ABSOLUTE MAXIMUM RATINGS

V _{CC} to GND V _L to GND	
V+ to GND	(= =)
V- to GND	
V+ + V- (Note 1)	+13V
Input Voltages	
T_IN, FORCEON, FORCEOFF to GND	
R_IN to GND	±25V
Output Voltages	
T_OUT to GND	±13.2V
R_OUT	0.3V to (V _L + 0.3V)

Short-Circuit Duration T_OUT to GNDContinuous	
Continuous Power Dissipation ($T_A = +70^{\circ}C$)	,
24-Pin TSSOP (derate 7.8mW/°C above +70°C)625mW	/
Operating Temperature Ranges	
MAX3387ECUG0°C to +70°C)
MAX3387EEUG40°C to +85°C)
Junction Temperature+150°C)
Storage Temperature Range65°C to +150°C)
Lead Temperature (soldering, 10s)+300°C)

Note 1: V+ and V- can have maximum magnitudes of 7V, but their absolute difference cannot exceed 13V.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = V_L = +3.0V \text{ to } +5.5V; C1-C4 = 0.1\mu\text{F}, \text{ tested at } +3.3V \pm 10\%; C1 = 0.047\mu\text{F}, C2-C4 = 0.33\mu\text{F}, \text{ tested at } +5.0V \pm 10\%; TA = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted}. Typical values are at V_{CC} = V_L = +3.3V, TA = +25^{\circ}\text{C}.)$

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS
DC CHARACTERISTICS (V _{CC} =	+3.3V or +5	$5V, T_A = +25^{\circ}C)$		-			
Supply Current, AutoShutdown Plus		All R_IN idle, FORCEON = GND, FORCEOFF = V_{CC} , all T_IN idle			1.0	10	μA
Supply Current		FORCEOFF = FORC	EON = V _{CC} , no load		0.3	1	mA
LOGIC INPUTS	1						
Input Logic Threshold Low		T_IN, FORCEON,	V _L = 3.3V or 5.0V			0.8	V
Input Logic Threshold Low		FORCEOFF	$V_L = 2.5V$			0.6	v
			$V_L = 5.0V$	2.4			
Input Logia Thrashold High		T_IN, FORCEON,	$V_{L} = 3.3V$	2.0			V
Input Logic Threshold High	FORCEOFF	$V_L = 2.5V$	1.4			v	
			$V_{L} = 1.8V$		0.9]
Transmitter Input Hysteresis			·		0.5		V
Input Leakage Current		T_IN, FORCEON, FORCEOFF			±0.01	±1	μΑ
RECEIVER OUTPUTS	-						
Output Voltage Low		IOUT = 1.6mA				0.4	V
Output Voltage High		I _{OUT} = -1mA		V _L - 0.6	V _L - 0.1		V
RECEIVER INPUTS	1						
Input Voltage Range				-25		+25	V
	т.	T 05°C	$V_L = 5.0V$	0.8	1.5		V
Input Threshold Low T _A = +25		$T_A = +25^{\circ}C$	$V_L = 3.3V$	0.6	1.2		v
Input Throshold High		T _A = +25°C	$V_L = 5.0V$		1.8	2.4	V
Input Threshold High		IA = +20 C	$V_{L} = 3.3V$		1.5	2.4	v

DC ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = V_L = +3.0V to +5.5V; C1–C4 = 0.1 μ F, tested at +3.3V ±10%; C1 = 0.047 μ F, C2–C4 = 0.33 μ F, tested at +5.0V ±10%; T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = V_L = +3.3V, T_A = +25°C.)

