## **ABSOLUTE MAXIMUM RATINGS**

(Voltages referenced to GND.)

V <sub>CC</sub>	+7V
RĒ, DE, DE, DI, TXD	0.3V to $(V_{CC} + 0.3V)$
A, B (Note 1) (MAX13442E/MAX13444E)	±80V
A, B (Note 1) (MAX13443E)	±60V
RO	0.3V to $(V_{CC} + 0.3V)$
Short-Circuit Duration (RO, A, B)	Continuous

Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
SO (derate 5.9mW/°C above +70°C)	471mW
Operating Temperature Range40°C	to +125°C
Storage Temperature Range65°C	to +150°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	

**Note 1:** During normal operation, a termination resistor must be connected between A and B in order to guarantee overvoltage protection up to the absolute maximum rating of this device. When not in operation, these devices can withstand fault voltages up to the maximum rating without a termination resistor and will not be damaged.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## DC ELECTRICAL CHARACTERISTICS

(V<sub>CC</sub> = +4.75V to +5.25V,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at V<sub>CC</sub> = +5V and  $T_A = +25^{\circ}$ C.)

PARAMETER	SYMBOL		CONDITIONS	MIN	ТҮР	MAX	UNITS
DRIVER							
Differential Driver Output		Figure 1,	$R_L = 100\Omega$	2		VCC	V
Differential Driver Output	VOD	Figure 1,	$R_L = 54\Omega$	1.5		V <sub>C</sub> C	V
Change in Magnitude of Differential Output Voltage	$\Delta V_{OD}$	Figure 1,	$R_L = 100\Omega$ or $54\Omega$ (Note 2)			0.2	V
Driver Common-Mode Output Voltage	Voc	Figure 1,	$R_L = 100\Omega$ or $54\Omega$		V <sub>CC</sub> / 2	3	V
Change in Magnitude of Common-Mode Voltage	ΔV <sub>OC</sub>	<b>U</b>	$R_{L} = 100\Omega \text{ or } 54\Omega \text{ (Note 2)}$ 42E/MAX13443E)			0.2	V
DRIVER LOGIC	•	•					
Driver-Input High Voltage	VDIH			2			V
Driver-Input Low Voltage	VDIL					0.8	V
Driver-Input Current	I <sub>DIN</sub>					±2	μΑ
Driver Short-Circuit Output Current	lean	$0 \le V_{OUT}$	≤ +12V			+350	<u>س</u> ۸
(Note 3)	IOSD	-7V ≤ V <sub>OL</sub>	$JT \le V_{CC}$	-350			mA
Driver Short-Circuit Foldback	loopr	(V <sub>CC</sub> - 1V	) ≤ V <sub>OUT</sub> ≤ +12V (Note 3)	+25			mA
Output Current	IOSDF	-7V ≤ V <sub>OL</sub>	JT ≤ +1V (Note 3)			-25	IIIA
Driver-Limit Short-Circuit Foldback		V <sub>OUT</sub> ≥ +	20V, $R_{L} = 100\Omega$	+6			mA
Output Current	IOSDL	$V_{OUT} \leq -15V, R_L = 100\Omega$			-6	IIIA	
RECEIVER							
		А, В	$V_{CC} = GND, V_{A, B} = 12V$			250	μA
Input Current	I <sub>A,B</sub>	receive	· · · · · · · · · · · · · · · · · · ·			-150	μA
		mode	$V_{A, B} = \pm 80V$			±6	mA
Receiver-Differential Threshold Voltage	V <sub>TH</sub>	$-7V \le V_{CM} \le +12V$		-200		-50	mV
Receiver-Input Hysteresis	$\Delta V_{TH}$				25		mV

## DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = +4.75V \text{ to } +5.25V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{CC} = +5V \text{ and } T_A = +25^{\circ}C.)$ 

