### **ABSOLUTE MAXIMUM RATINGS**

Voltage Range on VCC	0.5V to +6V
Voltage Range on Voute, Voute,	
VINP, VINN, SEN, DEN	$-0.5V$ to $(V_{CC} + 0.5V)^*$
Continuous Power Dissipation (T <sub>A</sub> = +70°	C)
μDFN (derate 4.8mW/°C above +70°C).	380.6mW

Operating Temperature Range	40°C to +85°C
Storage Temperature Range	55°C to +125°C
Lead Temperature (soldering,	10s)+300°C
	+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### RECOMMENDED OPERATING CONDITIONS

 $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ unless otherwise noted.})$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP MAX	UNITS
Supply Voltage	Vcc	(Note 1)	+2.97	+5.5	V

### DC ELECTRICAL CHARACTERISTICS

( $V_{CC}$  = +2.97V to +5.5V,  $T_A$  = -40°C to +85°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		TYP	MAX	UNITS
Supply Current	Icc	(Note 1)		5.7	9	mA
Input Capacitance	CIN	All pins (Note 2)			7	pF
Sample Capacitance	Cs	V <sub>INN</sub> and V <sub>INP</sub> (Note 2)		5		pF
Logic-Input Low	V <sub>IL</sub>	SEN and DEN inputs			0.3 x V <sub>CC</sub>	V
Logic-Input High	VIH	SEN and DEN inputs				V
Input Leakage	I <sub>IN</sub>	V <sub>INN</sub> or V <sub>INP</sub> , SEN = 0			1	μA
Input Voltage	V <sub>IN</sub>	$V_{IN} = V_{INP} - V_{INN}$	0		1.0	V
Output Voltage	V <sub>OUT</sub>	$V_{OUT} = V_{OUTP} - V_{OUTN}$ ; 100k $\Omega$ load on each output pin	0		1.0	V
Output Impedance	ROUTMAX	(Note 2)		1	1.3	kΩ
Output Capacitive Load	Cout	Capacitance for stable operation			50	pF
Total Input Referenced Voltage	Voo Diee	$V_{CC} = 2.9V$ , 1µs sample time, $V_{IN} = 6mV$		3.6	6.1	mV
Offset: Differential	Vos-diff	Voltco ( $V_{CC} = 2.9V \text{ to } 5.5V$ )			1	mV/V
Total Input Referenced Voltage	V00.05	$V_{CC} = 2.9V$ , 1µs sample time, $V_{IN} = 6mV$		3.4	8	mV
Offset: Single-Ended	Vos-se	Voltco (V <sub>CC</sub> = 2.9V to 5.5V)			1	mV/V

<sup>\*</sup>Subject to not exceeding +6V.

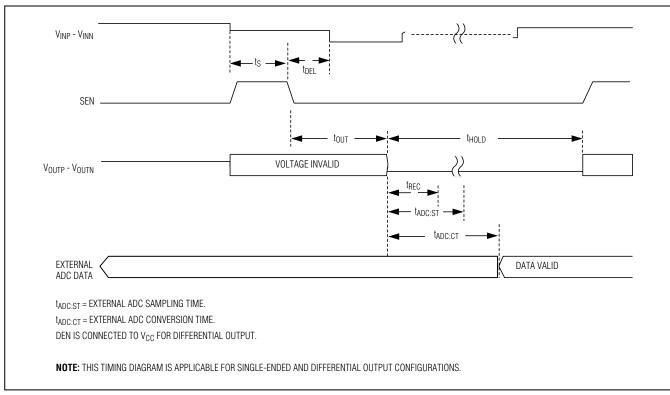
#### **AC ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = +2.97 \text{V to } +5.5 \text{V}, T_A = -40 ^{\circ}\text{C} \text{ to } +85 ^{\circ}\text{C}, \text{ unless otherwise noted.})$  (See the *Timing Diagram.*)

PARAMETER	SYMBOL	CONDITIONS		TYP	MAX	UNITS
Sample Time Minimum	ts	ts V <sub>OUT</sub> is within 0.4dB (Note 3)				ns
Delay Time Minimum	tDEL	t <sub>DEL</sub> (Note 4)				ns
Output Time  tout  Delay from SEN falling edge until valid output at Vout to 1% accuracy				2	μs	
Hold Time	tHOLD	(Note 5)	tout		100	μs
Output Step Recovery Time	toro	1V step, DEN = high			2	110
(Note 6)	†REC	3V step, DEN = high or low			3.5	μs