PARAMETER	SYMBOL	COND	MIN	ТҮР	MAX	UNITS		
Input Hysteresis					0.5		V	
Input Resistance		$T_A = +25^{\circ}C$		3	5	7	kΩ	
TRANSMITTER OUTPUTS								
Output Voltage Swing		All transmitter outputs ground	loaded with $3k\Omega$ to	±5	±5.4		V	
Output Resistance		V _{CC} = V+ = V- = 0, tra	nsmitter output = $\pm 2V$	300	10M		Ω	
Output Short-Circuit Current		$V_{T_OUT} = 0$				±60	mA	
Output Leakage Current		$V_{T_OUT} = \pm 12V$, transmitters disabled; $V_{CC} = 0$ or 3.0V to 5.5V				±25	μA	
ESD PROTECTION								
		Human Body Model			±15			
R_IN, T_OUT ESD Protection		IEC 1000-4-2 Air-Gap Discharge method IEC 1000-4-2 Contact Discharge method			±15		kV	
					±8			
AutoShutdown Plus (FORCEON	$I = GND, \overline{FC}$	$\overrightarrow{RCEOFF} = V_{CC}$		·				
Receiver Input Threshold to		Figure 3a Positive threshold				2.7	V	
INVALID Output High		rigule ba	Negative threshold				v	
Receiver Input Threshold to INVALID Output Low		Figure 3a		-0.3		0.3	V	
INVALID Output Voltage Low		I _{OUT} = -1.6mA				0.4	V	
INVALID Output Voltage High		$I_{OUT} = -1.0 mA$		V _L - 0.6			V	
Receiver Positive or Negative Threshold to INVALID High	tinvh	V _{CC} = 5V, Figure 3b			1		μs	
Receiver Positive or Negative Threshold to INVALID Low	tinvl	V _{CC} = 5V, Figure 3b			30		μs	
Receiver or Transmitter Edge to Transmitters Enabled	tw∪	V _{CC} = 5V, Figure 3b			100		μs	
Receiver or Transmitter Edge to Transmitters Shutdown	tautoshdn	V _{CC} = 5V, Figure 3b		15	30	60	S	

TIMING CHARACTERISTICS

(V_{CC} = V_L = +3V to +5.5V; C1–C4 = 0.1 μ F, tested at +3.3V ±10%; C1 = 0.047 μ F, C2–C4 = 0.33 μ F, tested at +5.0V ±10%; T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = V_L = +3.3V, T_A = +25°C.)

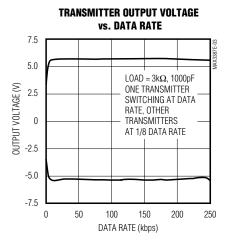
PARAMETER	SYMBOL	CONDITION	MIN	TYP	MAX	UNITS		
Maximum Data Rate		$R_L = 3k\Omega$, $C_L = 1000pF$, one transmitter switching		250			kbps	
Receiver Propagation Delay	t _{PHL}	Receiver input to receiver output, $C_L = 150pF$			0.15			
neceiver riopagation Delay	t _{PLH}				0.15		μs	
Time to Exit Shutdown		VT_OUT > 3.7V		100		μs		
Transmitter Skew	tphl - tplh	(Note 2)			100		ns	
Receiver Skew	tphl - tplh				50		ns	
Transition-Region Slew		VCC = $3.3V$, CL = $150pF$ to TA = $+25^{\circ}C$, $1000pF$ R _L = $3k\Omega$ to $7k\Omega$. $1000pF$		6		30	V/µs	
Rate measured from +3V to -3V or -3V to +3V	C _L = 150pF to 2500pF	4		30	ν/μο			

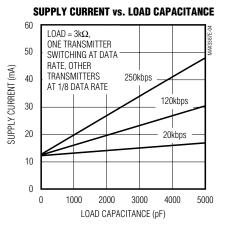
Typical Operating Characteristics

TRANSMITTER OUTPUT VOLTAGE vs. LOAD CAPACITANCE 7.5 5.0 DATA RATE = 250kbps OUTPUT VOLTAGE (V) $LOAD = 3k\Omega IN PARALLEL WITH C_1$ 2.5 0 -2.5 -5.0 -7.5 0 1000 2000 3000 4000 5000 LOAD CAPACITANCE (pF)

 $(V_{CC} = V_L = +3.3V, T_A = +25^{\circ}C, unless otherwise noted.)$

SLEW RATE vs. LOAD CAPACITANCE 16 14 12 SLEW RATE -SLEW RATE (V/µs) 10 8 SLEW RATE + 6 4 2 0 0 1000 2000 3000 4000 5000 LOAD CAPACITANCE (pF)