PARAMETER	SYMBOL		CONDITIONS	MIN	ТҮР	MAX	UNITS
RECEIVER LOGIC							
Output-High Voltage	VOH	Figure 2,	lон = -1.6mA	V <sub>CC</sub> - 0.	6		V
Output-Low Voltage	Vol	Figure 2,	I <sub>OL</sub> = 1mA			0.4	V
Three-State Output Current at Receiver	I <sub>OZR</sub>	0V ≤ V <sub>A, B</sub>	3 ≤ V <sub>CC</sub>			±1	μA
Receiver Input Resistance	R <sub>IN</sub>	$-7V \le V_{CN}$	1 ≤ +12V	48			kΩ
Receiver Output Short-Circuit Current	IOSR	$0V \le V_{RO}$	≤ V <sub>CC</sub>			±95	mA
CONTROL							
Control-Input High Voltage	VCIH	DE, DE, R	Ē	2			V
Input-Current Latch During First Rising Edge	lın	DE, RE			90		μA
SUPPLY CURRENT							
Normal Operation		No load, DI = V <sub>CC</sub> or GND	$\begin{array}{l} DE=V_{CC}, \overline{RE}=GND \mbox{(MAX13442E)} \\ (\overline{DE}=\overline{RE}=GND) \mbox{(MAX13444E)} \end{array}$			30	
Normal Operation	ICC		$(DE = V_{CC}, \overline{RE} = GND)$ (MAX13443E)			10	mA
			D, <del>RE</del> = V <sub>CC</sub> 42E/MAX13443E)			20	
Supply Current in Shutdown Mode	ISHDN	DE = GND, $\overline{RE}$ = V <sub>CC</sub> , T <sub>A</sub> = +25°C (MAX13442E/MAX13443E)				10	μA
		$\overline{\text{DE}} = \overline{\text{RE}} = V_{\text{CC}} (\text{MAX13444E})$				100	
		$\overline{\text{DE}} = \overline{\text{RE}} =$	$\overline{\text{DE}} = \overline{\text{RE}} = V_{\text{CC}}, T_{\text{A}} = +25^{\circ}\text{C} (\text{MAX13444E})$			10	
Supply Current with Output Shorted to ±60V	ISHRT		D, <del>RE</del> = GND, no load hree-state (MAX13443E)			±15	mA

# MAX13442E/MAX13443E/MAX13444E

## **PROTECTION SPECIFICATIONS**

 $(V_{CC} = +4.75V \text{ to } +5.25V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{CC} = +5V \text{ and } T_A = +25^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS	
Overvoltage Protection				MAX13442E/ MAX13444E	±80			V
		MAX13443E		±60				
ESD Protection		А, В	Human B	Body Model		±15		kV

