

The AT89C51ID2 retains all features of the Atmel 80C52 with 256 bytes of internal RAM, a 10-source 4-level interrupt controller and three timer/counters.

In addition, the AT89C51ID2 has a Programmable Counter Array, an XRAM of 1792 bytes, a Hardware Watchdog Timer, SPI and Keyboard, a more versatile serial channel that facilitates multiprocessor communication (EUART) and a speed improvement mechanism (X2 mode).

The fully static design of the AT89C51ID2 allows to reduce system power consumption by bringing the clock frequency down to any value, even DC, without loss of data.

The AT89C51ID2 has 2 software-selectable modes of reduced activity and 8-bit clock prescaler for further reduction in power consumption. In the Idle mode the CPU is frozen while the peripherals and the interrupt system are still operating. In the power-down mode the RAM is saved and all other functions are inoperative.

The added features of the AT89C51ID2 make it more powerful for applications that need pulse width modulation, high speed I/O and counting capabilities such as alarms, motor control, corded phones, smart card readers.

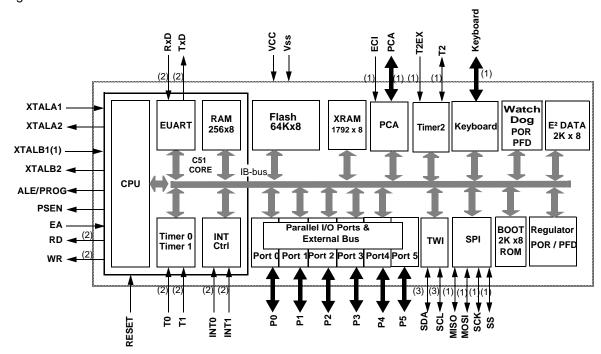
Table 1. Memory Size and I/O pins

_	AT89C51ID2	Flash (bytes)	XRAM (bytes)	TOTAL RAM (bytes)	I/O
	PLCC44/VQFP44	64K	1792	2048	34

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# **Block Diagram**

Figure 1. Block Diagram



- (1): Alternate function of Port 1
- (2): Alternate function of Port 3
- (3): Alternate function of Port I2



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# **SFR Mapping**

The Special Function Registers (SFRs) of the AT89C51ID2 fall into the following categories:

- C51 core registers: ACC, B, DPH, DPL, PSW, SP
- I/O port registers: P0, P1, P2, P3, Pl2
- Timer registers: T2CON, T2MOD, TCON, TH0, TH1, TH2, TMOD, TL0, TL1, TL2, RCAP2L, RCAP2H
- Serial I/O port registers: SADDR, SADEN, SBUF, SCON
- PCA (Programmable Counter Array) registers: CCON, CCAPMx, CL, CH, CCAPxH, CCAPxL (x: 0 to 4)
- Power and clock control registers: PCON
- Hardware Watchdog Timer registers: WDTRST, WDTPRG
- Interrupt system registers: IE0, IPL0, IPH0, IE1, IPL1, IPH1
- · Keyboard Interface registers: KBE, KBF, KBLS
- SPI registers: SPCON, SPSTR, SPDAT
- 2-wire Interface registers: SSCON, SSCS, SSDAT, SSADR
- BRG (Baud Rate Generator) registers: BRL, BDRCON
- Flash register: FCON
- Clock Prescaler register: CKRL
- 32 kHz Sub Clock Oscillator registers: CKSEL, OSSCON
- Others: AUXR, AUXR1, CKCON0, CKCON1





#### Table 2. C51 Core SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
ACC	E0h	Accumulator								
В	F0h	B Register								
PSW	D0h	Program Status Word	CY	AC	F0	RS1	RS0	OV	F1	Р
SP	81h	Stack Pointer								
DPL	82h	Data Pointer Low byte								
DPH	83h	Data Pointer High byte								

Table 3. System Management SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
PCON	87h	Power Control	SMOD1	SMOD0	-	POF	GF1	GF0	PD	IDL
AUXR	8Eh	Auxiliary Register 0	-	-	МО		XRS1	XRS0	EXTRA M	АО
AUXR1	A2h	Auxiliary Register 1	-	-	ENBOO T	-	GF3	0	-	DPS
CKRL	97h	Clock Reload Register	-	-	-	-	=	-	-	=
CKSEL	85h	Clock Selection Register	-	-	-	-	-	-	-	CKS
OSCON	86h	Oscillator Control Register	-	-	-	-	-	SCLKT0	OscBEn	OscAEn
CKCKON0	8Fh	Clock Control Register 0	TWIX2	WDTX2	PCAX2	SIX2	T2X2	T1X2	T0X2	X2
CKCKON1	AFh	Clock Control Register 1	-	-	-	-	-	-	-	SPIX2

#### Table 4. Interrupt SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
IEN0	A8h	Interrupt Enable Control 0	EA	EC	ET2	ES	ET1	EX1	ET0	EX0
IEN1	B1h	Interrupt Enable Control 1	-	-	-	-	-	ESPI	ETWI	EKBD
IPH0	B7h	Interrupt Priority Control High 0	-	PPCH	PT2H	PSH	PT1H	PX1H	PT0H	PX0H
IPL0	B8h	Interrupt Priority Control Low 0	-	PPCL	PT2L	PSL	PT1L	PX1L	PT0L	PX0L
IPH1	B3h	Interrupt Priority Control High 1	-	-	-	-	-	SPIH	IE2CH	KBDH
IPL1	B2h	Interrupt Priority Control Low 1	-	-	-	-	-	SPIL	IE2CL	KBDL

#### Table 5. Port SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
P0	80h	8-bit Port 0								
P1	90h	8-bit Port 1								
P2	A0h	8-bit Port 2								
P3	B0h	8-bit Port 3								
P4	C0h	8-bit Port 4								
P5	E8h	8-bit Port 5	-	-	-	-				
P5	C7h	8-bit Port 5 (byte addressable)								

# Table 6. Flash and EEPROM Data Memory SFR

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
FCON	D1h	Flash Control	FPL3	FPL2	FPL1	FPL0	FPS	FMOD1	FMOD0	FBUSY
EECON		EEPROM data Control								

#### Table 7. Timer SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
TCON	88h	Timer/Counter 0 and 1 Control	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0
TMOD	89h	Timer/Counter 0 and 1 Modes	GATE1	C/T1#	M11	M01	GATE0	C/T0#	M10	M00
TL0	8Ah	Timer/Counter 0 Low Byte								
TH0	8Ch	Timer/Counter 0 High Byte								
TL1	8Bh	Timer/Counter 1 Low Byte								
TH1	8Dh	Timer/Counter 1 High Byte								
WDTRST	A6h	WatchDog Timer Reset								
WDTPRG	A7h	WatchDog Timer Program	-	-	-	-	-	WTO2	WTO1	WTO0
T2CON	C8h	Timer/Counter 2 control	TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2#	CP/RL2#
T2MOD	C9h	Timer/Counter 2 Mode	-	-	-	-	-	-	T2OE	DCEN
RCAP2H	CBh	Timer/Counter 2 Reload/Capture High byte								
RCAP2L	CAh	Timer/Counter 2 Reload/Capture Low byte								
TH2	CDh	Timer/Counter 2 High Byte								
TL2	CCh	Timer/Counter 2 Low Byte								





#### Table 8. PCA SFRs

Mnemo -nic	Add	Name	7	6	5	4	3	2	1	0
CCON	D8h	PCA Timer/Counter Control	CF	CR	-	CCF4	CCF3	CCF2	CCF1	CCF0
CMOD	D9h	PCA Timer/Counter Mode	CIDL	WDTE	-	-	-	CPS1	CPS0	ECF
CL	E9h	PCA Timer/Counter Low byte								
СН	F9h	PCA Timer/Counter High byte								
CCAPM0 CCAPM1 CCAPM2 CCAPM3 CCAPM4	DDh	PCA Timer/Counter Mode 0 PCA Timer/Counter Mode 1 PCA Timer/Counter Mode 2 PCA Timer/Counter Mode 3 PCA Timer/Counter Mode 4	-	ECOM0 ECOM1 ECOM2 ECOM3 ECOM4	CAPP0 CAPP1 CAPP2 CAPP3 CAPP4	CAPN0 CAPN1 CAPN2 CAPN3 CAPN4	MAT0 MAT1 MAT2 MAT3 MAT4	TOG0 TOG1 TOG2 TOG3 TOG4	PWM0 PWM1 PWM2 PWM3 PWM4	ECCF0 ECCF1 ECCF2 ECCF3 ECCF4
CCAP0H CCAP1H CCAP2H CCAP3H CCAP4H	FBh FCh FDh	PCA Compare Capture Module 0 H PCA Compare Capture Module 1 H PCA Compare Capture Module 2 H PCA Compare Capture Module 3 H PCA Compare Capture Module 4 H	CCAP1H7 CCAP2H7 CCAP3H7	CCAP2H6	CCAP1H5 CCAP2H5 CCAP3H5	CCAP0H4 CCAP1H4 CCAP2H4 CCAP3H4 CCAP4H4	CCAP0H3 CCAP1H3 CCAP2H3 CCAP3H3 CCAP4H3	CCAP0H2 CCAP1H2 CCAP2H2 CCAP3H2 CCAP4H2	CCAP0H1 CCAP1H1 CCAP2H1 CCAP3H1 CCAP4H1	CCAP0H0 CCAP1H0 CCAP2H0 CCAP3H0 CCAP4H0
CCAP0L CCAP1L CCAP2L CCAP3L CCAP4L	EAh EBh ECh EDh EEh	PCA Compare Capture Module 0 L PCA Compare Capture Module 1 L PCA Compare Capture Module 2 L PCA Compare Capture Module 3 L PCA Compare Capture Module 4 L	CCAP0L7 CCAP1L7 CCAP2L7 CCAP3L7 CCAP4L7	CCAP0L6 CCAP1L6 CCAP2L6 CCAP3L6 CCAP4L6	CCAP0L5 CCAP1L5 CCAP2L5 CCAP3L5 CCAP4L5	CCAP0L4 CCAP1L4 CCAP2L4 CCAP3L4 CCAP4L4	CCAP0L3 CCAP1L3 CCAP2L3 CCAP3L3 CCAP4L3	CCAP0L2 CCAP1L2 CCAP2L2 CCAP3L2 CCAP4L2	CCAP0L1 CCAP1L1 CCAP2L1 CCAP3L1 CCAP4L1	CCAP0L0 CCAP1L0 CCAP2L0 CCAP3L0 CCAP4L0

#### Table 9. Serial I/O Port SFRs

		1 OIL OI ILO								
Mnemonic	Add	Name	7	6	5	4	3	2	1	0
SCON	98h	Serial Control	FE/SM0	SM1	SM2	REN	TB8	RB8	TI	RI
SBUF	99h	Serial Data Buffer								
SADEN	B9h	Slave Address Mask								
SADDR	A9h	Slave Address								
BDRCON	9Bh	Baud Rate Control				BRR	TBCK	RBCK	SPD	SRC
BRL	9Ah	Baud Rate Reload								

# Table 10. SPI Controller SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
SPCON	C3h	SPI Control	SPR2	SPEN	SSDIS	MSTR	CPOL	СРНА	SPR1	SPR0
SPSTA	C4h	SPI Status	SPIF	WCOL	SSERR	MODF	-	-	-	-
SPDAT	C5h	SPI Data	SPD7	SPD6	SPD5	SPD4	SPD3	SPD2	SPD1	SPD0

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#### Table 11. Two-Wire Interface Controller SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
SSCON	93h	Synchronous Serial control	SSCR2	SSPE	SSSTA	SSSTO	SSI	SSAA	SSCR1	SSCR0
SSCS	94h	Synchronous Serial Status	SSC4	SSC3	SSC2	SSC1	SSC0	0	0	0
SSDAT	95h	Synchronous Serial Data	SSD7	SSD6	SSD5	SSD4	SSD3	SSD2	SSD1	SSD0
SSADR	96h	Synchronous Serial Address	SSA7	SSA6	SSA5	SSA4	SSA3	SSA2	SSA1	SSGC

Table 12. Keyboard Interface SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
KBLS	9Ch	Keyboard Level Selector	KBLS7	KBLS6	KBLS5	KBLS4	KBLS3	KBLS2	KBLS1	KBLS0
KBE	9Dh	Keyboard Input Enable	KBE7	KBE6	KBE5	KBE4	KBE3	KBE2	KBE1	KBE0
KBF	9Eh	Keyboard Flag Register	KBF7	KBF6	KBF5	KBF4	KBF3	KBF2	KBF1	KBF0

 Table 13.
 EEPROM data Memory SFR

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
EECON	D2h	EEPROM Data Control							EEE	EEBUSY





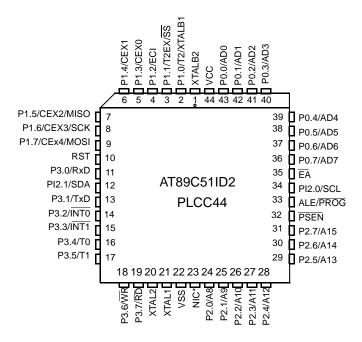
Table below shows all SFRs with their address and their reset value.

Table 14. SFR Mapping

	Bit addressable			No	on Bit addressat	ole			
	0/8	1/9	2/A	3/B	4/C	5/D	6/E	7/F	
F8h	PI2 XXXX XX11	CH 0000 0000	CCAP0H XXXX XXXX	CCAP1H XXXX XXXX	CCAP2H XXXX XXXX	CCAP3H XXXX XXXX	CCAP4H XXXX XXXX		FFh
F0h	B 0000 0000								F7h
E8h	P5 bit addressable 1111 1111	CL 0000 0000	CCAP0L XXXX XXXX	CCAP1L XXXX XXXX	CCAP2L XXXX XXXX	CCAP3L XXXX XXXX	CCAP4L XXXX XXXX		EFh
E0h	ACC 0000 0000								E7h
D8h	CCON 00X0 0000	CMOD 00XX X000	CCAPM0 X000 0000	CCAPM1 X000 0000	CCAPM2 X000 0000	CCAPM3 X000 0000	CCAPM4 X000 0000		DFh
D0h	PSW 0000 0000	FCON (1) XXXX 0000	EECON xxxx xx00						D7h
C8h	T2CON 0000 0000	T2MOD XXXX XX00	RCAP2L 0000 0000	RCAP2H 0000 0000	TL2 0000 0000	TH2 0000 0000			CFh
C0h	P4 1111 1111			SPCON 0001 0100	SPSTA 0000 0000	SPDAT XXXX XXXX		P5 byte Addressable 1111 1111	C7h
B8h	IPL0 X000 000	SADEN 0000 0000							BFh
B0h	P3 1111 1111	IEN1 XXXX X000	IPL1 XXXX X000	IPH1 XXXX X111				IPH0 X000 0000	B7h
A8h	IEN0 0000 0000	SADDR 0000 0000						CKCON1 XXXX XXX0	AFh
A0h	P2 1111 1111		AUXR1 XXXX X0X0				WDTRST XXXX XXXX	WDTPRG XXXX X000	A7h
98h	SCON 0000 0000	SBUF XXXX XXXX	BRL 0000 0000	BDRCON XXX0 0000	KBLS 0000 0000	KBE 0000 0000	KBF 0000 0000		9Fh
90h	P1 1111 1111			SSCON 0000 0000	SSCS 1111 1000	SSDAT 1111 1111	SSADR 1111 1110	CKRL 1111 1111	97h
88h	TCON 0000 0000	TMOD 0000 0000	TL0 0000 0000	TL1 0000 0000	TH0 0000 0000	TH1 0000 0000	AUXR XX00 1000	CKCON0 0000 0000	8Fh
80h	P0 1111 1111	SP 0000 0111	DPL 0000 0000	DPH 0000 0000		CKSEL XXXX XXX0	OSSCON XXXX X001	PCON 00X1 0000	87h
	0/8	1/9	2/A	3/B	4/C	5/D	6/E	7/F	

Reserved

# **Pin Configurations**



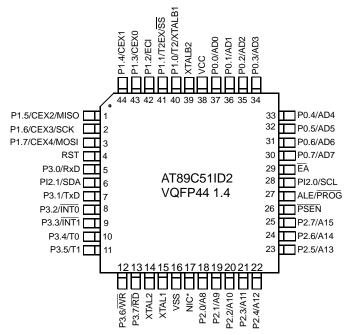






 Table 15. Pin Description

	Pin N	umber	T				
Mnemonic	PLCC44	VQFP44	Туре	Name and Function			
V <sub>SS</sub>	22	16	I	Ground: 0V reference			
V <sub>CC</sub>	44	38	I	<b>Power Supply:</b> This is the power supply voltage for normal, idle and power-down operation			
P0.0 - P0.7	43 - 36	37 - 30	I/O	<b>Port 0</b> : Port 0 is an open-drain, bidirectional I/O port. Port 0 pins that have 1s written to them float and can be used as high impedance inputs. Port 0 must be polarized to $V_{CC}$ or $V_{SS}$ in order to prevent any parasitic current consumption. Port 0 is also the multiplexed low-order address and data bus during access to external program and data memory. In this application, it uses strong internal pull-up when emitting 1s. Port 0 also inputs the code bytes during EPROM programming. External pull-ups are required during program verification during which P0 outputs the code bytes.			
P1.0 - P1.7	2 - 9	40 - 44 1 - 3	I/O	Port 1: Port 1 is an 8-bit bidirectional I/O port with internal pull-ups. Port 1 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 1 pins that are externally pulled low will source current because of the internal pull-ups. Port 1 also receives the low-order address byte during memory programming and verification.  Alternate functions for AT89C51ID2 Port 1 include:			
	2	40	I/O	P1.0: Input/Output			
			I/O	T2 (P1.0): Timer/Counter 2 external count input/Clockout			
			I	XTALB1 (P1.0): Sub Clock input to the inverting oscillator amplifier			
	3	41	I/O	P1.1: Input/Output			
			1	T2EX: Timer/Counter 2 Reload/Capture/Direction Control			
			1	SS: SPI Slave Select			
	4	42	I/O	P1.2: Input/Output			
			I	ECI: External Clock for the PCA			
	5	43	I/O	P1.3: Input/Output			
			I/O	CEX0: Capture/Compare External I/O for PCA module 0			
	6	44	I/O	P1.4: Input/Output			
			I/O	CEX1: Capture/Compare External I/O for PCA module 1			
	7	1	I/O	P1.5: Input/Output			
			I/O	CEX2: Capture/Compare External I/O for PCA module 2			
			I/O	MISO: SPI Master Input Slave Output line			
				When SPI is in master mode, MISO receives data from the slave peripheral. When SPI is in slave mode, MISO outputs data to the master controller.			
	8	2	I/O	P1.6: Input/Output			
			I/O	CEX3: Capture/Compare External I/O for PCA module 3			
			I/O	SCK: SPI Serial Clock			
	9	3	I/O	P1.7: Input/Output:			
			I/O	CEX4: Capture/Compare External I/O for PCA module 4			

Table 15. Pin Description (Continued)

	Pin N	umber	T		
Mnemonic	PLCC44	VQFP44	Туре	Name and Function	
			I/O	MOSI: SPI Master Output Slave Input line	
				When SPI is in master mode, MOSI outputs data to the slave peripheral. When SPI is in slave mode, MOSI receives data from the master controller.	
XTALA1	21	15	I	<b>Crystal A 1:</b> Input to the inverting oscillator amplifier and input to the internal clock generator circuits.	
XTALA2	20	14	0	Crystal A 2: Output from the inverting oscillator amplifier	
XTALB1	2	40	I	Crystal B 1: (Sub Clock) Input to the inverting oscillator amplifier and input to the internal clock generator circuits.	
XTALB2	1	39	0	Crystal B 2: (Sub Clock) Output from the inverting oscillator amplifier	
P2.0 - P2.7	24 - 31	18 - 25	I/O	Port 2: Port 2 is an 8-bit bidirectional I/O port with internal pull-ups. Port 2 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 2 pins that are externally pulled low will source current because of the internal pull-ups. Port 2 emits the high-order address byte during fetches from external program memory and during accesses to external data memory that use 16-bit addresses (MOVX @DPTR).In this application, it uses strong internal pull-ups emitting 1s. During accesses to external data memory that use 8-bit addresses (MOVX @Ri), port 2 emits the contents of the P2 SFR.	
P3.0 - P3.7	11, 13 - 19	5, 7 - 13	I/O	<b>Port 3:</b> Port 3 is an 8-bit bidirectional I/O port with internal pull-ups. Port 3 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 3 pins that are externally pulled low will source current because of the internal pull-ups. Port 3 also serves the special features of the 80C51 family, as listed below.	
	11	5	1	RXD (P3.0): Serial input port	
	13	7	0	TXD (P3.1): Serial output port	
	14	8	1	INT0 (P3.2): External interrupt 0	
	15	9	1	INT1 (P3.3): External interrupt 1	
	16	10	1	T0 (P3.4): Timer 0 external input	
	17	11	1	T1 (P3.5): Timer 1 external input	
	18	12	0	WR (P3.6): External data memory write strobe	
	19	13	0	RD (P3.7): External data memory read strobe	
P4.0 - P4.7	-	-	I/O	<b>Port 4:</b> Port 4 is an 8-bit bidirectional I/O port with internal pull-ups. Port 5 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 4 pins that are externally pulled low will source current because of the internal pull-ups.	
P5.0 - P5.7	-	-	I/O	<b>Port 5:</b> Port 5 is an 8-bit bidirectional I/O port with internal pull-ups. Port 3 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 5 pins that are externally pulled low will source current because of the internal pull-ups.	
Pl2.0 - Pl2.1	34, 12	28, 6		Port 12: Port 12 is an open drain. It can be used as inputs (must be polarized to Vcc with external resistor to prevent any parasitic current consumption).	
	34	28	I/O	SCL (PI2.0): 2-wire Serial Clock	
				SCL output the serial clock to slave peripherals SCL input the serial clock from master	





Table 15. Pin Description (Continued)

	Pin N	umber	_	
Mnemonic	PLCC44	VQFP44	Туре	Name and Function
	12	6	I/O	SDA (PI2.1): 2-wire Serial Data
	12	6	1/0	SDA is the bidirectional 2-wire data line
RST	10	4	I	<b>Reset:</b> A high on this pin for two machine cycles while the oscillator is running, resets the device. An internal diffused resistor to $V_{SS}$ permits a power-on reset using only an external capacitor to $V_{CC}$ . This pin is an output when the hardware watchdog forces a system reset.
ALE/PROG	33	27	O (I)	Address Latch Enable/Program Pulse: Output pulse for latching the low byte of the address during an access to external memory. In normal operation, ALE is emitted at a constant rate of 1/6 (1/3 in X2 mode) the oscillator frequency, and can be used for external timing or clocking. Note that one ALE pulse is skipped during each access to external data memory. This pin is also the program pulse input (PROG) during Flash programming. ALE can be disabled by setting SFR's AUXR.0 bit. With this bit set, ALE will be inactive during internal fetches.
PSEN	32	26	0	Program Strobe ENable: The read strobe to external program memory. When executing code from the external program memory, PSEN is activated twice each machine cycle, except that two PSEN activations are skipped during each access to external data memory. PSEN is not activated during fetches from internal program memory.
EA	35	29	ı	External Access Enable: EA must be externally held low to enable the device to fetch code from external program memory locations 0000H to FFFFH. If security level 1 is programmed, EA will be internally latched on Reset.

### **Oscillators**

#### Overview

Two oscillators are available (for AT8xC511xD2 devices only, the others part number provide only the main high frequency oscillator):

- OSCA used for high frequency: Up to 40 MHz
- OSCB used for low frequency: 32.768 kHz

Several operating modes are available and programmable by software:

- to switch OSCA to OSCB and vice-versa
- to stop OSCA or OSCB to reduce consumption

In order to optimize the power consumption and the execution time needed for a specific task, an internal prescaler feature has been implemented between the selected oscillator and the CPU.

#### Registers

Table 16. CKSEL Register (for AT8xC51Ix2 only)

CKSEL - Clock Selection Register (85h)

7	6	5	4	3	2	1	0
-	-	-	-	-	-	-	CKS

Bit Number	Bit Mnemonic	Description
7	-	Reserved
6	-	Reserved
5	-	Reserved
4	-	Reserved
3	-	Reserved
2	-	Reserved
1	-	Reserved
0	CKS	CPU Oscillator Select Bit: (CKS)  Cleared, CPU and peripherals connected to OSCB  Set, CPU and peripherals connected to OSCA  Programmed by hardware after a Power-up regarding Hardware Security Byte (HSB).HSB.OSC (Default setting, OSCA selected)

Reset Value = 0000 000'HSB.OSC'b (see Hardware Security Byte (HSB)) Not bit addressable





Table 17. OSCCON Register (for AT8xC51Ix2 only)

OSCCON- Oscillator Control Register (86h)

•	-	-	•	-	SCLKIO	OscBEn	OSCAEN
Bit Number	Bit Mnemonic	Description					
7	-	Reserved					
6	-	Reserved					
5	-	Reserved					
4	_	Reserved					

SCLKT0 Cleared by software to select T0 pin
Set by software to select T0 Sub Clock
Cleared by hardware after a Power Up

Set by software to run OscB

Set by software to run OscA

Sub Clock Timer0

OscB enable bit

OscA enable bit

Reserved

OscBEn Cleared by software to stop OscB
Programmed by hardware after a Power-up regarding HSB.OSC (Default cleared, OSCB stopped)

OscAEn Cleared by software to stop OscA
Programmed by hardware after a Power-up regarding HSB.OSC(Default Set, OSCA runs)

Reset Value = XXXX X0'HSB.OSC"HSB.OSC'b (see Hardware Security Byte (HSB)) Not bit addressable

Table 18. CKRL Register

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CKRL - Clock Reload Register

7 6 5 4 3 2 1 0

Bit Number	Mnemonic	Description
7:0	CKRL	Clock Reload Register: Prescaler value

Reset Value = 1111 1111b Not bit addressable

Table 19. PCON Register

PCON - Power Control Register (87h)

7	6	5	4	3	2	1	0
SMOD1	SMOD0	-	POF	GF1	GF0	PD	IDL

Bit Number	Bit Mnemonic	Description			
7	SMOD1	Serial port Mode bit 1 Set to select double baud rate in mode 1, 2 or 3.			
6	SMOD0	Serial port Mode bit 0 Cleared to select SM0 bit in SCON register. Set to select FE bit in SCON register.			
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.			
4	POF	Power-Off Flag Cleared to recognize next reset type. Set by hardware when VCC rises from 0 to its nominal voltage. Can also be set by software.			
3	GF1	General purpose Flag Cleared by software for general purpose usage. Set by software for general purpose usage.			
2	GF0	General purpose Flag Cleared by software for general purpose usage. Set by software for general purpose usage.			
1	PD	Power-Down mode bit Cleared by hardware when reset occurs. Set to enter power-down mode.			
0	IDL	Idle mode bit Cleared by hardware when interrupt or reset occurs. Set to enter idle mode.			

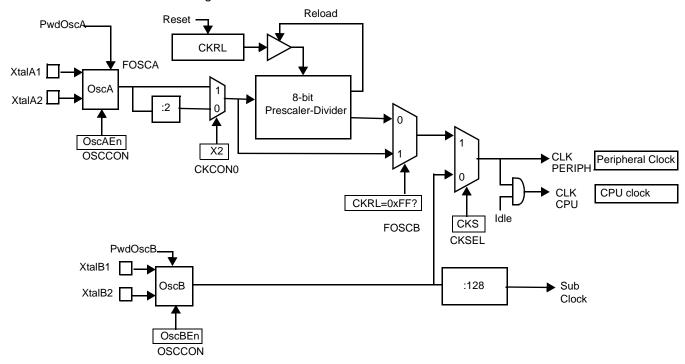
Reset Value = 00X1 0000b Not bit addressable





# Functional Block Diagram

Figure 2. Functional Oscillator Block Diagram



## **Operating Modes**

Reset

A hardware RESET puts the Clock generator in the following state:

The selected oscillator depends on OSC bit in Hardware Security Byte (HSB).

HSB.OSC = 1 (Oscillator A selected)

- OscAEn = 1 & OscBEn = 0: OscA is running, OscB is stopped.
- CKS = 1: OscA is selected for CPU.

HSB.OSC = 0 (Oscillator B selected)

- OscAEn = 0 & OscBEn = 1: OscB is running, OscA is stopped.
- CKS = 0: OscB is selected for CPU.

#### **Functional Modes**

Normal Modes

- CPU and Peripherals clock depend on the software selection using CKCON0, CKCON1 and CKRL registers
- CKS bit in CKSEL register selects either OscA or OscB
- CKRL register determines the frequency of the OscA clock.
- It is always possible to switch dynamically by software from OscA to OscB, and vice versa by changing CKS bit.

