.45	-	0.45	-	0.50	ns
t <sub>CLZ</sub>	t <sub>CHQX1</sub>	Clock (C/C) Rise to Low-Z [26, 27]	-0.45	-	-0.45	-	-0.45	_	-0.45	_	-0.50	_	ns
DLL Timing													
t <sub>KC Var</sub>	t <sub>KC Var</sub>	Clock Phase Jitter	_	0.20	_	0.20	_	0.20	_	0.20	_	0.20	ns
t <sub>KC lock</sub>	t <sub>KC lock</sub>	DLL Lock Time (K, C)	1024	-	1024	-	1024	-	1024	_	1024	-	Cycles
t <sub>KC Reset</sub>	t <sub>KC Reset</sub>	K Static to DLL Reset	30		30		30		30		30		ns

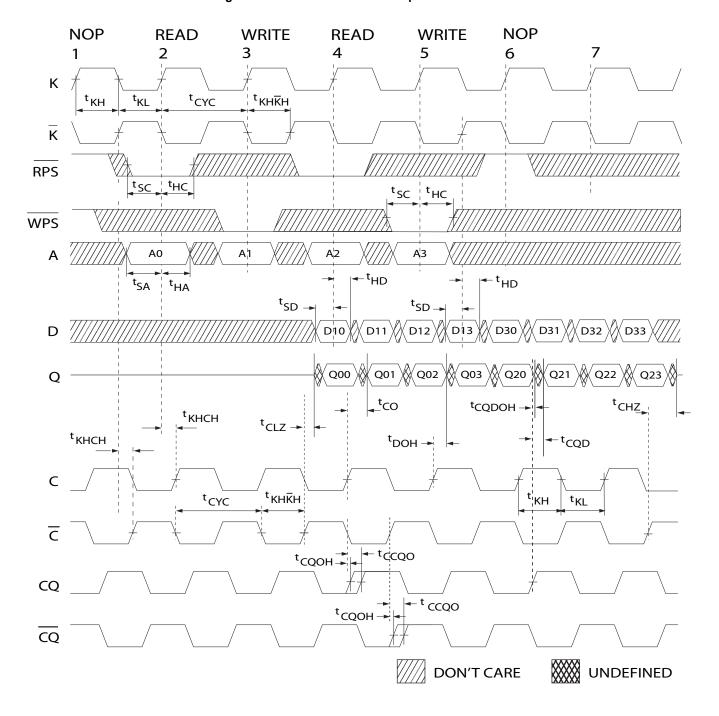
Document Number: 38-05620 Rev. \*F Page 26 of 32

<sup>26.</sup> t<sub>CHZ</sub>, t<sub>CLZ</sub>, are specified with a load capacitance of 5 pF as in (b) of AC Test Loads and Waveforms on page 24. Transition is measured ± 100 mV from steady-state voltage. 27. At any voltage and temperature t<sub>CHZ</sub> is less than t<sub>CLZ</sub> and t<sub>CHZ</sub> less than t<sub>CO</sub>.



### **Switching Waveforms**

Figure 5. Read/Write/Deselect Sequence [28, 29, 30]



#### Notes

- 28. Q00 refers to output from address A0. Q01 refers to output from the next internal burst address following A0, that is, A0+1.
- 29. Outputs are disabled (High-Z) one clock cycle after a NOP.
- 30. In this example, if address A2 = A1, then data Q20 = D10, Q21 = D11, Q22 = D12, and Q23 = D13. Write data is forwarded immediately as read results. This note applies to the whole diagram.

Document Number: 38-05620 Rev. \*F Page 27 of 32



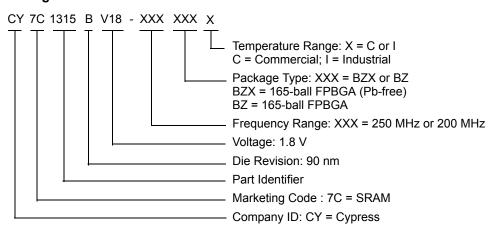
### **Ordering Information**

Cypress offers other versions of this type of product in many different configurations and features. The below table contains only the list of parts that are currently available. For a complete listing of all options, visit the Cypress website at www.cypress.com and refer to the product summary page at http://www.cypress.com/products or contact your local sales representative.

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Speed (MHz)	Ordering Code	Package Diagram		Operating Range
250	CY7C1315BV18-250BZXC	51-85180	165-Ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm) Pb-Free	Commercial
200	CY7C1315BV18-200BZXC	51-85180	165-Ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm) Pb-Free	Commercial
	CY7C1315BV18-200BZI	51-85180	165-Ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm)	Industrial
	CY7C1315BV18-200BZXI	51-85180	165-Ball Fine Pitch Ball Grid Array (13 x 15 x 1.4 mm) Pb-Free	