## SWITCHING CHARACTERISTICS (MAX13442E/MAX13444E)

(V<sub>CC</sub> = +4.75V to +5.25V,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $V_{CC}$  = +5V and  $T_A$  = +25°C.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Driver Propagation Dolay	tplha,	Figure 3, $R_L = 54\Omega$ , $C_L = 50pF$ (MAX13442E)			2000	20
Driver Propagation Delay	<b>t</b> PLHB	$R_{DIFF} = 60\Omega$ , $C_{DIFF} = 100$ pF (MAX13444E)			2000	ns
Driver Differential Propagation Delay	tdplh, tdphl	$R_L = 54\Omega$ , $C_L = 50pF$ , Figure 4			2000	ns
Driver Differential Output Transition Time	tLH,tHL	$R_L = 54\Omega$ , $C_L = 50pF$ , Figure 4	200		2000	ns
Driver Output Skew	<sup>t</sup> SKEWAB, <sup>t</sup> SKEWBA	$R_L = 54\Omega, C_L = 50pF,$ tskewab = Itplha - tphlbl, tskewba = Itplhb - tphlal			350	ns
Differential Driver Output Skew	<sup>t</sup> DSKEW	$R_L = 54\Omega, C_L = 50pF,$ toskew = ltoplh - tophLl			200	ns
Maximum Data Rate	fMAX		250			kbps
Driver Enable Time to Output High	tpdzh	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 5			2000	ns
Driver Disable Time from Output High	<b>t</b> PDHZ	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 5			2000	ns
Driver Enable Time from Shutdown to Output High	<b>t</b> PDHS	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 5			4.2	μs
Driver Enable Time to Output Low	<b>t</b> PDZL	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 6			2000	ns
Driver Disable Time from Output Low	t <sub>PDLZ</sub>	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 6			2000	ns
Driver Enable Time from Shutdown to Output Low	<b>t</b> PDLS	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 6			4.2	μs
Driver Time to Shutdown	t <sub>SHDN</sub>	$R_{L} = 500\Omega, C_{L} = 50pF$			800	ns
Receiver Propagation Delay	t <sub>RPLH</sub> , t <sub>RPHL</sub>	$C_L = 20pF$ , $V_{ID} = 2V$ , $V_{CM} = 0V$ , Figure 7			2000	ns
Receiver Output Skew	<b>t</b> RSKEW	CL = 20pF, t <sub>RSKEW</sub> = lt <sub>RPLH</sub> - t <sub>RPHL</sub> l			200	ns
Receiver Enable Time to Output High	t <sub>RPZH</sub>	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			2000	ns
Receiver Disable Time from Output High	t <sub>RPHZ</sub>	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			2000	ns
Receiver Wake Time from Shutdown	<b>t</b> RPWAKE	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			4.2	μs
Receiver Enable Time to Output Low	t <sub>RPZL</sub>	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			2000	ns
Receiver Disable Time from Output Low	t <sub>RPLZ</sub>	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			2000	ns
Receiver Time to Shutdown	<b>t</b> SHDN	$R_{L} = 500\Omega, C_{L} = 50pF$			800	ns

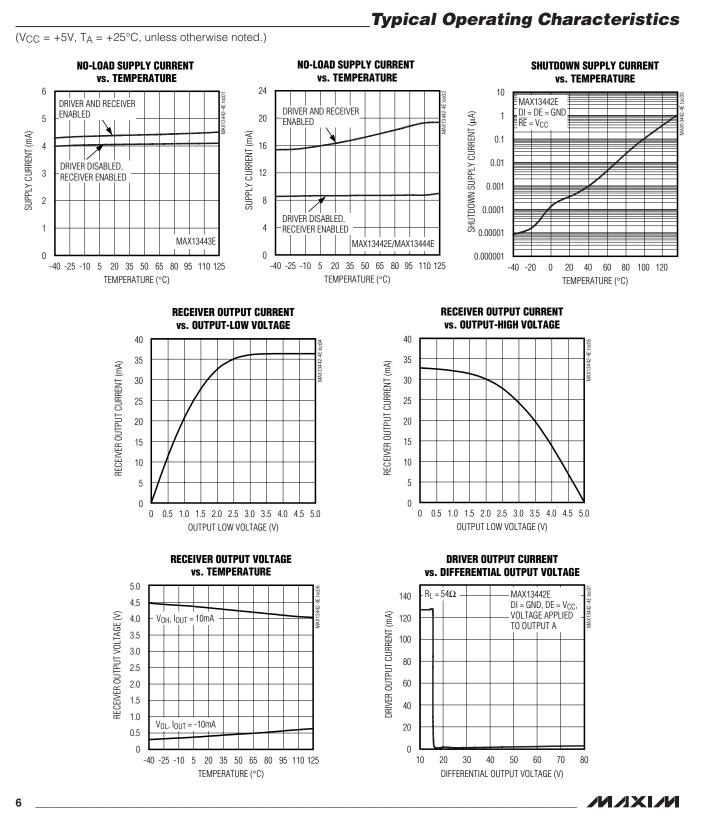
# SWITCHING CHARACTERISTICS (MAX13443E)