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Idle Modes

- IDLE modes are achieved by using any instruction that writes into PCON.0 bit (IDL)
- IDLE modes A and B depend on previous software sequence, prior to writing into PCON.0 bit:
- IDLE MODE A: OscA is running (OscAEn = 1) and selected (CKS = 1)
- IDLE MODE B: OscB is running (OscBEn = 1) and selected (CKS = 0)
- The unused oscillator OscA or OscB can be stopped by software by clearing OscAEn or OscBEn respectively.
- IDLE mode can be canceled either by Reset, or by activation of any enabled interruption
- In both cases, PCON.0 bit (IDL) is cleared by hardware
- Exit from IDLE modes will leave Oscillators control bits (OscEnA, OscEnB, CKS) unchanged.

Power Down Modes

- POWER DOWN modes are achieved by using any instruction that writes into PCON.1 bit (PD)
- POWER DOWN modes A and B depend on previous software sequence, prior to writing into PCON.1 bit:
- Both OscA and OscB will be stopped.
- POWER DOWN mode can be cancelled either by a hardware Reset, an external interruption, or the keyboard interrupt.
- By Reset signal: The CPU will restart according to OSC bit in Hardware Security Bit (HSB) register.
- By INT0 or INT1 interruption, if enabled: (standard behavioral), request on Pads must be driven low enough to ensure correct restart of the oscillator which was selected when entering in Power down.
- By keyboard Interrupt if enabled: a hardware clear of the PCON.1 flag ensure the restart of the oscillator which was selected when entering in Power down.

Table 20. Overview

PCON.1	PCON.0	OscBEn	OscAEn	CKS Selected Mode		Comment
0	0	0	1	1	NORMAL MODE A, OscB stopped	Default mode after power-up or Warm Reset
0	0	1	1	1	NORMAL MODE A, OscB running	Default mode after power-up or Warm Reset + OscB running
0	0	1	0	0	NORMAL MODE B, OscA stopped	OscB running and selected
0	0	1	1	0	NORMAL MODE B, OscA running	OscB running and selected + OscA running
Х	Х	0	0	Х	INVALID	OscA & OscB cannot be stopped at the same time
Х	Х	Х	0	1	INVALID	OscA must not be stopped, as used for CPU and peripherals
Х	Х	0	Х	0	INVALID	OscB must not be stopped as used for CPU and peripherals
0	1	Х	1	1	IDLE MODE A	The CPU is off, OscA supplies the peripherals, OscB can be disabled (OscBEn = 0)





Table 20. Overview (Continued)

PCON.1	PCON.0	OscBEn	OscAEn	CKS	Selected Mode	Comment
0	1	1	х	0	IDLE MODE B	The CPU is off, OscB supplies the peripherals, OscA can be disabled (OscAEn = 0)
1	Х	Х	1	Х	POWER DOWN MODE	The CPU and peripherals are off, OscA and OscB are stopped

#### **Design Considerations**

#### **Oscillators Control**

- PwdOscA and PwdOscB signals are generated in the Clock generator and used to control the hard blocks of oscillators A and B.
- PwdOscA ='1' stops OscA
- PwdOscB ='1' stops OscB
- The following tables summarize the Operating modes:

PCON.1	OscAEn	PwdOscA	Comments
0	1	0	OscA running
1	Х	1	OscA stopped by Power-down mode
0	0	1	OscA stopped by clearing OscAEn

PCON.1	OscBEn	PwdOscB	Comments
0	1	0	OscB running
1	Х	1	OscB stopped by Power-down mode
0	0	1	OscB stopped by clearing OscBEn

#### **Prescaler Divider**

- A hardware RESET puts the prescaler divider in the following state:
  - CKRL = FFh: F<sub>CLK CPU</sub> = F<sub>CLK PERIPH</sub> = F<sub>OSCA</sub>/2 (Standard C51 feature)
- CKS signal selects OSCA or OSCB:  $F_{CLK OUT} = F_{OSCA}$  or  $F_{OSCB}$
- Any value between FFh down to 00h can be written by software into CKRL register in order to divide frequency of the selected oscillator:
  - CKRL = 00h: minimum frequency

 $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSCA}/1020$  (Standard Mode)  $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSCA}/510$  (X2 Mode)

CKRL = FFh: maximum frequency

 $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSCA}/2$  (Standard Mode)  $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSCA}$  (X2 Mode)

 F<sub>CLK CPU</sub> and F<sub>CLK PERIPH</sub>, for CKRL≠0xFF In X2 Mode:

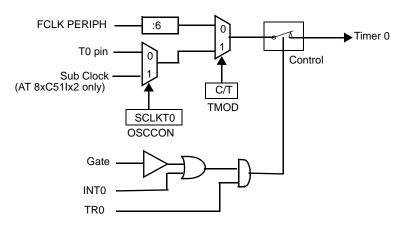
$$F_{CPU} = F_{CLKPERIPH} = \frac{F_{OSCA}}{2 \times (255 - CKRL)}$$

In X1 Mode:

$$F_{CPU} = F_{CLKPERIPH} = \frac{F_{OSCA}}{4 \times (255 - CKRL)}$$

#### **Timer 0: Clock Inputs**

Figure 1. Timer 0: Clock Inputs



Note: The SCLKT0 bit in OSCCON register allows to select Timer 0 Subsidiary clock.

SCLKT0 = 0: Timer 0 uses the standard T0 pin as clock input (Standard mode)

SCLKT0 = 1: Timer 0 uses the special Sub Clock as clock input, this feature can be use as periodic interrupt for time clock.



## **Enhanced Features**

In comparison to the original 80C52, the AT89C51ID2 implements some new features, which are:

- X2 option
- Dual Data Pointer
- Extended RAM
- Programmable Counter Array (PCA)
- Hardware Watchdog
- SPI interface
- 4-level interrupt priority system
- power-off flag
- ONCE mode
- ALE disabling
- Enhanced features on the UART and the timer 2

#### X2 Feature

The AT89C51ID2 core needs only 6 clock periods per machine cycle. This feature called 'X2' provides the following advantages:

- Divide frequency crystals by 2 (cheaper crystals) while keeping same CPU power.
- Save power consumption while keeping same CPU power (oscillator power saving).
- Save power consumption by dividing dynamically the operating frequency by 2 in operating and idle modes.
- Increase CPU power by 2 while keeping same crystal frequency.

In order to keep the original C51 compatibility, a divider by 2 is inserted between the XTAL1 signal and the main clock input of the core (phase generator). This divider may be disabled by software.

#### Description

The clock for the whole circuit and peripherals is first divided by two before being used by the CPU core and the peripherals.

This allows any cyclic ratio to be accepted on XTAL1 input. In X2 mode, as this divider is bypassed, the signals on XTAL1 must have a cyclic ratio between 40 to 60%.

Figure 3 shows the clock generation block diagram. X2 bit is validated on the rising edge of the XTAL1÷2 to avoid glitches when switching from X2 to STD mode. Figure 4 shows the switching mode waveforms.

Figure 3. Clock Generation Diagram

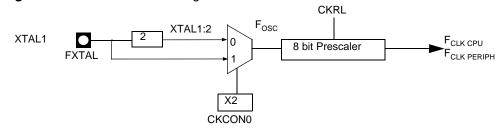
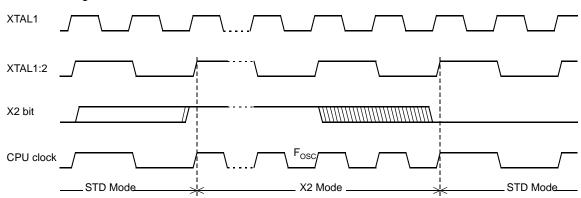






Figure 4. Mode Switching Waveforms



The X2 bit in the CKCON0 register (see Table 21) allows a switch from 12 clock periods per instruction to 6 clock periods and vice versa. At reset, the speed is set according to X2 bit of Hardware Security Byte (HSB). By default, Standard mode is active. Setting the X2 bit activates the X2 feature (X2 mode).

The T0X2, T1X2, T2X2, UartX2, PcaX2, and WdX2 bits in the CKCON0 register (See Table 21.) and SPIX2 bit in the CKCON1 register (see Table 22) allows a switch from standard peripheral speed (12 clock periods per peripheral clock cycle) to fast peripheral speed (6 clock periods per peripheral clock cycle). These bits are active only in X2 mode.

Table 21. CKCON0 Register

CKCON0 - Clock Control Register (8Fh)

7	6	5	4	3	2	1	0
TWIX2	WDX2	PCAX2	SIX2	T2X2	T1X2	T0X2	X2

Bit Number	Bit Mnemonic	Description
7	TWIX2	2-wire clock (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect) Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.
6	WDX2	Watchdog Clock (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.
5	PCAX2	Programmable Counter Array Clock (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.
4	SIX2	Enhanced UART Clock (Mode 0 and 2) (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.
3	T2X2	Timer2 Clock (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.
2	T1X2	Timer1 Clock (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.
1	T0X2	Timer0 Clock (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.
0	X2	CPU Clock Cleared to select 12 clock periods per machine cycle (STD mode) for CPU and all the peripherals. Set to select 6clock periods per machine cycle (X2 mode) and to enable the individual peripherals'X2' bits. Programmed by hardware after Power-up regarding Hardware Security Byte (HSB), Default setting, X2 is cleared.

Reset Value = 0000 000'HSB. X2'b (See "Hardware Security Byte") Not bit addressable





Table 22. CKCON1 Register

CKCON1 - Clock Control Register (AFh)

,	0	5	4	ა	2	I	U
-	-	-	-	-	-	-	SPIX2
Bit Number	Bit Mnemonic	Description					
7	-	Reserved					
6	-	Reserved					
5	-	Reserved					
4	-	Reserved					
3	-	Reserved					
2	-	Reserved					
1	-	Reserved					
0	SPIX2	this bit has n Clear to sele	o effect). ct 6 clock peri	ods per peripl	ne CPU clock cyceral clock cyceral clock cyc	cle.	en X2 is low,

Reset Value = XXXX XXX0b Not bit addressable

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# Dual Data Pointer Register DPTR

The additional data pointer can be used to speed up code execution and reduce code size.

The dual DPTR structure is a way by which the chip will specify the address of an external data memory location. There are two 16-bit DPTR registers that address the external memory, and a single bit called DPS = AUXR1.0 (see Table 23) that allows the program code to switch between them (Refer to Figure 5).

Figure 5. Use of Dual Pointer

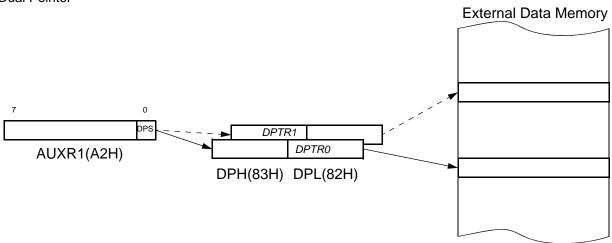






Table 23. AUXR1 register

AUXR1- Auxiliary Register 1(0A2h)

7	6	5	4	3	2	1	0
-	-	ENBOOT	-	GF3	0	•	DPS

Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
5	ENBOOT	Enable Boot Flash Cleared to disable boot ROM. Set to map the boot ROM between F800h - 0FFFFh.
4	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
3	GF3	This bit is a general purpose user flag. *
2	0	Always cleared.
1	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
0	DPS	Data Pointer Selection Cleared to select DPTR0. Set to select DPTR1.

Reset Value: XXXX XX0X0b

Not bit addressable

Note: \*Bit 2 stuck at 0; this allows to use INC AUXR1 to toggle DPS without changing GF3.

#### ASSEMBLY LANGUAGE

```
; Block move using dual data pointers
; Modifies DPTR0, DPTR1, A and PSW
; note: DPS exits opposite of entry state
; unless an extra INC AUXR1 is added
;
00A2 AUXR1 EQU 0A2H
;
0000 909000MOV DPTR,#SOURCE; address of SOURCE
0003 05A2 INC AUXR1; switch data pointers
0005 90A000 MOV DPTR,#DEST; address of DEST
0008 LOOP:
0008 05A2 INC AUXR1; switch data pointers
000A E0 MOVX A,@DPTR; get a byte from SOURCE
000B A3 INC DPTR; increment SOURCE address
000C 05A2 INC AUXR1; switch data pointers
000E F0 MOVX @DPTR,A; write the byte to DEST
000F A3 INC DPTR; increment DEST address
```

0010 70F6JNZ LOOP; check for 0 terminator 0012 05A2 INC AUXR1; (optional) restore DPS

INC is a short (2 bytes) and fast (12 clocks) way to manipulate the DPS bit in the AUXR1 SFR. However, note that the INC instruction does not directly force the DPS bit to a particular state, but simply toggles it. In simple routines, such as the block move example, only the fact that DPS is toggled in the proper sequence matters, not its actual value. In other words, the block move routine works the same whether DPS is '0' or '1' on entry. Observe that without the last instruction (INC AUXR1), the routine will exit with DPS in the opposite state.



# Expanded RAM (XRAM)

The AT89C51ID2 provides additional Bytes of random access memory (RAM) space for increased data parameter handling and high level language usage.

AT89C51ID2 devices have expanded RAM in external data space configurable up to 1792bytes (see Table 24.).

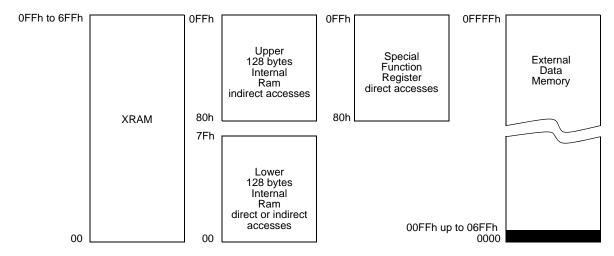
The AT89C51ID2 has internal data memory that is mapped into four separate segments.

The four segments are:

- 1. The Lower 128 bytes of RAM (addresses 00h to 7Fh) are directly and indirectly addressable.
- 2. The Upper 128 bytes of RAM (addresses 80h to FFh) are indirectly addressable only.
- 3. The Special Function Registers, SFRs, (addresses 80h to FFh) are directly addressable only.
- 4. The expanded RAM bytes are indirectly accessed by MOVX instructions, and with the EXTRAM bit cleared in the AUXR register (see Table 24).

The lower 128 bytes can be accessed by either direct or indirect addressing. The Upper 128 bytes can be accessed by indirect addressing only. The Upper 128 bytes occupy the same address space as the SFR. That means they have the same address, but are physically separate from SFR space.

Figure 6. Internal and External Data Memory Address



When an instruction accesses an internal location above address 7Fh, the CPU knows whether the access is to the upper 128 bytes of data RAM or to SFR space by the addressing mode used in the instruction.

- Instructions that use direct addressing access SFR space. For example: MOV 0A0H, # data, accesses the SFR at location 0A0h (which is P2).
- Instructions that use indirect addressing access the Upper 128 bytes of data RAM. For example: MOV @R0, # data where R0 contains 0A0h, accesses the data byte at address 0A0h, rather than P2 (whose address is 0A0h).
- The XRAM bytes can be accessed by indirect addressing, with EXTRAM bit cleared and MOVX instructions. This part of memory which is physically located on-chip, logically occupies the first bytes of external data memory. The bits XRSO and XRS1 are used to hide a part of the available XRAM as explained in Table 24. This can be





useful if external peripherals are mapped at addresses already used by the internal XRAM.

- With EXTRAM = 0, the XRAM is indirectly addressed, using the MOVX instruction in combination with any of the registers R0, R1 of the selected bank or DPTR. An access to XRAM will not affect ports P0, P2, P3.6 (WR) and P3.7 (RD). For example, with EXTRAM = 0, MOVX @R0, # data where R0 contains 0A0H, accesses the XRAM at address 0A0H rather than external memory. An access to external data memory locations higher than the accessible size of the XRAM will be performed with the MOVX DPTR instructions in the same way as in the standard 80C51, with P0 and P2 as data/address busses, and P3.6 and P3.7 as write and read timing signals. Accesses to XRAM above 0FFH can only be done by the use of DPTR.
- With EXTRAM = 1, MOVX @Ri and MOVX @DPTR will be similar to the standard 80C51.MOVX @ Ri will provide an eight-bit address multiplexed with data on Port0 and any output port pins can be used to output higher order address bits. This is to provide the external paging capability. MOVX @DPTR will generate a sixteen-bit address. Port2 outputs the high-order eight address bits (the contents of DPH) while Port0 multiplexes the low-order eight address bits (DPL) with data. MOVX @ Ri and MOVX @DPTR will generate either read or write signals on P3.6 (WR) and P3.7 (RD).

The stack pointer (SP) may be located anywhere in the 256 bytes RAM (lower and upper RAM) internal data memory. The stack may not be located in the XRAM.

The M0 bit allows to stretch the XRAM timings; if M0 is set, the read and write pulses are extended from 6 to 30 clock periods. This is useful to access external slow peripherals.

# Registers

Table 24. AUXR Register

AUXR - Auxiliary Register (8Eh)

7	6	5	4	3	2	1	0
-	-	МО	XRS2	XRS1	XRS0	EXTRAM	AO

Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
5	MO	Pulse length Cleared to stretch MOVX control: the RD/ and the WR/ pulse length is 6 clock periods (default). Set to stretch MOVX control: the RD/ and the WR/ pulse length is 30 clock periods.
4	XRS2	XRAM Size
3	XRS1	XRS2XRS1XRS0XRAM size 0 0 0256 bytes
2	XRS0	0 0 1512 bytes 0 1 0768 bytes(default) 0 1 11024 bytes 1 0 01792 bytes
1	EXTRAM	EXTRAM bit Cleared to access internal XRAM using movx @ Ri/ @ DPTR. Set to access external memory. Programmed by hardware after Power-up regarding Hardware Security Byte (HSB), default setting, XRAM selected.
0	AO	ALE Output bit Cleared, ALE is emitted at a constant rate of 1/6 the oscillator frequency (or 1/3 if X2 mode is used). (default) Set, ALE is active only during a MOVX or MOVC instruction is used.

Reset Value = XX00 10'HSB. XRAM'0b Not bit addressable

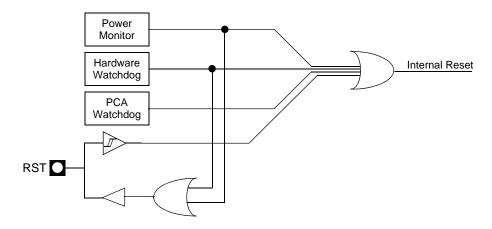


#### Reset

#### Introduction

The reset sources are : Power Management, Hardware Watchdog, PCA Watchdog and Reset input.

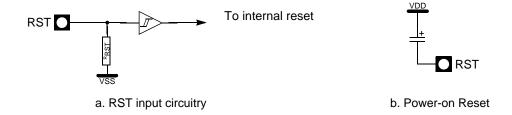
Figure 7. Reset schematic



### **Reset Input**

The Reset input can be used to force a reset pulse longer than the internal reset controlled by the Power Monitor. RST input has a pull-down resistor allowing power-on reset by simply connecting an external capacitor to  $V_{CC}$  as shown in Figure 8. Resistor value and input characteristics are discussed in the Section "DC Characteristics" of the AT89C51ID2 datasheet.

Figure 8. Reset Circuitry and Power-On Reset



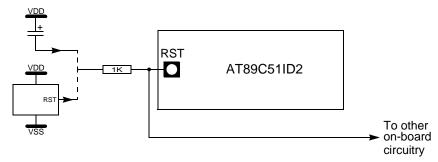




# **Reset Output**

As detailed in Section "Hardware Watchdog Timer", page 107, the WDT generates a 96-clock period pulse on the RST pin. In order to properly propagate this pulse to the rest of the application in case of external capacitor or power-supply supervisor circuit, a 1 k $\Omega$  resistor must be added as shown Figure 9.

Figure 9. Recommended Reset Output Schematic



#### **Power Monitor**

The POR/PFD function monitors the internal power-supply of the CPU core memories and the peripherals, and if needed, suspends their activity when the internal power supply falls below a safety threshold. This is achieved by applying an internal reset to them.

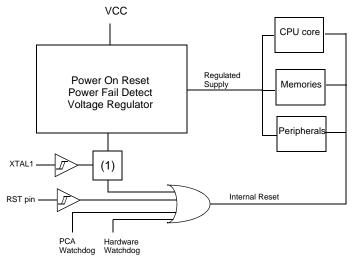
By generating the Reset the Power Monitor insures a correct start up when AT89C51ID2 is powered up.

#### **Description**

In order to startup and maintain the microcontroller in correct operating mode,  $V_{CC}$  has to be stabilized in the  $V_{CC}$  operating range and the oscillator has to be stabilized with a nominal amplitude compatible with logic level VIH/VIL.

These parameters are controlled during the three phases: power-up, normal operation and power going down. See Figure 10.

Figure 10. Power Monitor Block Diagram



Note: 1. Once XTAL1 High and low levels reach above and below VIH/VIL. a 1024 clock period delay will extend the reset coming from the Power Fail Detect. If the power falls below the Power Fail Detect threshold level, the Reset will be applied immediately.

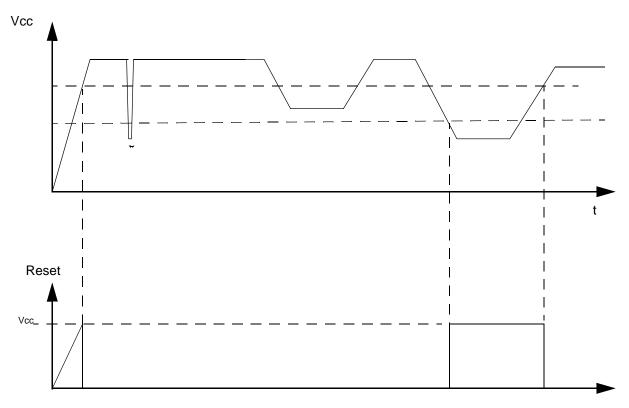
The Voltage regulator generates a regulated internal supply for the CPU core the memories and the peripherals. Spikes on the external Vcc are smoothed by the voltage regulator.





The Power fail detect monitor the supply generated by the voltage regulator and generate a reset if this supply falls below a safety threshold as illustrated in the Figure 11 below.

Figure 11. Power Fail Detect



When the power is applied, the Power Monitor immediately asserts a reset. Once the internal supply after the voltage regulator reach a safety level, the power monitor then looks at the XTAL clock input. The internal reset will remain asserted until the Xtal1 levels are above and below VIH and VIL. Further more. An internal counter will count 1024 clock periods before the reset is de-asserted.

If the internal power supply falls below a safety level, a reset is immediately asserted.

#### Timer 2

The Timer 2 in the AT89C51ID2 is the standard C52 Timer 2.

It is a 16-bit timer/counter: the count is maintained by two eight-bit timer registers, TH2 and TL2 are cascaded. It is controlled by T2CON (Table 25) and  $\underline{\text{T2}\text{MOD}}$  (Table 26) registers. Timer 2 operation is similar to Timer 0 and Timer 1.C/ $\overline{\text{T2}}$  selects  $F_{\text{OSC}}/12$  (timer operation) or external pin T2 (counter operation) as the timer clock input. Setting TR2 allows TL2 to increment by the selected input.

Timer 2 has 3 operating modes: capture, autoreload and Baud Rate Generator. These modes are selected by the combination of RCLK, TCLK and CP/RL2 (T2CON).

Refer to the Atmel 8-bit Microcontroller Hardware description for the description of Capture and Baud Rate Generator Modes.

Timer 2 includes the following enhancements:

- · Auto-reload mode with up or down counter
- Programmable clock-output

#### **Auto-Reload Mode**

The auto-reload mode configures Timer 2 as a 16-bit timer or event counter with auto-matic reload. If DCEN bit in T2MOD is cleared, Timer 2 behaves as in 80C52 (refer to the Atmel C51 Microcontroller Hardware description). If DCEN bit is set, Timer 2 acts as an Up/down timer/counter as shown in Figure 12. In this mode the T2EX pin controls the direction of count.

When T2EX is high, Timer 2 counts up. Timer overflow occurs at FFFFh which sets the TF2 flag and generates an interrupt request. The overflow also causes the 16-bit value in RCAP2H and RCAP2L registers to be loaded into the timer registers TH2 and TL2.

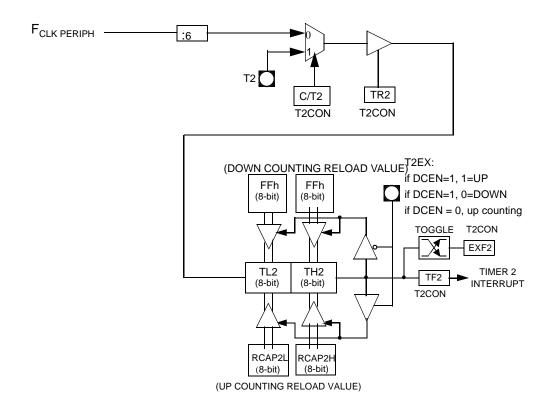
When T2EX is low, Timer 2 counts down. Timer underflow occurs when the count in the timer registers TH2 and TL2 equals the value stored in RCAP2H and RCAP2L registers. The underflow sets TF2 flag and reloads FFFFh into the timer registers.

The EXF2 bit toggles when Timer 2 overflows or underflows according to the direction of the count. EXF2 does not generate any interrupt. This bit can be used to provide 17-bit resolution.





Figure 12. Auto-Reload Mode Up/Down Counter (DCEN = 1)



# Programmable Clock-Output

In the clock-out mode, Timer 2 operates as a 50%-duty-cycle, programmable clock generator (See Figure 13). The input clock increments TL2 at frequency  $F_{\text{CLK PERIPH}}/2$ . The timer repeatedly counts to overflow from a loaded value. At overflow, the contents of RCAP2H and RCAP2L registers are loaded into TH2 and TL2. In this mode, Timer 2 overflows do not generate interrupts. The formula gives the clock-out frequency as a function of the system oscillator frequency and the value in the RCAP2H and RCAP2L registers:

$$Clock-OutFrequency = \frac{F_{CLKPERIPH}}{4 \times (65536 - RCAP2H/RCAP2L)}$$

For a 16 MHz system clock, Timer 2 has a programmable frequency range of 61 Hz  $(F_{CLK\ PERIPH}/2^{16})$  to 4 MHz  $(F_{CLK\ PERIPH}/4)$ . The generated clock signal is brought out to T2 pin (P1.0).

Timer 2 is programmed for the clock-out mode as follows:

- Set T2OE bit in T2MOD register.
- Clear C/T2 bit in T2CON register.
- Determine the 16-bit reload value from the formula and enter it in RCAP2H/RCAP2L registers.
- Enter a 16-bit initial value in timer registers TH2/TL2.It can be the same as the reload value or a different one depending on the application.
- To start the timer, set TR2 run control bit in T2CON register.

It is possible to use Timer 2 as a baud rate generator and a clock generator simultaneously. For this configuration, the baud rates and clock frequencies are not independent since both functions use the values in the RCAP2H and RCAP2L registers.

FCLK PERIPH —— TR2 T2CON TL2 (8-bit) TH2 (8-bit) OVER-FLOW RCAP2H (8-bit) RCAP2L (8-bit) Toggle T2 Q T2OE T2MOD TIMER 2
INTERRUPT EXF2 T2CON EXEN2 T2CON

Figure 13. Clock-Out Mode  $C/\overline{T2} = 0$ 





# Registers

Table 25. T2CON Register

T2CON - Timer 2 Control Register (C8h)

7 6 5 4 3 2 1 0

TF2 EXF2 RCLK TCLK EXEN2 TR2 C/T2# CP/RL2#

	1	
Bit	Bit	
Number	Mnemonic	Description
7	TF2	Timer 2 overflow Flag  Must be cleared by software.  Set by hardware on Timer 2 overflow, if RCLK = 0 and TCLK = 0.
6	EXF2	Timer 2 External Flag Set when a capture or a reload is caused by a negative transition on T2EX pin if EXEN2=1. When set, causes the CPU to vector to Timer 2 interrupt routine when Timer 2 interrupt is enabled. Must be cleared by software. EXF2 doesn't cause an interrupt in Up/down counter mode (DCEN = 1).
5	RCLK	Receive Clock bit Cleared to use timer 1 overflow as receive clock for serial port in mode 1 or 3. Set to use Timer 2 overflow as receive clock for serial port in mode 1 or 3.
4	TCLK	Transmit Clock bit Cleared to use timer 1 overflow as transmit clock for serial port in mode 1 or 3. Set to use Timer 2 overflow as transmit clock for serial port in mode 1 or 3.
3	EXEN2	Timer 2 External Enable bit Cleared to ignore events on T2EX pin for Timer 2 operation. Set to cause a capture or reload when a negative transition on T2EX pin is detected, if Timer 2 is not used to clock the serial port.
2	TR2	Timer 2 Run control bit Cleared to turn off Timer 2. Set to turn on Timer 2.
1	C/T2#	Timer/Counter 2 select bit Cleared for timer operation (input from internal clock system: F <sub>CLK PERIPH</sub> ). Set for counter operation (input from T2 input pin, falling edge trigger). Must be 0 for clock out mode.
0	CP/RL2#	Timer 2 Capture/Reload bit  If RCLK=1 or TCLK=1, CP/RL2# is ignored and timer is forced to auto-reload on Timer 2 overflow.  Cleared to auto-reload on Timer 2 overflows or negative transitions on T2EX pin if EXEN2=1.  Set to capture on negative transitions on T2EX pin if EXEN2=1.