Pin Description

PIN	NAME	FUNCTION			
1	C1+	Positive Terminal of the Voltage-Doubler Charge-Pump Capacitor			
2	V+	+5.5V Supply Generated by the Charge Pump			
3	C1-	Negative Terminal of the Voltage-Doubler Charge-Pump Capacitor			
4	C2+	Positive Terminal of the Inverting Charge-Pump Capacitor			
5	C2-	Negative Terminal of the Inverting Charge-Pump Capacitor			
6	V-	-5.5V Generated by the Charge Pump			
7	T1IN	TTL/CMOS Transmitter Inputs			
8	T2IN				
9	INVALID	Output of the Valid Signal Detector. INVALID is high if a valid RS-232 signal is present on the receiver inputs.			
10	T3IN	TTL/CMOS Transmitter Inputs			
11	FORCEON	Force-On Input. Drive high to override automatic circuitry keeping transmitters on (FORCEOFF must be high) (Table 1).			
12	R3OUT				
13	R2OUT	TTL/CMOS Receiver Outputs. Swing between 0 and VL.			
14	R1OUT				
15	VL	Logic-Level Supply. All CMOS inputs and outputs are referenced to this supply.			
16	R3IN				
17	R2IN	RS-232 Receiver Inputs			
18	R1IN				
19	T3OUT				
20	T2OUT	RS-232 Transmitter Outputs			
21	T1OUT				
22	GND	Ground			
23	Vcc	+3.0V to +5.5V Supply Voltage			
24	FORCEOFF	Force-Off Input. Drive low to shut down transmitters and on-board power supply. This over- rides all automatic circuitry and FORCEON (Table 1).			

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Detailed Description

Dual Charge-Pump Voltage Converter

The MAX3387E's internal power supply consists of a regulated dual charge pump that provides output voltages of +5.5V (doubling charge pump) and -5.5V (inverting charge pump), regardless of the input voltage (VCC) over a +3.0V to +5.5V range. The charge pumps operate in a discontinuous mode: if the output voltages are less than 5.5V, the charge pumps are enabled; if the output voltages exceed 5.5V, the charge pumps are disabled. Each charge pump requires a flying capacitor (C1, C2) and a reservoir capacitor (C3, C4) to generate the V+ and V- supplies.

RS-232 Transmitters

The transmitters are inverting level translators that convert CMOS-logic levels to 5.0V EIA/TIA-232 levels.

The MAX3387E transmitters guarantee a 250kbps data rate with worst-case loads of $3k\Omega$ in parallel with 1000pF, providing compatibility with PC-to-PC communication software (such as LapLinkTM). Transmitters can be paralleled to drive multiple receivers or mice. Figure 1 shows a complete system connection.

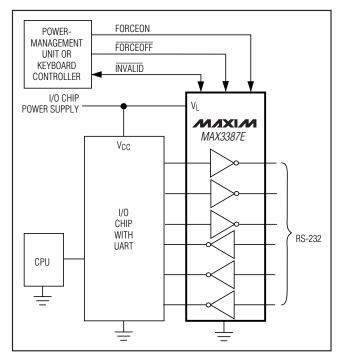


Figure 1. Interface Under Control of PMU

LapLink is a trademark of Traveling Software.

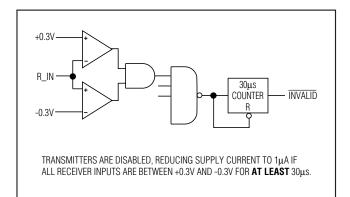
These RS-232 output stages are turned off (high impedance) when the device is in shutdown mode. When the power is off, the MAX3387E permits the outputs to be driven up to $\pm 12V$.

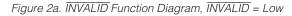
The transmitter inputs do not have pull-up resistors. Connect unused inputs to GND or $V_{\rm L}.$

RS-232 Receivers

The receivers convert RS-232 signals to CMOS-logic output levels. The MAX3387E's receivers are always active, even when the device is in shutdown.

The MAX3387E features an INVALID output that indicates when no signal is present on any RS-232 receiver inputs. INVALID is independent of other control logic functions; it indicates the receiver input conditions only (Figures 2 and 3).





