Vishay Siliconix



ABSOLUTE MAXIMUM RATINGS

Absolute Maximum Ratings

Input Voltage, V _{IN} to GND	–6.0 to 6.5 V
V _{SD} (See Detailed Description)	–0.3 V to V _{IN}
Output Current, I _{OUT}	Short Circuit Protected
Output Voltage, V _{OUT}	0.3 V to V _{IN} + 0.3 V
Package Power Dissipation, (P _d) ^b	440 mW

Package Thermal Resistance, $(\theta_{JA})^{a}\dots$. 180°C/W
Maximum Junction Temperature, T _{J(max)}	150°C
Storage Temperature, T _{STG} 65°C	C to 150°C
Notes	

Operating Ambient Temperature, T_A $\ldots \ldots \ldots -40^\circ C$ to $85^\circ C$

a. Device mounted with all leads soldered or welded to PC board.

b. Derate 5.5 mW/°C above $T_A = 70°C$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING RANGE

Input Voltage, V _{IN}	2 V to 6 V
Input Voltage, V _{SD}	0 V to V _{IN}

 $C_{IN} = C_{OUT} = 1 \,\mu F$ (ceramic), $C_{BP} = 0.01 \,\mu F$ (ceramic) Maximum ESR of C_{OUT} : 0.4 Ω

SPECIFICATIONS

		Test Conditions Unless Specified $T_A = 25^{\circ}C$, $V_{IN} = V_{OUT(nom)} + 1 V$ $I_{OUT} = 1 mA$, $C_{IN} = 1 \mu F$, $C_{OUT} = 1.0 \mu F$		Limits -40 to 85°C				
Parameter	Symbol	$V_{SD} = 1.5 V$	Temp ^a	Min ^b	Тур ^с	Max ^b	Unit	
Start-Up BP Current	I _{OUT}	ON/OFF = High	Room		1		mA	
Input Voltage Range	V _{IN}		Full	2		6	V	
Output Voltage Accuracy	N.	1 mA < 1. < 150 mA	Room	-1.5	1	1.5	%	
Oulput voltage Accuracy	V _{OUT}	$1 \text{ mA} \le I_{OUT} \le 150 \text{ mA}$	Full	-2.5	1	2.5	70	
Line Regulation ($V_{OUT} \le 3 V$)			Full	-0.06		0.18		
Line Regulation (3.0 V < V _{OUT} ≤3.6 V)	$\frac{\Delta V_{OUT} \times 100}{\Delta V_{IN} \times V_{OUT(nom)}}$	From $V_{IN} = V_{OUT(nom)} + 1 V \text{ to } V_{OUT(nom)} + 2 V$		0		0.3	%/V	
Line Regulation (5-V Version)		From V _{IN} = 5.5 V to 6 V		0		0.4		
	V _{IN} – V _{OUT}	I _{OUT} = 1 mA	Room		1			
		I _{OUT} = 50 mA	Room		45	80	mV	
Dropout Voltage ^{d, g} (V _{OUT(nom)} ≥ 2.6 V)			Full		50	90		
$(\mathbf{v} \mathbf{O} \mathbf{U} \mathbf{I} (\mathbf{nom}) = \mathbf{Z} \mathbf{O} \mathbf{v} \mathbf{V}$			Room		130	180		
		I _{OUT} = 150 mA	Full			220		
			Room		65	100		
Dropout Voltage ^{d, g}		I _{OUT} = 50 mA	Full			120		
$(V_{OUT(nom)} < 2.6 V, V_{IN} \ge 2 V)$		150 4	Room		190	250		
		I _{OUT} = 150 mA	Full			300		
		L 0 A	Room		100	150		
Ground Pin Current ^{e, g}		I _{OUT} = 0 mA	Full			180	μΑ	
$(V_{OUT(nom)} \le 3 V)$		150	Room		110	200		
		$I_{OUT} = 150 \text{ mA}$	Full			230		
Ground Pin Currente	- I _{GND}		Room		110	170		
		I _{OUT} = 0 mA	Full			200		
$(V_{OUT(nom)} > 3 V)$		150 mA	Room		120	200		
		$I_{OUT} = 150 \text{ mA}$	Full			230		
Peak Output current	I _{O (peak)}	$V_{OUT} \ge 0.95 \text{ x } V_{OUT(nom)}$. t _{PW} = 2 ms	Full	300			mA	

