Characteristics HDMIULC6-4SC6

1 Characteristics

Table 1. Absolute ratings

Symbol	Pa	Value	Unit	
V _{PP}	Peak pulse voltage IEC 61000-4-2 air discharge IEC 61000-4-2 contact discharge MIL STD883C-Method 3015-6		±15 ±15 ±25	kV
T _{stg}	Storage temperature range		-55 to +150	°C
Tj	Maximum junction temperature		125	°C
T _L	Lead solder temperature (10 seconds duration)		260	°C

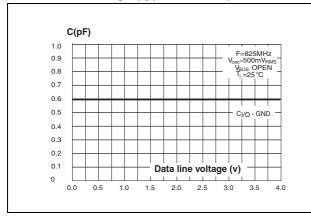
Table 2. Electrical characteristics ($T_{amb} = 25^{\circ} C$)

Symbol	Dovemeter	Took Conditions	Value			11
	Parameter	Test Conditions	Min.	Тур.	Max	Unit
I _{RM}	Leakage current	V _{RM} = 5 V			0.5	μΑ
V _{BR}	Breakdown voltage between V _{BUS} and GND	I _R = 1 mA	6			V
V _{CL}	Clamping voltage	$I_{PP} = 1 \text{ A}, t_p = 8/20 \mu s$ Any I/O pin to GND			12	V
		$I_{PP} = 5 \text{ A}, t_p = 8/20 \mu s$ Any I/O pin to GND			17	V
C _{i/o-GND}	Capacitance between I/O and GND	V _R = 0 V, F = 1 MHz			1	
		V _R = 0 V, F = 825 MHz		0.6		pF
$\Delta C_{i/o\text{-GND}}$	Capacitance variation between I/O and GND			0.015		F .
C _{i/o-i/o}	Capacitance between I/O	V _R = 0 V, F = 1 MHz		0.42	0.5	
		V _R = 0 V, F = 825 MHz		0.3		pF
$\Delta C_{i/o-i/o}$	Capacitance variation between I/O			0.007		

HDMIULC6-4SC6 Characteristics

Figure 1. Line capacitance versus line voltage (typical values)

Figure 2. Line capacitance versus frequency (typical values)



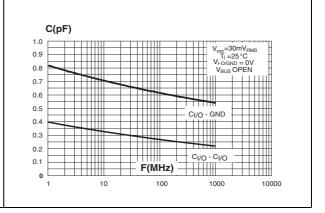
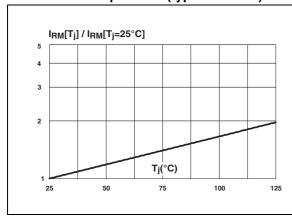
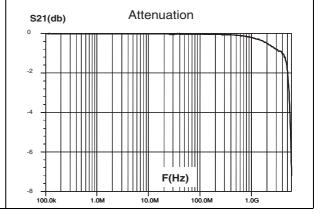


Figure 3. Relative variation of leakage current versus junction temperature (typical values)

Figure 4. Frequency response



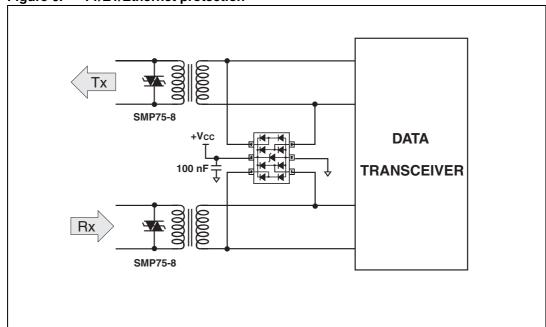


2 Application examples

HDMI Display (TV, flat panel, monitor, Host Tx0-Rx0-(Set Top Box, DVD player, PC) projector) Rx0+ Tx0+ TMDS HDMI connectors TMDS vided Rx1-Tx1-Tx1+ Rx1+ Multimedia controller Rx2-Tx2controller Ctrl / Rx2+ Ctrl / Tx2+ HDMIULC6-4SC6 RC-TC-RC+ TC+ TMDS links CEC CEC SCL SCL -Vcc 5V Vcc 5V HDMIULC6-4SC6 SDA SDA HPD HPD. Control links

Figure 5. HDMI Digital single link application using HDMIULC6-4SC6





477

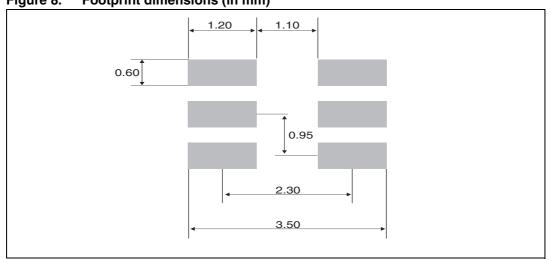
PCB layout considerations 2.1

For HDMI applications, $V_{\mbox{\footnotesize{CC}}}$ should not be connected. In this case the capacitor C in Figure 7. is not needed.