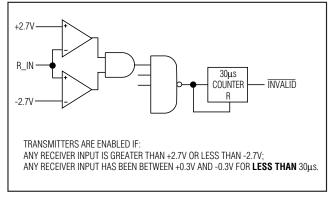


Figure 2b. INVALID Function Diagram, INVALID = High

Figure 2c. AutoShutdown Plus Logic

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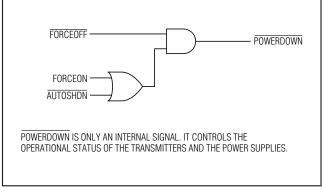


Figure 2d. Power-Down Logic

AutoShutdown Plus Mode

The MAX3387E achieves a1µA supply current with Maxim's AutoShutdown Plus feature, which operates when FORCEOFF is high and a FORCEON is low. When these devices do not sense a valid signal transition on any receiver and transmitter input for 30sec, the onboard charge pumps are shut down, reducing supply current to 1µA. This occurs if the RS-232 cable is disconnected or if the connected peripheral transmitter inputs is inactive. The system turns on again when a valid transition is applied to any RS-232 receiver or transmitter input. As a result, the system saves power without changes to the existing BIOS or operating system.

Figures 2a and 2b show valid and invalid RS-232 receiver voltage levels. INVALID indicates the receiver input's condition, and is independent of FORCEON and

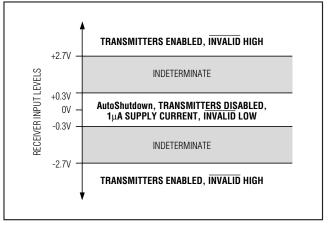


Figure 3a. Receiver Thresholds for INVALID

FORCEOFF states. Figure 2 and Table 1 summarize the MAX3387E's operating modes. FORCEON and FORCE-OFF override AutoShutdown Plus circuitry. When neither control is asserted, the IC selects between these states automatically based on the last receiver or transmitter input edge received.

By connecting FORCEON to INVALID, the MAX3387E shuts down when no valid receiver level and no receiver or transmitter edge is detected for 30sec, and wakes up when a valid receiver level or receiver or transmitter edge is detected.

By connecting FORCEON and FORCEOFF to INVALID, the MAX3387E shuts down when no valid receiver level is detected.

A mouse or other system with AutoShutdown Plus may need time to wake up. Figure 4 shows a circuit that forces the transmitters on for 100ms, allowing enough time for the other system to realize that the MAX3387E is awake. If the other system outputs valid RS-232 signal transitions within that time, the RS-232 ports on both systems remain enabled.

V_L Logic Supply Input

Unlike other RS-232 interface devices where the receiver outputs swing between 0 and V_{CC}, the MAX3387E features a separate logic supply input (V_L) that sets V_{OH} for the receiver outputs and sets thresholds for the receiver inputs. This feature allows a great deal of flexibility in interfacing to many different types of systems with different logic levels. Connect this input to the host logic supply (1.8V \leq V_L \leq V_{CC}). Also, see the *Typical PDA/Cell-Phone Application* section.



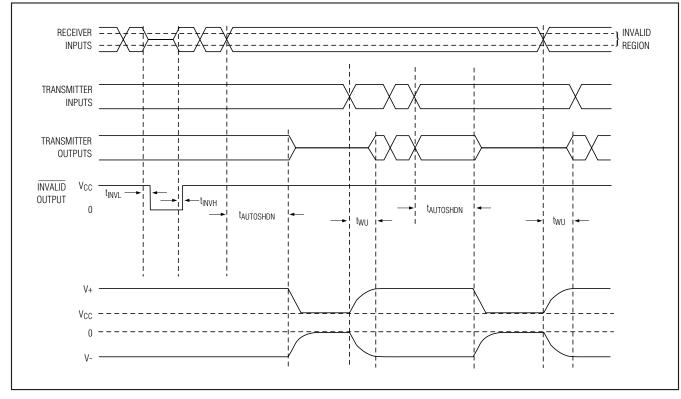


Figure 3b. AutoShutdown Plus/INVALID Timing Diagram

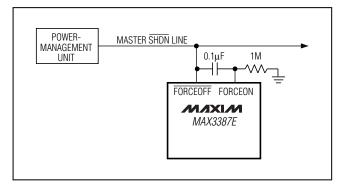


Figure 4. AutoShutdown with Initial Turn-On to Wake Up a System

Software-Controlled Shutdown

If direct software control is desired, INVALID can be used to indicate a DTR or ring indicator signal. Connect FORCEOFF and FORCEON together to bypass AutoShutdown so the line acts like a SHDN input.