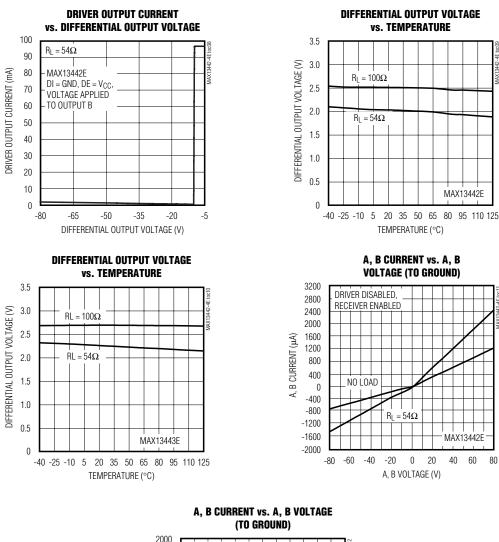
 $(V_{CC} = +4.75V \text{ to } +5.25V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } V_{CC} = +5V \text{ and } T_A = +25^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ΤΥΡ	MAX	UNITS
Driver Propagation Delay	tplha, tplhb	$R_L = 27\Omega$ , $C_L = 50pF$ , Figure 3			60	ns
Driver Differential Propagation Delay	tdplh, tdphl	$R_L = 54\Omega$ , $C_L = 50pF$ , Figure 4			60	ns
Driver Differential Output Transition Time	t <sub>LH</sub> ,t <sub>HL</sub>	$R_L = 54\Omega$ , $C_L = 50pF$ , Figure 4			25	ns
Driver Output Skew	tskewab, tskewba	$ \begin{array}{l} R_{L} = 54\Omega,  C_{L} = 50 pF, \\ t_{SKEWAB} = lt_{PLHA} \; \text{-} \; t_{PHLB}l, \\ t_{SKEWBA} = lt_{PLHB} \; \text{-} \; t_{PHLA}l \end{array} $			10	ns
Differential Driver Output Skew	<b>t</b> DSKEW	$R_L = 54\Omega, C_L = 50pF,$ tDSKEW = ltDPLH - tDPHLl			10	ns
Maximum Data Rate	fMAX		10			Mbps
Driver Enable Time to Output High	t <sub>PDZH</sub>	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 5			1200	ns
Driver Disable Time from Output High	t <sub>PDHZ</sub>	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 5			1200	ns
Driver Enable Time from Shutdown to Output High	<b>t</b> PDHS	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 5			4.2	μs
Driver Enable Time to Output Low	t <sub>PDZL</sub>	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 6			1200	ns
Driver Disable Time from Output Low	<b>t</b> PDLZ	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 6			1200	ns
Driver Enable Time from Shutdown to Output Low	<b>t</b> PDLS	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 6			4.2	μs
Driver Time to Shutdown	<b>t</b> SHDN	$R_L = 500\Omega$ , $C_L = 50pF$ , Figure 6			800	ns
Receiver Propagation Delay	t <sub>RPLH</sub> , t <sub>RPHL</sub>	$C_L$ = 20pF, $V_{ID}$ = 2V, $V_{CM}$ = 0V, Figure 7			85	ns
Receiver Output Skew	<b>t</b> RSKEW	CL = 20pF, t <sub>RSKEW</sub> = lt <sub>RPLH</sub> - t <sub>RPHL</sub> l			15	ns
Receiver Enable Time to Output High	t <sub>RPZH</sub>	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			400	ns
Receiver Disable Time from Output High	t <sub>RPHZ</sub>	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			400	ns
Receiver Wake Time from Shutdown	<b>t</b> RPWAKE	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			4.2	μs
Receiver Enable Wake Time from Shutdown	tRPSH	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			400	ns
Receiver Disable Time from Output Low	t <sub>RPLZ</sub>	$R_L = 1k\Omega$ , $C_L = 20pF$ , Figure 8			400	ns
Receiver Time to Shutdown	t <sub>SHDN</sub>	$R_L = 500\Omega$ , $C_L = 50pF$			800	ns

**Note 2:**  $\Delta V_{OD}$  and  $\Delta V_{OC}$  are the changes in  $V_{OD}$  and  $V_{OC}$ , respectively, when the DI input changes state.

