### MAX626/7/8-TSC426/7/8

### **Dual Power MOSFET Drivers**

#### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage V <sub>DD</sub> to GND	. +20V	CERDIP (derate above 70°C
Input Voltage V <sub>DD</sub> +0.3V to GNI	D -0.3V	by 8.0mW/°C) 640mW
Package Dissipation		Maximum Chip Temperature +150°C
Plastic DIP (derate above 70°C		Storage Temperature55°C to +160°C
by 6.25mW/°C)	500mW	Lead Temperature (10 seconds) +300°C
Small Outline (derate above 70°C		
by 5.88mW/°C)	450mW	

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### ELECTRICAL CHARACTERISTICS—MAX626/7/8 and TSC426/7/8

(V<sub>DD</sub> = 4.5V to 18V, Over Temperature unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
Logic 1 Input Voltage	V <sub>IH</sub>			2.4			V	
Logic 0 Input Voltage	V <sub>IL</sub>					0.8	٧	
Input Current	I <sub>IN</sub>	V <sub>IN</sub> = 0V to 18V, T <sub>A</sub> = 25°C V <sub>IN</sub> = 0V to 18V		-1 -10		1 10	μΑ	
Output High Voltage	V <sub>OH</sub>	No Load		V <sub>DD</sub> -25			V	
Output Low Voltage	V <sub>OL</sub>	No Load				+25	mV	
Output Resistance		V <sub>DD</sub> = 18V, T <sub>A</sub> = 25°C I <sub>LOAD</sub> = 10mA	MAX626/7/8 TSC426/7/8		4 10	15 15		
	(Note 1)	V <sub>DD</sub> = 18V Over Temp I <sub>LOAD</sub> = 10mA	MAX626/7/8 TSC426/7/8		6 13	20 20		
	R <sub>OUT</sub>	V <sub>DD</sub> = 18V, T <sub>A</sub> = 25°C I <sub>LOAD</sub> = 10mA	MAX626/7/8 TSC426/7/8	į	4 6	10 10	Ω	
	(Note 2)	V <sub>DD</sub> = 18V Over Temp I <sub>LOAD</sub> = 10mA	MAX626/7/8 TSC426/7/8		6 8	15 15		
Peak Output Current	l <sub>PK</sub>	MAX626/7/8, V <sub>DD</sub> = 18V TSC426/7/8, V <sub>DD</sub> = 18V			2 1.5		A	
Power Supply Current	I <sub>SUPP</sub>	V <sub>IN</sub> = +3V both inputs, T <sub>A</sub> = 25°C V <sub>IN</sub> = +3V both inputs V <sub>IN</sub> = +0V both inputs, T <sub>A</sub> = 25°C V <sub>IN</sub> = +0V both inputs				8 12 0.4 0.6	mA	
Bio Time ts		T <sub>A</sub> = 25°C			20	30	T	
Rise Time (Note 3)	Over Temp	MAX626/7/8 TSC426/7/8		25 25	40 60			
Fall Time	t <sub>F</sub> (Note 3)	T <sub>A</sub> = 25°C Over Temp			20 25	30 40		
Delay Time t <sub>D1</sub> (Note 3)	T <sub>A</sub> = 25°C	MAX626/7/8 TSC426/7/8		20 20	30 40			
		Over Temp	MAX626/7/8 TSC426/7/8		25 25	40 60	ns ns	
Delay Time t <sub>D2</sub> (Note 3)	T <sub>A</sub> = 25°C	MAX626/7/8 TSC426/7/8		25 25	50 75			
		Over Temp	MAX626/7/8 TSC426/7/8		30 30	60 120		

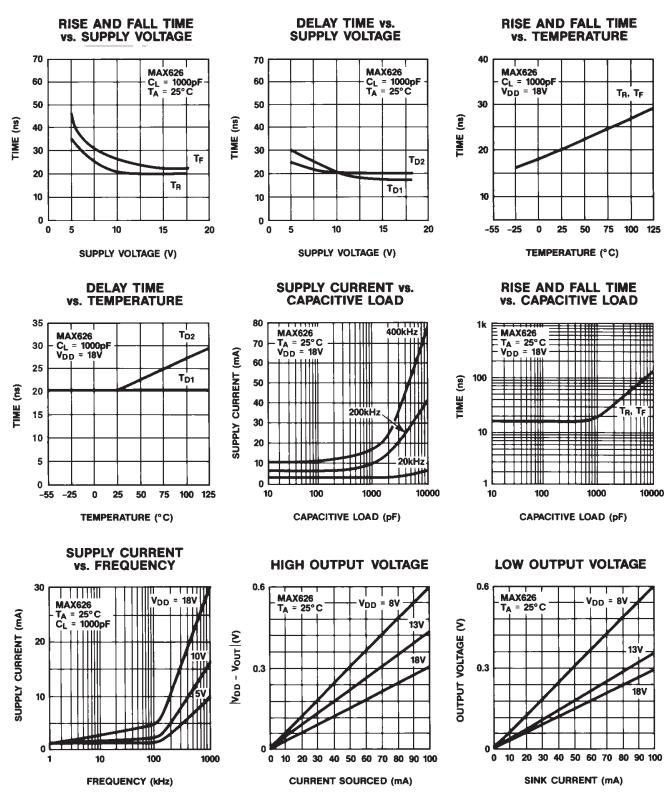
**Note 1:**  $V_{IN}$  = 0.8V for inverting stages,  $V_{IN}$  = 2.4V for non-inverting stages. **Note 2:**  $V_{IN}$  = 2.4V for inverting stages,  $V_{IN}$  = 0.8V for non-inverting stages.