±15kV ESD Protection

As with all Maxim devices, ESD-protection structures are incorporated on all pins to protect against electrostatic discharges (ESDs) encountered during handling and assembly. The MAX3387E driver outputs and receiver inputs have extra protection against static electricity. Maxim has developed state-of-the-art structures to protect these pins against ESD of ±15kV without damage. The ESD structures withstand high ESD in all states: normal operation, shutdown, and powered down. After an ESD event, Maxim's "E" version devices keep working without latchup, whereas competing RS-232 products can latch and must be powered down to remove latchup. ESD protection can be tested in various ways. The transmitter outputs and receiver inputs of this product family are characterized for protection to the following limits:

- 1) ±15kV using the Human Body Model
- 2) ±8kV using the Contact-Discharge method specified in IEC 1000-4-2
- 3) ±15kV using IEC 1000-4-2's Air-Gap method



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Table 1. Output Control Truth Table

OPERATION STATUS	FORCEON	FORCEOFF	VALID RECEIVER LEVEL	RECEIVER OR TRANSMITTER EDGE WITHIN 30	T_OUT	R_OUT
Shutdown (Forced Off)	Х	0	Х	Х	High-Z	Active
Normal Operation (Forced On)	1	1	Х	Х	Active	Active
Normal Operation (AutoShutdown Plus)	0	1	Х	Yes	Active	Active
Shutdown (AutoShutdown Plus)	0	1	Х	No	High-Z	Active
Normal Operation	INVALID	1	Yes	Х	Active	Active
Normal Operation	INVALID	1	Х	Yes	Active	Active
Shutdown	INVALID	1	No	No	High-Z	Active
Normal Operation (AutoShutdown)	INVALID	INVALID	Yes	Х	Active	Active
Shutdown (AutoShutdown)	INVALID	INVALID	No	Х	High-Z	Active

X = Don't care

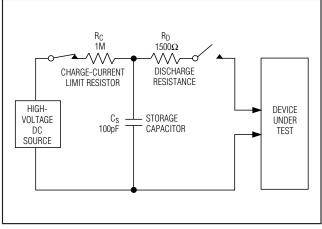


Figure 5a. Human Body ESD Test Model

ESD Test Conditions

ESD performance depends on a variety of conditions. Contact Maxim for a reliability report that documents test setup, methodology, and results.

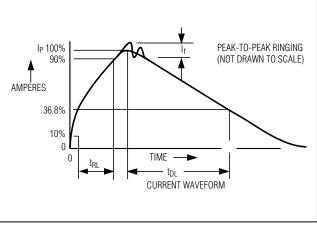


Figure 5b. Human Body Current Waveform

Human Body Model

Figure 5a shows the Human Body Model, and Figure 5b shows the current waveform it generates when discharged into a low impedance. This model consists of a 100pF capacitor charged to the ESD voltage of interest, which is then discharged into the test device through a $1.5 \mathrm{k}\Omega$ resistor.



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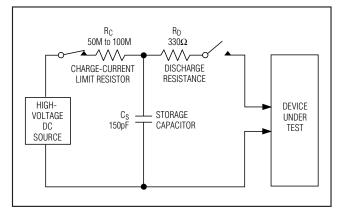


Figure 6a. IEC 1000-4-2 ESD Test Model

IEC 1000-4-2

The IEC 1000-4-2 standard covers ESD testing and performance of finished equipment; it does not specifically refer to ICs. The MAX3387E helps you design equipment that meets Level 4 (the highest level) of IEC 1000-4-2, without the need for additional ESD-protection components.