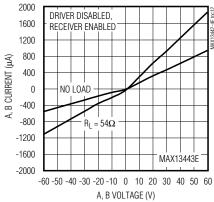
Note 3: The short-circuit output current applies to peak current just before foldback current limiting. The short-circuit foldback output current applies during current limiting to allow a recovery from bus contention.





## \_Typical Operating Characteristics (continued)

MAX13442E/MAX13443E/MAX13444E





 $(V_{CC} = +5V, T_A = +25^{\circ}C, unless otherwise noted.)$ 

# MAX13442E/MAX13443E/MAX13444E

## **Test Circuits and Waveforms**

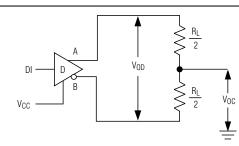


Figure 1. Driver VOD and VOC

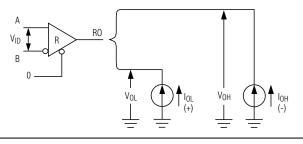
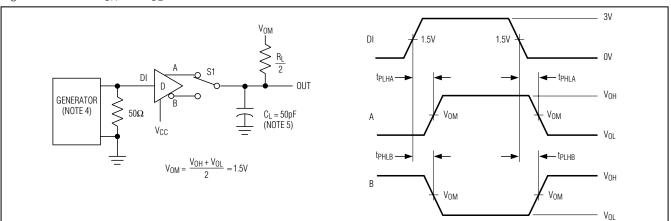


Figure 2. Receiver VOH and VOL





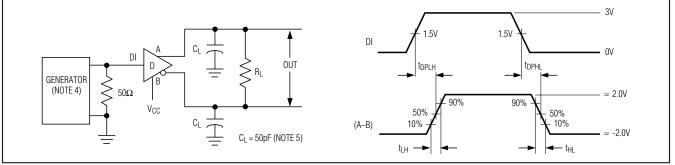
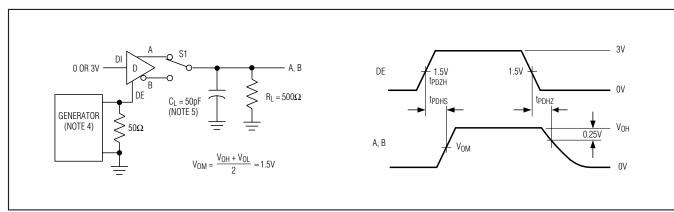


Figure 4. Driver Differential Output Delay and Transition Times





## **\_Test Circuits and Waveforms (continued)**

Figure 5. Driver Enable and Disable Times

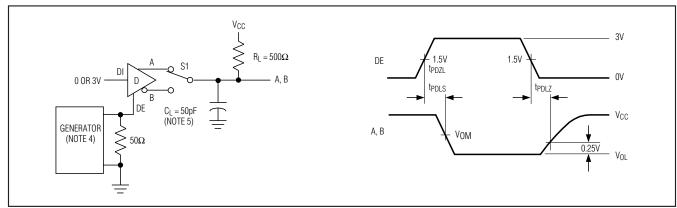


Figure 6. Driver Enable and Disable Times

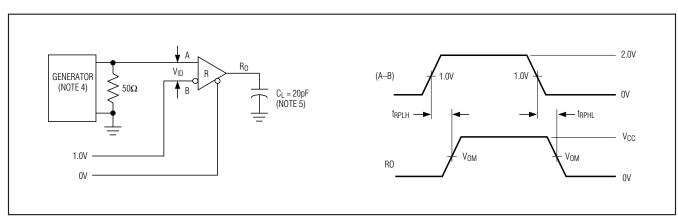
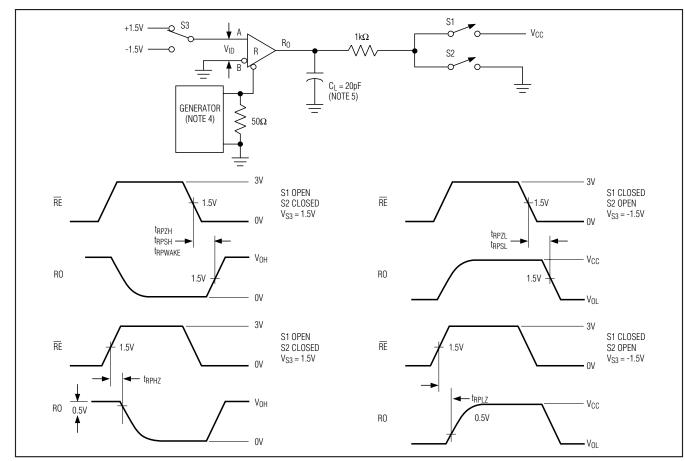