- Note 1: All voltages are referenced to ground. Currents entering the IC are specified positive and currents exiting the IC are negative.
- Note 2: Guaranteed by design.
- **Note 3:**  $V_{OUT}$  at the end of the 10µs hold time is within specified level of  $V_{IN}$  during the sample window; a 50 $\Omega$  resistor connected in series to both  $V_{INP}$  and  $V_{INN}$  ( $V_{INP}$   $V_{INN}$  = 1V). External capacitance to ground for both  $V_{INP}$  and  $V_{INN}$  is approximately 10pF.
- Note 4: The sampling capacitor must be removed from the input signal before the input signal changes. Therefore, the SEN pin must be low for a short period of time, t<sub>DEL</sub>, before the input changes.
- Note 5: V<sub>OUT</sub> at the end of the hold time is within 1% of V<sub>IN</sub> during the sample window (V<sub>INP</sub> V<sub>INN</sub> = 1V).
- **Note 6:** Voltage step applied across V<sub>OUTP</sub> to V<sub>OUTN</sub> through a 5pF capacitor connected to each pin. This models the load presented by an ADC while it is sampling the DS1843's output. See the *Output Buffer* section. Settled within 1% of initial voltage.

## Timing Diagram

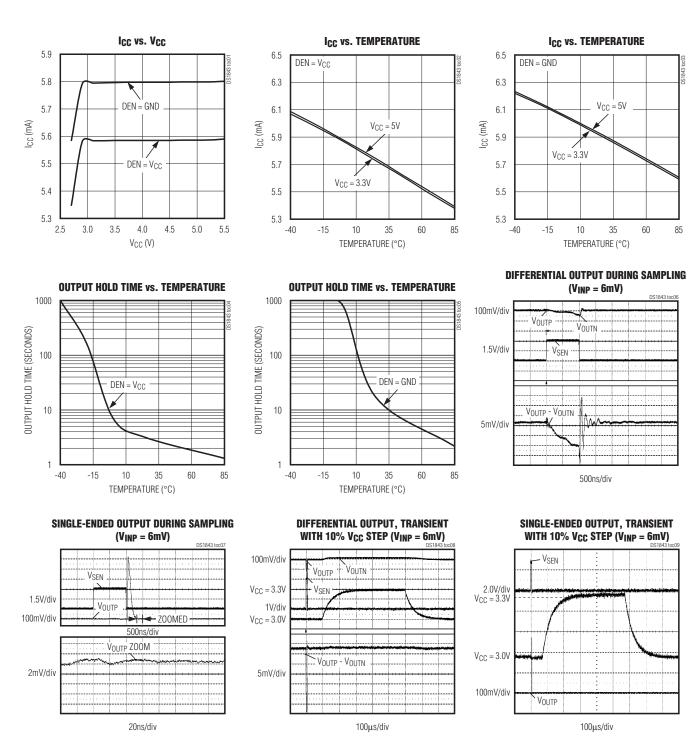


N/XX/N

## **Typical Operating Characteristics**

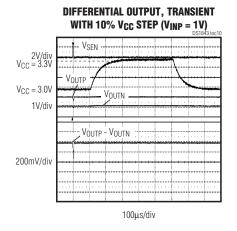
MIXIM

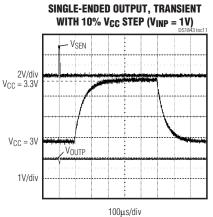
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$ 

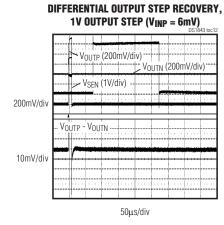


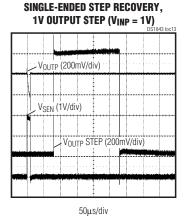
## Typical Operating Characteristics (continued)

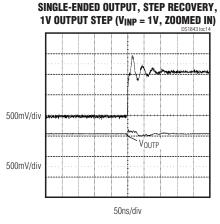
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$ 