Reset Value = 0000 0000b Bit addressable

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Table 26. T2MOD Register

T2MOD - Timer 2 Mode Control Register (C9h)

7	6	5	4	3	2	1	0
-	-	-	-	-	-	T2OE	DCEN

Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
4	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
3	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
2	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
1	T2OE	Timer 2 Output Enable bit Cleared to program P1.0/T2 as clock input or I/O port. Set to program P1.0/T2 as clock output.
0	DCEN	Down Counter Enable bit Cleared to disable Timer 2 as up/down counter. Set to enable Timer 2 as up/down counter.

Reset Value = XXXX XX00b Not bit addressable



# Programmable Counter Array PCA

The PCA provides more timing capabilities with less CPU intervention than the standard timer/counters. Its advantages include reduced software overhead and improved accuracy. The PCA consists of a dedicated timer/counter which serves as the time base for an array of five compare/capture modules. Its clock input can be programmed to count any one of the following signals:

- Peripheral clock frequency (F<sub>CLK PERIPH</sub>) ÷ 6
- Peripheral clock frequency (F<sub>CLK PERIPH</sub>) ÷ 2
- Timer 0 overflow
- External input on ECI (P1.2)

Each compare/capture modules can be programmed in any one of the following modes:

- Rising and/or falling edge capture
- Software timer
- · High-speed output
- Pulse width modulator

Module 4 can also be programmed as a watchdog timer (See Section "PCA Watchdog Timer", page 51).

When the compare/capture modules are programmed in the capture mode, software timer, or high speed output mode, an interrupt can be generated when the module executes its function. All five modules plus the PCA timer overflow share one interrupt vector.

The PCA timer/counter and compare/capture modules share Port 1 for external I/O. These pins are listed below. If the port is not used for the PCA, it can still be used for standard I/O.

PCA component	External I/O Pin			
16-bit Counter	P1.2 / ECI			
16-bit Module 0	P1.3 / CEX0			
16-bit Module 1	P1.4 / CEX1			
16-bit Module 2	P1.5 / CEX2			
16-bit Module 3	P1.6 / CEX3			

The PCA timer is a common time base for all five modules (See Figure 14). The timer count source is determined from the CPS1 and CPS0 bits in the CMOD register (Table 27) and can be programmed to run at:

- 1/6 the peripheral clock frequency (F<sub>CLK PERIPH</sub>)
- 1/2 the peripheral clock frequency (F<sub>CLK PERIPH</sub>)
- The Timer 0 overflow
- The input on the ECI pin (P1.2)





Figure 14. PCA Timer/Counter

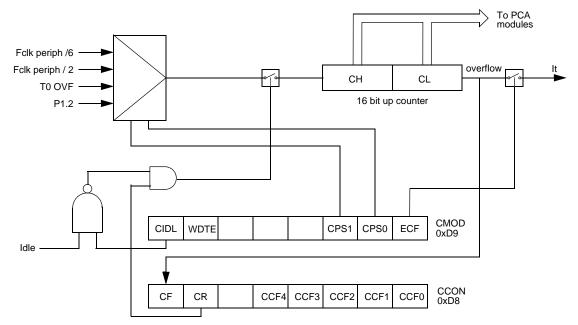


Table 27. CMOD Register

CMOD - PCA Counter Mode Register (D9h)

	7	6	5	4	3	2	1	0
	CIDL	WDTE	•	-	-	CPS1	CPS0	ECF
Ī								

Bit Number	Bit Mnemonic	Description
7	CIDL	Counter Idle Control Cleared to program the PCA Counter to continue functioning during idle Mode. Set to program PCA to be gated off during idle.
6	WDTE	Watchdog Timer Enable Cleared to disable Watchdog Timer function on PCA Module 4. Set to enable Watchdog Timer function on PCA Module 4.
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
4	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
3	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
2	CPS1	PCA Count Pulse Select
1	CPS0	CPS1 CPS0Selected PCA input 0 0 Internal clock fCLK PERIPH/6 0 1Internal clock fCLK PERIPH/2 1 0Timer 0 Overflow 1 1 External clock at ECI/P1.2 pin (max rate = fCLK PERIPH/ 4)
0	ECF	PCA Enable Counter Overflow Interrupt Cleared to disable CF bit in CCON to inhibit an interrupt. Set to enable CF bit in CCON to generate an interrupt.

Reset Value = 00XX X000b Not bit addressable

The CMOD register includes three additional bits associated with the PCA (See Figure 14 and Table 27).

- The CIDL bit which allows the PCA to stop during idle mode.
- The WDTE bit which enables or disables the watchdog function on module 4.
- The ECF bit which when set causes an interrupt and the PCA overflow flag CF (in the CCON SFR) to be set when the PCA timer overflows.

The CCON register contains the run control bit for the PCA and the flags for the PCA timer (CF) and each module (Refer to Table 28).

- Bit CR (CCON.6) must be set by software to run the PCA. The PCA is shut off by clearing this bit.
- Bit CF: The CF bit (CCON.7) is set when the PCA counter overflows and an interrupt will be generated if the ECF bit in the CMOD register is set. The CF bit can only be cleared by software.





• Bits 0 through 4 are the flags for the modules (bit 0 for module 0, bit 1 for module 1, etc.) and are set by hardware when either a match or a capture occurs. These flags also can only be cleared by software.

3

2

0

Table 28. CCON Register

7

CCON - PCA Counter Control Register (D8h)

CF	CR	-	CCF4	CCF3	CCF2	CCF1	CCF0		
Bit Number	Bit Mnemonic	Description							
7	CF	Set by hardw CMOD is set	. CF	ag counter rolls ware or softwa					
6	CR	Must be clea	CA Counter Run control bit ust be cleared by software to turn the PCA counter off. et by software to turn the PCA counter on.						
5	-	Reserved The value re	Reserved The value read from this bit is indeterminate. Do not set this bit.						
4	CCF4	Must be clea	PCA Module 4 interrupt flag  Must be cleared by software.  Set by hardware when a match or capture occurs.						
3	CCF3	Must be clea	PCA Module 3 interrupt flag  Must be cleared by software.  Set by hardware when a match or capture occurs.						
2	CCF2	Must be clea	PCA Module 2 interrupt flag  Must be cleared by software.  Set by hardware when a match or capture occurs.						
1	CCF1	Must be clea	PCA Module 1 interrupt flag  Must be cleared by software.  Set by hardware when a match or capture occurs.						
0	CCF0		e 0 interrupt the	•					

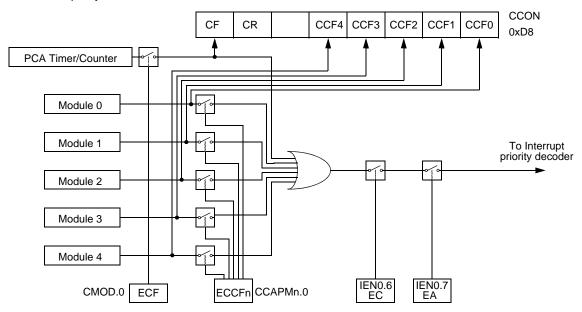
Reset Value = 00X0 0000b Not bit addressable

The watchdog timer function is implemented in module 4 (See Figure 17).

Set by hardware when a match or capture occurs.

The PCA interrupt system is shown in Figure 15.

Figure 15. PCA Interrupt System



**PCA Modules:** each one of the five compare/capture modules has six possible functions. It can perform:

- 16-bit Capture, positive-edge triggered
- 16-bit Capture, negative-edge triggered
- 16-bit Capture, both positive and negative-edge triggered
- 16-bit Software Timer
- 16-bit High Speed Output
- 8-bit Pulse Width Modulator

In addition, module 4 can be used as a Watchdog Timer.

Each module in the PCA has a special function register associated with it. These registers are: CCAPM0 for module 0, CCAPM1 for module 1, etc. (See Table 29). The registers contain the bits that control the mode that each module will operate in.

- The ECCF bit (CCAPMn.0 where n=0, 1, 2, 3, or 4 depending on the module) enables the CCF flag in the CCON SFR to generate an interrupt when a match or compare occurs in the associated module.
- PWM (CCAPMn.1) enables the pulse width modulation mode.
- The TOG bit (CCAPMn.2) when set causes the CEX output associated with the module to toggle when there is a match between the PCA counter and the module's capture/compare register.
- The match bit MAT (CCAPMn.3) when set will cause the CCFn bit in the CCON register to be set when there is a match between the PCA counter and the module's capture/compare register.
- The next two bits CAPN (CCAPMn.4) and CAPP (CCAPMn.5) determine the edge
  that a capture input will be active on. The CAPN bit enables the negative edge, and
  the CAPP bit enables the positive edge. If both bits are set both edges will be
  enabled and a capture will occur for either transition.
- The last bit in the register ECOM (CCAPMn.6) when set enables the comparator function.





Table 29 shows the CCAPMn settings for the various PCA functions.

**Table 29.** CCAPMn Registers (n = 0-4)

CCAPM0 - PCA Module 0 Compare/Capture Control Register (0DAh)

CCAPM1 - PCA Module 1 Compare/Capture Control Register (0DBh)

CCAPM2 - PCA Module 2 Compare/Capture Control Register (0DCh)

CCAPM3 - PCA Module 3 Compare/Capture Control Register (0DDh)

CCAPM4 - PCA Module 4 Compare/Capture Control Register (0DEh)

7	6	5	4	3	2	1	0			
-	ECOMn	CAPPn	CAPNn	MATn	TOGn	PWMn	ECCFn			
Bit Number	Bit Mnemonic	Description								
7	-	Reserved The value re	ad from this b	it is indetermi	nate. Do not s	et this bit.				
6	ECOMn	Enable Comparator Cleared to disable the comparator function. Set to enable the comparator function.								
5	CAPPn	Cleared to di	apture Positive leared to disable positive edge capture. et to enable positive edge capture.							
4	CAPNn	Capture Negative Cleared to disable negative edge capture. Set to enable negative edge capture.								
3	MATn	Match When MATn compare/cap CCFn bit in 0	unter with this	s module's						
2	TOGn	Toggle When TOGn = 1, a match of the PCA counter with this module's compare/capture register causes the CEXn pin to toggle.								
1	PWMn	Pulse Width Modulation Mode Cleared to disable the CEXn pin to be used as a pulse width modulated Set to enable the CEXn pin to be used as a pulse width modulated outp								
0	CCF0	Enable CCF interrupt Cleared to disable compare/capture flag CCFn in the CCON register to generate an interrupt. Set to enable compare/capture flag CCFn in the CCON register to generate an interrupt.								

Reset Value = X000 0000b Not bit addressable

Table 30. PCA Module Modes (CCAPMn Registers)

ECOMn	CAPPn	CAPNn	MATn	TOGn	PWMm	ECCFn	Module Function	
0	0	0	0	0	0	0	No Operation	
Х	1	0	0	0	0	Х	16-bit capture by a positive-edge trigger on CEXn	
Х	0	1	0	0	0	Х	16-bit capture by a negative trigger on CEXn	
Х	1	1	0	0	0	Х	16-bit capture by a transition on CEXn	
1	0	0	1	0	0	Х	16-bit Software Timer / Compare mode.	
1	0	0	1	1	0	Х	16-bit High Speed Output	
1	0	0	0	0	1	0	8-bit PWM	
1	0	0	1	Х	0	Х	Watchdog Timer (module 4 only)	

There are two additional registers associated with each of the PCA modules. They are CCAPnH and CCAPnL and these are the registers that store the 16-bit count when a capture occurs or a compare should occur. When a module is used in the PWM mode these registers are used to control the duty cycle of the output (See Table 31 & Table 32).

**Table 31.** CCAPnH Registers (n = 0-4)

CCAP0H - PCA Module 0 Compare/Capture Control Register High (0FAh)

CCAP1H - PCA Module 1 Compare/Capture Control Register High (0FBh)

CCAP2H - PCA Module 2 Compare/Capture Control Register High (0FCh)

CCAP3H - PCA Module 3 Compare/Capture Control Register High (0FDh)

CCAP4H - PCA Module 4 Compare/Capture Control Register High (0FEh)

7	6	5	4	3	2	1	0
-	-	-	-	-	-	-	-

Bit Number	Bit Mnemonic	Description
7-0	-	PCA Module n Compare/Capture Control CCAPnH Value

Reset Value = 0000 0000b Not bit addressable

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**Table 32.** CCAPnL Registers (n = 0-4)

CCAP0L - PCA Module 0 Compare/Capture Control Register Low (0EAh)

CCAP1L - PCA Module 1 Compare/Capture Control Register Low (0EBh)

CCAP2L - PCA Module 2 Compare/Capture Control Register Low (0ECh)

CCAP3L - PCA Module 3 Compare/Capture Control Register Low (0EDh)

CCAP4L - PCA Module 4 Compare/Capture Control Register Low (0EEh)

7	6	5	4	3	2	1	0
-	•	-	-	-	-	•	•

Bit Number	Bit Mnemonic	Description
7-0	_	PCA Module n Compare/Capture Control CCAPnL Value

Reset Value = 0000 0000b

Not bit addressable

Table 33. CH Register

CH - PCA Counter Register High (0F9h)

7	6	5	4	3	2	1	0
-	-	-	-	-	-	-	-

Bit Number	Bit Mnemonic	Description
7-0	-	PCA counter CH Value

Reset Value = 0000 0000b

Not bit addressable

#### Table 34. CL Register

CL - PCA Counter Register Low (0E9h)

7	6	5	4	3	2	1	0	
-	-	-	-	-	-	-	-	

Bit Number	Bit Mnemonic	Description
7-0	-	PCA Counter CL Value

Reset Value = 0000 0000b

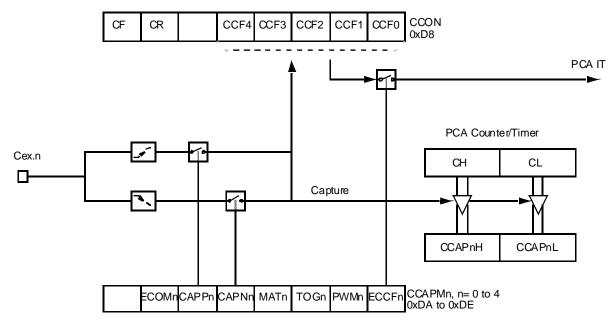
Not bit addressable

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### **PCA Capture Mode**

To use one of the PCA modules in the capture mode either one or both of the CCAPM bits CAPN and CAPP for that module must be set. The external CEX input for the module (on port 1) is sampled for a transition. When a valid transition occurs the PCA hardware loads the value of the PCA counter registers (CH and CL) into the module's capture registers (CCAPnL and CCAPnH). If the CCFn bit for the module in the CCON SFR and the ECCFn bit in the CCAPMn SFR are set then an interrupt will be generated (Refer to Figure 16).

Figure 16. PCA Capture Mode



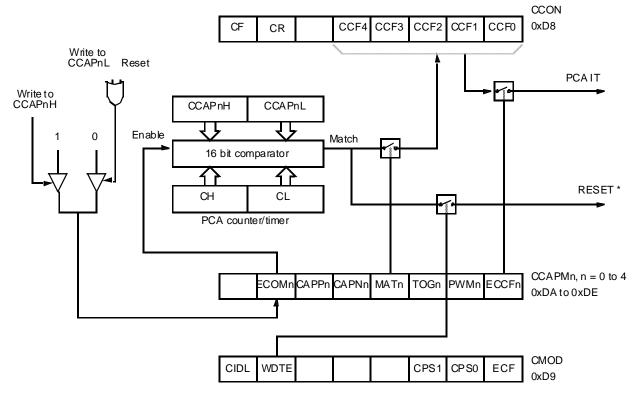
# 16-bit Software Timer/ Compare Mode

The PCA modules can be used as software timers by setting both the ECOM and MAT bits in the modules CCAPMn register. The PCA timer will be compared to the module's capture registers and when a match occurs an interrupt will occur if the CCFn (CCON SFR) and the ECCFn (CCAPMn SFR) bits for the module are both set (See Figure 17).





Figure 17. PCA Compare Mode and PCA Watchdog Timer



Before enabling ECOM bit, CCAPnL and CCAPnH should be set with a non zero value, otherwise an unwanted match could happen. Writing to CCAPnH will set the ECOM bit.

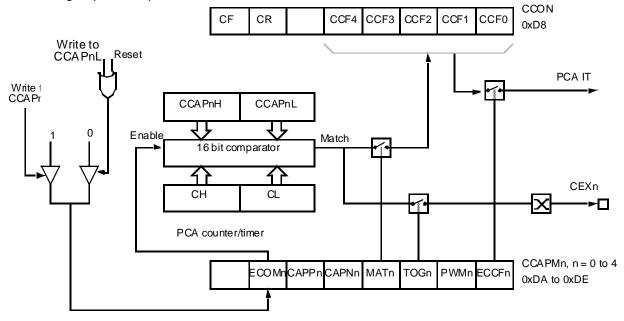
Once ECOM set, writing CCAPnL will clear ECOM so that an unwanted match doesn't occur while modifying the compare value. Writing to CCAPnH will set ECOM. For this reason, user software should write CCAPnL first, and then CCAPnH. Of course, the ECOM bit can still be controlled by accessing to CCAPMn register.

#### **High Speed Output Mode**

In this mode the CEX output (on port 1) associated with the PCA module will toggle each time a match occurs between the PCA counter and the module's capture registers. To activate this mode the TOG, MAT, and ECOM bits in the module's CCAPMn SFR must be set (See Figure 18).

A prior write must be done to CCAPnL and CCAPnH before writing the ECOMn bit.

Figure 18. PCA High Speed Output Mode



Before enabling ECOM bit, CCAPnL and CCAPnH should be set with a non zero value, otherwise an unwanted match could happen.

Once ECOM set, writing CCAPnL will clear ECOM so that an unwanted match doesn't occur while modifying the compare value. Writing to CCAPnH will set ECOM. For this reason, user software should write CCAPnL first, and then CCAPnH. Of course, the ECOM bit can still be controlled by accessing to CCAPMn register.

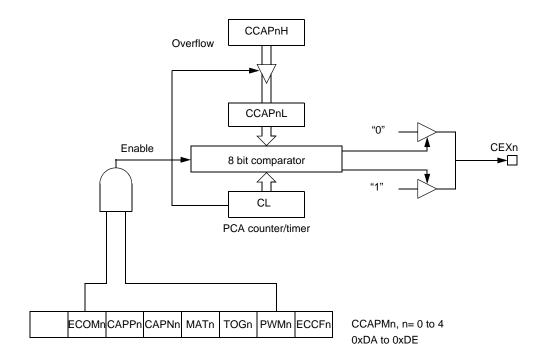
# Pulse Width Modulator Mode

All of the PCA modules can be used as PWM outputs. Figure 19 shows the PWM function. The frequency of the output depends on the source for the PCA timer. All of the modules will have the same frequency of output because they all share the PCA timer. The duty cycle of each module is independently variable using the module's capture register CCAPLn. When the value of the PCA CL SFR is less than the value in the module's CCAPLn SFR the output will be low, when it is equal to or greater than the output will be high. When CL overflows from FF to 00, CCAPLn is reloaded with the value in CCAPHn. This allows updating the PWM without glitches. The PWM and ECOM bits in the module's CCAPMn register must be set to enable the PWM mode.





Figure 19. PCA PWM Mode



# **PCA Watchdog Timer**

An on-board watchdog timer is available with the PCA to improve the reliability of the system without increasing chip count. Watchdog timers are useful for systems that are susceptible to noise, power glitches, or electrostatic discharge. Module 4 is the only PCA module that can be programmed as a watchdog. However, this module can still be used for other modes if the watchdog is not needed. Figure 17 shows a diagram of how the watchdog works. The user pre-loads a 16-bit value in the compare registers. Just like the other compare modes, this 16-bit value is compared to the PCA timer value. If a match is allowed to occur, an internal reset will be generated. This will not cause the RST pin to be driven high.

In order to hold off the reset, the user has three options:

- 1. periodically change the compare value so it will never match the PCA timer,
- 2. periodically change the PCA timer value so it will never match the compare values, or
- 3. disable the watchdog by clearing the WDTE bit before a match occurs and then reenable it.

The first two options are more reliable because the watchdog timer is never disabled as in option #3. If the program counter ever goes astray, a match will eventually occur and cause an internal reset. The second option is also not recommended if other PCA modules are being used. Remember, the PCA timer is the time base for all modules; changing the time base for other modules would not be a good idea. Thus, in most applications the first solution is the best option.

This watchdog timer won't generate a reset out on the reset pin.

#### Serial I/O Port

The serial I/O port in the AT89C51ID2 is compatible with the serial I/O port in the 80C52. It provides both synchronous and asynchronous communication modes. It operates as a Universal Asynchronous Receiver and Transmitter (UART) in three full-duplex modes (Modes 1, 2 and 3). Asynchronous transmission and reception can occur simultaneously and at different baud rates

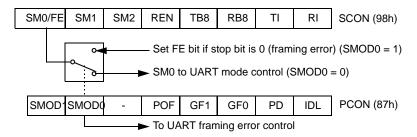
Serial I/O port includes the following enhancements:

- Framing error detection
- · Automatic address recognition

#### **Framing Error Detection**

Framing bit error detection is provided for the three asynchronous modes (modes 1, 2 and 3). To enable the framing bit error detection feature, set SMOD0 bit in PCON register (See Figure 20).

Figure 20. Framing Error Block Diagram



When this feature is enabled, the receiver checks each incoming data frame for a valid stop bit. An invalid stop bit may result from noise on the serial lines or from simultaneous transmission by two CPUs. If a valid stop bit is not found, the Framing Error bit (FE) in SCON register (See Table 38.) bit is set.

Software may examine FE bit after each reception to check for data errors. Once set, only software or a reset can clear FE bit. Subsequently received frames with valid stop bits cannot clear FE bit. When FE feature is enabled, RI rises on stop bit instead of the last data bit (See Figure 21. and Figure 22.).

Figure 21. UART Timings in Mode 1

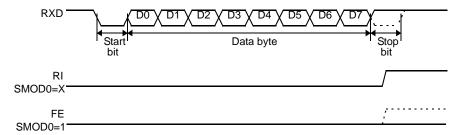
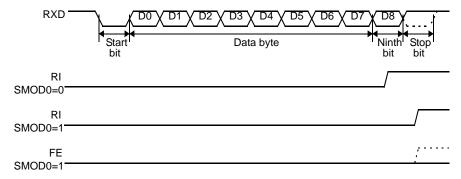






Figure 22. UART Timings in Modes 2 and 3



# Automatic Address Recognition

The automatic address recognition feature is enabled when the multiprocessor communication feature is enabled (SM2 bit in SCON register is set).

Implemented in hardware, automatic address recognition enhances the multiprocessor communication feature by allowing the serial port to examine the address of each incoming command frame. Only when the serial port recognizes its own address, the receiver sets RI bit in SCON register to generate an interrupt. This ensures that the CPU is not interrupted by command frames addressed to other devices.

If desired, the user may enable the automatic address recognition feature in mode 1.In this configuration, the stop bit takes the place of the ninth data bit. Bit RI is set only when the received command frame address matches the device's address and is terminated by a valid stop bit.

To support automatic address recognition, a device is identified by a given address and a broadcast address.

Note:

The multiprocessor communication and automatic address recognition features cannot be enabled in mode 0 (i. e. setting SM2 bit in SCON register in mode 0 has no effect).

### **Given Address**

Each device has an individual address that is specified in SADDR register; the SADEN register is a mask byte that contains don't-care bits (defined by zeros) to form the device's given address. The don't-care bits provide the flexibility to address one or more slaves at a time. The following example illustrates how a given address is formed.

To address a device by its individual address, the SADEN mask byte must be 1111 1111b.

For example:

SADDR0101 0110b SADEN1111 1100b

Given0101 01XXb

The following is an example of how to use given addresses to address different slaves:

Slave A:SADDR1111 0001b SADEN1111 1010b

Given1111 0X0Xb

Slave B:SADDR1111 0011b <u>SADEN1111 1001b</u>

Given1111 0XX1b

Slave C:SADDR1111 0010b <u>SADEN1111 1101b</u> Given1111 00X1b The SADEN byte is selected so that each slave may be addressed separately.

For slave A, bit 0 (the LSB) is a don't-care bit; for slaves B and C, bit 0 is a 1.To communicate with slave A only, the master must send an address where bit 0 is clear (e. g. 1111 0000b).

For slave A, bit 1 is a 1; for slaves B and C, bit 1 is a don't care bit. To communicate with slaves B and C, but not slave A, the master must send an address with bits 0 and 1 both set (e. g. 1111 0011b).

To communicate with slaves A, B and C, the master must send an address with bit 0 set, bit 1 clear, and bit 2 clear (e. g. 1111 0001b).

#### **Broadcast Address**

A broadcast address is formed from the logical OR of the SADDR and SADEN registers with zeros defined as don't-care bits, e. g. :

SADDR0101 0110b SADEN1111 1100b

Broadcast = SADDR OR SADEN1111 111Xb

The use of don't-care bits provides flexibility in defining the broadcast address, however in most applications, a broadcast address is FFh. The following is an example of using broadcast addresses:

Slave A:SADDR1111 0001b <u>SADEN1111 1010b</u> Broadcast1111 1X11b,

Slave B:SADDR1111 0011b <u>SADEN1111 1001b</u> Broadcast1111 1X11B,

Slave C:SADDR=1111 0011b <u>SADEN1111 1101b</u> Broadcast1111 1111b

For slaves A and B, bit 2 is a don't care bit; for slave C, bit 2 is set. To communicate with all of the slaves, the master must send an address FFh. To communicate with slaves A and B, but not slave C, the master can send and address FBh.

#### **Reset Addresses**

On reset, the SADDR and SADEN registers are initialized to 00h, i. e. the given and broadcast addresses are XXXX XXXXb (all don't-care bits). This ensures that the serial port will reply to any address, and so, that it is backwards compatible with the 80C51 microcontrollers that do not support automatic address recognition.





# Registers

Table 35. SADEN Register

SADEN - Slave Address Mask Register (B9h)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b

Not bit addressable

#### Table 36. SADDR Register

SADDR - Slave Address Register (A9h)

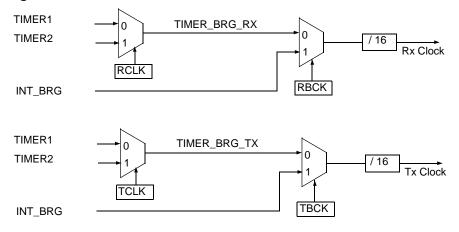
7	6	5	4	3	2	1	0

Reset Value = 0000 0000b Not bit addressable

# Baud Rate Selection for UART for Mode 1 and 3

The Baud Rate Generator for transmit and receive clocks can be selected separately via the T2CON and BDRCON registers.

Figure 23. Baud Rate Selection



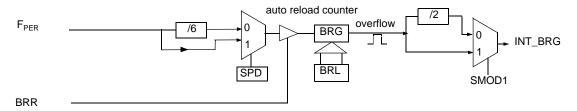
**TCLK RCLK TBCK RBCK Clock Source Clock Source** (T2CON) (T2CON) (BDRCON) (BDRCON) **UART Tx UART Rx** 0 0 0 0 Timer 1 Timer 1 1 0 0 0 Timer 2 Timer 1 0 1 0 0 Timer 1 Timer 2 1 1 0 0 Timer 2 Timer 2 0 Χ 1 0 INT\_BRG Timer 1 INT\_BRG Χ 1 1 0 Timer 2 0 Χ 0 1 Timer 1 INT\_BRG 1 Χ 0 1 Timer 2 INT\_BRG Χ Χ 1 1 INT\_BRG INT\_BRG

Table 37. Baud Rate Selection Table UART

# Internal Baud Rate Generator (BRG)

When the internal Baud Rate Generator is used, the Baud Rates are determined by the BRG overflow depending on the BRL reload value, the value of SPD bit (Speed Mode) in BDRCON register and the value of the SMOD1 bit in PCON register.