Figure 7. Receiver Propagation Delay



**Test Circuits and Waveforms (continued)** 

Figure 8. Receiver Enable and Disable Times

**Note 4:** The input pulse is supplied by a generator with the following characteristics: f = 5MHz, 50% duty cycle;  $t_r \le 6ns$ ;  $Z_0 = 50\Omega$ . **Note 5:**  $C_L$  includes probe and stray capacitance.

MAX13442E/MAX13443E/MAX13444E

# \_Pin Description

Р	IN				
MAX13442E MAX13443E	MAX13444E	NAME	FUNCTION		
1	1	RO	Receiver Output. If the receiver is enabled and $(A - B) \ge$ -50mV, RO = high; if $(A - B) \le$ -200mV, RO = low.		
2	2	RE	Receiver Output Enable. Pull RE low to enable RO.		
3	_	DE	Driver Output Enable. Force DE high to enable driver. Pull DE low to three-state the driver output. Drive $\overline{\text{RE}}$ high and pull DE low to enter low-power shutdown mode.		
4		DI	Driver Input. A logic-low on DI forces the noninverting output low and the inverting output high. A logic-high on DI forces the noninverting output high and the inverting output low.		
5	5	GND	Ground		
6	6	А	Noninverting Receiver Input/Driver Output		
7	7	В	Inverting Receiver Input/Driver Output		
8	8	V <sub>CC</sub>	Positive Supply, $V_{CC}$ = +4.75V to +5.25V. For normal operation, bypass $V_{CC}$ to GND with a 0.1µF ceramic capacitor. For full ESD protection, bypass $V_{CC}$ to GND with 1µF ceramic capacitor.		
_	3	DE	Driver Output Enable. Pull $\overline{\text{DE}}$ low to enable the outputs. Force $\overline{\text{DE}}$ high to three-state the outputs. Drive $\overline{\text{RE}}$ and $\overline{\text{DE}}$ high to enter low-power shutdown mode.		
_	4	TXD	J1708 Input. A logic-low on TXD forces outputs A and B to the dominant state. A logic-high on TXD forces outputs A and B to the recessive state.		



## **Function Tables**

## Table 1. MAX13442E/MAX13443E (RS-485/RS-422)

TRANSMITTING							
	INPUTS	OUTPUTS					
RE	DE	DI	Α	В			
0	0	Х	High-Z	High-Z			
0	1	0	0	1			
0	1	1	1	0			
1	0	Х	Shutdown	Shutdown			
1	1	0	0	1			
1	1	1	1	0			

X = Don't care.

## Table 2. MAX13444E (J1708) Application

TRANSMITTING							
INP	UTS	OUTF	PUTS	CONDITIONS			
TXD	DE	Α	В	—			
0	1	High-Z	High-Z	—			
1	1	High-Z	High-Z	—			
0	0	0	1	Dominant state			
1	0	High-Z	High-Z	Recessive state			

## Table 3. MAX13442E/MAX13443E (RS-485/RS-422)

RECEIVING						
	INPUTS					
RE	DE	(A - B)	RO			
0	Х	≥-0.05V	1			
0	Х	≤-0.2V	0			
0	Х	Open/shorted	1			
1	1	Х	High-Z			
1	0	Х	Shutdown			

X = Don't care.