The major difference between tests done using the Human Body Model and IEC 1000-4-2 is higher peak current in IEC 1000-4-2 because series resistance is lower in the IEC 1000-4-2 model. Hence, the ESD withstand voltage measured to IEC 1000-4-2 is generally lower than that measured using the Human Body Model. Figure 6a shows the IEC 1000-4-2 model, and Figure 6b shows the current waveform for the \pm 8kV IEC 1000-4-2 Level 4 ESD Contact Discharge test.

The air-gap test involves approaching the device with a charged probe. The contact-discharge method connects the probe to the device before the probe is energized.

Machine Model

The Machine Model for ESD tests all pins using a 200pF storage capacitor and zero discharge resistance. Its objective is to emulate the stress caused by contact that occurs with handling and assembly during manufacturing. All pins require this protection during manufacturing, not just RS-232 inputs and outputs. Therefore, after PC board assembly, the Machine Model is less relevant to I/O ports.

Applications Information

Capacitor Selection

The capacitor type used for C1–C4 is not critical for proper operation; polarized or nonpolarized capacitors can be used. The charge pump requires 0.1μ F capaci-



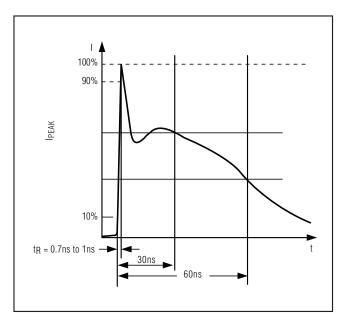


Figure 6b. IEC 1000-4-2 ESD Generator Current Waveform

Table 2. Minimum Required CapacitorValues

Vcc (V)	C1 (μF)	C2, C3, C4 (μF)
3.0 to 3.6	0.1	0.1
4.5 to 5.5	0.047	0.33
3.0 to 5.5	0.22	1

tors for 3.3V operation. For other supply voltages, see Table 2 for required capacitor values. Do not use values smaller than those listed in Table 2. Increasing the capacitor values (e.g., by a factor of 2) reduces ripple on the transmitter outputs and slightly reduces power consumption. C2, C3, and C4 can be increased without changing C1's value. However, do not increase C1 without also increasing the values of C2, C3, and C4 to maintain the proper ratios (C1 to the other capacitors).

When using the minimum required capacitor values, make sure the capacitor value does not degrade excessively with temperature. If in doubt, use capacitors with a larger nominal value. The capacitor's equivalent series resistance (ESR), which usually rises at low temperatures, influences the amount of ripple on V+ and V-.

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Power-Supply Decoupling

In most circumstances, a 0.1μ F bypass capacitor is adequate. In applications that are sensitive to power-supply noise, decouple V_{CC} to ground with a capacitor of the same value as charge-pump capacitor C1. Connect bypass capacitors as close to the IC as possible.

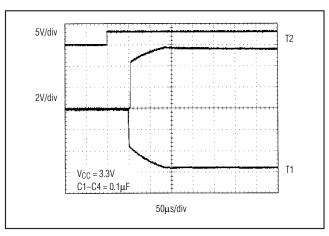


Figure 7. Transmitter Outputs when Exiting Shutdown

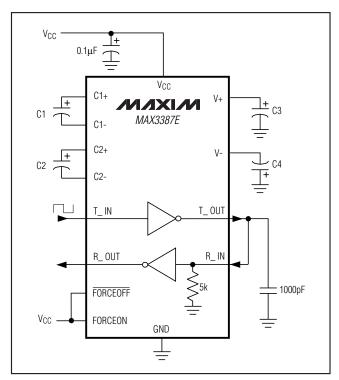


Figure 8. Loopback Test Circuit

Operation Down to 2.7V

Transmitter outputs will meet TIA/EIA-562 levels of $\pm 3.7V$ with supply voltages as low as $\pm 2.7V$.

Transmitter Outputs when Exiting Shutdown

Figure 7 shows two transmitter outputs when exiting shutdown mode. As they become active, the two transmitter outputs are shown going to opposite RS-232 levels (one transmitter input is high; the other is low). Each transmitter is loaded with $3k\Omega$ in parallel with 2500pF. The transmitter outputs display no ringing or undesirable transmitters are enabled only when the magnitude of V- exceeds approximately 3V.