Figure 24. Internal Baud Rate



• The baud rate for UART is token by formula:

$$Baud\_Rate = \frac{2^{SMOD1} \cdot F_{PER}}{6^{(1-SPD)} \cdot 32 \cdot (256 \cdot BRL)}$$

$$\mathsf{BRL} = 256 - \frac{2^{\mathsf{SMOD1}} \cdot \mathsf{F}_{\mathsf{PER}}}{6^{(1-\mathsf{SPD})} \cdot 32 \cdot \mathsf{Baud\_Rate}}$$





# Table 38. SCON Register

SCON - Serial Control Register (98h)

7	6	5	4	3	2	1	0
FE/SM0	SM1	SM2	REN	TB8	RB8	TI	RI

Bit Number	Bit Mnemonic	Description
7	FE	Framing Error bit (SMOD0=1)  Clear to reset the error state, not cleared by a valid stop bit.  Set by hardware when an invalid stop bit is detected.  SMOD0 must be set to enable access to the FE bit.
	SM0	Serial port Mode bit 0 Refer to SM1 for serial port mode selection. SMOD0 must be cleared to enable access to the SM0 bit.
6	SM1	Serial port Mode bit 1           SM0         SM1         Mode         Baud Rate           0         0         Shift Register         F <sub>XTAL</sub> /12 (or F <sub>XTAL</sub> /6 in mode X2)           0         1         8-bit UART         Variable           1         0         9-bit UART         F <sub>XTAL</sub> /64 or F <sub>XTAL</sub> /32           1         1         9-bit UART         Variable
5	SM2	Serial port Mode 2 bit / Multiprocessor Communication Enable bit Clear to disable multiprocessor communication feature. Set to enable multiprocessor communication feature in mode 2 and 3, and eventually mode 1. This bit should be cleared in mode 0.
4	REN	Reception Enable bit Clear to disable serial reception. Set to enable serial reception.
3	TB8	Transmitter Bit 8 / Ninth bit to transmit in modes 2 and 3 Clear to transmit a logic 0 in the 9th bit. Set to transmit a logic 1 in the 9th bit.
2	RB8	Receiver Bit 8 / Ninth bit received in modes 2 and 3 Cleared by hardware if 9th bit received is a logic 0. Set by hardware if 9th bit received is a logic 1. In mode 1, if SM2 = 0, RB8 is the received stop bit. In mode 0 RB8 is not used.
1	ТІ	Transmit Interrupt flag Clear to acknowledge interrupt. Set by hardware at the end of the 8th bit time in mode 0 or at the beginning of the stop bit in the other modes.
0	RI	Receive Interrupt flag Clear to acknowledge interrupt. Set by hardware at the end of the 8th bit time in mode 0, see Figure 21. and Figure 22. in the other modes.

Reset Value = 0000 0000b Bit addressable

Table 39. Example of Computed Value When X2=1, SMOD1=1, SPD=1

Baud Rates	F <sub>osc</sub> = 16	. 384 MHz	F <sub>OSC</sub> = 24MHz		
	BRL	Error (%)	BRL	Error (%)	
115200	247	1.23	243	0.16	
57600	238	1.23	230	0.16	
38400	229	1.23	217	0.16	
28800	220	1.23	204	0.16	
19200	203	0.63	178	0.16	
9600	149	0.31	100	0.16	
4800	43	1.23	-	-	

Table 40. Example of Computed Value When X2=0, SMOD1=0, SPD=0

Baud Rates	F <sub>osc</sub> = 16	. 384 MHz	F <sub>OSC</sub> = 24MHz		
	BRL	Error (%)	BRL	Error (%)	
4800	247	1.23	243	0.16	
2400	238	1.23	230	0.16	
1200	220	1.23	202	3.55	
600	185	0.16	152	0.16	

The baud rate generator can be used for mode 1 or 3 (refer to Figure 23.), but also for mode 0 for UART, thanks to the bit SRC located in BDRCON register (Table 47.)

### **UART Registers**

Table 41. SADEN Register

SADEN - Slave Address Mask Register for UART (B9h)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b

#### Table 42. SADDR Register

SADDR - Slave Address Register for UART (A9h)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b





Table 43. SBUF Register

SBUF - Serial Buffer Register for UART (99h)

7	6	5	4	3	2	1	0

Reset Value = XXXX XXXXb

# Table 44. BRL Register

BRL - Baud Rate Reload Register for the internal baud rate generator, UART (9Ah)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b

Table 45. T2CON Register

T2CON - Timer 2 Control Register (C8h)

7	6	5	4	3	2	1	0
TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2#	CP/RL2#

D'	D::					
Bit Number	Bit Mnemonic	Description				
7	TF2	Timer 2 overflow Flag  Must be cleared by software.  Set by hardware on timer 2 overflow, if RCLK = 0 and TCLK = 0.				
6	EXF2	Timer 2 External Flag Set when a capture or a reload is caused by a negative transition on T2EX pin if EXEN2=1. When set, causes the CPU to vector to timer 2 interrupt routine when timer 2 interrupt is enabled. Must be cleared by software. EXF2 doesn't cause an interrupt in Up/down counter mode (DCEN = 1)				
5	RCLK	Receive Clock bit for UART Cleared to use timer 1 overflow as receive clock for serial port in mode 1 or 3. Set to use timer 2 overflow as receive clock for serial port in mode 1 or 3.				
4	TCLK	Transmit Clock bit for UART Cleared to use timer 1 overflow as transmit clock for serial port in mode 1 or 3. Set to use timer 2 overflow as transmit clock for serial port in mode 1 or 3.				
3	EXEN2	Timer 2 External Enable bit Cleared to ignore events on T2EX pin for timer 2 operation. Set to cause a capture or reload when a negative transition on T2EX pin is detected, if timer 2 is not used to clock the serial port.				
2	TR2	Timer 2 Run control bit Cleared to turn off timer 2. Set to turn on timer 2.				
1	C/T2#	Timer/Counter 2 select bit Cleared for timer operation (input from internal clock system: F <sub>CLK PERIPH</sub> ). Set for counter operation (input from T2 input pin, falling edge trigger). Must be 0 for clock out mode.				
0	CP/RL2#	Timer 2 Capture/Reload bit  If RCLK=1 or TCLK=1, CP/RL2# is ignored and timer is forced to auto-reload on timer 2 overflow.  Cleared to auto-reload on timer 2 overflows or negative transitions on T2EX pin if EXEN2=1.  Set to capture on negative transitions on T2EX pin if EXEN2=1.				

Reset Value = 0000 0000b Bit addressable





Table 46. PCON Register

PCON - Power Control Register (87h)

7	6	5	4	3	2	1	0
SMOD1	SMOD0	-	POF	GF1	GF0	PD	IDL

Bit Number	Bit Mnemonic	Description
7	SMOD1	Serial port Mode bit 1 for UART Set to select double baud rate in mode 1, 2 or 3.
6	SMOD0	Serial port Mode bit 0 for UART Cleared to select SM0 bit in SCON register. Set to select FE bit in SCON register.
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
4	POF	Power-Off Flag Cleared to recognize next reset type. Set by hardware when VCC rises from 0 to its nominal voltage. Can also be set by software.
3	GF1	General purpose Flag Cleared by user for general purpose usage. Set by user for general purpose usage.
2	GF0	General purpose Flag Cleared by user for general purpose usage. Set by user for general purpose usage.
1	PD	Power-Down mode bit Cleared by hardware when reset occurs. Set to enter power-down mode.
0	IDL	Idle mode bit Cleared by hardware when interrupt or reset occurs. Set to enter idle mode.

Reset Value = 00X1 0000b

Not bit addressable

Power-off flag reset value will be 1 only after a power on (cold reset). A warm reset doesn't affect the value of this bit.

Table 47. BDRCON Register

BDRCON - Baud Rate Control Register (9Bh)

7	6	5	4	3	2	1	0
-	-	-	BRR	твск	RBCK	SPD	SRC

Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
4	BRR	Baud Rate Run Control bit Cleared to stop the internal Baud Rate Generator. Set to start the internal Baud Rate Generator.
3	TBCK	Transmission Baud rate Generator Selection bit for UART Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.
2	RBCK	Reception Baud Rate Generator Selection bit for UART Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.
1	SPD	Baud Rate Speed Control bit for UART Cleared to select the SLOW Baud Rate Generator. Set to select the FAST Baud Rate Generator.
0	SRC	Baud Rate Source select bit in Mode 0 for UART  Cleared to select F <sub>OSC</sub> /12 as the Baud Rate Generator (F <sub>CLK PERIPH</sub> /6 in X2 mode).  Set to select the internal Baud Rate Generator for UARTs in mode 0.

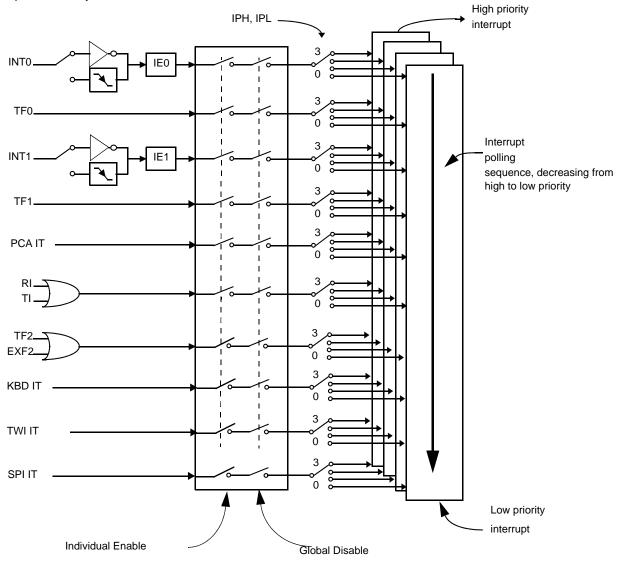
Reset Value = XXX0 0000b Not bit addressablef



# **Interrupt System**

<u>The AT89C51ID2</u> has a total of 10 interrupt vectors: two external interrupts (INT0 and INT1), three timer interrupts (timers 0, 1 and 2), the serial port interrupt, SPI interrupt, Keyboard interrupt and the PCA global interrupt. These interrupts are shown in Figure 25.

Figure 25. Interrupt Control System



Each of the interrupt sources can be individually enabled or disabled by setting or clearing a bit in the Interrupt Enable register (Table 52 and Table 50). This register also contains a global disable bit, which must be cleared to disable all interrupts at once.

Each interrupt source can also be individually programmed to one out of four priority levels by setting or clearing a bit in the Interrupt Priority register (Table 53) and in the Interrupt Priority High register (Table 51 and Table 52) shows the bit values and priority levels associated with each combination.





### Registers

The PCA interrupt vector is located at address 0033H, the SPI interrupt vector is located at address 0043H and Keyboard interrupt vector is located at address 004BH. All other vectors addresses are the same as standard C52 devices.

Table 48. Priority Level Bit Values

IPH. x	IPL. x	Interrupt Level Priority
0	0	0 (Lowest)
0	1	1
1	0	2
1	1	3 (Highest)

A low-priority interrupt can be interrupted by a high priority interrupt, but not by another low-priority interrupt. A high-priority interrupt can't be interrupted by any other interrupt source.

If two interrupt requests of different priority levels are received simultaneously, the request of higher priority level is serviced. If interrupt requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. Thus within each priority level there is a second priority structure determined by the polling sequence.

Table 49. IENO Register

IEN0 - Interrupt Enable Register (A8h)

7	6	5	4	3	2	1	0
EA	EC	ET2	ES	ET1	EX1	ET0	EX0

	<u> </u>					
Bit Number	Bit Mnemonic	Description				
7	EA	Enable All interrupt bit Cleared to disable all interrupts. Set to enable all interrupts.				
6	EC	PCA interrupt enable bit Cleared to disable. Set to enable.				
5	ET2	imer 2 overflow interrupt Enable bit leared to disable timer 2 overflow interrupt. et to enable timer 2 overflow interrupt.				
4	ES	Serial port Enable bit Cleared to disable serial port interrupt. Set to enable serial port interrupt.				
3	ET1	Timer 1 overflow interrupt Enable bit Cleared to disable timer 1 overflow interrupt. Set to enable timer 1 overflow interrupt.				
2	EX1	External interrupt 1 Enable bit Cleared to disable external interrupt 1. Set to enable external interrupt 1.				
1	ET0	Timer 0 overflow interrupt Enable bit Cleared to disable timer 0 overflow interrupt. Set to enable timer 0 overflow interrupt.				
0	EX0	External interrupt 0 Enable bit Cleared to disable external interrupt 0. Set to enable external interrupt 0.				

Reset Value = 0000 0000b Bit addressable





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Table 50. IPL0 Register

IPL0 - Interrupt Priority Register (B8h)

-	PPCL	PT2L	PSL	PT1L	PX1L	PT0L	PX0L	
Bit Number	Bit Mnemonic	Description						
7	-	Reserved The value re	Reserved The value read from this bit is indeterminate. Do not set this bit.					
6	PPCL		PCA interrupt Priority bit Refer to PPCH for priority level.					
5	PT2L		Timer 2 overflow interrupt Priority bit Refer to PT2H for priority level.					
4	PSL	•	Serial port Priority bit Refer to PSH for priority level.					
3	PT1L	Timer 1 overflow interrupt Priority bit Refer to PT1H for priority level.						
2	PX1L		errupt 1 Prion H for priority I	•				

Timer 0 overflow interrupt Priority bit Refer to PT0H for priority level.

External interrupt 0 Priority bit

Refer to PX0H for priority level.

3

2

1

0

Reset Value = X000 0000b Bit addressable

0

PT0L

PX0L

Table 51. IPH0 Register

IPH0 - Interrupt Priority High Register (B7h)

7	6	5	4	3	2	1	0
-	PPCH	PT2H	PSH	PT1H	PX1H	PT0H	PX0H

	<u> </u>	
Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
6	PPCH	PCA interrupt Priority high bit.  PPCHPPCLPriority Level  0 0Lowest  0 1  1 0  1 1Highest
5	PT2H	Timer 2 overflow interrupt Priority High bit PT2HPT2LPriority Level 0
4	PSH	Serial port Priority High bit PSH PSLPriority Level 0
3	PT1H	Timer 1 overflow interrupt Priority High bit PT1HPT1L Priority Level 0 0 Lowest 0 1 1 0 1 1Highest
2	PX1H	External interrupt 1 Priority High bit  PX1HPX1LPriority Level  0
1	PT0H	Timer 0 overflow interrupt Priority High bit PT0HPT0LPriority Level 0 0Lowest 0 1 1 0 1 1Highest
0	PX0H	External interrupt 0 Priority High bit  PX0H PX0LPriority Level  0

Reset Value = X000 0000b Not bit addressable





Table 52. IEN1 Register

IEN1 - Interrupt Enable Register (B1h)

,	6	э	4	3	2	1	U	
-	-	-	-	-	ESPI	ETWI	EKBD	
Bit Number	Bit Mnemonic	Description						
7	-	Reserved						
6	-	Reserved						
5	-	Reserved						
4	-	Reserved						
3	-	Reserved						
2	ESPI	Cleared to di	SPI interrupt Enable bit Cleared to disable SPI interrupt. Set to enable SPI interrupt.					
1	ETWI		t Enable bit isable TWI int e TWI interrup	•				
0	EKBD	Cleared to di	nterrupt Enak isable keyboa e keyboard in	ırd interrupt.				

Reset Value = XXXX X000b Bit addressable

Table 53. IPL1 Register

IPL1 - Interrupt Priority Register (B2h)

#### Table 54.

7	6	5	4	3	2	1	0
-	-	-	-	-	SPIL	TWIL	KBDL

Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
4	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
3	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
2	SPIL	SPI interrupt Priority bit Refer to SPIH for priority level.
1	TWIL	TWI interrupt Priority bit Refer to TWIH for priority level.
0	KBDL	Keyboard interrupt Priority bit Refer to KBDH for priority level.

Reset Value = XXXX X000b Bit addressable





Table 55. IPH1 Register

IPH1 - Interrupt Priority High Register (B3h)

7	6	5	4	3	2	1	0
-	-	-	•	•	SPIH	TWIH	KBDH

Bit Number	Bit Mnemonic	Description					
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.					
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.					
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.					
4	-	Reserved The value read from this bit is indeterminate. Do not set this bit.					
3	-	Reserved The value read from this bit is indeterminate. Do not set this bit.					
2	SPIH	SPI interrupt Priority High bit SPIH SPILPriority Level 0 OLowest 0 1 1 0 1 Highest					
1	TWIH	TWI interrupt Priority High bit  TWIH TWILPriority Level 0 OLowest 0 1 1 0 1 Highest					
0	KBDH	Keyboard interrupt Priority High bitKB DHKBDLPriority Level00 Lowest011011 Highest					

Reset Value = XXXX X000b Not bit addressable

# Interrupt Sources and Vector Addresses

Table 56. Interrupt Sources and Vector Addresses

Number	Polling Priority	Interrupt Source	Interrupt Request	Vector Address
0	0	Reset		0000h
1	1	INT0	IE0	0003h
2	2	Timer 0	TF0	000Bh
3	3	INT1	IE1	0013h
4	4	Timer 1	IF1	001Bh
5	6	UART	RI+TI	0023h
6	7	Timer 2	TF2+EXF2	002Bh
7	5	PCA	CF + CCFn (n = 0-4)	0033h
8	8	Keyboard	KBDIT	003Bh
9	9	TWI	TWIIT	0043h
9	9	SPI	SPIIT	004Bh



# **Power Management**

#### Introduction

Two power reduction modes are implemented in the AT89C51ID2. The Idle mode and the Power-Down mode. These modes are detailed in the following sections. In addition to these power reduction modes, the clocks of the core and peripherals can be dynamically divided by 2 using the X2 mode detailed in Section "Enhanced Features", page 21.

#### **Idle Mode**

Idle mode is a power reduction mode that reduces the power consumption. In this mode, program execution halts. Idle mode freezes the clock to the CPU at known states while the peripherals continue to be clocked. The CPU status before entering Idle mode is preserved, i.e., the program counter and program status word register retain their data for the duration of Idle mode. The contents of the SFRs and RAM are also retained. The status of the Port pins during Idle mode is detailed in Table 57.

#### **Entering Idle Mode**

To enter Idle mode, set the IDL bit in PCON register (see Table 58). The AT89C51ID2 enters Idle mode upon execution of the instruction that sets IDL bit. The instruction that sets IDL bit is the last instruction executed.

Note: If IDL bit and PD bit are set simultaneously, the AT89C51ID2 enters Power-Down mode. Then it does not go in Idle mode when exiting Power-Down mode.

#### **Exiting Idle Mode**

There are two ways to exit Idle mode:

- 1. Generate an enabled interrupt.
  - Hardware clears IDL bit in PCON register which restores the clock to the CPU. Execution resumes with the interrupt service routine. Upon completion of the interrupt service routine, program execution resumes with the instruction immediately following the instruction that activated Idle mode. The general purpose flags (GF1 and GF0 in PCON register) may be used to indicate whether an interrupt occurred during normal operation or during Idle mode. When Idle mode is exited by an interrupt, the interrupt service routine may examine GF1 and GF0.

#### 2. Generate a reset.

A logic high on the RST pin clears IDL bit in PCON register directly and asynchronously. This restores the clock to the CPU. Program execution momentarily resumes with the instruction immediately following the instruction that activated the Idle mode and may continue for a number of clock cycles before the internal reset algorithm takes control. Reset initializes the AT89C51ID2 and vectors the CPU to address C:0000h.

Note: During the time that execution resumes, the internal RAM cannot be accessed; however, it is possible for the Port pins to be accessed. To avoid unexpected outputs at the Port pins, the instruction immediately following the instruction that activated Idle mode should not write to a Port pin or to the external RAM.

#### **Power-Down Mode**

The Power-Down mode places the AT89C51ID2 in a very low power state. Power-Down mode stops the oscillator, freezes all clock at known states. The CPU status prior to entering Power-Down mode is preserved, i.e., the program counter, program status word register retain their data for the duration of Power-Down mode. In addition, the SFR





and RAM contents are preserved. The status of the Port pins during Power-Down mode is detailed in Table 57.

Note

VCC may be reduced to as low as  $V_{RET}$  during Power-Down mode to further reduce power dissipation. Take care, however, that VDD is not reduced until Power-Down mode is invoked.

#### **Entering Power-Down Mode**

To enter Power-Down mode, set PD bit in PCON register. The AT89C51ID2 enters the Power-Down mode upon execution of the instruction that sets PD bit. The instruction that sets PD bit is the last instruction executed.

#### **Exiting Power-Down Mode**

Note: If VCC was reduced during the Power-Down mode, do not exit Power-Down mode until VCC is restored to the normal operating level.

There are two ways to exit the Power-Down mode:

- Generate an enabled external interrupt.
  - The AT89C51ID2 provides capability to exit from Power-Down using INT0#, INT1#.

Hardware clears PD bit in PCON register which starts the oscillator and restores the clocks to the CPU and peripherals. Using INTx# input, execution resumes when the input is released (see Figure 26). Execution resumes with the interrupt service routine. Upon completion of the interrupt service routine, program execution resumes with the instruction immediately following the instruction that activated Power-Down mode.

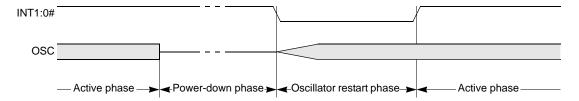
Note:

The external interrupt used to exit Power-Down mode must be configured as level sensitive (INT0# and INT1#) and must be assigned the highest priority. In addition, the duration of the interrupt must be long enough to allow the oscillator to stabilize. The execution will only resume when the interrupt is deasserted.

Note:

Exit from power-down by external interrupt does not affect the SFRs nor the internal RAM content.

Figure 26. Power-Down Exit Waveform Using INT1:0#



#### 2. Generate a reset.

A logic high on the RST pin clears PD bit in PCON register directly and asynchronously. This starts the oscillator and restores the clock to the CPU and peripherals. Program execution momentarily resumes with the instruction immediately following the instruction that activated Power-Down mode and may continue for a number of clock cycles before the internal reset algorithm takes control. Reset initializes the AT89C51ID2 and vectors the CPU to address 0000h.

Note:

During the time that execution resumes, the internal RAM cannot be accessed; however, it is possible for the Port pins to be accessed. To avoid unexpected outputs at the Port

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pins, the instruction immediately following the instruction that activated the Power-Down mode should not write to a Port pin or to the external RAM.

Note: Exit from power-down by reset redefines all the SFRs, but does not affect the internal RAM content.

Table 57. Pin Conditions in Special Operating Modes

Mode	Port 0	Port 1	Port 2	Port 3	Port 4	ALE	PSEN#
Reset	Floating	High	High	High	High	High	High
Idle (internal code)	Data	Data	Data	Data	Data Data		High
Idle (external code)	Floating	Data	Data	Data	Data	High	High
Power- Down(inter nal code)	Data	Data	Data	Data	Data	Low	Low
Power- Down (external code)	Floating	Data	Data	Data	Data	Low	Low





#### Registers

**Table 58.** PCON Register PCON (S87:h) Power configuration Register

7	6	5	4	3	2	1	0
-	-	-	POF	GF1	GF0	PD	IDL

Bit Number	Bit Mnemonic	Description
7-5	-	Reserved The value read from these bits is indeterminate. Do not set these bits.
4	POF	Power-Off Flag Cleared to recognize next reset type. Set by hardware when VCC rises from 0 to its nominal voltage. Can also be set by software.
3	GF1	General Purpose flag 1 One use is to indicate whether an interrupt occurred during normal operation or during Idle mode.
2	GF0	General Purpose flag 0 One use is to indicate whether an interrupt occurred during normal operation or during Idle mode.
1	PD	Power-Down Mode bit Cleared by hardware when an interrupt or reset occurs. Set to activate the Power-Down mode. If IDL and PD are both set, PD takes precedence.
0	IDL	Idle Mode bit Cleared by hardware when an interrupt or reset occurs. Set to activate the Idle mode. If IDL and PD are both set, PD takes precedence.

Reset Value= XXXX 0000b

#### **Keyboard Interface**

The AT89C51ID2 implements a keyboard interface allowing the connection of a 8 x n matrix keyboard. It is based on 8 inputs with programmable interrupt capability on both high or low level. These inputs are available as alternate function of P1 and allow to exit from idle and power down modes.

The keyboard interface interfaces with the C51 core through 3 special function registers: KBLS, the Keyboard Level Selection register (Table 61), KBE, The Keyboard interrupt Enable register (Table 60), and KBF, the Keyboard Flag register (Table 59).

Interrupt

The keyboard inputs are considered as 8 independent interrupt sources sharing the same interrupt vector. An interrupt enable bit (KBD in IE1) allows global enable or disable of the keyboard interrupt (see Figure 27). As detailed in Figure 28 each keyboard input has the capability to detect a programmable level according to KBLS. x bit value. Level detection is then reported in interrupt flags KBF. x that can be masked by software using KBE. x bits.

This structure allow keyboard arrangement from 1 by n to 8 by n matrix and allow usage of P1 inputs for other purpose.

Figure 27. Keyboard Interface Block Diagram

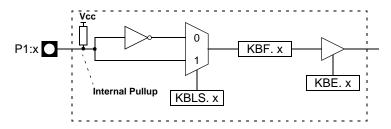
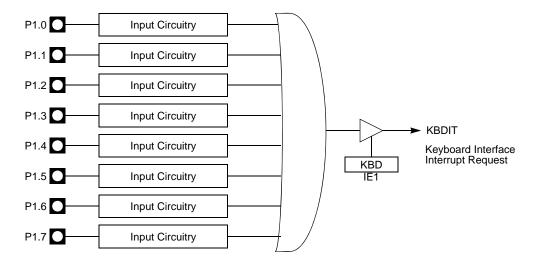


Figure 28. Keyboard Input Circuitry



**Power Reduction Mode** 

P1 inputs allow exit from idle and power down modes as detailed in Section "Power Management", page 72.





#### Registers

#### Table 59. KBF Register

KBF-Keyboard Flag Register (9Eh)

7	6	5	4	3	2	1	U
KBF7	KBF6	KBF5	KBF4	KBF3	KBF2	KBF1	KBF0

Bit Number	Bit Mnemonic	Description
7	KBF7	Keyboard line 7 flag Set by hardware when the Port line 7 detects a programmed level. It generates a Keyboard interrupt request if the KBKBIE. 7 bit in KBIE register is set. Must be cleared by software.
6	KBF6	Keyboard line 6 flag Set by hardware when the Port line 6 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 6 bit in KBIE register is set. Must be cleared by software.
5	KBF5	Keyboard line 5 flag Set by hardware when the Port line 5 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 5 bit in KBIE register is set. Must be cleared by software.
4	KBF4	Keyboard line 4 flag Set by hardware when the Port line 4 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 4 bit in KBIE register is set. Must be cleared by software.
3	KBF3	Keyboard line 3 flag Set by hardware when the Port line 3 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 3 bit in KBIE register is set. Must be cleared by software.
2	KBF2	Keyboard line 2 flag Set by hardware when the Port line 2 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 2 bit in KBIE register is set. Must be cleared by software.
1	KBF1	Keyboard line 1 flag Set by hardware when the Port line 1 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 1 bit in KBIE register is set. Must be cleared by software.
0	KBF0	Keyboard line 0 flag Set by hardware when the Port line 0 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 0 bit in KBIE register is set. Must be cleared by software.

#### Reset Value= 0000 0000b

This register is read only access, all flags are automatically cleared by reading the register.

Table 60. KBE Register

KBE-Keyboard Input Enable Register (9Dh)

,	6	5	4	3	2	1	0			
KBE7	KBE6	KBE5	KBE4	KBE3	KBE2	KBE1	KBE0			
Bit Number	Bit Mnemonic	Description	Description							
7	KBE7	Cleared to e	Keyboard line 7 Enable bit Cleared to enable standard I/O pin. Set to enable KBF. 7 bit in KBF register to generate an interrupt request.							
6	KBE6	Cleared to e	Keyboard line 6 Enable bit Cleared to enable standard I/O pin. Set to enable KBF. 6 bit in KBF register to generate an interrupt request.							
5	KBE5	Cleared to er	Keyboard line 5 Enable bit Cleared to enable standard I/O pin. Set to enable KBF. 5 bit in KBF register to generate an interrupt request.							
4	KBE4	Keyboard line 4 Enable bit Cleared to enable standard I/O pin. Set to enable KBF. 4 bit in KBF register to generate an interrupt request.								
3	KBE3	Cleared to e	ne 3 Enable t nable standar e KBF. 3 bit in		to generate ar	n interrupt req	uest.			
2	KBE2	Cleared to er	ne 2 Enable t nable standar e KBF. 2 bit in		to generate ar	n interrupt req	uest.			
1	KBE1	Cleared to e	ne 1 Enable t nable standar e KBF. 1 bit in		to generate ar	n interrupt req	uest.			
0	KBE0	Cleared to e	ne 0 Enable t nable standar e KBF. 0 bit in		to generate ar	n interrupt req	uest.			

**Reset Value= 0000 0000b** 





Table 61. KBLS Register

KBLS-Keyboard Level Selector Register (9Ch)

7	6	5	4	3	2	1	0		
KBLS7	KBLS6	KBLS5	KBLS4	KBLS3	KBLS2	KBLS1	KBLS0		
Bit Number	Bit Mnemonic	Description							
7	KBLS7	Cleared to e	Keyboard line 7 Level Selection bit Cleared to enable a low level detection on Port line 7. Set to enable a high level detection on Port line 7.						
6	KBLS6	Cleared to e	Keyboard line 6 Level Selection bit Cleared to enable a low level detection on Port line 6. Set to enable a high level detection on Port line 6.						
5	KBLS5	Cleared to e	Keyboard line 5 Level Selection bit Cleared to enable a low level detection on Port line 5. Set to enable a high level detection on Port line 5.						
4	KBLS4	Cleared to e	Keyboard line 4 Level Selection bit Cleared to enable a low level detection on Port line 4. Set to enable a high level detection on Port line 4.						
3	KBLS3	Cleared to e			on Port line 3. Port line 3.				
2	KBLS2	Cleared to e			on Port line 2. Port line 2.				
1	KBLS1	Cleared to e	Keyboard line 1 Level Selection bit Cleared to enable a low level detection on Port line 1. Set to enable a high level detection on Port line 1.						
0	KBLS0	Cleared to e			on Port line 0. Port line 0.				