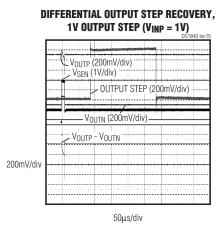


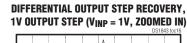


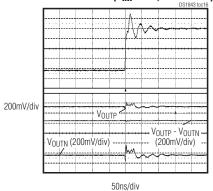










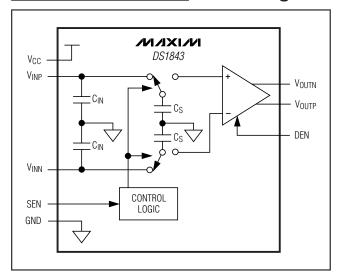


/VIXI/VI

### Pin Description

PIN	NAME	FUNCTION				
1	Vcc	Power-Supply Input				
2	VINP	sitive Voltage Input. Input to sample circuit.				
3	VINN	Negative Voltage Input. Input to sample circuit.				
4	DEN	Differential Output Enable. Connect to V <sub>CC</sub> for differential output or GND for single-ended output.				
5	GND	Ground Terminal				
6	Voutn	Sampled Voltage Negative Output. Buffered output of the hold capacitor. Keep unconnected or connect to GND for single-ended output mode.				
7	Voutp	Sampled Voltage Positive Output and Single-Ended Output. Buffered output of the hold capacitor.				
8	SEN	Sample Enable. Enables input sampling. This input is pulsed.				

### **Block Diagram**



## **Detailed Description**

The DS1843 consists of a fully differential sampling capacitor, switches, and a differential output buffer. It is designed to operate in fiber optic burst-mode systems; however, it can be used in other applications requiring a fast sample-and-hold circuit. The output can be configured for single-ended operations.

#### **Input Sampling Capacitor**

The input voltage is sampled using a 5pF capacitor on the positive input and another on the negative input. The capacitors are connected to the input when SEN is high. In addition to the sampling capacitors, the inputs also have parasitic capacitance (C<sub>IN</sub>). These capacitors must fully charge before SEN is switched to low in order to ensure accurate sampling. An RC time constant is created by the resistance of the voltage source connected to the DS1843's input and the capacitances on this node. See the *Applications Information* section for details.

### **Output Buffer**

After sampling is complete, the sampling capacitor is switched to the output buffer. This buffer requires a small amount of time to settle, tout. When an ADC is used to measure the DS1843's output, a step occurs at the ADC's input caused by the ADC's internal sampling capacitor. The DS1843's recovery time, trec, is dependent on the size of the ADC's sampling capacitor and the voltage applied across the ADC. To maximize accuracy, the ADC's sampling speed (ADC clock frequency) should be reduced until the ADC's conversion window (tadc:st, as shown in the *Timing Diagram*) is larger than the DS1843's recovery time. Refer to the ADC's documentation for tadc:st.

### **Sampling Time and Output Error**

As the sampling time (ts) is decreased, the output error increases. The output error is largely dependent on the settling time of the sampling capacitor and, to a lesser degree, the output buffer's gain error and offset voltage. Settling time can be reduced by driving the DS1843 with a lower impedance. In a typical fiber optic application, a current is applied across a  $5k\Omega$  resistor. By using a stronger current source, the resistance and the settling time can be reduced (see the *Applications Information* section for details).

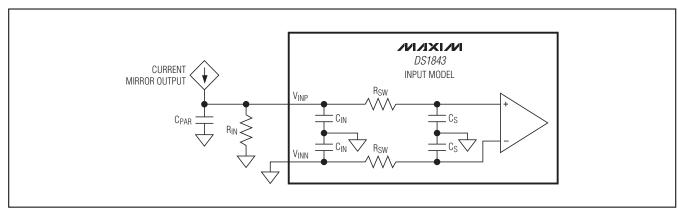


Figure 1. Input Impedances for Settling Time Calculations Diagram

## **Applications Information**

### **Power-Supply Decoupling**

To achieve the best results when using the DS1843, decouple the power-supply pin, VCC, with a 0.01µF or 0.1µF capacitor. Use a high-quality X7R or equivalent ceramic surface-mount capacitor.