#### **Ordering Code Definitions**

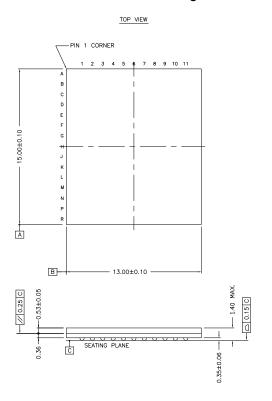


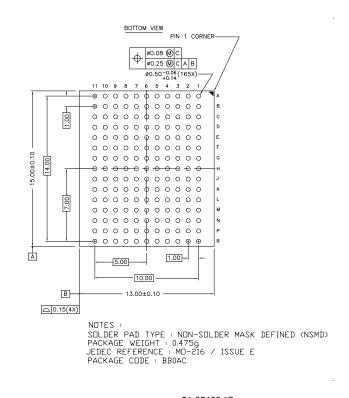
Document Number: 38-05620 Rev. \*F Page 28 of 32



# **Package Diagram**

Figure 6. 165-Ball FBGA (13 x 15 x 1.4 mm), 51-85180





51-85180 \*C

Document Number: 38-05620 Rev. \*F Page 29 of 32



# **Document History Page**

Rev.	ECN No.	Submission Date	Orig. of Change	Description of Change
**	252474	See ECN	SYT	New data sheet
*A	325581	See ECN	SYT	Removed CY7C1911BV18 from the title Included 300-MHz Speed Bin Added Industrial Temperature Grade Replaced TBDs for $I_{DD}$ and $I_{SB1}$ specs Replaced the TBDs on the Thermal Characteristics Table to $\Theta_{JA}$ = 28.51°C/W and $\Theta_{JC}$ = 5.91°C/W Replaced TBDs in the Capacitance Table for the 165 FBGA Package Changed the package diagram from BB165E (15 x 17 x 1.4 mm) to BB165D (13 x 15 x 1.4 mm) Added Lead-Free Product Information Updated the Ordering Information by Shading and Unshading MPNs as per availability
*B	413997	See ECN	NXR	Converted from Preliminary to Final Added CY7C1911BV18 to the title Added 278-MHz speed Bin Changed address of Cypress Semiconductor Corporation on Page# 1 from "3901 North First Street" to "198 Champion Court" Changed C/C Description in the features section Added power-up sequence details and waveforms Added foot notes# 17, 18, 19 on page# 19 Changed the description of $\rm I_X$ from Input Load Current to Input Leakage Current on page# 20 Modified the $\rm I_{DD}$ and $\rm I_{SB}$ values Modified test condition in Footnote # 22 on page# 20 from $\rm V_{DDQ} < \rm V_{DD}$ to $\rm V_{DDQ} \leq \rm V_{DD}$ Replaced Package Name column with Package Diagram in the Ordering Information table Updated Ordering Information Table
*C	472384	See ECN	NXR	Modified the ZQ Definition from Alternately, this pin can be connected directly to $V_{DD}$ to Alternately, this pin can be connected directly to $V_{DDQ}$ Included Maximum Ratings for Supply Voltage on $V_{DDQ}$ Relative to GND Changed the Maximum Ratings for DC Input Voltage from $V_{DDQ}$ to $V_{DD}$ Changed $t_{TH}$ and $t_{TL}$ from 40 ns to 20 ns, changed $t_{TMSS}$ , $t_{TDIS}$ , $t_{CS}$ , $t_{TMSH}$ , $t_{TDIH}$ , $t_{CH}$ from 10 ns to 5 ns and changed $t_{TDOV}$ from 20 ns to 10 ns in TAP AC Switching Characteristics table Modified Power-Up waveform Changed the Maximum rating of Ambient Temperature with Power Applied from –10 C to +85 C to –55 C to +125 C Added additional notes in the AC parameter section Modified AC Switching Waveform Corrected the typo In the AC Switching Characteristics Table Updated the Ordering Information Table

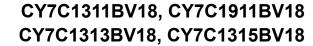
Document Number: 38-05620 Rev. \*F Page 30 of 32



# **Document History Page**

Archi	Document Title: CY7C1311BV18/CY7C1911BV18/CY7C1313BV18/CY7C1315BV18, 18-Mbit QDR™-II SRAM 4-Word Burst Architecture Document Number: 38-05620								
Rev.	ECN No.	Submission Date	Orig. of Change	Description of Change					
*D	2511674	06/03/08	VKN/PYRS	Updated Logic Block diagrams Updated $I_{DD}/I_{SB}$ specs Added footnote# 21 related to $I_{DD}$ Updated power up sequence waveform and its description Changed DLL minimum operating frequency from 80 MHz to 120 MHz Changed $\Theta_{JA}$ spec from 28.51 to 18.7 Changed $\Theta_{JC}$ spec from 5.91 to 4.5 Changed $t_{CYC}$ maximum spec to 8.4 ns for all speed bins Modified footnotes 23 and 30					
*E	2898958	03/25/10	NJY	Removed inactive parts from the ordering information table. Updated package diagram.					
*F	3160393	02/02/2011	NJY	Updated Ordering Information and added Ordering Code Definitions.					

Document Number: 38-05620 Rev. \*F Page 31 of 32





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Document Number: 38-05620 Rev. \*F Revised February 02, 2011 Page 32 of 32

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