Reset Value= 0000 0000b

#### 2-wire Interface (TWI)

This section describes the 2-wire interface. The 2-wire bus is a bi-directional 2-wire serial communication standard. It is designed primarily for simple but efficient integrated circuit (IC) control. The system is comprised of two lines, SCL (Serial Clock) and SDA (Serial Data) that carry information between the ICs connected to them. The serial data transfer is limited to 400 Kbit/s in standard mode. Various communication configuration can be designed using this bus. Figure 29 shows a typical 2-wire bus configuration. All the devices connected to the bus can be master and slave.

Figure 29. 2-wire Bus Configuration

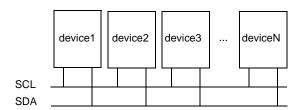
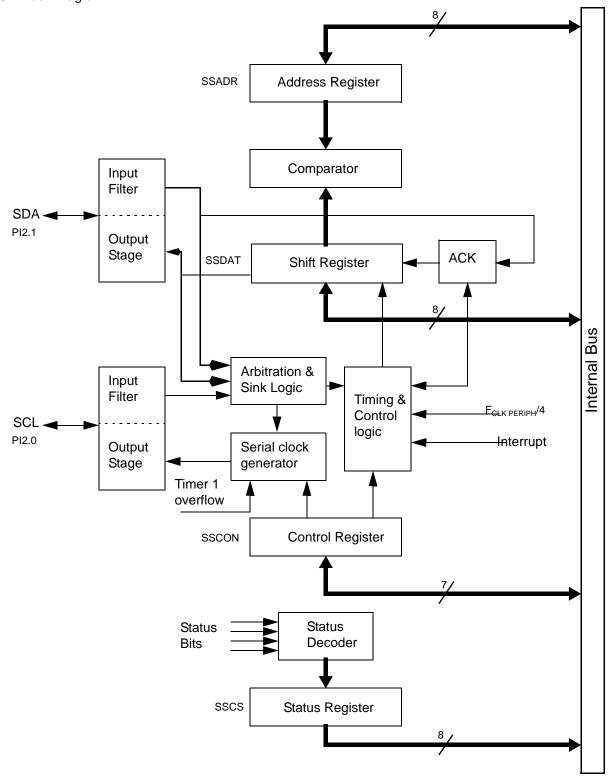




Figure 30. Block Diagram



#### **Description**

The CPU interfaces to the 2-wire logic via the following four 8-bit special function registers: the Synchronous Serial Control register (SSCON; Table 71), the Synchronous Serial Data register (SSDAT; Table 72), the Synchronous Serial Control and Status register (SSCS; Table 73) and the Synchronous Serial Address register (SSADR Table 76).

SSCON is used to enable the TWI interface, to program the bit rate (see Table 64), to enable slave modes, to acknowledge or not a received data, to send a START or a STOP condition on the 2-wire bus, and to acknowledge a serial interrupt. A hardware reset disables the TWI module.

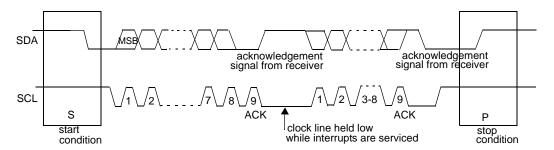
SSCS contains a status code which reflects the status of the 2-wire logic and the 2-wire bus. The three least significant bits are always zero. The five most significant bits contains the status code. There are 26 possible status codes. When SSCS contains F8h, no relevant state information is available and no serial interrupt is requested. A valid status code is available in SSCS one machine cycle after SI is set by hardware and is still present one machine cycle after SI has been reset by software. to Table 70. give the status for the master modes and miscellaneous states.

SSDAT contains a byte of serial data to be transmitted or a byte which has just been received. It is addressable while it is not in process of shifting a byte. This occurs when 2-wire logic is in a defined state and the serial interrupt flag is set. Data in SSDAT remains stable as long as SI is set. While data is being shifted out, data on the bus is simultaneously shifted in; SSDAT always contains the last byte present on the bus.

SSADR may be loaded with the 7-bit slave address (7 most significant bits) to which the TWI module will respond when programmed as a slave transmitter or receiver. The LSB is used to enable general call address (00h) recognition.

Figure 31 shows how a data transfer is accomplished on the 2-wire bus.

Figure 31. Complete Data Transfer on 2-wire Bus



The four operating modes are:

- Master Transmitter
- Master Receiver
- Slave transmitter
- Slave receiver

Data transfer in each mode of operation is shown in Table to Table 70 and Figure 32. to Figure 35.. These figures contain the following abbreviations:

S: START condition

R: Read bit (high level at SDA)





W: Write bit (low level at SDA)

A: Acknowledge bit (low level at SDA)

A: Not acknowledge bit (high level at SDA)

Data: 8-bit data byte
P: STOP condition

In Figure 32 to Figure 35, circles are used to indicate when the serial interrupt flag is set. The numbers in the circles show the status code held in SSCS. At these points, a service routine must be executed to continue or complete the serial transfer. These service routines are not critical since the serial transfer is suspended until the serial interrupt flag is cleared by software.

When the serial interrupt routine is entered, the status code in SSCS is used to branch to the appropriate service routine. For each status code, the required software action and details of the following serial transfer are given in Table to Table 70.

#### **Master Transmitter Mode**

In the master transmitter mode, a number of data bytes are transmitted to a slave receiver (Figure 32). Before the master transmitter mode can be entered, SSCON must be initialised as follows:

Table 62. SSCON Initialization

CR2	SSIE	STA	STO	SI	AA	CR1	CR0
bit rate	1	0	0	0	Х	bit rate	bit rate

CR0, CR1 and CR2 define the internal serial bit rate if external bit rate generator is not used. SSIE must be set to enable TWI. STA, STO and SI must be cleared.

The master transmitter mode may now be entered by setting the STA bit. The 2-wire logic will now test the 2-wire bus and generate a START condition as soon as the bus becomes free. When a START condition is transmitted, the serial interrupt flag (SI bit in SSCON) is set, and the status code in SSCS will be 08h. This status must be used to vector to an interrupt routine that loads SSDAT with the slave address and the data direction bit (SLA+W).

When the slave address and the direction bit have been transmitted and an acknowledgement bit has been received, SI is set again and a number of status code in SSCS are possible. There are 18h, 20h or 38h for the master mode and also 68h, 78h or B0h if the slave mode was enabled (AA=logic 1). The appropriate action to be taken for each of these status code is detailed in Table . This scheme is repeated until a STOP condition is transmitted.

SSIE, CR2, CR1 and CR0 are not affected by the serial transfer and are referred to Table 7 to Table 11. After a repeated START condition (state 10h) the TWI module may switch to the master receiver mode by loading SSDAT with SLA+R.

#### **Master Receiver Mode**

In the master receiver mode, a number of data bytes are received from a slave transmitter (Figure 33). The transfer is initialized as in the master transmitter mode. When the START condition has been transmitted, the interrupt routine must load SSDAT with the 7-bit slave address and the data direction bit (SLA+R). The serial interrupt flag SI must then be cleared before the serial transfer can continue.

When the slave address and the direction bit have been transmitted and an acknowledgement bit has been received, the serial interrupt flag is set again and a number of status code in SSCS are possible. There are 40h, 48h or 38h for the master mode and also 68h, 78h or B0h if the slave mode was enabled (AA=logic 1). The appropriate action to be taken for each of these status code is detailed in Table . This scheme is repeated until a STOP condition is transmitted.

SSIE, CR2, CR1 and CR0 are not affected by the serial transfer and are referred to Table 7 to Table 11. After a repeated START condition (state 10h) the TWI module may switch to the master transmitter mode by loading SSDAT with SLA+W.

#### Slave Receiver Mode

In the slave receiver mode, a number of data bytes are received from a master transmitter (Figure 34). To initiate the slave receiver mode, SSADR and SSCON must be loaded as follows:

Table 63. SSADR: Slave Receiver Mode Initialization

A6	A5	A4	А3	A2	A1	A0	GC		
	own slave address								

The upper 7 bits are the address to which the TWI module will respond when addressed by a master. If the LSB (GC) is set the TWI module will respond to the general call address (00h); otherwise it ignores the general call address.

Table 64. SSCON: Slave Receiver Mode Initialization

CR2	SSIE	STA	STO	SI	AA	CR1	CR0
bit rate	1	0	0	0	1	bit rate	bit rate

CR0, CR1 and CR2 have no effect in the slave mode. SSIE must be set to enable the TWI. The AA bit must be set to enable the own slave address or the general call address acknowledgement. STA, STO and SI must be cleared.

When SSADR and SSCON have been initialised, the TWI module waits until it is addressed by its own slave address followed by the data direction bit which must be at logic 0 (W) for the TWI to operate in the slave receiver mode. After its own slave address and the W bit have been received, the serial interrupt flag is set and a valid status code can be read from SSCS. This status code is used to vector to an interrupt service routine. The appropriate action to be taken for each of these status code is detailed in Table . The slave receiver mode may also be entered if arbitration is lost while TWI is in the master mode (states 68h and 78h).

If the AA bit is reset during a transfer, TWI module will return a not acknowledge (logic 1) to SDA after the next received data byte. While AA is reset, the TWI module does not respond to its own slave address. However, the 2-wire bus is still monitored and address recognition may be resume at any time by setting AA. This means that the AA bit may be used to temporarily isolate the module from the 2-wire bus.

#### **Slave Transmitter Mode**

In the slave transmitter mode, a number of data bytes are transmitted to a master receiver (Figure 35). Data transfer is initialized as in the slave receiver mode. When SSADR and SSCON have been initialized, the TWI module waits until it is addressed by





its own slave address followed by the data direction bit which must be at logic 1 (R) for TWI to operate in the slave transmitter mode. After its own slave address and the R bit have been received, the serial interrupt flag is set and a valid status code can be read from SSCS. This status code is used to vector to an interrupt service routine. The appropriate action to be taken for each of these status code is detailed in Table . The slave transmitter mode may also be entered if arbitration is lost while the TWI module is in the master mode.

If the AA bit is reset during a transfer, the TWI module will transmit the last byte of the transfer and enter state C0h or C8h. the TWI module is switched to the not addressed slave mode and will ignore the master receiver if it continues the transfer. Thus the master receiver receiver all 1's as serial data. While AA is reset, the TWI module does not respond to its own slave address. However, the 2-wire bus is still monitored and address recognition may be resume at any time by setting AA. This means that the AA bit may be used to temporarily isolate the TWI module from the 2-wire bus.

#### Miscellaneous States

There are two SSCS codes that do not correspond to a define TWI hardware state (Table 70). These codes are discuss hereafter.

Status F8h indicates that no relevant information is available because the serial interrupt flag is not set yet. This occurs between other states and when the TWI module is not involved in a serial transfer.

Status 00h indicates that a bus error has occurred during a TWI serial transfer. A bus error is caused when a START or a STOP condition occurs at an illegal position in the format frame. Examples of such illegal positions happen during the serial transfer of an address byte, a data byte, or an acknowledge bit. When a bus error occurs, SI is set. To recover from a bus error, the STO flag must be set and SI must be cleared. This causes the TWI module to enter the not addressed slave mode and to clear the STO flag (no other bits in SSCON are affected). The SDA and SCL lines are released and no STOP condition is transmitted.

**Notes** 

the TWI module interfaces to the external 2-wire bus via two port pins: SCL (serial clock line) and SDA (serial data line). To avoid low level asserting on these lines when the TWI module is enabled, the output latches of SDA and SLC must be set to logic 1.

Table 65. Bit Frequency Configuration

			Bit Freque	ency ( kHz)	
CR2	CR1	CR0	F <sub>OSCA</sub> = 12 MHz	F <sub>OSCA</sub> = 16 MHz	F <sub>OSCA</sub> divided by
0	0	0	47	62.5	256
0	0	1	53.5	71.5	224
0	1	0	62.5	83	192
0	1	1	75	100	160
1	0	0	-	-	Unused
1	0	1	100	133.3	120
1	1	0	200	266.6	60
1	1	1	0.5 <. < 62.5	0.67 <. < 83	96 · (256 - reload valueTimer 1) (reload value range: 0-254 in mode 2)

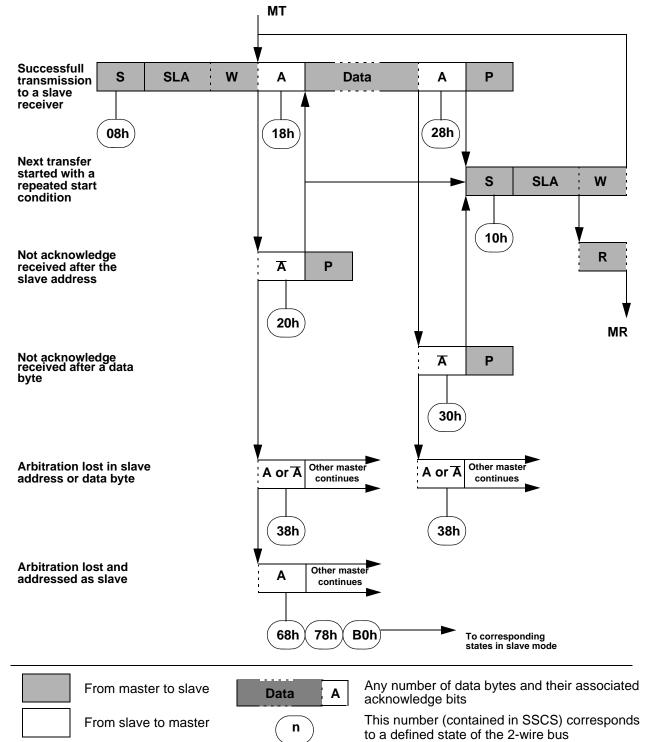


Figure 32. Format and State in the Master Transmitter Mode





Table 66. Status in Master Transmitter Mode

		Appli	cation soft	ware respo	nse		
Status Code	Status of the Two- wire Bus and Two-			To SSC	CON		
SSSTA	wire Hardware	To/From SSDAT	SSSTA	SSSTO	SSI	SSAA	Next Action Taken by Two-wire Hardware
08h	A START condition has been transmitted	Write SLA+W	Х	0	0	Х	SLA+W will be transmitted.
10h	A repeated START	Write SLA+W	Х	0	0	х	SLA+W will be transmitted.
10h condition has been transmitted	Write SLA+R	Х	0	0	Х	SLA+R will be transmitted. Logic will switch to master receiver mode	
18h	SLA+W has been transmitted; ACK has been received	Write data byte No SSDAT action No SSDAT action No SSDAT action	0 1 0	0 0 1	0 0 0	X X X	Data byte will be transmitted. Repeated START will be transmitted. STOP condition will be transmitted and SSSTO flag will be reset. STOP condition followed by a START condition will
20h	SLA+W has been transmitted; NOT ACK has been received	Write data byte No SSDAT action No SSDAT action No SSDAT action	0 1 0	0 0 1	0 0 0	X X X	be transmitted and SSSTO flag will be reset.  Data byte will be transmitted. Repeated START will be transmitted. STOP condition will be transmitted and SSSTO flag will be reset. STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.
28h	Data byte has been transmitted; ACK has been received	Write data byte No SSDAT action No SSDAT action No SSDAT action	0 1 0	0 0 1	0 0 0	X X X	Data byte will be transmitted. Repeated START will be transmitted. STOP condition will be transmitted and SSSTO flag will be reset. STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.
30h	Data byte has been transmitted; NOT ACK has been received	Write data byte No SSDAT action No SSDAT action No SSDAT action	0 1 0	0 0 1	0 0 0	X X X	Data byte will be transmitted. Repeated START will be transmitted. STOP condition will be transmitted and SSSTO flag will be reset. STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.
38h	Arbitration lost in SLA+W or data bytes	No SSDAT action	0	0	0	х	Two-wire bus will be released and not addressed slave mode will be entered.  A START condition will be transmitted when the bus
	SEATW OF Gala bytes	No SSDAT action	1	0	0	Х	becomes free.

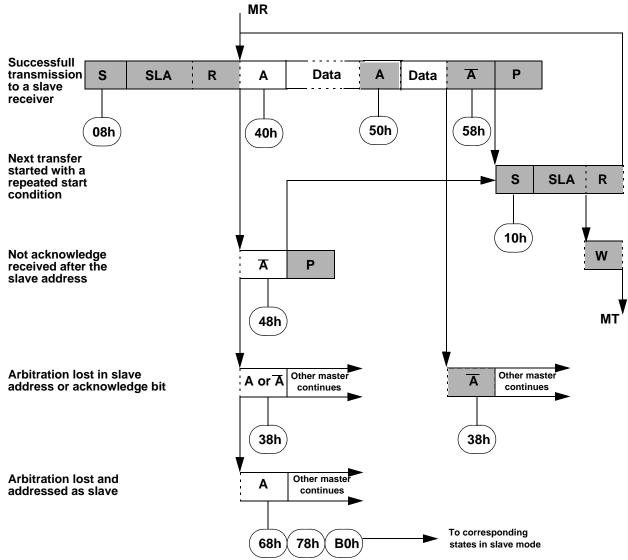


Figure 33. Format and State in the Master Receiver Mode

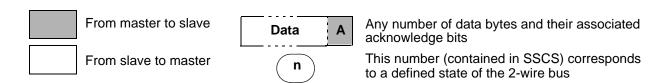






Table 67. Status in Master Receiver Mode

		Appli	Application software response					
Status Code	Status of the Two- wire Bus and Two-			To SSC	CON			
SSSTA	wire Hardware	To/From SSDAT	SSSTA	SSSTO	SSI	SSAA	Next Action Taken by Two-wire Hardware	
08h	A START condition has been transmitted	Write SLA+R	х	0	0	Х	SLA+R will be transmitted.	
1.01	A repeated START	Write SLA+R	Х	0	0	Х	SLA+R will be transmitted.	
10h	condition has been transmitted	Write SLA+W	X	0	0	Х	SLA+W will be transmitted. Logic will switch to master transmitter mode.	
38h	Arbitration lost in SLA+R or NOT ACK	No SSDAT action	0	0	0	Х	Two-wire bus will be released and not addressed slave mode will be entered.	
3011	bit	No SSDAT action	1	0	0	Х	A START condition will be transmitted when the bus becomes free.	
40h	SLA+R has been transmitted; ACK has	No SSDAT action	0	0	0	0	Data byte will be received and NOT ACK will be returned.	
	been received	No SSDAT action	0	0	0	1	Data byte will be received and ACK will be returned.	
401	SLA+R has been	No SSDAT action No SSDAT action	1	0	0	X X	Repeated START will be transmitted. STOP condition will be transmitted and SSSTO flag	
48h	transmitted; NOT ACK has been received	No SSDAT action	1	1	0	x	will be reset.  STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.	
50h	Data byte has been received; ACK has	Read data byte	0	0	0	0	Data byte will be received and NOT ACK will be returned.	
5011	been returned	Read data byte	0	0	0	1	Data byte will be received and ACK will be returned.	
	Data huta haa haas	Read data byte	1	0	0	Х	Repeated START will be transmitted.	
58h	Data byte has been received; NOT ACK	Read data byte	0	1	0	Х	STOP condition will be transmitted and SSSTO flag will be reset.	
	has been returned	Read data byte	1	1	0	Х	STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.	

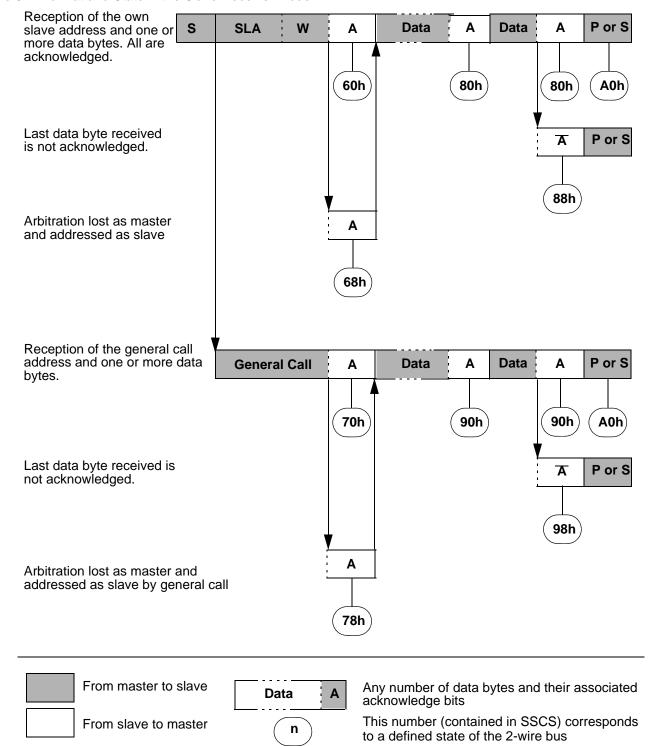


Figure 34. Format and State in the Slave Receiver Mode





Table 68. Status in Slave Receiver Mode

		Application Software Response					
Status		To/from SSDAT		To SS	CON		
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	ѕто	SI	AA	Next Action Taken By 2-wire Software
COL	Own SLA+W has been	No SSDAT action or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
60h	received; ACK has been returned	No SSDAT action	Х	0	0	1	Data byte will be received and ACK will be returned
COL	Arbitration lost in SLA+R/W as master; own SLA+W has been	No SSDAT action or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
68h	received; ACK has been returned	No SSDAT action	Х	0	0	1	Data byte will be received and ACK will be returned
701-	General call address has been	No SSDAT action or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
70h	received; ACK has been returned	No SSDAT action	Х	0	0	1	Data byte will be received and ACK will be returned
701	Arbitration lost in SLA+R/W as master; general call address	No SSDAT action or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
78h	has been received; ACK has been returned	No SSDAT action	Х	0	0	1	Data byte will be received and ACK will be returned
001	Previously addressed with own SLA+W; data has been	No SSDAT action or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
80h	received; ACK has been returned	No SSDAT action	Х	0	0	1	Data byte will be received and ACK will be returned
		Dood data buta an	0		0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA
		Read data byte or  Read data byte or	0	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if
	Previously addressed with	redu data byte of					GC=logic 1 Switched to the not addressed slave mode; no
88h	own SLA+W; data has been received; NOT ACK has been returned	Read data byte or	1	0	0	0	recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free
		Read data byte	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free
651	Previously addressed with general call; data has been	Read data byte or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
90h	received; ACK has been returned	Read data byte	Х	0	0	1	Data byte will be received and ACK will be returned

Table 68. Status in Slave Receiver Mode (Continued)

		Application S	oftware	Respo	nse		
Status		To/from SSDAT		To SS	CON		
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	STA STO SI		AA	Next Action Taken By 2-wire Software
	Previously addressed with	Read data byte or Read data byte or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1 Switched to the not addressed slave mode; no
98h	general call; data has been received; NOT ACK has been returned	Read data byte or	1	0	0	0	recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free
		Read data byte	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free
		No SSDAT action or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own
	A 0.T.O.D	No SSDAT action or	0	0	0	1	SLA will be recognised; GCA will be recognised if GC=logic 1
A0h	A STOP condition or repeated START condition has been received while still addressed as slave	No SSDAT action or	1	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free
		No SSDAT action	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free





Figure 35. Format and State in the Slave Transmitter Mode

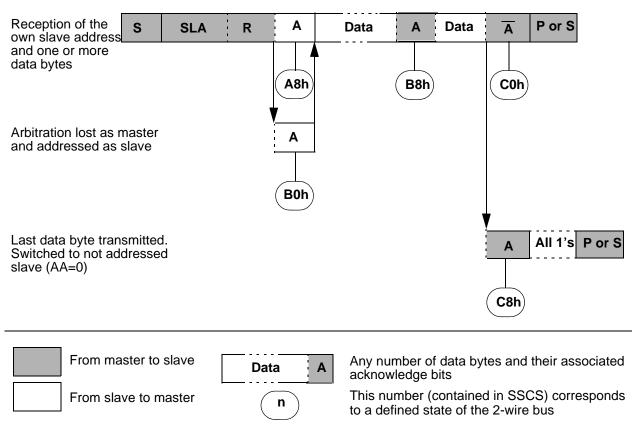


Table 69. Status in Slave Transmitter Mode

		Application S	Software	Respo				
Status		To/from SSDAT		To SS	CON			
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	STA STO SI AA		AA	Next Action Taken By 2-wire Software	
A OL	Own SLA+R has been	Load data byte or	Х	0	0	0	Last data byte will be transmitted and NOT ACK will be received	
A8h	received; ACK has been returned	Load data byte	Х	0	0	1	Data byte will be transmitted and ACK will be received	
Dok	Arbitration lost in SLA+R/W as master; own SLA+R has been	Load data byte or	Х	0	0	0	Last data byte will be transmitted and NOT ACK will be received	
B0h	received; ACK has been returned	Load data byte	Х	0	0	1	Data byte will be transmitted and ACK will be received	
Dob	Data byte in SSDAT has been	Load data byte or	Х	0	0	0	Last data byte will be transmitted and NOT ACK will be received	
B8h	transmitted; NOT ACK has been received	Load data byte	Х	0	0	1	Data byte will be transmitted and ACK will be received	

Table 69. Status in Slave Transmitter Mode (Continued)

		Application S	oftware	Respo	nse		
Status		To/from SSDAT		To SS	CON		
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	sto	SI	AA	Next Action Taken By 2-wire Software
	Data byte in SSDAT has been transmitted; NOT ACK has been received	No SSDAT action or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1
C0h		No SSDAT action or	1	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free
		No SSDAT action	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free
		No SSDAT action or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if
		No SSDAT action or	0	0	0	1	GC=logic 1
C8h	Last data byte in SSDAT has been transmitted (AA=0); ACK has been received	No SSDAT action or	1	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free
		No SSDAT action	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free

Table 70. Miscellaneous Status

		Application	Softwa	are Res			
		To/from		To SS	CON		
Status Code (SSCS)	Status of the 2-wire bus and 2-wire hardware	SSDAT	STA	sто	SI	AA	Next Action Taken By 2-wire Software
F8h	No relevant state information available; SI= 0	No SSDAT action	No SSCON action			ion	Wait or proceed current transfer
00h	Bus error due to an illegal START or STOP condition	No SSDAT action	0 1 0		0	Х	Only the internal hardware is affected, no STOP condition is sent on the bus. In all cases, the bus is released and STO is reset.





#### Registers

Table 71. SSCON Register

SSCON - Synchronous Serial Control register (93h)

7	6	5	4	3	2	1	0
CR2	SSIE	STA	STO	SI	AA	CR1	CR0

Bit Number	Bit Mnemonic	Description						
7	CR2	Control Rate bit 2 See Table 65.						
6	SSIE	Synchronous Serial Interface Enable bit Clear to disable the TWI module. Set to enable the TWI module.						
5	STA	Start flag Set to send a START condition on the bus.						
4	ST0	Stop flag Set to send a STOP condition on the bus.						
3	SI	Synchronous Serial Interrupt flag Set by hardware when a serial interrupt is requested. Must be cleared by software to acknowledge interrupt.						
2	AA	Assert Acknowledge flag Clear in master and slave receiver modes, to force a not acknowledge (high level on SDA). Clear to disable SLA or GCA recognition. Set to recognise SLA or GCA (if GC set) for entering slave receiver or transmitter modes. Set in master and slave receiver modes, to force an acknowledge (low level on SDA). This bit has no effect when in master transmitter mode.						
1	CR1	Control Rate bit 1 See Table 65.						
0	CR0	Control Rate bit 0 See Table 65.						

#### Table 72. SSDAT (095h) - Syncrhonous Serial Data register (read/write)

SD7	SD6	SD5	SD4	SD3	SD2	SD1	SD0
7	6	5	4	3	2	1	0

Bit Number	Bit Mnemonic	Description
7	SD7	Address bit 7 or Data bit 7.
6	SD6	Address bit 6 or Data bit 6.
5	SD5	Address bit 5 or Data bit 5.
4	SD4	Address bit 4 or Data bit 4.
3	SD3	Address bit 3 or Data bit 3.
2	SD2	Address bit 2 or Data bit 2.

Bit Number	Bit Mnemonic	Description
1	SD1	Address bit 1 or Data bit 1.
0	SD0	Address bit 0 (R/W) or Data bit 0.

#### Table 73. SSCS (094h) read - Synchronous Serial Control and Status Register

7	6	5	4	3	2	1	0
SC4	SC3	SC2	SC1	SC0	0	0	0

Table 74. SSCS Register: Read Mode - Reset Value = F8h

Bit Number	Bit Mnemonic	Description
0	0	Always zero
1	0	Always zero
2	0	Always zero
3	SC0	Status Code bit 0 See to Table 70.
4	SC1	Status Code bit 1 See to Table 70.
5	SC2	Status Code bit 2 See to Table 70.
6	SC3	Status Code bit 3 See to Table 70.
7	SC4	Status Code bit 4 See to Table 70.