High Data Rates

The MAX3387E maintains the RS-232 \pm 5.0V minimum transmitter output voltage even at high data rates. Figure 8 shows a transmitter loopback test circuit. Figure 9 shows a loopback test result at 120kbps, and Figure 10 shows the same test at 250kbps. For Figure 9, all transmitters were driven simultaneously at 120kbps into RS-232 loads in parallel with 1000pF. For Figure 10, a single transmitter was driven at 250kbps, and all transmitters were loaded with an RS-232 receiver in parallel with 1000pF.

Interconnection with 3V and 5V Logic

The MAX3387E can directly interface with various 5V logic families, including ACT and HCT CMOS. The logic voltage power-supply pin (V_L) sets the output voltage level of the receivers and the input thresholds of the transmitters.

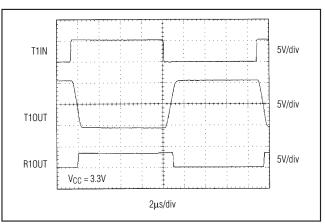


Figure 9. Loopback Test Results at 120kbps



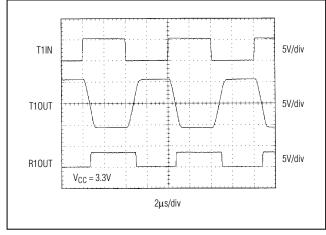


Figure 10. Loopback Test Results at 250kbps

Typical PDA/Cell-Phone Application

The MAX3387E is designed with PDA applications in mind. Two transmitters and two receivers handle standard full-duplex communication protocol, while an extra transmitter allows a ring indicator signal to alert the UART on the PC. Without the ring indicator transmitter, solutions for these applications would require softwareintensive polling of the cradle inputs.

The ring indicate (RI) signal is generated when a PDA, phone, or other "cradled" device is plugged into its cradle.

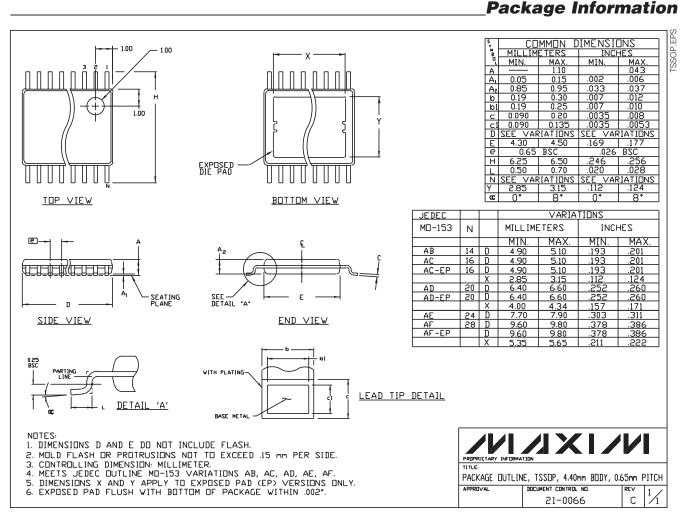
This generates a logic-low signal to the RI transmitter input, creating +6V at the ring indicate pin. The PC's UART RI input is the only pin that can generate an interrupt from signals arriving through the RS-232 port. The interrupt routine for this UART will then service the RS-232 full-duplex communication between the PDA and the PC.

As cell phone design becomes more like that of PDAs, cell phones will require similar docking ability and communication protocol. Cell phones operate on a single lithium-ion (Li+) battery and work with a power-supply voltage of +2.7V to +4V. The baseband logic coming from the phone connector can be as low as 1.8V at the transceivers. To prevent forward biasing of a device internal to the cell phone, the MAX3387E comes with a logic power-supply pin (VL) that limits the logic levels presented to the phone. The receiver outputs will sink to zero for low outputs, but will not exceed VL for logic highs. The input logic levels for the transmitters are also altered, scaled by the magnitude of the VL input. The device will work with V₁ as low as 1.8V before the charge-pump noise will begin to cause the transmitter outputs to oscillate. This is useful with cell phones and other power-efficient devices with core logic voltage levels that go as low as 1.8V.

Chip Information

TRANSISTOR COUNT: 1267





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