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		Test Conditions Un	less Specified			Limits			
		T _A = 25°C, V _{IN} = V _C		–40 to 85°C					
Parameter	Symbol	$I_{OUT} = 1 \text{ mA}, C_{IN} = 1 \mu F, C_{OUT} = 1.0 \mu F$		Temp ^a	Min ^b	Тур ^с	Max ^b	Unit	
Output Noise Voltage	e _N	V _{NOM} = 2.6 V, BW = 10 0 mA < I _{OUT} < 150 mA,	Room		30		μV(rms		
			f = 1 kHz	Room		60		dB	
Ripple Rejection	$\Delta V_{OUT} / \Delta V_{IN}$	I _{OUT} = 150 mA	f = 10 kHz	Room		40			
			f = 100 kHz	Room		30			
Dynamic Line Regulation	$\Delta V_{O(line)}$	$ \begin{array}{c} V_{IN}:V_{OUT(nom)}+1~V~to~V_{OUT(nom)}+2~V\\ t_{f}/t_{f}=2~\mu s,~I_{OUT}=150~mA \end{array} $		Room		20		mV	
Dynamic Load Regulation	$\Delta V_{O(load)}$	I_{OUT} : 1 mA to 150 mA, t_r/t_f = 2 μ s		Room		20			
Thermal Shutdown Junction Temperature	T _{J(S/D)}			Room		150		°C	
Thermal Hysteresis	T _{HYST}			Room		20			
Reverse current	I _R	V _{IN} = -6.0 V		Room		1		μA	
Short Circuit Current	I _{SC}	V _{OUT} = 0	V	Room		700		mA	
Shutdown	•						•		
Shutdown Supply Current	I _{CC(off)}	$V_{SD} = 0 V$		Room		0.1	1	μA	
		High = Regulator ON (Rising)		Full	1.5		V _{IN}		
SD Pin Input Voltage	V _{SD}	Low = Regulator OFF (Falling)		Full			0.4	- V	
Auto Discharge Resistance	R_DIS	Si91841 Only		Room		100		Ω	
SD Pin Input Current ^f	I _{IN(SD)}	$V_{SD} = 1.5 \text{ V}, V_{IN} = 6 \text{ V}$		Room		0.7		μΑ	
SD Hysteresis	V _{HYST(SD)}	1		Full		150		mV	
V _{OUT} Turn-On Time	t _{ON}	V _{SD} (See Figure 1), I _{I OAD} = 100 nA				50		μS	

Notes

Room = 25°C, Full = -40 to 85°C. a.

b. The algebraic convention whereby the most negative value is a minimum and the most positive a maximum.

Typical values are for DESIGN AID ONLY, not guaranteed nor subject to production testing. Typical values for dropout voltage at $V_{OUT} \ge 2$ V are measured at $V_{OUT} = 3.3$ V, while typical values for dropout voltage at $V_{OUT} < 2$ V are measured at $V_{OUT} = 1.8$ V. Dropout voltage is defined as the input to output differential voltage at which the output voltage drops 2% below the output voltage measured with a 1-V c.

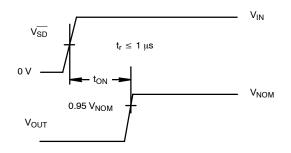
d. differential, provided that VIN does not not drop below 2.0 V.

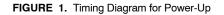
e.

Ground current is specified for normal operation as well as "drop-out" operation. The device's shutdown pin includes a typical 2-MΩ internal pull-down resistor connected to ground. f.

 $V_{OUT(nom)}$ is V_{OUT} when measured with a 1-V differential to V_{IN} g.

TIMING WAVEFORMS



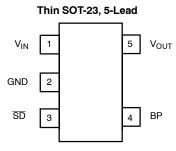


Si91841

Vishay Siliconix



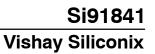
PIN CONFIGURATION



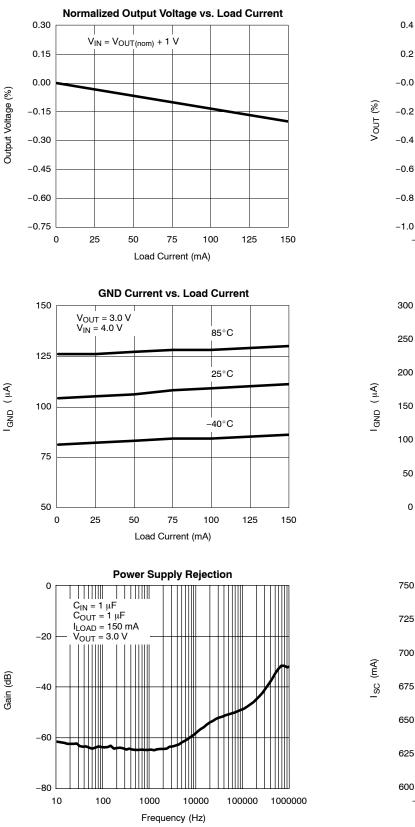
PIN DESCRIPTION					
Pin No.	Name	Function			
1	V _{IN}	Input supply pin. Bypass this pin with a 1- μ F ceramic or tantalum capacitor to ground			
2	GND	Ground pin. For better thermal capability, directly connected to large ground plane			
3	SD	By applying less than 0.4 V to this pin, the device will be turned off. Connect this pin to V_{IN} if unused			
4	BP	Noise bypass pin. For low noise applications, a 0.01 μF ceramic capacitor should be connected from this pin to ground.			
5	V _{OUT}	Output voltage. Connect C _{OUT} between this pin and ground.			