D+1 **HDMI D-1** Vcc Connector **GND** C = 100 nFD+2 Side D-2 **HDMIULC6-4SC6**

PCB layout considerations (V_{CC} connection is application dependent) Figure 7.

A differential impedance of 100 Ω must be respected in the layout. Both lines of the differential pair should have the same length.



Footprint dimensions (in mm) Figure 8.

Technical information HDMIULC6-4SC6

3 Technical information

3.1 Surge protection

The HDMIULC6-4SC6 is particularly optimized to perform ESD surge protection based on the rail to rail topology.

The clamping voltage V_{CL} can be calculated as follows:

$$V_{CL}$$
+ = $V_{TRANSIL}$ + V_F for positive surges
 V_{Cl} - = - V_F for negative surges

with:
$$V_F = V_T + R_d I_p$$

(V_F forward drop voltage) / (V_T forward drop threshold voltage)

and
$$V_{TRANSIL} = V_{BR} + R_{d_TRANSIL}$$
 . I_{P}

Calculation example

We assume that the value of the dynamic resistance of the clamping diode is typically: R_{d} = 0.5 Ω and V_{T} = 1.1 V.

We assume that the value of the dynamic resisteance of the transil diode is typically R_{d TRANSIL} = 0.5 Ω and V_{BR} = 6.1 V

For an IEC 61000-4-2 surge Level 4 (Contact Discharge: V_g = 8 kV, R_g = 330 Ω), V_{BUS} = +5 V, and, in first approximation, we assume that: I_p = V_g / R_g = 24 A.

We find:

$$V_{CL}$$
+ = +31.2 V V_{CL} - = -13.1 V

Note: The calculations do not take into account phenomena due to parasitic inductances.

3.2 Surge protection application example

If we consider that the connections from the pin V_{BUS} to V_{CC} , from I/O to data line, and from GND to PCB GND plane are two tracks 10 mm long and 0.5 mm wide, we can assume that the parasitic inductances, L_{VBUS} , $L_{I/O}$, and L_{GND} , of these tracks are about 6 nH. So when an IEC 61000-4-2 surge occurs on the data line, due to the rise time of this spike (tr = 1 ns), the voltage V_{CL} has an extra value equal to $L_{I/O}$.dI/dt + L_{GND} .dI/dt.

The dl/dt is calculated as: dl/dt = lp/tr = 24 A/ns for an IEC 61000-4-2 surge level 4 (contact discharge V_q = 8 kV, R_q = 330 Ω

The over voltage due to the parasitic inductances is:

$$L_{I/O}.dI/dt = L_{GND}.dI/dt = 6 \times 24 = 144 \text{ V}$$

By taking into account the effect of these parasitic inductances due to unsuitable layout, the clamping voltage will be:

We can reduce as much as possible these phenomena with simple layout optimization.

477

HDMIULC6-4SC6 Technical information

It's the reason why some recommendations have to be followed (see *Section 3.3: How to ensure good ESD protection*).

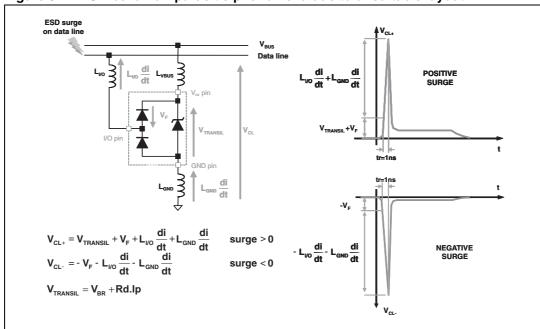


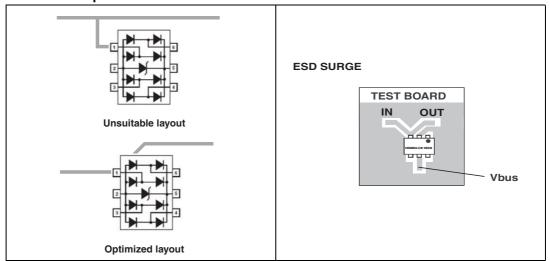
Figure 9. ESD behavior: parasitic phenomena due to unsuitable layout

3.3 How to ensure good ESD protection

While the HDMIULC6-4SC6 provides a high immunity to ESD surge, an efficient protection depends on the layout of the board. In the same way, with the rail to rail topology, the track from data lines to I/O pins, from V_{CC} to V_{BUS} pin, and from GND plane to GND pin must be as short as possible to avoid over voltages due to parasitic phenomena (see *Figure 9* and *Figure 10* for layout considerations).