#### Table 75. SSADR (096h) - Synchronus Serial Address Register (read/write)

7	6	5	4	3	2	1	0
A7	A6	A5	A4	А3	A2	A1	A0

#### **Table 76.** SSADR Register - Reset value = FEh

Bit Number	Bit Mnemonic	Description
7	A7	Slave Address bit 7
6	A6	Slave Address bit 6
5	A5	Slave Address bit 5
4	A4	Slave Address bit 4
3	А3	Slave Address bit 3
2	A2	Slave Address bit 2
1	A1	Slave Address bit 1





Bit Number	Bit Mnemonic	Description
		General Call bit
0	GC	Clear to disable the general call address recognition.
		Set to enable the general call address recognition.

## Serial Port Interface (SPI)

The Serial Peripheral Interface Module (SPI) allows full-duplex, synchronous, serial communication between the MCU and peripheral devices, including other MCUs.

#### **Features**

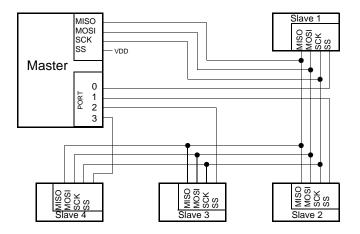
Features of the SPI Module include the following:

- Full-duplex, three-wire synchronous transfers
- Master or Slave operation
- Eight programmable Master clock rates
- Serial clock with programmable polarity and phase
- · Master Mode fault error flag with MCU interrupt capability
- Write collision flag protection

#### **Signal Description**

Figure 36 shows a typical SPI bus configuration using one Master controller and many Slave peripherals. The bus is made of three wires connecting all the devices.

Figure 36. SPI Master/Slaves Interconnection



The Master device selects the individual Slave devices by using four pins of a parallel port to control the four SS pins of the Slave devices.

### Master Output Slave Input (MOSI)

This 1-bit signal is directly connected between the Master Device and a Slave Device. The MOSI line is used to transfer data in series from the Master to the Slave. Therefore, it is an output signal from the Master, and an input signal to a Slave. A Byte (8-bit word) is transmitted most significant bit (MSB) first, least significant bit (LSB) last.

## Master Input Slave Output (MISO)

This 1-bit signal is directly connected between the Slave Device and a Master Device. The MISO line is used to transfer data in series from the Slave to the Master. Therefore, it is an output signal from the Slave, and an input signal to the Master. A Byte (8-bit word) is transmitted most significant bit (MSB) first, least significant bit (LSB) last.

#### SPI Serial Clock (SCK)

This signal is used to synchronize the data movement both in and out of the devices through their MOSI and MISO lines. It is driven by the Master for eight clock cycles which allows to exchange one Byte on the serial lines.

#### Slave Select (SS)

Each Slave peripheral is selected by one Slave Select pin (SS). This signal must stay low for any message for a Slave. It is obvious that only one Master (SS high level) can





drive the network. The Master may select each Slave device by software through port pins (Figure 37). To prevent bus conflicts on the MISO line, only one slave should be selected at a time by the Master for a transmission.

In a Master configuration, the  $\overline{SS}$  line can be used in conjunction with the MODF flag in the SPI Status register (SPSTA) to prevent multiple masters from driving MOSI and SCK (see Error conditions).

A high level on the  $\overline{SS}$  pin puts the MISO line of a Slave SPI in a high-impedance state.

The SS pin could be used as a general-purpose if the following conditions are met:

- The device is configured as a Master and the SSDIS control bit in SPCON is set. This kind of configuration can be found when only one Master is driving the network and there is no way that the SS pin could be pulled low. Therefore, the MODF flag in the SPSTA will never be set<sup>(1)</sup>.
- The Device is configured as a Slave with CPHA and SSDIS control bits set<sup>(2)</sup>. This kind of configuration can happen when the system comprises one Master and one Slave only. Therefore, the device should always be selected and there is no reason that the Master uses the SS pin to select the communicating Slave device.

Note: 1. Clearing SSDIS control bit does not clear MODF.

2. Special care should be taken not to set SSDIS control bit when CPHA = '0' because in this mode, the  $\overline{SS}$  is used to start the transmission.

In Master mode, the baud rate can be selected from a baud rate generator which is controlled by three bits in the SPCON register: SPR2, SPR1 and SPR0. The Master clock is selected from one of seven clock rates resulting from the division of the internal clock by 2, 4, 8, 16, 32, 64 or 128.

Table 77 gives the different clock rates selected by SPR2:SPR1:SPR0.

Table 77. SPI Master Baud Rate Selection

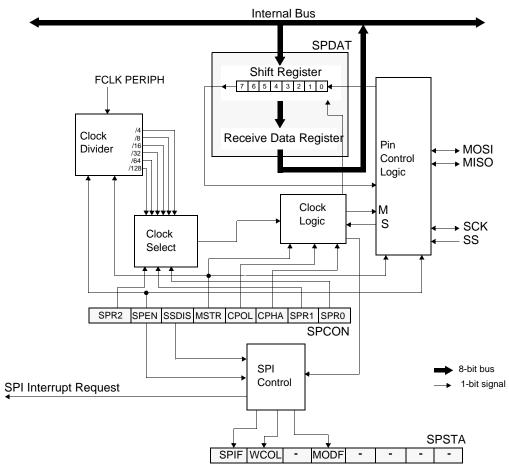
SPR2	SPR1	SPR0	Clock Rate	Baud Rate Divisor (BD)
0	0	0	F <sub>CLK PERIPH</sub> /2	2
0	0	1	F <sub>CLK PERIPH</sub> /4	4
0	1	0	F <sub>CLK PERIPH</sub> /8	8
0	1	1	F <sub>CLK PERIPH</sub> /16	16
1	0	0	F <sub>CLK PERIPH</sub> /32	32
1	0	1	F <sub>CLK PERIPH</sub> /64	64
1	1	0	F <sub>CLK PERIPH</sub> /128	128
1	1	1	Don't Use	No BRG

**Baud Rate** 

#### **Functional Description**

Figure 37 shows a detailed structure of the SPI Module.

Figure 37. SPI Module Block Diagram



#### **Operating Modes**

The Serial Peripheral Interface can be configured in one of the two modes: Master mode or Slave mode. The configuration and initialization of the SPI Module is made through one register:

The Serial Peripheral Control register (SPCON)

Once the SPI is configured, the data exchange is made using:

- SPCON
- The Serial Peripheral STAtus register (SPSTA)
- The Serial Peripheral DATa register (SPDAT)

During an SPI transmission, data is simultaneously transmitted (shifted out serially) and received (shifted in serially). A serial clock line (SCK) synchronizes shifting and sampling on the two serial data lines (MOSI and MISO). A Slave Select line (SS) allows individual selection of a Slave SPI device; Slave devices that are not selected do not interfere with SPI bus activities.

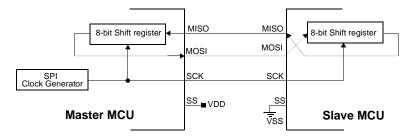
When the Master device transmits data to the Slave device via the MOSI line, the Slave device responds by sending data to the Master device via the MISO line. This implies full-duplex transmission with both data out and data in synchronized with the same clock (Figure 38).



100



Figure 38. Full-Duplex Master-Slave Interconnection



Master Mode

Slave Mode

**Transmission Formats** 

The SPI operates in Master mode when the Master bit, MSTR <sup>(1)</sup>, in the SPCON register is set. Only one Master SPI device can initiate transmissions. Software begins the transmission from a Master SPI Module by writing to the Serial Peripheral Data Register (SPDAT). If the shift register is empty, the Byte is immediately transferred to the shift register. The Byte begins shifting out on MOSI pin under the control of the serial clock, SCK. Simultaneously, another Byte shifts in from the Slave on the Master's MISO pin. The transmission ends when the Serial Peripheral transfer data flag, SPIF, in SPSTA becomes set. At the same time that SPIF becomes set, the received Byte from the Slave is transferred to the receive data register in SPDAT. Software clears SPIF by reading the Serial Peripheral Status register (SPSTA) with the SPIF bit set, and then reading the SPDAT.

The SPI operates in Slave mode when the Master bit, MSTR <sup>(2)</sup>, in the SPCON register is cleared. Before a data transmission occurs, the Slave Select pin, SS, of the Slave device must be set to '0'. SS must remain low until the transmission is complete.

In a Slave SPI Module, data enters the shift register under the control of the SCK from the Master SPI Module. After a Byte enters the shift register, it is immediately transferred to the receive data register in SPDAT, and the SPIF bit is set. To prevent an overflow condition, Slave software must then read the SPDAT before another Byte enters the shift register  $^{(3)}$ . A Slave SPI must complete the write to the SPDAT (shift register) at least one bus cycle before the Master SPI starts a transmission. If the write to the data register is late, the SPI transmits the data already in the shift register from the previous transmission. The maximum SCK frequency allowed in slave mode is  $F_{CLK PERIPH}$  /4.

Software can select any of four combinations of serial clock (SCK) phase and polarity using two bits in the SPCON: the Clock Polarity (CPOL <sup>(4)</sup>) and the Clock Phase (CPHA<sup>4</sup>). CPOL defines the default SCK line level in idle state. It has no significant effect on the transmission format. CPHA defines the edges on which the input data are sampled and the edges on which the output data are shifted (Figure 39 and Figure 40). The clock phase and polarity should be identical for the Master SPI device and the communicating Slave device.

- The SPI Module should be configured as a Master before it is enabled (SPEN set). Also, the Master SPI should be configured before the Slave SPI.
- The SPI Module should be configured as a Slave before it is enabled (SPEN set).
- The maximum frequency of the SCK for an SPI configured as a Slave is the bus clock speed.
- 4. Before writing to the CPOL and CPHA bits, the SPI should be disabled (SPEN = '0').

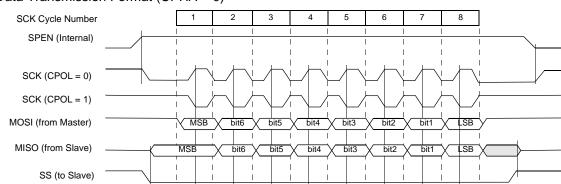


Figure 39. Data Transmission Format (CPHA = 0)

Figure 40. Data Transmission Format (CPHA = 1)

Capture Point

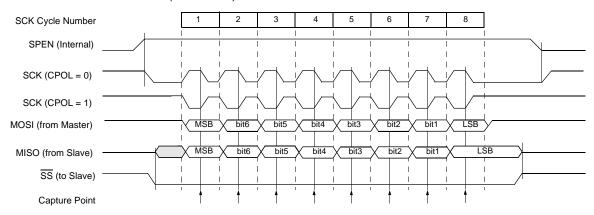
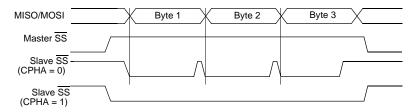


Figure 41. CPHA/SS Timing



As shown in Figure 39, the first SCK edge is the MSB capture strobe. Therefore, the Slave must begin driving its data before the first SCK edge, and a falling edge on the SS pin is used to start the transmission. The SS pin must be toggled high and then low between each Byte transmitted (Figure 41).

Figure 40 shows an SPI transmission in which CPHA is '1'. In this case, the Master begins driving its MOSI pin on the first SCK edge. Therefore, the Slave uses the first SCK edge as a start transmission signal. The SS pin can remain low between transmissions (Figure 41). This format may be preffered in systems having only one Master and only one Slave driving the MISO data line.





#### **Error Conditions**

The following flags in the SPSTA signal SPI error conditions:

Mode Fault (MODF)

Mode Fault error in Master mode SPI indicates that the level on the Slave Select  $(\overline{SS})$  pin is inconsistent with the actual mode of the device. MODF is set to warn that there may be a multi-master conflict for system control. In this case, the SPI system is affected in the following ways:

- An SPI receiver/error CPU interrupt request is generated
- The SPEN bit in SPCON is cleared. This disables the SPI
- The MSTR bit in SPCON is cleared.

When SS Disable (SSDIS) bit in the SPCON register is cleared, the MODF flag is set when the SS signal becomes '0'.

However, as stated before, for a system with one Master, if the SS pin of the Master device is pulled low, there is no way that another Master attempts to drive the network. In this case, to prevent the MODF flag from being set, software can set the SSDIS bit in the SPCON register and therefore making the SS pin as a general-purpose I/O pin.

Clearing the MODF bit is accomplished by a read of SPSTA register with MODF bit set, followed by a write to the SPCON register. SPEN Control bit may be restored to its original set state after the MODF bit has been cleared.

Write Collision (WCOL)

A Write Collision (WCOL) flag in the SPSTA is set when a write to the SPDAT register is done during a transmit sequence.

WCOL does not cause an interruption, and the transfer continues uninterrupted.

Clearing the WCOL bit is done through a software sequence of an access to SPSTA and an access to SPDAT.

Overrun Condition

An overrun condition occurs when the Master device tries to send several data Bytes and the Slave devise has not cleared the SPIF bit issuing from the previous data Byte transmitted. In this case, the receiver buffer contains the Byte sent after the SPIF bit was last cleared. A read of the SPDAT returns this Byte. All others Bytes are lost.

This condition is not detected by the SPI peripheral.

SS Error Flag (SSERR)

A Synchronous Serial Slave Error occurs when  $\overline{SS}$  goes high before the end of a received data in slave mode. SSERR does not cause in interruption, this bit is cleared by writing 0 to SPEN bit (reset of the SPI state machine).

Interrupts

Two SPI status flags can generate a CPU interrupt requests:

Table 78. SPI Interrupts

Flag	Request		
SPIF (SP data transfer)	SPI Transmitter Interrupt request		
MODF (Mode Fault)	SPI Receiver/Error Interrupt Request (if SSDIS = '0')		

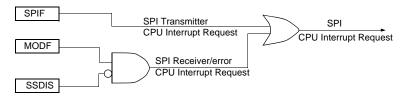
Serial Peripheral data transfer flag, SPIF: This bit is set by hardware when a transfer has been completed. SPIF bit generates transmitter CPU interrupt requests.

Mode Fault flag, MODF: This bit becomes set to indicate that the level on the SS is inconsistent with the mode of the SPI. MODF with SSDIS reset, generates receiver/error CPU interrupt requests. When SSDIS is set, no MODF interrupt request is generated.

Figure 42 gives a logical view of the above statements.

AT89C51ID2 =

Figure 42. SPI Interrupt Requests Generation



#### Registers

Serial Peripheral Control Register (SPCON) There are three registers in the Module that provide control, status and data storage functions. These registers are describes in the following paragraphs.

- The Serial Peripheral Control Register does the following:
- Selects one of the Master clock rates
- Configure the SPI Module as Master or Slave
- Selects serial clock polarity and phase
- Enables the SPI Module
- Frees the SS pin for a general-purpose

Table 79 describes this register and explains the use of each bit

Table 79. SPCON Register

SPCON - Serial Peripheral Control Register (0C3H)

Table 1.

CPOL

MSTR

OI ILL	OI EIT GODIO		0.02	<b>0111</b> A	OI IXI		
Bit Number	Bit Mnemonic	Description					
7	SPR2	Serial Periphe Bit with SPR1 a		ine the clock	rate.		
6	SPEN	Serial Peripheral Enable Cleared to disable the SPI interface. Set to enable the SPI interface.					
5	SSDIS	SS Disable  Cleared to enable SS in both Master and Slave modes.  Set to disable SS in both Master and Slave modes. In Slave mode, this bit has no effect if CPHA ='0'. When SSDIS is set, no MODF interrupt request is generated.					
4	MSTR	Serial Periphe Cleared to configure Set to configure	figure the SPI				
3	CPOL	Clock Polarity Cleared to have the SCK set to '0' in idle state. Set to have the SCK set to '1' in idle low.					
2	СРНА	Clock Phase Cleared to have the data sampled when the SCK leaves the idle state (see CPOL). Set to have the data sampled when the SCK returns to idle state (see CPOL).					





Bit Number	Bit Mnemonic	Description				
		SPR2	SPR1	SPR0	Serial Peripheral Rate	
1	SPR1	0	0	0	F <sub>CLK PERIPH</sub> /2	
'		0	0	1	F <sub>CLK PERIPH</sub> /4	
		0	1	0	F <sub>CLK PERIPH</sub> /8	
		0	1	1	F <sub>CLK PERIPH</sub> /16	
		1	0	0	F <sub>CLK PERIPH</sub> /32	
0	SPR0	1	0	1	F <sub>CLK PERIPH</sub> /64	
		1	1	0	F <sub>CLK PERIPH</sub> /128	
		1	1	1	Invalid	

Reset Value = 0001 0100b

Not bit addressable

Serial Peripheral Status Register (SPSTA)

The Serial Peripheral Status Register contains flags to signal the following conditions:

- Data transfer complete
- Write collision
- Inconsistent logic level on SS pin (mode fault error)

Table 80 describes the SPSTA register and explains the use of every bit in the register.

Table 80. SPSTA Register

SPSTA - Serial Peripheral Status and Control register (0C4H)

#### Table 2.

7	6	5	4	3	2	1	0
SPI	F WCOL	SSERR	MODF	-	-	-	-

Bit Number	Bit Mnemonic	Description
7	SPIF	Serial Peripheral Data Transfer Flag Cleared by hardware to indicate data transfer is in progress or has been approved by a clearing sequence. Set by hardware to indicate that the data transfer has been completed.
6	WCOL	Write Collision Flag Cleared by hardware to indicate that no collision has occurred or has been approved by a clearing sequence. Set by hardware to indicate that a collision has been detected.
5	SSERR	Synchronous Serial Slave Error Flag Set by hardware when SS is deasserted before the end of a received data. Cleared by disabling the SPI (clearing SPEN bit in SPCON).
4	MODF	Mode Fault  Cleared by hardware to indicate that the SS pin is at appropriate logic level, or has been approved by a clearing sequence.  Set by hardware to indicate that the SS pin is at inappropriate logic level.
3	-	Reserved The value read from this bit is indeterminate. Do not set this bit
2	-	Reserved The value read from this bit is indeterminate. Do not set this bit.

Bit Number	Bit Mnemonic	Description
1	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
0	-	Reserved The value read from this bit is indeterminate. Do not set this bit.

Reset Value = 00X0 XXXXb

Not Bit addressable

### Serial Peripheral DATa Register (SPDAT)

The Serial Peripheral Data Register (Table 81) is a read/write buffer for the receive data register. A write to SPDAT places data directly into the shift register. No transmit buffer is available in this model.

A Read of the SPDAT returns the value located in the receive buffer and not the content of the shift register.

Table 81. SPDAT Register

SPDAT - Serial Peripheral Data Register (0C5H)

			_	_
4	٦.	_	_	7

7	6	5	4	3	2	1	0
R7	R6	R5	R4	R3	R2	R1	R0

Reset Value = Indeterminate

R7:R0: Receive data bits

SPCON, SPSTA and SPDAT registers may be read and written at any time while there is no on-going exchange. However, special care should be taken when writing to them while a transmission is on-going:

- Do not change SPR2, SPR1 and SPR0
- Do not change CPHA and CPOL
- Do not change MSTR
- Clearing SPEN would immediately disable the peripheral
- Writing to the SPDAT will cause an overflow.



## Hardware Watchdog Timer

The WDT is intended as a recovery method in situations where the CPU may be subjected to software upset. The WDT consists of a 14-bit counter and the WatchDog Timer ReSeT (WDTRST) SFR. The WDT is by default disabled from exiting reset. To enable the WDT, user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When WDT is enabled, it will increment every machine cycle while the oscillator is running and there is no way to disable the WDT except through reset (either hardware reset or WDT overflow reset). When WDT overflows, it will drive an output RESET HIGH pulse at the RST-pin.

#### **Using the WDT**

To enable the WDT, user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When WDT is enabled, the user needs to service it by writing to 01EH and 0E1H to WDTRST to avoid WDT overflow. The 14-bit counter overflows when it reaches 16383 (3FFFH) and this will reset the device. When WDT is enabled, it will increment every machine cycle while the oscillator is running. This means the user must reset the WDT at least every 16383 machine cycle. To reset the WDT the user must write 01EH and 0E1H to WDTRST. WDTRST is a write only register. The WDT counter cannot be read or written. When WDT overflows, it will generate an output RESET pulse at the RST-pin. The RESET pulse duration is 96 x  $T_{CLK PERIPH}$ , where  $T_{CLK PERIPH} = 1/F_{CLK PERIPH}$ . To make the best use of the WDT, it should be serviced in those sections of code that will periodically be executed within the time required to prevent a WDT reset.

To have a more powerful WDT, a  $2^7$  counter has been added to extend the Time-out capability, ranking from 16ms to 2s @  $F_{OSCA}$  = 12MHz. To manage this feature, refer to WDTPRG register description, Table 82.

Table 82. WDTRST Register

WDTRST - Watchdog Reset Register (0A6h)

7	6	5	4	3	2	1	0
-	-	-	-	-	-	•	-

Reset Value = XXXX XXXXb

Write only, this SFR is used to reset/enable the WDT by writing 01EH then 0E1H in sequence.





Table 83. WDTPRG Register

WDTPRG - Watchdog Timer Out Register (0A7h)

7	6	5	4	3	2	1	0
-	-	-	-	-	S2	S1	S0

Bit Number	Bit Mnemonic	Description					
7	-						
6	-	Reserved The value read from this bit is undetermined. Do not try to set this bit.					
5	-						
4	-						
3	-						
2	S2	WDT Time-out select bit 2					
1	S1	WDT Time-out select bit 1					
0	S0	WDT Time-out select bit 0					
		$ \begin{array}{cccccccccccccccccccccccccccccccccccc$					

Reset value = XXXX X000

## WDT During Power Down and Idle

In Power Down mode the oscillator stops, which means the WDT also stops. While in Power Down mode the user does not need to service the WDT. There are 2 methods of exiting Power Down mode: by a hardware reset or via a level activated external interrupt which is enabled prior to entering Power Down mode. When Power Down is exited with hardware reset, servicing the WDT should occur as it normally should whenever the AT89C51ID2 is reset. Exiting Power Down with an interrupt is significantly different. The interrupt is held low long enough for the oscillator to stabilize. When the interrupt is brought high, the interrupt is serviced. To prevent the WDT from resetting the device while the interrupt pin is held low, the WDT is not started until the interrupt is pulled high. It is suggested that the WDT be reset during the interrupt service routine.

To ensure that the WDT does not overflow within a few states of exiting of powerdown, it is better to reset the WDT just before entering powerdown.

In the Idle mode, the oscillator continues to run. To prevent the WDT from resetting the AT89C51ID2 while in Idle mode, the user should always set up a timer that will periodically exit Idle, service the WDT, and re-enter Idle mode.

# ONCE<sup>(TM)</sup> Mode (ON Chip Emulation)

The ONCE mode facilitates testing and debugging of systems using AT89C51ID2 without removing the circuit from the board. The ONCE mode is invoked by driving certain pins of the AT89C51ID2; the following sequence must be exercised:

- Pull ALE low while the device is in reset (RST high) and PSEN is high.
- Hold ALE low as RST is deactivated.

While the AT89C51ID2 is in ONCE mode, an emulator or test CPU can be used to drive the circuit Table 84 shows the status of the port pins during ONCE mode.

Normal operation is restored when normal reset is applied.

Table 84. External Pin Status during ONCE Mode

ALE	PSEN	Port 0	Port 1	Port 2	Port 3	Port I2	XTALA1/2	XTALB1/2
Weak pull-up	Weak pull-up	Float	Weak pull-up	Weak pull-up	Weak pull-up	Float	Active	Active

(a) "Once" is a registered trademark of Intel Corporation.



# **Power-off Flag**

The power-off flag allows the user to distinguish between a "cold start" reset and a "warm start" reset.

A cold start reset is the one induced by  $V_{CC}$  switch-on. A warm start reset occurs while  $V_{CC}$  is still applied to the device and could be generated for example by an exit from power-down.

The power-off flag (POF) is located in PCON register (Table 85). POF is set by hardware when  $V_{CC}$  rises from 0 to its nominal voltage. The POF can be set or cleared by software allowing the user to determine the type of reset.

Table 85. PCON Register

PCON - Power Control Register (87h)

7	6	5	4	3	2	1	0
SMOD1	SMOD0	-	POF	GF1	GF0	PD	IDL

Bit Number	Bit Mnemonic	Description					
7	SMOD1	Serial port Mode bit 1 Set to select double baud rate in mode 1, 2 or 3.					
6	SMOD0	Serial port Mode bit 0 Cleared to select SM0 bit in SCON register. Set to select FE bit in SCON register.					
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.					
4	POF	Power-Off Flag Cleared to recognize next reset type. Set by hardware when V <sub>CC</sub> rises from 0 to its nominal voltage. Can also be set by software.					
3	GF1	General purpose Flag Cleared by user for general purpose usage. Set by user for general purpose usage.					
2	GF0	General purpose Flag Cleared by user for general purpose usage. Set by user for general purpose usage.					
1	PD	Power-Down mode bit Cleared by hardware when reset occurs. Set to enter power-down mode.					
0	IDL	Idle mode bit Cleared by hardware when interrupt or reset occurs. Set to enter idle mode.					

Reset Value = 00X1 0000b Not bit addressable



# **EEPROM Data Memory**

The 2K bytes on-chip EEPROM memory block is located at addresses 0000h to 07FFh of the XRAM/ERAM memory space and is selected by setting control bits in the EECON register.

A read or write access to the EEPROM memory is done with a MOVX instruction.

#### Write Data

Data is written by byte to the EEPROM memory block as for an external RAM memory.

The following procedure is used to write to the EEPROM memory:

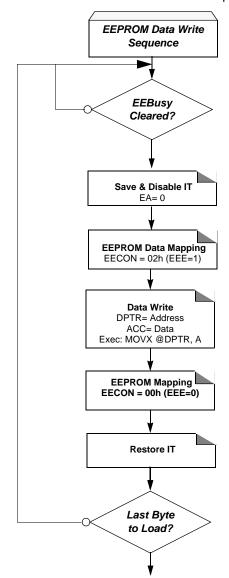
- Check EEBUSY flag
- If the user application interrupts routines use XRAM memory space: Save and disable interrupts.
- Load DPTR with the address to write
- Store A register with the data to be written
- Set bit EEE of EECON register
- Execute a MOVX @DPTR, A
- Clear bit EEE of EECON register
- Restore interrupts.
- EEBUSY flag in EECON is then set by hardware to indicate that programming is in progress and that the EEPROM segment is not available for reading or writing.
- The end of programming is indicated by a hardware clear of the EEBUSY flag.

Figure 43 represents the optimal write sequence to the on-chip EEPROM data memory.





Figure 43. Recommended EEPROM Data Write Sequence

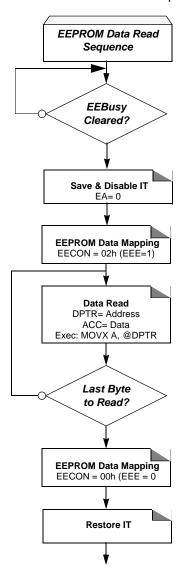


### **Read Data**

The following procedure is used to read the data stored in the EEPROM memory:

- Check EEBUSY flag
- If the user application interrupts routines use XRAM memory space: Save and disable interrupts.
- Load DPTR with the address to read
- Set bit EEE of EECON register
- Execute a MOVX A, @DPTR
- Clear bit EEE of EECON register
- Restore interrupts.

Figure 44. Recommended EEPROM Data Read Sequence







# Registers

Table 86. EECON Register

EECON (0D2h) EEPROM Control Register

7	6	5	4	3	2	1	0
-	-	-	-	-	-	EEE	EEBUSY

Bit Number	Bit Mnemonic	Description
7 - 2	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
1	EEE	Enable EEPROM Space bit Set to map the EEPROM space during MOVX instructions (Write or Read to the EEPROM . Clear to map the XRAM space during MOVX.
0	EEBUSY	Programming Busy flag Set by hardware when programming is in progress. Cleared by hardware when programming is done. Can not be set or cleared by software.

Reset Value = XXXX XX00b Not bit addressable

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# **Reduced EMI Mode**

The ALE signal is used to demultiplex address and data buses on port 0 when used with external program or data memory. Nevertheless, during internal code execution, ALE signal is still generated. In order to reduce EMI, ALE signal can be disabled by setting AO bit.

The AO bit is located in AUXR register at bit location 0. As soon as AO is set, ALE is no longer output but remains active during MOVX and MOVC instructions and external fetches. During ALE disabling, ALE pin is weakly pulled high.