ORDERING INFORMATION—Si91841							
Part Number	Lead (Pb)-Free Part Number	Marking	Voltage	Temperature Range	Package		
Si91841DT-18-T1	Si91841DT-18-T1—E3	B4LL	1.8				
Si91841DT-25-T1	Si91841DT-25-T1—E3	B7LL	2.5				
Si91841DT-26-T1	Si91841DT-26-T1—E3	B8LL	2.6				
Si91841DT-28-T1	Si91841DT-28-T1—E3	B0LL	2.8				
Si91841DT-285-T1	Si91841DT-285—E3	C1LL	2.85	–40 to 85°C	Thin SOT23-5		
Si91841DT-29-T1	Si91841DT-29-T1—E3	C2LL	2.9				
Si91841DT-30-T1	Si91841DT-30-T1—E3	C3LL	3.0	1			
Si91841DT-33-T1	Si91841DT-33-T1—E3	C4LL	3.3	1			
Si91841DT-50-T1	Si91841DT-50-T1—E3	C7LL	5.0				

Note: LL = Lot Code

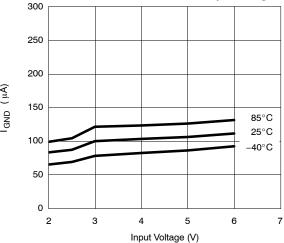


TYPICAL CHARACTERISTICS (INTERNALLY REGULATED, 25° C UNLESS NOTED)

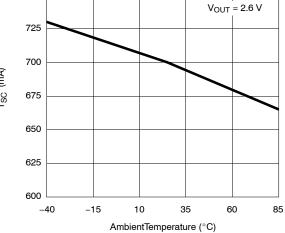


Normalized VOUT vs. Temperature 0.4 $V_{IN} = V_{OUT(nom)} + 1 V$ 0.2 I_{OUT} = 0 mA -0.0 I_{OUT} = 75 mA -0.2 l_{OUT} = 150 mA -0.4 -0.6 -0.8 -1.0-40 -15 10 35 60 85 Ambient Temperature (°C)

No Load GND Pin Current vs. Input Voltage







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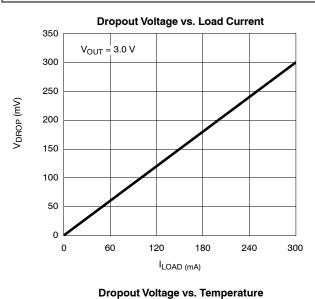
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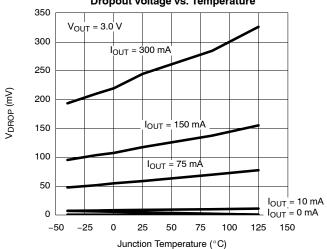
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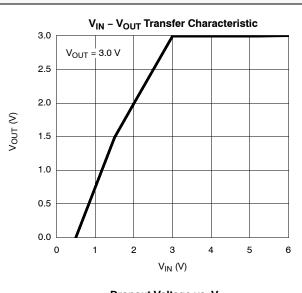
Si91841

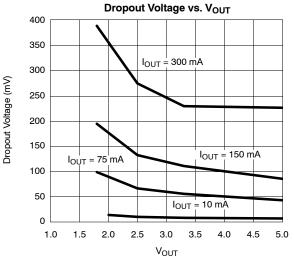


TYPICAL CHARACTERISTICS (INTERNALLY REGULATED, 25°C UNLESS NOTED)





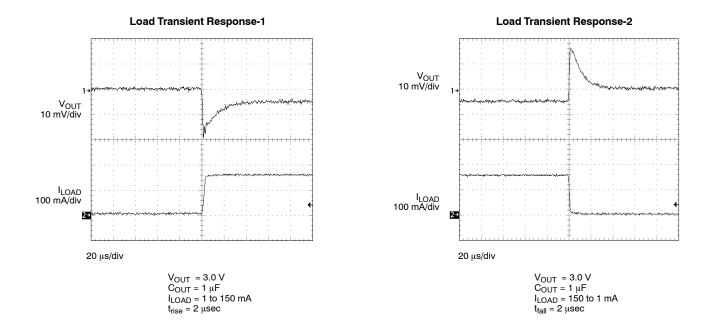




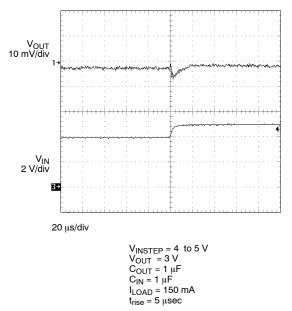
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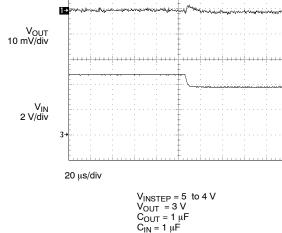