Note 3: Switching times guaranteed by design, not tested.

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## MAX626/7/8-TSC426/7/8 Dual Power MOSFET Drivers

### **Typical Operating Characteristics**



## MAX626/7/8-TSC426/7/8 Dual Power MOSFET Drivers

### **Application Hints**

The MAX626/7/8 have an easy to drive input, however, the input must never be allowed to stay between  $V_{\rm IH}$  and  $V_{\rm IL}$  for more than 50ns. Unused inputs should always be connected to ground to minimize supply current. Drivers can be paralleled on the MAX626 or MAX627 by tying both inputs together and both outputs together.

Supply bypassing and grounding are extremely important with the MAX626/7/8 as the peak supply and output currents can be greater than 2 Amps. Ground drops are a form of negative feedback with inverters and, hence, will degrade the delay and transition time of the MAX626/8. Ringing may also be a problem with large dV/dt and/or large AC currents.

Suggested bypass capacitors are a  $4.7\mu\text{F}$  (low ESR) capacitor in parallel with a  $0.1\mu\text{F}$  ceramic capacitor, mounted as close as possible to the MAX626/7/8. Use a ground plane if possible, or separate ground returns for inputs and outputs. Ringing can be minimized with a 5-20 $\Omega$  resistor in series with the output but this may degrade output transition time.

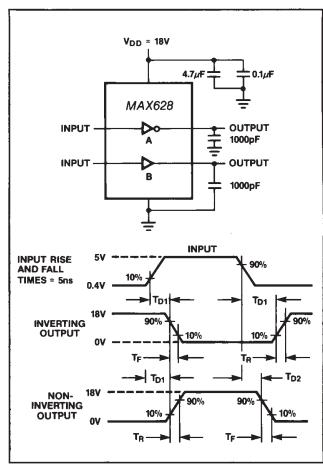


Figure 1. Inverting and Non-inverting Test Circuit

### **Power Dissipation**

Power dissipation of the MAX626/7/8 consists of:

- 1) Input inverter losses:
- 2) Crowbar current through the output devices;
- 3) Output current (either capacitive or resistive).

The sum of these must be kept below the maximum power dissipation limit.

The DC input inverter losses are 0.4mA when both inputs are low and 4mA when both inputs are high.

The Crowbar current through an output device making a transition is approximately 100mA for a few nanoseconds. This is a small portion of the total supply current, except for high switching frequencies or a small load capacitance (100pF).

The MAX626/7/8 power dissipation when driving a ground referenced resistive load is:

$$P = D \times R_{ON(MAX)} \times I_{LOAD}^2$$

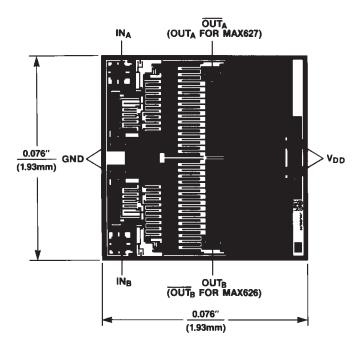
where D is the percentage of time the MAX626/7/8 output pulls high,  $R_{ON(MAX)}$  is the maximum on resistance of the inverting sections of MAX626/8 with  $V_{IN} = V_{IL}$  for non-inverting sections), the  $I_{LOAD}$  is the load current of the MAX626/7/8.

For capacitive loads, the power dissipation is:

$$P = C_{LOAD} \times V_{DD}^2 \times FREQ$$

where  $C_{\rm LOAD}$  is the capacitive load,  $V_{\rm DD}$  is the MAX626/7/8 supply voltage, and FREQ is the toggle frequency.

### Chip Topography



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# MAX626/7/8-TSC426/7/8 Dual Power MOSFET Drivers



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