Table 87. AUXR Register

AUXR - Auxiliary Register (8Eh)

7	6	5	4	3	2	1	0
-	-	МО	XRS2	XRS1	XRS0	EXTRAM	AO

Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
5	МО	Pulse length Cleared to stretch MOVX control: the RD/ and the WR/ pulse length is 6 clock periods (default). Set to stretch MOVX control: the RD/ and the WR/ pulse length is 30 clock periods.
4	XRS2	XRAM Size
3	XRS1	XRS2 XRS1XRS0XRAM size 0 0 0 256 bytes
2	XRS0	0 0 1 512 bytes 0 1 0 768 bytes(default) 0 1 1 1024 bytes 1 0 0 1792 bytes
1	EXTRAM	EXTRAM bit Cleared to access internal XRAM using movx @ Ri/ @ DPTR. Set to access external memory. Programmed by hardware after Power-up regarding Hardware Security Byte (HSB), default setting, XRAM selected.
0	AO	ALE Output bit Cleared, ALE is emitted at a constant rate of 1/6 the oscillator frequency (or 1/3 if X2 mode is used). (default) Set, ALE is active only during a MOVX or MOVC instruction is used.

Reset Value = XX00 10'HSB. XRAM'0b Not bit addressable



# Flash Memory

The Flash memory increases EPROM and ROM functionality with in-circuit electrical erasure and programming. It contains 64K bytes of program memory organized respectively in 512 pages of 128 bytes. This memory is both parallel and serial In-System Programmable (ISP). ISP allows devices to alter their own program memory in the actual end product under software control. A default serial loader (bootloader) program allows ISP of the Flash.

The programming **does not require** external dedicated programming voltage. The necessary high programming voltage is generated on-chip using the standard  $V_{CC}$  pins of the microcontroller.

### **Features**

- Flash internal program memory.
- Boot vector allows user provided Flash loader code to reside anywhere in the Flash memory space. This configuration provides flexibility to the user.
- Default loader in Boot ROM allows programming via the serial port without the need of a user provided loader.
- Up to 64K byte external program memory if the internal program memory is disabled (EA = 0).
- Programming and erase voltage with standard power supply.
- Read/Programming/Erase:
- Byte-wise read without wait state
- Byte or page erase and programming (10 ms)
- Typical programming time (64K bytes) is 22s with on chip serial bootloader
- Parallel programming with 87C51 compatible hardware interface to programmer
- Programmable security for the code in the Flash
- 100k write cycles
- 10 years data retention

# Flash Programming and Erasure

The 64K bytes Flash is programmed by bytes or by pages of 128 bytes. It is not necessary to erase a byte or a page before programming. The programming of a byte or a page includes a self erase before programming.

There are three methods of programming the Flash memory:

- First, the on-chip ISP bootloader may be invoked which will use low level routines to program the pages. The interface used for serial downloading of Flash is the UART.
- Second, the Flash may be programmed or erased in the end-user application by calling low-level routines through a common entry point in the Boot ROM.
- Third, the Flash may be programmed using the parallel method by using a
  conventional EPROM programmer. The parallel programming method used by
  these devices is similar to that used by EPROM 87C51 but it is not identical and the
  commercially available programmers need to have support for the AT89C51ID2.
  The bootloader and the Application Programming Interface (API) routines are
  located in the BOOT ROM.





# Flash Registers and Memory Map

The AT89C51ID2 Flash memory uses several registers for his management:

- Hardware registers can only be accessed through the parallel programming modes which are handled by the parallel programmer.
- Software registers are in a special page of the Flash memory which can be accessed through the API or with the parallel programming modes. This page, called "Extra Flash Memory", is not in the internal Flash program memory addressing space.

### **Hardware Register**

The only hardware register of the AT89C51ID2 is called Hardware Security Byte (HSB).

Table 88. Hardware Security Byte (HSB)

7	6	5	4	3	2	1	0
X2	BLJB	osc	-	XRAM	LB2	LB1	LB0

Bit Number	Bit Mnemonic	Description
7	X2	X2 Mode Programmed ('0' value) to force X2 mode (6 clocks per instruction) after reset. Unprogrammed ('1' Value) to force X1 mode, Standard Mode, after reset (Default).
6	BLJB	Boot Loader Jump Bit Unprogrammed ('1' value) to start the user's application on next reset at address 0000h. Programmed ('0' value) to start the boot loader at address F800h on next reset (Default).
5	osc	Oscillator Bit Programmed to allow oscillator B at startup Unprogrammed this bit to allow oscillator A at startup ( Default).
4	-	Reserved
3	XRAM	XRAM config bit (only programmable by programmer tools) Programmed to inhibit XRAM Unprogrammed, this bit to valid XRAM (Default)
2-0	LB2-0	User Memory Lock Bits (only programmable by programmer tools) See Table 89

Boot Loader Jump Bit (BLJB)

One bit of the HSB, the BLJB bit, is used to force the boot address:

- When this bit is programmed ('1' value) the boot address is 0000h.
- When this bit is unprogrammed ('1' value) the boot address is F800h. By default, this bit is unprogrammed and the ISP is enabled.

### Flash Memory Lock Bits

The three lock bits provide different levels of protection for the on-chip code and data, when programmed as shown in Table 89.

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Table 89. Program Lock Bits

Program Lock Bits				
Security level	LB0	LB1	LB2	Protection description
1	U	U	U	No program lock features enabled.
2	Р	U	U	MOVC instruction executed from external program memory is disabled from fetching code bytes from internal memory, EA is sampled and latched on reset, and further parallel programming of the on chip code memory is disabled.  ISP and software programming with API are still allowed.  Writing EEprom Data from external parallel programmer is disabled but still allowed from internal code execution.
3	Х	Р	U	Same as 2, also verify code memory through parallel programming interface is disabled.  Writing And Reading EEPROM Data from external parallel programmer is disabled but still allowed from internal code execution
4	Х	Х	Р	Same as 3, also external execution is disabled. (Default)

Note:

U: unprogrammed or "one" level.

P: programmed or "zero" level.

X: do not care

WARNING: Security level 2 and 3 should only be programmed after Flash and code verification.

These security bits protect the code access through the parallel programming interface. They are set by default to level 4. The code access through the ISP is still possible and is controlled by the "software security bits" which are stored in the extra Flash memory accessed by the ISP firmware.

To load a new application with the parallel programmer, a chip erase must first be done. This will set the HSB in its inactive state and will erase the Flash memory. The part reference can always be read using Flash parallel programming modes.

### **Default Values**

The default value of the HSB provides parts ready to be programmed with ISP:

- BLJB: Programmed force ISP operation.
- X2: Unprogrammed to force X1 mode (Standard Mode).
- XRAM: Unprogrammed to valid XRAM
- LB2-0: Security level four to protect the code from a parallel access with maximum security.

### **Software Registers**

Several registers are used, in factory and by parallel programmers. These values are used by Atmel ISP.

These registers are in the "Extra Flash Memory" part of the Flash memory. This block is also called "XAF" or eXtra Array Flash. They are accessed in the following ways:

- Commands issued by the parallel memory programmer.
- Commands issued by the ISP software.
- Calls of API issued by the application software.

Several software registers described in Table 90.



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Table 90. Default Values

Mnemonic	Definition	Default value	Description
SBV	Software Boot Vector	FCh	
HSB	Copy of the Hardware security byte	101x 1011b	
BSB	Boot Status Byte	0FFh	
SSB	Software Security Byte	FFh	
	Copy of the Manufacturer Code	58h	ATMEL
	Copy of the Device ID #1: Family Code	D7h	C51 X2, Electrically Erasable
	Copy of the Device ID #2: memories size and type	ECh	AT89C51ID2 64KB
	Copy of the Device ID #3: name and revision	EFh	AT89C51ID2 64KB, Revision 0

After programming the part by ISP, the BSB must be cleared (00h) in order to allow the application to boot at 0000h.

The content of the Software Security Byte (SSB) is described in Table 90 and Table 93.

To assure code protection from a parallel access, the HSB must also be at the required level.

Table 91. Software Security Byte

Table 92.

7	6	5	4	3	2	1	0
-	-	-	-	•	-	LB1	LB0

Bit Number	Bit Mnemonic	Description
7	-	Reserved Do not clear this bit.
6	-	Reserved Do not clear this bit.
5	-	Reserved Do not clear this bit.
4	-	Reserved Do not clear this bit.
3	-	Reserved Do not clear this bit.
2	-	Reserved Do not clear this bit.
1-0	LB1-0	User Memory Lock Bits See Table 93

The two lock bits provide different levels of protection for the on-chip code and data, when programmed as shown to Table 93.

Table 93. Program Lock bits of the SSB

Program Lock Bits		Bits		
Security level	LB0	LB1	Protection description	
1	U	U	No program lock features enabled.	
2	Р	U	ISP programming of the Flash is disabled.	
3	Х	Р	Same as 2, also verify through ISP programming interface is disabled.	

Note:

U: unprogrammed or "one" level. P: programmed or "zero" level.

X: do not care

WARNING: Security level 2 and 3 should only be programmed after Flash and code

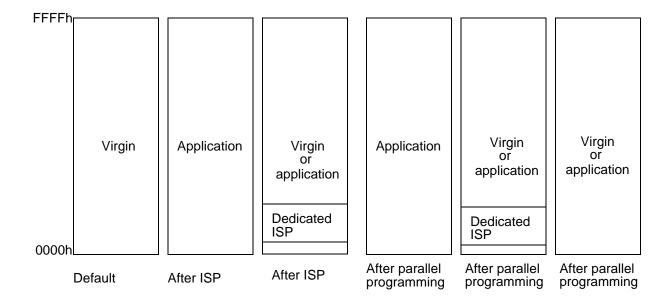
verification.

# Flash Memory Status

AT89C51ID2 parts are delivered in standard with the ISP rom bootloader.

After ISP or parallel programming, the possible contents of the Flash memory are summarized on the figure below:

Figure 45. Flash memory possible contents



# **Memory Organization**

When the  $\overline{\mathsf{EA}}$  pin high, the processor fetches instructions from internal program Flash. . If the  $\overline{EA}$  pin is tied low, all program memory fetches are from external memory.



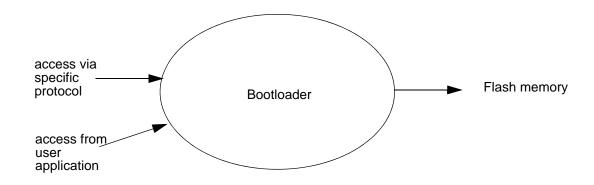


### **Bootloader Architecture**

### Introduction

The bootloader manages a communication according to a specific defined protocol to provide the whole access and service on Flash memory. Furthermore, all accesses and routines can be called from the user application.

Figure 46. Diagram Context Description



### **Acronyms**

ISP: In-System Programming

SBV: Software Boot Vector

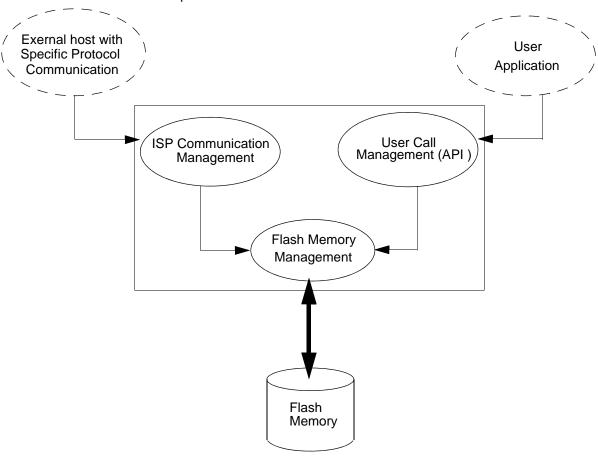
BSB: Boot Status Byte

SSB: Software Security Bit

HW: Hardware Byte

### **Functional Description**

Figure 47. Bootloader Functional Description



On the above diagram, the on chip bootloader processes are:

ISP Communication Management

The purpose of this process is to manage the communication and its protocol between the on-chip bootloader and a external device. The on-chip ROM implement a serial protocol (see section Bootloader Protocol). This process translate serial communication frame (UART) into flash memory acess (read, write, erase ...).

User Call Management

Several Application Program Interface (API) calls are available for use by an application program to permit selective erasing and programming of Flash pages. All calls are made through a common interface (API calls), included in the ROM bootloader. The programming functions are selected by setting up the microcontroller's registers before making a call to a common entry point (0xFFF0). Results are returned in the registers. The purpose on this process is to translate the registers values into internal Flash Memory Management.

· Flash Memory Management

This process manages low level access to flash memory (performs read and write access).





### **Bootloader Functionality**

Introduction

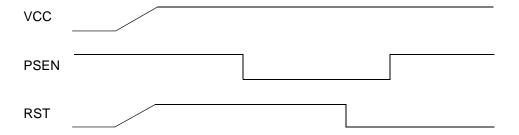
The bootloader can be activated by two means: Hardware conditions or regular boot process.

The Hardware conditions (EA = 1, PSEN = 0) during the Reset# falling edge force the on-chip bootloader execution. This allows an application to be built that will normally execute the end user's code but can be manually forced into default ISP operation.

As PSEN is a an output port in normal operating mode after reset, user application should take care to release PSEN after falling edge of reset signal. The hardware conditions are sampled at reset signal falling edge, thus they can be released at any time when reset input is low.

To ensure correct microcontroller startup, the PSEN pin should not be tied to ground during power-on (See Figure 48).

Figure 48. Hardware conditions typical sequence during power-on.



The on-chip bootloader boot process is shown Figure 49

	Purpose
Hardware Conditions	The Hardware Conditions force the bootloader execution whatever BLJB, BSB and SBV values.
	The Boot Loader Jump Bit forces the application execution.  BLJB = 0 => Boot loader execution.  BLJB = 1 => Application execution
BLJB	The BLJB is a fuse bit in the Hardware Byte.  That can be modified by hardware (programmer) or by software (API).
	Note: The BLJB test is perform by hardware to prevent any program execution

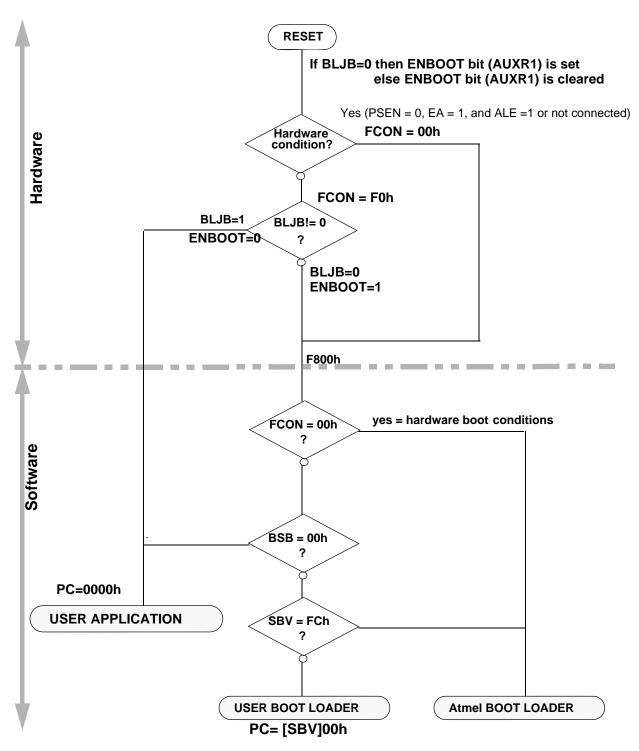
	Purpose
SBV	The Software Boot Vector contains the high address of custumer bootloader stored in the application.  SBV = FCh (default value) if no custumer bootloader in user Flash.
	Note: The costumer bootloader is called by JMP [SBV]00h instruction.





### **Boot Process**

Figure 49. Bootloader Process



### **ISP Protocol Description**

### **Physical Layer**

The UART used to transmit information has the following configuration:

· Character: 8-bit data

Parity: noneStop: 2 bits

Flow control: none

 Baudrate: autobaud is performed by the bootloader to compute the baudrate choosen by the host.

### **Frame Description**

The Serial Protocol is based on the Intel Hex-type records.

Intel Hex records consist of ASCII characters used to represent hexadecimal values and are summarized below

Figure 50. Intel Hex Type Frame

Record Mark ':'	Reclen	Load Offset	Record Type	Data or Info	Checksum
1-byte	1-byte	2-bytes	1-byte	n-bytes	1-byte

#### Record Mark:

Record Mark is the start of frame. This field must contain ':'.

Reclen:

Reclen specifies the number of bytes of information or data which follows the Record Type field of the record.

Load Offset:

Load Offset specifies the 16-bit starting load offset of the data bytes, therefore this field is used only for

Data Program Record (see Section "ISP Commands Summary").

Record Type:

Record Type specifies the command type. This field is used to interpret the remaining information within the frame. The encoding for all the current record types is described in Section "ISP Commands Summary".

Data/Info:

Data/Info is a variable length field. It consists of zero or more bytes encoded as pairs of hexadecimal digits. The meaning of data depends on the **Record Type**.

Checksum:

The two's complement of the 8-bit bytes that result from converting each pair of ASCII hexadecimal digits to one byte of binary, and including the **Reclen** field to and including the last byte of the **Data/Info** field. Therefore, the sum of all the ASCII pairs in a record after converting to binary, from the **Reclen** field to and including the **Checksum** field, is zero.





# **Functional Description**

### **Software Security Bits (SSB)**

The SSB protects any Flash access from ISP command.

The command "Program Software Security bit" can only write a higher priority level.

There are three levels of security:

• level 0: **NO SECURITY** (FFh)

This is the default level.

From level 0, one can write level 1 or level 2.

level 1: WRITE\_SECURITY (FEh )

For this level it is impossible to write in the Flash memory, BSB and SBV.

The Bootloader returns 'P' on write access.

From level 1, one can write only level 2.

level 2: RD\_WR\_SECURITY (FCh

The Post lead and write accesses to/from the Flash/EEPROM memory.

The Bootloader returns 'L' on read or write access.

Only a full chip erase in parallel mode (using a programmer) or ISP command can reset the software security bits.

From level 2, one cannot read and write anything.

Table 94. Software Security Byte Behavior

	Level 0	Level 1	Level 2
Flash/EEprom	Any access allowed	Read only access allowed	Any access not allowed
Fuse bit	Any access allowed	Read only access allowed	Any access not allowed
BSB & SBV	Any access allowed	Read only access allowed	Any access not allowed
SSB	Any access allowed	Write level 2 allowed	Read only access allowed
Manufacturer info	Read only access allowed	Read only access allowed	Read only access allowed
Bootloader info	Read only access allowed	Read only access allowed	Read only access allowed
Erase block	Allowed	Not allowed	Not allowed
Full chip erase	Allowed	Allowed	Allowed
Blank Check	Allowed	Allowed	Allowed

### **Full Chip Erase**

The ISP command "Full Chip Erase" erases all User Flash memory (fills with FFh) and sets some bytes used by the bootloader at their default values:

- BSB = FFh
- SBV = FCh
- SSB = FFh and finally erase the Software Security Bits

The Full Chip Erase does not affect the bootloader.

### **Checksum Error**

When a checksum error is detected send 'X' followed with CR&LF.





### **Flow Description**

### Overview

An initialization step must be performed after each Reset. After microcontroller reset, the bootloader waits for an autobaud sequence ( see section 'autobaud performance').

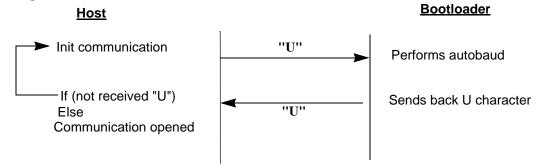
When the communication is initialized the protocol depends on the record type requested by the host.

FLIP, a software utility to implement ISP programming with a PC, is available from the Atmel the web site.

### **Communication Initialization**

The host initializes the communication by sending a 'U' character to help the bootloader to compute the baudrate (autobaud).

Figure 51. Initialization



### **Autobaud Performances**

The ISP feature allows a wide range of baud rates in the user application. It is also adaptable to a wide range of oscillator frequencies. This is accomplished by measuring the bit-time of a single bit in a received character. This information is then used to program the baud rate in terms of timer counts based on the oscillator frequency. The ISP feature requires that an initial character (an uppercase U) be sent to the AT89C51ID2 to establish the baud rate. Table show the autobaud capability.

Table 95. Autobaud Performances

Frequency (MHz) Baudrate (kHz)	1.8432	2	2.4576	3	3.6864	4	5	6	7.3728
2400	OK	ОК	OK	ОК	ОК	ОК	ОК	ОК	OK
4800	OK	-	OK	ОК	OK	ОК	OK	OK	OK
9600	ОК	-	OK	ОК	ОК	ОК	ОК	ОК	ОК
19200	ОК	-	OK	ОК	ОК	-	-	ОК	ОК
38400	-	-	OK		OK	-	OK	OK	ОК
57600	-	-	-	-	OK	-	-	-	ОК
115200	-	-	-	-	-	-	-	-	OK
Frequency (MHz) Baudrate (kHz)	8	10	11.0592	12	14.746	16	20	24	26.6
2400	OK	ОК	OK	ОК	ОК	ОК	OK	ОК	OK

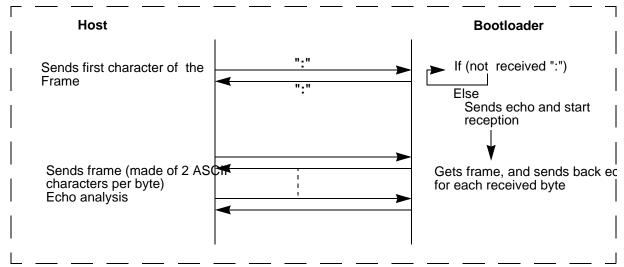
Table 95. Autobaud Performances (Continued)

Frequency (MHz) Baudrate (kHz)	1.8432	2	2.4576	3	3.6864	4	5	6	7.3728
4800	ОК	ОК	ОК	ОК	ОК	ОК	ОК	OK	ОК
9600	ОК	ОК	ОК	OK	OK	OK	OK	OK	OK
19200	OK	ОК	OK	OK	OK	OK	OK	OK	OK
38400	-	-	OK	OK	OK	OK	OK	OK	OK
57600	-	-	ОК	-	OK	OK	OK	OK	OK
115200	-	-	ОК	-	OK	-	-	-	-

# Command Data Stream Protocol

All commands are sent using the same flow. Each frame sent by the host is echoed by the bootloader.

Figure 52. Command Flow







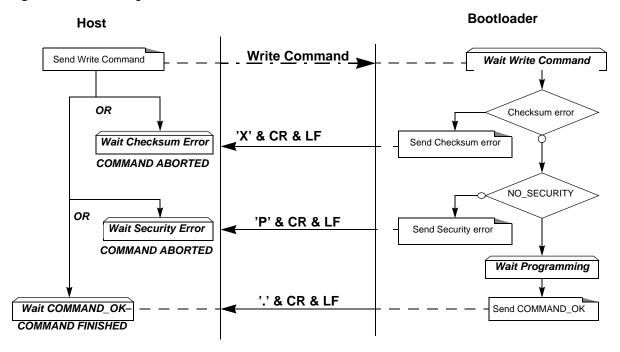
### **Write / Program Commands**

This flow is common to the following frames:

- Flash / Eeprom Programming Data Frame
- EOF or Atmel Frame (only Programming Atmel Frame)
- · Config Byte Programming Data Frame
- · Baud Rate Frame

### Description

Figure 53. Write/Program Flow



### Example

# Programming Data (write 55h at address 0010h in the Flash)

**HOST** : 01 0010 00 55 9A **BOOTLOADER** : 01 0010 00 55 9A . CR LF

Programming Atmel function (write SSB to level 2)

**HOST** : 02 0000 03 05 01 F5

**BOOTLOADER** : 02 0000 03 05 01 F5. CR LF

### Writing Frame (write BSB to 55h)

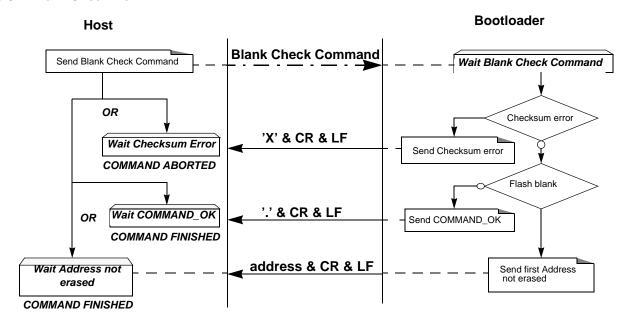
**HOST** : 03 0000 03 06 00 55 9F

**BOOTLOADER** : 03 0000 03 06 00 55 9F . CR LF

### **Blank Check Command**

### Description

Figure 54. Blank Check Flow



### Example

### Blank Check ok

**HOST** : 05 0000 04 0000 7FFF 01 78

**BOOTLOADER** : 05 0000 04 0000 7FFF 01 78 . CR LF

### Blank Check ko at address xxxx

**HOST** : 05 0000 04 0000 7FFF 01 78

**BOOTLOADER** : 05 0000 04 0000 7FFF 01 78 xxxx CR LF

### Blank Check with checksum error

**HOST** : 05 0000 04 0000 7FFF 01 70

**BOOTLOADER** : 05 0000 04 0000 7FFF 01 70 X CR LF CR LF

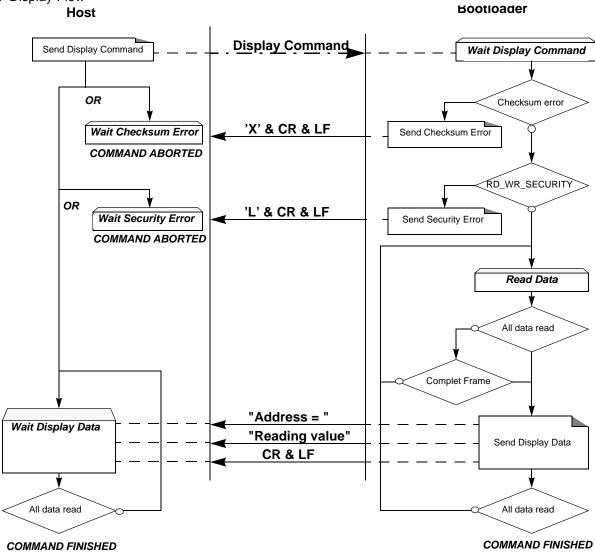




### **Display Data**

### Description

Figure 55. Display Flow



Note: The maximum size of block is 400h. To read more than 400h bytes, the Host must send a new command.

### Example

### Display data from address 0000h to 0020h

HOST	: 05 0000 04 0000 0020 00 D7	
BOOTLOADER	: 05 0000 04 0000 0020 00 D7	
BOOTLOADER	0000=data CR LF	(16 data)
BOOTLOADER	0010=data CR LF	(16 data)
BOOTLOADER	0020=data CR LF	( 1 data)

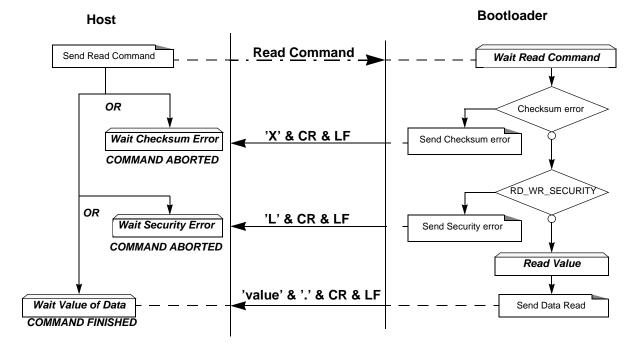
### **Read Function**

This flow is similar for the following frames:

- Reading Frame
- EOF Frame/ Atmel Frame (only reading Atmel Frame)

Description

Figure 56. Read Flow







### Example

### Read function (read SBV)

**HOST** : 02 0000 05 07 02 F0

**BOOTLOADER** : 02 0000 05 07 02 F0 Value . CR LF

Atmel Read function (read Bootloader version)

**HOST** : 02 0000 01 02 00 FB

**BOOTLOADER** : 02 0000 01 02 00 FB Value . CR LF

# **ISP Commands Summary**

Table 96. ISP Commands Summary

Command	Command Name	data[0]	data[1]	Command Effect
00h	Program Data			Program Nb Data Byte. Bootloader will accept up to 128 (80h) data bytes. The data bytes should be 128 byte page flash boundary.
			00h	Erase block0 (0000h-1FFFh)
			20h	Erase block1 (2000h-3FFFh)
		01h	40h	Erase block2 (4000h-7FFFh)
			80h	Erase block3 (8000h- BFFFh)
			C0h	Erase block4 (C000h- FFFFh)
		03h	00h	Hardware Reset
		04h	00h	Erase SBV & BSB
		05h	00h	Program SSB level 1
		0311	01h	Program SSB level 2
03h	Write Function	06h	00h	Program BSB (value to write in data[2])
		OOH	01h	Program SBV (value to write in data[2])
		07h	-	Full Chip Erase (This command needs about 6 sec to be executed)
			02h	Program Osc fuse (value to write in data[2])
		0Ah	04h	Program BLJB fuse (value to write in data[2])
			08h	Program X2 fuse (value to write in data[2])
			start address	Display Data
04h	Display Function		end address -> Display data	Blank Check
0411	Display I unction	Data[4] = 01h	-> Blank check Display EEPROMk	Display EEPROM data





Table 96. ISP Commands Summary (Continued)

Carrinary			1	
Command	Command Name	data[0]	data[1]	Command Effect
			00h	Manufacturer Id
		006	01h	Device Id #1
		00h	02h	Device Id #2
			03h	Device Id #3
			00h	Read SSB
054	Dood Function	07h	01h	Read BSB
05h	Read Function		02h	Read SBV
			06h	Read Extra Byte
		0Bh	00h	Read Hardware Byte
		051	00h	Read Device Boot ID1
		0Eh	01h	Read Device Boot ID2
		0Fh	00h	Read Bootloader Version
07h	Program EEPROM data			Program Nb EEProm Data Byte. Bootloader will accept up to 128 (80h) data bytes.