Figure 10. ESD behavior: layout optimization

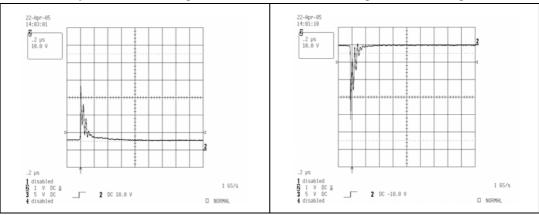
Figure 11. ESD behavior: measurement conditions



Technical information HDMIULC6-4SC6

Figure 12. Remaining voltage after the HDMIULC6-4SC6 during positive ESD surge

Figure 13. Remaining voltage after the HDMIULC6-4SC6 during negative ESD surge



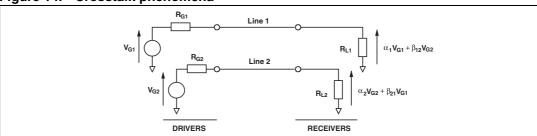
Note: The measurements have been done with the HDMIULC6-4SC6 in open circuit.

IMPORTANT:

An important precaution to take is to put the protection device as close as possible to the disturbance source (generally the connector).

3.4 Crosstalk behavior

Figure 14. Crosstalk phenomena



The crosstalk phenomena is due to the coupling between 2 lines. The coupling factor (β_{12} or β_{21}) increases when the gap across lines decreases, particularly in silicon dice. In the example above the expected signal on load R_{L2} is $\alpha_2 V_{G2}$, in fact the real voltage at this point has got an extra value $\beta_{21} V_{G1}$. This part of the V_{G1} signal represents the effect of the crosstalk phenomenon of the line 1 on the line 2. This phenomenon has to be taken into account when the drivers impose fast digital data or high frequency analog signals in the disturbing line. The perturbed line will be more affected if it works with low voltage signal or high load impedance (few $k\Omega$).

577

HDMIULC6-4SC6 Technical information

Figure 15. Analog crosstalk measurements

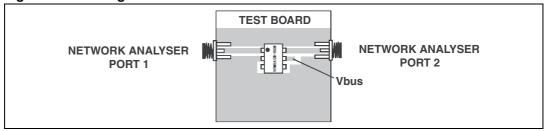
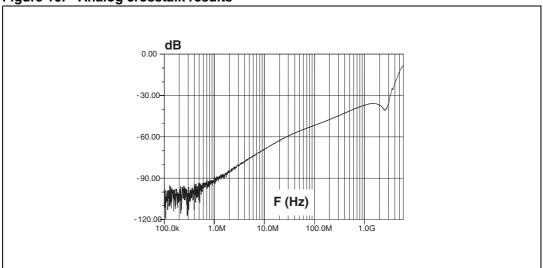


Figure 15 gives the measurement circuit for the analog application. In usual frequency range of analog signals (up to 240 MHz) the effect on disturbed line is less than -45 dB (see Figure 16).

Figure 16. Analog crosstalk results

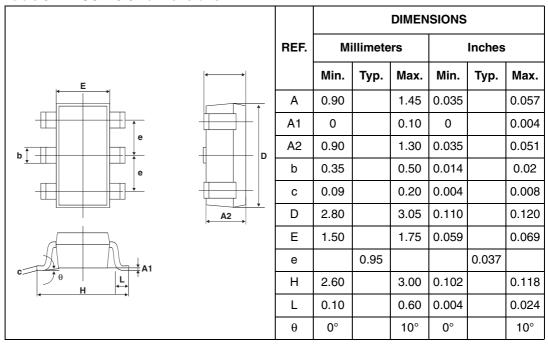


As the HDMIULC6-4SC6 is designed to protect high speed data lines, it must ensure a good transmission of operating signals. The frequency response (*Figure 4.*) gives attenuation information and shows that the HDMIULC6-4SC6 is well suitable for data line transmission up to 3.2 Gb/s.

Package information HDMIULC6-4SC6

4 Package information

Table 3. SOT23-6L dimensions



In order to meet environmental requirements, ST offers these devices in ECOPACK® packages. These packages have a lead-free second level interconnect. The category of second level interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com.

5 Ordering information

Ordering code	Marking	Package	Weight	Base qty	Delivery mode
HDMIULC6-4SC6	DL46	SOT23-6L	16.7 mg	3000	Tape and reel

6 Revision history

Date	Revision	Description of Changes
28-Mar-2006	1	First Issue
26-Jul-2006	2	Replaced technical information section.

57

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477