TYPICAL WAVEFORMS



LineTransient Response-1





 $\begin{array}{l} V_{INSTEP}=5 \ to \ 4 \ V \\ V_{OUT}=3 \ V \\ C_{OUT}=1 \ \mu F \\ C_{IN}=1 \ \mu F \\ I_{LOAD}=150 \ mA \\ t_{fall}=5 \ \mu sec \end{array}$

LineTransient Respons-2

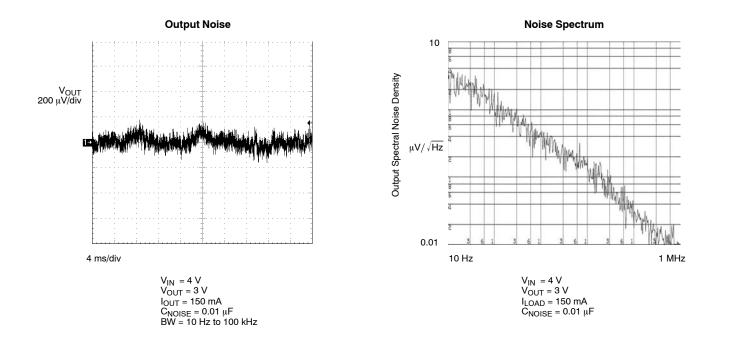
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Si91841

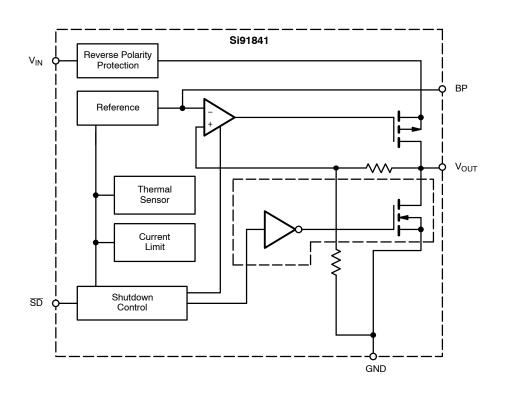
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TYPICAL WAVEFORMS



BLOCK DIAGRAM



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DETAILED DESCRIPTION

The Si91841 is a low-noise, low drop-out and low quiescent current linear voltage regulator, packaged in a small footprint Thin SOT23-5 package. The Si91841 can supply loads up to 150 mA. As shown in the block diagram, the circuit consists of a bandgap reference error, amplifier, p-channel pass transistor and feedback resistor string. An external bypass capacitor connected to the BP pin reduces noise at the output. Additional blocks, not shown in the block diagram, include a precise current limiter, reverse battery and current protection and thermal sensor.

Thermal Overload Protection

The thermal overload protection limits the total power dissipation and protects the device from being damaged. When the junction temperature exceeds 150° , the device turns the p-channel pass transistor off.

Reverse Battery Protection

The Si91841 has a battery reverse protection circuitry that disconnects the internal circuitry when V_{IN} drops below the GND voltage. There is no current drawn in such an event. When the SD pin is hardwired to V_{IN}, the user must connect the SD pin to V_{IN} via a 100-k Ω resistor if reverse battery

protection is desired. Hardwiring the $\overline{\text{SD}}$ pin directly to the V_{IN} pin is allowed when reverse battery protection is not desired.

Noise Reduction

An external 10-nF bypass capacitor at BP is used to create a low pass filter for noise reduction. The start-up time is fast, since a power-on circuit pre-charges the bypass capacitor. After the power-up sequence the pre-charge circuit is switched to standby mode in order to save current. It is therefore not recommended to use larger bypass capacitor values than 50 nF. When the circuit is used without a capacitor, stable operation is guaranteed.

Auto-Discharge/No-Discharge

For Si91841 only, V_{OUT} has an internal 100- Ω (typ.) discharge path to ground when the \overline{SD} pin is low.

Stability

The circuit is stable with only a small output capacitor equal to 6 nF/mA (= 1 μ F @ 150 mA). Since the bandwidth of the error amplifier is around 1–3 MHz and the dominant pole is at the output node, the capacitor should be capacitive in this range, i.e., for 150-mA load current, an ESR <0.4 Ω is necessary. Parasitic inductance of about 10 nH can be tolerated.

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