# **API Call Description**

Several Application Program Interface (API) calls are available for use by an application program to permit selective erasing and programming of Flash pages. All calls are made through a common interface, PGM\_MTP. The programming functions are selected by setting up the microcontroller's registers before making a call to PGM\_MTP at FFF0h. Results are returned in the registers.

When several bytes have to be programmed, it is highly recommended to use the Atmel API "PROGRAM DATA PAGE" call. Indeed, this API call writes up to 128 bytes in a single command.

All routines for software access are provided in the C Flash driver available on Atmel web site.

The API calls description and arguments are shown in Table

Table 97. API Call Summary

Command	R1	Α	DPTR0	DPTR1	Returned Value	Command Effect												
READ MANUF ID	00h	XXh	0000h	XXh	ACC = Manufacturer Id	Read Manufacturer identifier												
READ DEVICE ID1	00h	XXh	0001h	XXh	ACC = Device Id 1	Read Device identifier 1												
READ DEVICE ID2	00h	XXh	0002h	XXh	ACC = Device Id 2	Read Device identifier 2												
READ DEVICE ID3	00h	XXh	0003h	XXh	ACC = Device Id 3	Read Device identifier 3												
			DPH = 00h			Erase block 0												
			DPH = 20h			Erase block 1												
			DPH = 40h			Erase block 2												
ERASE BLOCK	01h	XXh	Address of byte to program	00h	ACC = DPH	Program one Data Byte in user Flash												
			XXh			Erase Software boot vector and boot status byte. (SBV = FCh and BSB = FFh)												
			DPH = 00h DPL = 00h			Set SSB level 1												
PROGRAM SSB		XXh	DPH = 00h DPL = 01h	00h	ACC = SSB value	Set SSB level 2												
PROGRAM 55B	05h	XXII	DPH = 00h DPL = 10h		OON	oon	OUT	oon	OOH	0011	OOH	OOH	0011	OON	OOH	OOH	0011	ACC = 33b value
			DPH = 00h DPL = 11h			Set SSB level 1												
PROGRAM BSB	06h	New BSB value	0000h	XXh	none	Program boot status byte												
PROGRAM SBV	06h	New SBV value	0001h	XXh	none	Program software boot vector												
READ SSB	07h	XXh	0000h	XXh	ACC = SSB	Read Software Security Byte												
READ BSB	07h	XXh	0001h	XXh	ACC = BSB	Read Boot Status Byte												
READ SBV	07h	XXh	0002h	XXh	ACC = SBV	Read Software Boot Vector												





Table 97. API Call Summary (Continued)

Command	R1	Α	DPTR0	DPTR1	Returned Value	Command Effect
PROGRAM DATA PAGE	09h	Number of byte to program	Address of the first byte to program in the Flash memory	Address in XRAM of the first data to program	ACC = 0: DONE	Program up to 128 bytes in user Flash. Remark: number of bytes to program is limited such as the Flash write remains in a single 128 bytes page. Hence, when ACC is 128, valid values of DPL are 00h, or, 80h.
PROGRAM X2 FUSE	0Ah	Fuse value 00h or 01h	0008h	XXh	none	Program X2 fuse bit with ACC
PROGRAM BLJB FUSE	0Ah	Fuse value 00h or 01h	0004h	XXh	none	Program BLJB fuse bit with ACC
READ HSB	0Bh	XXh	XXXXh	XXh	ACC = HSB	Read Hardware Byte
READ BOOT ID1	0Eh	XXh	DPL = 00h	XXh	ACC = ID1	Read boot ID1
READ BOOT ID2	0Eh	XXh	DPL = 01h	XXh	ACC = ID2	Read boot ID2
READ BOOT VERSION	0Fh	XXh	XXXXh	XXh	ACC = Boot_Version	Read bootloader version

### **Electrical Characteristics**

### **Absolute Maximum Ratings**

I = industrial	40°C to 85°C
Storage Temperature	65°C to + 150°C
Voltage on V <sub>CC</sub> to V <sub>SS</sub> (standard voltage).	0.5V to + 6.5V
Voltage on V <sub>CC</sub> to V <sub>SS</sub> (low voltage)	0.5V to + 4.5V
Voltage on Any Pin to V <sub>SS</sub>	
Power Dissipation	1 W <sup>(2)</sup>

Note:

Stresses at or above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Power dissipation is based on the maximum allowable die temperature and the thermal resistance of the package.

# **DC Parameters**

 $T_A = -40^{\circ}C \text{ to } +85^{\circ}C; V_{SS} = 0V;$ 

 $V_{CC}$  =2.7V to 5.5V and F = 0 to 40 MHz (both internal and external code execution)

 $V_{CC}$  =4.5V to 5.5V and F = 0 to 60 MHz (internal code execution only)

Symbol	Parameter	Min	Тур	Max	Unit	Test Conditions
V <sub>IL</sub>	Input Low Voltage	-0.5		0.2 V <sub>CC</sub> - 0.1	V	
V <sub>IH</sub>	Input High Voltage except RST, XTAL1	0.2 V <sub>CC</sub> + 0.9		V <sub>CC</sub> + 0.5	V	
V <sub>IH1</sub>	Input High Voltage RST, XTAL1	0.7 V <sub>CC</sub>		V <sub>CC</sub> + 0.5	V	
V <sub>OL</sub>	Output Low Voltage, ports 1, 2, 3, 4 <sup>(6)</sup>			0.3 0.45 1.0	V V V	$\begin{split} &V_{CC} = 4.5 V \text{ to } 5.5 V \\ &I_{OL} = 100 \ \mu\text{A}^{(4)} \\ &I_{OL} = 1.6 \ \text{mA}^{(4)} \\ &I_{OL} = 3.5 \ \text{mA}^{(4)} \end{split}$
				0.45	V	$V_{CC} = 2.7V \text{ to } 5.5V$ $I_{OL} = 0.8 \text{ mA}^{(4)}$
V <sub>OL1</sub>	Output Low Voltage, port 0, ALE, PSEN (6)			0.3 0.45 1.0	V V V	$\begin{split} &V_{CC} = 4.5 V \text{ to } 5.5 V \\ &I_{OL} = 200 \ \mu\text{A}^{(4)} \\ &I_{OL} = 3.2 \ \text{mA}^{(4)} \\ &I_{OL} = 7.0 \ \text{mA}^{(4)} \end{split}$
				0.45	V	$V_{CC} = 2.7V \text{ to } 5.5V$ $I_{OL} = 1.6 \text{ mA}^{(4)}$
$V_{OH}$	Output High Voltage, ports 1, 2, 3, 4	V <sub>CC</sub> - 0.3 V <sub>CC</sub> - 0.7 V <sub>CC</sub> - 1.5			V V V	$V_{CC} = 5V \pm 10\%$ $I_{OH} = -10 \ \mu A$ $I_{OH} = -30 \ \mu A$ $I_{OH} = -60 \ \mu A$
		0.9 V <sub>CC</sub>			V	$V_{CC} = 2.7V \text{ to } 5.5V$ $I_{OH} = -10 \ \mu\text{A}$





 $T_A = -40^{\circ}C \text{ to } +85^{\circ}C; V_{SS} = 0V;$ 

 $V_{CC}$  =2.7V to 5.5V and F = 0 to 40 MHz (both internal and external code execution)

 $V_{CC}$  =4.5V to 5.5V and F = 0 to 60 MHz (internal code execution only) (Continued)

Symbol	Parameter	Min	Тур	Max	Unit	Test Conditions
V <sub>OH1</sub>	Output High Voltage, port 0, ALE, PSEN	V <sub>CC</sub> - 0.3 V <sub>CC</sub> - 0.7 V <sub>CC</sub> - 1.5			V V V	$V_{CC} = 5V \pm 10\%$ $I_{OH} = -200 \mu A$ $I_{OH} = -3.2 \text{ mA}$ $I_{OH} = -7.0 \text{ mA}$
		0.9 V <sub>CC</sub>			V	$V_{CC} = 2.7V \text{ to } 5.5V$ $I_{OH} = -10  \mu\text{A}$
R <sub>RST</sub>	RST Pull-down Resistor	50	200 <sup>(5)</sup>	250	kΩ	
I <sub>IL</sub>	Logical 0 Input Current ports 1, 2, 3, 4 and 5			-50	μА	V <sub>IN</sub> = 0.45V
I <sub>LI</sub>	Input Leakage Current			±10	μΑ	$0.45 \text{V} < \text{V}_{\text{IN}} < \text{V}_{\text{CC}}$
I <sub>TL</sub>	Logical 1 to 0 Transition Current, ports 1, 2, 3, 4			-650	μА	V <sub>IN</sub> = 2.0V
C <sub>IO</sub>	Capacitance of I/O Buffer			10	pF	$F_C = 3 \text{ MHz}$ $T_A = 25^{\circ}\text{C}$
I <sub>PD</sub>	Power-down Current		75	150	μА	$2.7 < V_{CC} < 5.5V^{(3)}$
I <sub>CCOP</sub>	Power Supply Current on normal mode			0.4 x Frequency (MHz) + 5	mA	$V_{CC} = 5.5V^{(1)}$
I <sub>CCIDLE</sub>	Power Supply Current on idle mode			0.3 x Frequency (MHz) + 5	mA	$V_{CC} = 5.5V^{(2)}$
I <sub>CCWRITE</sub>	Power Supply Current on flash or EEdata write			0.8 x Frequency (MHz) + 15	mA	V <sub>CC</sub> = 5.5V
t <sub>WRITE</sub>	Flash or EEdata programming time		7	10	ms	2.7 < V <sub>CC &lt;</sub> 5.5V

Notes:

- Operating I<sub>CC</sub> is measured with all output pins disconnected; XTAL1 driven with T<sub>CLCH</sub>, T<sub>CHCL</sub> = 5 ns (see Figure 60), V<sub>IL</sub> = V<sub>SS</sub> + 0.5V, V<sub>IH</sub> = V<sub>CC</sub> 0.5V; XTAL2 N.C.; EA = RST = Port 0 = V<sub>CC</sub>. I<sub>CC</sub> would be slightly higher if a crystal oscillator used (see Figure 57).
- 2. Idle  $I_{CC}$  is measured with all out<u>put</u> pins disconnected; XTAL1 driven with  $T_{CLCH}$ ,  $T_{CHCL} = 5$  ns,  $V_{IL} = V_{SS} + 0.5$ V,  $V_{IH} = V_{CC} 0.5$ V; XTAL2 N.C; Port  $0 = V_{CC}$ ; EA = RST =  $V_{SS}$  (see Figure 58).
- 3. Power-down I<sub>CC</sub> is measured with all output pins disconnected; EA = V<sub>CC</sub>, PORT 0 = V<sub>CC</sub>; XTAL2 NC.; RST = V<sub>SS</sub> (see Figure 59).
- 4. Capacitance loading on Ports 0 and 2 may cause spurious noise pulses to be superimposed on the V<sub>OLS</sub> of ALE and Ports 1 and 3. The noise is due to external bus capacitance discharging into the Port 0 and Port 2 pins when these pins make 1 to 0 transitions during bus operation. In the worst cases (capacitive loading 100 pF), the noise pulse on the ALE line may exceed 0.45V with maxi V<sub>OL</sub> peak 0.6V. A Schmitt Trigger use is not necessary.
- 5. Typical values are based on a limited number of samples and are not guaranteed. The values listed are at room temperature and 5V.
- 6. Under steady state (non-transient) conditions, I<sub>OL</sub> must be externally limited as follows:

Maximum  $I_{OL}$  per port pin: 10 mA

Maximum I<sub>OL</sub> per 8-bit port:

Port 0: 26 mA

Ports 1, 2 and 3: 15 mA

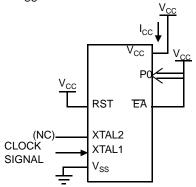
Maximum total I<sub>OL</sub> for all output pins: 71 mA

If  $I_{OL}$  exceeds the test condition,  $V_{OL}$  may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.

AT89C51ID2

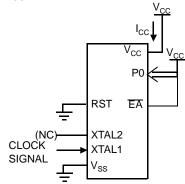
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Figure 57.  $I_{CC}$  Test Condition, Active Mode  $\underline{V_{CC}}$ 



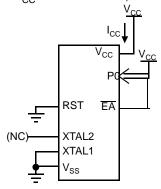
All other pins are disconnected.

Figure 58.  $I_{CC}$  Test Condition, Idle Mode



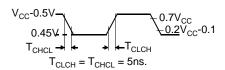
All other pins are disconnected.

Figure 59. I<sub>CC</sub> Test Condition, Power-down Mode



All other pins are disconnected.

Figure 60. Clock Signal Waveform for  $\rm I_{\rm CC}$  Tests in Active and Idle Modes



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### **AC Parameters**

# Explanation of the AC Symbols

Each timing symbol has 5 characters. The first character is always a "T" (stands for time). The other characters, depending on their positions, stand for the name of a signal or the logical status of that signal. The following is a list of all the characters and what they stand for.

Example: T<sub>AVLL</sub> = Time for Address Valid to ALE Low.

 $T_{IIPI}$  = Time for ALE Low to  $\overline{PSEN}$  Low.

(Load Capacitance for port 0, ALE and PSEN = 100 pF; Load Capacitance for all other outputs = 80 pF.)

Table 98 Table 101, and Table 104 give the description of each AC symbols.

Table 99, Table 100, Table 102 and Table 105 gives the range for each AC parameter.

Table 99, Table 100 and Table 106 give the frequency derating formula of the AC parameter for each speed range description. To calculate each AC symbols. take the x value in the correponding column (-M or -L) and use this value in the formula.

Example:  $T_{LLIU}$  for -M and 20 MHz, Standard clock.

x = 35 nsT 50 ns

 $T_{CCIV} = 4T - x = 165 \text{ ns}$ 

# External Program Memory Characteristics

Table 98. Symbol Description

Symbol	Parameter
Т	Oscillator clock period
T <sub>LHLL</sub>	ALE pulse width
T <sub>AVLL</sub>	Address Valid to ALE
T <sub>LLAX</sub>	Address Hold After ALE
T <sub>LLIV</sub>	ALE to Valid Instruction In
T <sub>LLPL</sub>	ALE to PSEN
T <sub>PLPH</sub>	PSEN Pulse Width
T <sub>PLIV</sub>	PSEN to Valid Instruction In
T <sub>PXIX</sub>	Input Instruction Hold After PSEN
T <sub>PXIZ</sub>	Input Instruction Float After PSEN
T <sub>AVIV</sub>	Address to Valid Instruction In
T <sub>PLAZ</sub>	PSEN Low to Address Float

Table 99. AC Parameters for a Fix Clock

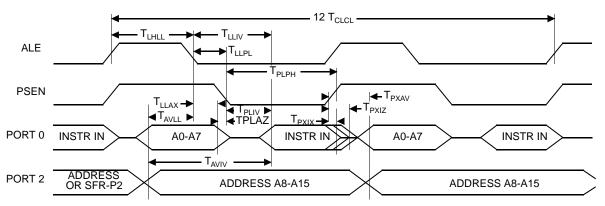
Symbol	-м			-L	Units
	Min	Max	Min	Max	
Т	25		25		ns
T <sub>LHLL</sub>	35		35		ns
T <sub>AVLL</sub>	5		5		ns
T <sub>LLAX</sub>	5		5		ns
T <sub>LLIV</sub>		n 65		65	ns
T <sub>LLPL</sub>	5		5		ns
T <sub>PLPH</sub>	50		50		ns
T <sub>PLIV</sub>		30		30	ns
T <sub>PXIX</sub>	0		0		ns
T <sub>PXIZ</sub>		10		10	ns
T <sub>AVIV</sub>		80		80	ns
T <sub>PLAZ</sub>	AC Parameters for a Veriable C	10		10	ns

**Table 100.** AC Parameters for a Variable Clock

Symbol	Туре	Standard Clock	X2 Clock	X parameter for -M range	X parameter for -L range	Units
T <sub>LHLL</sub>	Min	2 T - x	T - x	15	15	ns
T <sub>AVLL</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>LLAX</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>LLIV</sub>	Max	4 T - x	2 T - x	35	35	ns
T <sub>LLPL</sub>	Min	T - x	0.5 T - x	15	15	ns
T <sub>PLPH</sub>	Min	3 T - x	1.5 T - x	25	25	ns
T <sub>PLIV</sub>	Max	3 T - x	1.5 T - x	45	45	ns
T <sub>PXIX</sub>	Min	х	Х	0	0	ns
T <sub>PXIZ</sub>	Max	T - x	0.5 T - x	15	15	ns
T <sub>AVIV</sub>	Max	5 T - x	2.5 T - x	45	45	ns
T <sub>PLAZ</sub>	Max	х	х	10	10	ns



# **External Program Memory** Read Cycle



# **External Data Memory** Characteristics

Table 101. Symbol Description

Symbol	Parameter
T <sub>RLRH</sub>	RD Pulse Width
T <sub>WLWH</sub>	WR Pulse Width
T <sub>RLDV</sub>	RD to Valid Data In
T <sub>RHDX</sub>	Data Hold After RD
$T_{RHDZ}$	Data Float After RD
T <sub>LLDV</sub>	ALE to Valid Data In
T <sub>AVDV</sub>	Address to Valid Data In
T <sub>LLWL</sub>	ALE to WR or RD
T <sub>AVWL</sub>	Address to WR or RD
$T_{QVWX}$	Data Valid to WR Transition
$T_{QVWH}$	Data Set-up to WR High
$T_{WHQX}$	Data Hold After WR
T <sub>RLAZ</sub>	RD Low to Address Float
T <sub>WHLH</sub>	RD or WR High to ALE high

Table 102. AC Parameters for a Fix Clock

	-M			-L	
Symbol	Min	Max	Min	Max	Units
T <sub>RLRH</sub>	125		125		ns
$T_{WLWH}$	125		125		ns
$T_{RLDV}$		95		95	ns
$T_RHDX$	0		0		ns
$T_{RHDZ}$		25		25	ns
$T_LLDV$		155		155	ns
T <sub>AVDV</sub>		160		160	ns
$T_LLWL$	45	105	45	105	ns
T <sub>AVWL</sub>	70		70		ns
$T_{QVWX}$	5		5		ns
$T_{QVWH}$	155		155		ns
$T_{WHQX}$	10		10		ns
T <sub>RLAZ</sub>	0		0		ns
T <sub>WHLH</sub>	5	45	5	45	ns

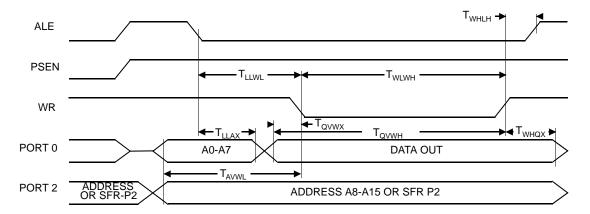
Table 103. AC Parameters for a Variable Clock

Symbol	Туре	Standard Clock	X2 Clock	X parameter for -M range	X parameter for -L range	Units
T <sub>RLRH</sub>	Min	6 T - x	3 T - x	25	25	ns
T <sub>WLWH</sub>	Min	6 T - x	3 T - x	25	25	ns
T <sub>RLDV</sub>	Max	5 T - x	2.5 T - x	30	30	ns
T <sub>RHDX</sub>	Min	х	Х	0	0	ns
T <sub>RHDZ</sub>	Max	2 T - x	T - x	25	25	ns
T <sub>LLDV</sub>	Max	8 T - x	4T -x	45	45	ns
T <sub>AVDV</sub>	Max	9 T - x	4.5 T - x	65	65	ns
T <sub>LLWL</sub>	Min	3 T - x	1.5 T - x	30	30	ns
T <sub>LLWL</sub>	Max	3 T + x	1.5 T + x	30	30	ns
T <sub>AVWL</sub>	Min	4 T - x	2 T - x	30	30	ns
T <sub>QVWX</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>QVWH</sub>	Min	7 T - x	3.5 T - x	20	20	ns
T <sub>WHQX</sub>	Min	T - x	0.5 T - x	15	15	ns
T <sub>RLAZ</sub>	Max	х	Х	0	0	ns
T <sub>WHLH</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>WHLH</sub>	Max	T + x	0.5 T + x	20	20	ns

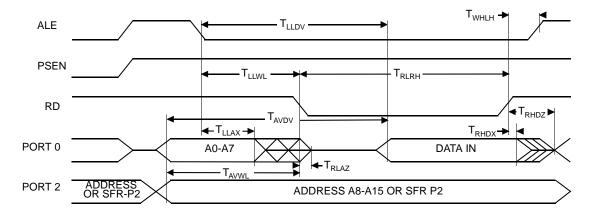




## **External Data Memory Write** Cycle



## **External Data Memory Read Cycle**



## **Serial Port Timing - Shift Register Mode**

Table 104. Symbol Description

Symbol	Parameter
T <sub>XLXL</sub>	Serial port clock cycle time
T <sub>QVHX</sub>	Output data set-up to clock rising edge
T <sub>XHQX</sub>	Output data hold after clock rising edge
T <sub>XHDX</sub>	Input data hold after clock rising edge
T <sub>XHDV</sub>	Clock rising edge to input data valid

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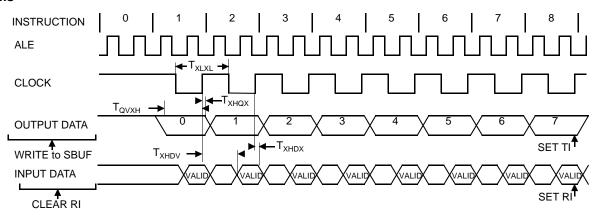
Table 105. AC Parameters for a Fix Clock

	-M				
Symbol	Min	Max	Min	Max	Units
T <sub>XLXL</sub>	300		300		ns
T <sub>QVHX</sub>	200		200		ns
T <sub>XHQX</sub>	30		30		ns
T <sub>XHDX</sub>	0		0		ns
T <sub>XHDV</sub>		117		117	ns

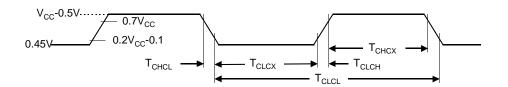
Table 106. AC Parameters for a Variable Clock

Symbol	Туре	Standard Clock	X2 Clock	X Parameter For -M Range	X Parameter For -L Range	Units
T <sub>XLXL</sub>	Min	12 T	6 T			ns
T <sub>QVHX</sub>	Min	10 T - x	5 T - x	50	50	ns
T <sub>XHQX</sub>	Min	2 T - x	T - x	20	20	ns
T <sub>XHDX</sub>	Min	х	х	0	0	ns
T <sub>XHDV</sub>	Max	10 T - x	5 T- x	133	133	ns

# Shift Register Timing Waveforms



# **External Clock Drive** Waveforms



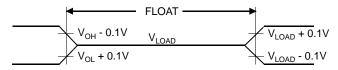




# AC Testing Input/Output Waveforms

AC inputs during testing are driven at  $V_{CC}$  - 0.5 for a logic "1" and 0.45V for a logic "0". Timing measurement are made at  $V_{IH}$  min for a logic "1" and  $V_{IL}$  max for a logic "0".

#### **Float Waveforms**

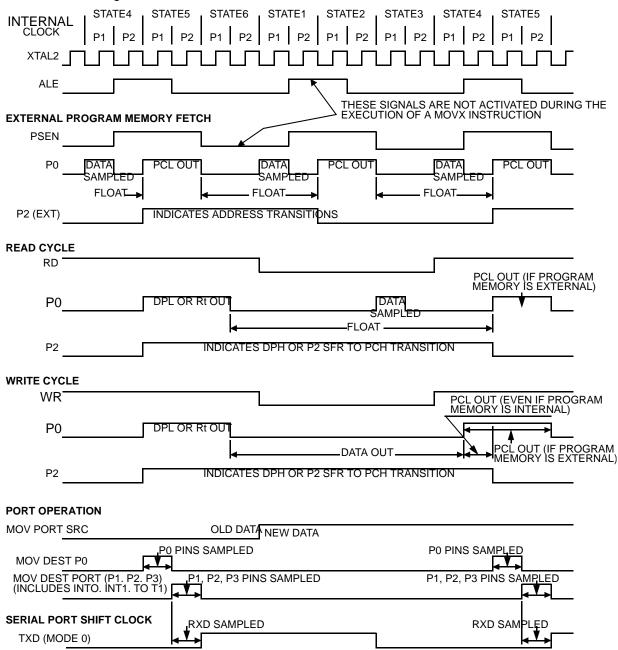


For timing purposes as port pin is no longer floating when a 100 mV change from load voltage occurs and begins to float when a 100 mV change from the loaded  $V_{OH}/V_{OL}$  level occurs.  $I_{OL}/I_{OH} \ge \pm$  20 mA.

#### **Clock Waveforms**

Valid in normal clock mode. In X2 mode XTAL2 must be changed to XTAL2/2.

Figure 61. Internal Clock Signals



This diagram indicates when signals are clocked internally. The time it takes the signals to propagate to the pins, however, ranges from 25 to 125 ns. This propagation delay is dependent on variables such as temperature and pin loading. Propagation also varies from output to output and component. Typically though ( $T_A = 25^{\circ}$ C fully loaded)  $\overline{RD}$  and  $\overline{WR}$  propagation delays are approximately 50 ns. The other signals are typically 85 ns. Propagation delays are incorporated in the AC specifications.



# **Ordering Information**

Table 107. Possible Order Entries

Part Number	Supply Voltage	Temperature Range	Package	Packing	Product Marking	
AT89C51ID2-SLSIM	2.7V-5.5V	Industrial	PLCC44	Stick	AT89C51ID2-IM	
AT89C51ID2-RLTIM		2.7\/.5.5\/	ilidustilai	VQFP44	Tray	AT89C51ID2-IM
AT89C51ID2-SLSUM		Industrial &	PLCC44	Stick	AT89C51ID2-UM	
AT89C51ID2-RLTUM		Green	VQFP44	Tray	AT89C51ID2-UM	

Change Log for 4289A - 09/03 to 4289B - 12/03

1. Improvement of explanations throughout the document.

4289B - 12/03 to 4289C - 11/05

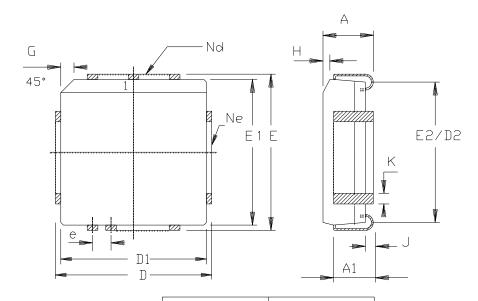
1. Added 'Industrial & Green" product versions.





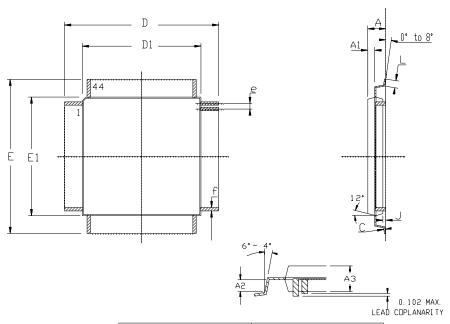
# **Packaging Information**

# PLCC44



	M	IM ·	ΙN	CH
А	4. 20	4. 57	. 165	. 180
A1	2. 29	3. 04	. 090	. 120
D	17.40	17.65	. 685	. 695
D1	16. 44	16.66	. 647	. 656
D2	14. 99	16.00	. 590	. 630
E	17.40	17.65	. 685	. 695
E1	16. 44	16.66	. 647	. 656
E2	14. 99	16.00	. 590	. 630
е	1.27	BSC	. 050	BSC
G	1.07	1.22	. 042	. 048
Н	1.07	1.42	. 042	. 056
J	0. 51	-	. 020	-
К	0.33	0. 53	. 013	. 021
Nd	1 1		1	1
Ne	1	1	1	1
Р	KG STD	0.0		

# VQFP44



	М	М	IN	СН
	Min	Max	Min	Max
А	_	1.60	_	. 063
A1	0.	64 REF	. 0	25 REF
A2	0.	64 REF	. 025 REF	
А3	1.35	1.45	. 053	. 057
D	11.90	12.10	. 468	. 476
D1	9. 90	10.10	. 390	. 398
E	11.90	12.10	. 468	. 476
E1	9. 90	10.10	. 390	. 398
J	0.05	_	. 002	-
L	0. 45	0.75	. 018	. 030
е	0.8	0 BSC	. 03	15 BSC
f	0.35 BSC		. 01	4 BSC



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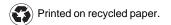
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