### **DS1843 Estimated Settling Time**

The settling time is dependent on the gain ratio of the current mirror used at the input of the DS1843. For example, the MAX4007 includes a 10:1 ratio current mirror. This requires a  $5k\Omega$  resistor to create a 1V full-scale output with 2mA current input to the MAX4007. This resistor can be decreased to  $2.5k\Omega$  by using the DS1842, which has a 5:1 ratio current mirror.

#### **Variable Definitions:**

R<sub>IN</sub>: Input resistor. The current mirror creates a voltage across this resistor.

Rsw: Resistance of series switch that connects internal circuitry to input pins after t<sub>IST</sub> time.

CIN: 7pF parasitic (ESD) capacitor.

CPAR: External parasitic capacitance. A current mirror's output and typical trace capacitance are less than 10pF.

Cs: 5pF sample capacitor.

t<sub>IST</sub>: Internal settling time based on ts from the AC electrical specification. The minimum ts includes one time constant. t<sub>IST</sub> removes this time constant.

tRC: RC settling time of the input.

Figure 1 shows the simplified diagram of input impedances for settling time calculations. Sample time is divided into two parts:

 t<sub>IST</sub>: Internal settling time (max 250ns). During this time, voltage V<sub>IN</sub> (V<sub>INP</sub> - V<sub>INN</sub>) rises with a time constant of:

2) tRC: During this period two things happen:

a. Input  $V_{\text{IN}}$  keeps increasing from its value at t<sub>IST</sub> to its final value with a new time constant of:

$$\sqrt{\left\{\left(\mathsf{R}_{\mathsf{IN}}\times\left(\mathsf{C}_{\mathsf{IN}}+\mathsf{C}_{\mathsf{PAR}}\right)\right)^{2}+\left(\mathsf{R}_{\mathsf{SW}}\times\mathsf{C}_{\mathsf{S}}\right)^{2}\right\}}$$

b. Rsw and Cs track this  $V_{IN}$  (input) with a time constant of Rsw x Cs, which is 12.5ns (worst case).

#### Example:

Approximate accuracy calculations can be done for an input voltage based on the above impedance values. These calculations can be divided into three parts.

Accuracy of input at t<sub>IST</sub> (250ns):

$$Accuracy = 1 - e^{\frac{-t_1}{\left[R_{IN} \times \left(C_{IN} + C_{PAR}\right)\right]}}$$

where  $t_1 = t_{IST} = 250$ ns.

At tIST the internal circuit tags input impedance. This causes charge redistribution to occur, which causes a dip in the input voltage. The worst-case value of the input voltage at tIST is:

$$V_{\text{IN@t}_{\text{IST}}} = \left[1 - \frac{C_{\text{S}}}{\left(C_{\text{IN}} + C_{\text{PAR}} + C_{\text{S}}\right)}\right] \times \left[1 - e^{\left[\frac{-t_{\text{IST}}}{\left[R_{\text{IN}} \times \left(C_{\text{IN}} + C_{\text{PAR}}\right)\right]}\right]} \times V_{\text{IN}}\right]$$

2) Accuracy of internal circuitry between ts - tist:

$$Accuracy = 1 - e^{\frac{-t_2}{(R_{SW} \times C_S)}}$$

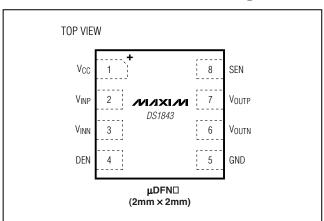
where  $t_2 = (t_S - t_{IST})$  and  $(R_{SW} \times C_S) \sim = 12ns$ .

3) Total accuracy of input at sampling time, ts:

$$Accuracy = \left[1 - \left(1 - V_{IN@t_{IST}}\right) \times e^{\frac{-t_2}{newRC}}\right] \times \left[1 - e^{\frac{-t_2}{\left(R_{SW} \times C_S\right)}}\right]$$

where newRC = 
$$\sqrt{\left\{\left(R_{IN} \times \left(C_{IN} + C_{PAR}\right)\right)^2 + \left(R_{SW} \times C_S\right)^2\right\}}$$

## Pin Configuration



## Package Information

For the latest package outline information and land patterns (footprints), go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8 μDFN	L822+1	<u>21-0164</u>	<u>90-0005</u>

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.

B \_\_\_\_\_\_Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 408-737-7600