## Table 4. MAX13444E (RS-485/RS-422)

RECEIVING						
	INPUTS		OUTPUTS			
RE	DE	(A - B)	RO			
0	Х	≥-0.05V	1			
0	Х	≤-0.2V	0			
0	Х	Open/shorted	1			
1	0	Х	High-Z			
1	1	Х	Shutdown			

X = Don't care.

## **Detailed Description**

The MAX13442E/MAX13443E/MAX13444E fault-protected transceivers for RS-485/RS-422 and J1708 communication contain one driver and one receiver. These devices feature fail-safe circuitry, which guarantees a logic-high receiver output when the receiver inputs are open or shorted, or when they are connected to a terminated transmission line with all drivers disabled (see the True Fail-Safe section). All devices have a hot-swap input structure that prevents disturbances on the differential signal lines when a circuit board is plugged into a hot backplane (see the Hot-Swap Capability section). The MAX13442E/MAX13444E feature a reduced slew-rate driver that minimizes EMI and reduces reflections caused by improperly terminated cables, allowing error-free data transmission up to 250kbps (see the Reduced EMI and Reflections section). The MAX13443E driver is not slewrate limited, allowing transmit speeds up to 10Mbps.

#### Driver

The driver accepts a single-ended, logic-level input (DI) and transfers it to a differential, RS-485/RS-422 level output (A and B). Deasserting the driver enable places the driver outputs (A and B) into a high-impedance state.

#### Receiver

The receiver accepts a differential, RS-485/RS-422 level input (A and B), and transfers it to a single-ended, logic-level output (RO). Deasserting the receiver enable places the receiver inputs (A and B) into a high-impedance state (see Tables 1–4).

#### **Low-Power Shutdown**

The MAX13442E/MAX13443E/MAX13444E offer a lowpower shutdown mode. Force DE low and RE high to shut down the MAX13442E/MAX13443E. Force DE and RE high to shut down the MAX13444E. A time delay of 50ns prevents the device from accidentally entering shutdown due to logic skews when switching between transmit and receive modes. Holding DE low and RE high for at least 800ns guarantees that the MAX13442E/MAX13443E enter shutdown. In shutdown, the devices consume a maximum 20µA supply current.

#### **±80V Fault Protection**

The driver outputs/receiver inputs of RS-485 devices in industrial network applications often experience voltage faults resulting from shorts to the power grid that exceed the -7V to +12V range specified in the EIA/TIA-485 standard. In these applications, ordinary RS-485 devices (typical absolute maximum -8V to +12.5V) require costly external protection devices. To reduce system complexity and eliminate this need for external protection, the dri-



ver outputs/receiver inputs of the MAX13442E/ MAX13444E withstand voltage faults up to ±80V (±60V for the MAX13443E) with respect to ground without damage. Protection is guaranteed regardless whether the device is active, shut down, or without power.

#### **True Fail-Safe**

The MAX13442E/MAX13443E/MAX13444E use a -50mV to -200mV differential input threshold to ensure true fail-safe receiver inputs. This threshold guarantees the receiver outputs a logic-high for shorted, open, or idle data lines. The -50mV to -200mV threshold complies with the ±200mV threshold EIA/TIA-485 standard.

#### ±15kV ESD Protection

As with all Maxim devices, ESD-protection structures are incorporated on all pins to protect against ESD encountered during handling and assembly. The MAX13442E/MAX13443E/MAX13444E receiver inputs/ driver outputs (A, B) have extra protection against static electricity found in normal operation. Maxim's engineers have developed state-of-the-art structures to protect these pins against ±15kV ESD without damage. After an ESD event, the MAX13442E/MAX13443E/ MAX13444E continue working without latchup.

ESD protection can be tested in several ways. The receiver inputs are characterized for protection to  $\pm 15 \text{kV}$  using the Human Body Model.

#### ESD Test Conditions

ESD performance depends on a number of conditions. Contact Maxim for a reliability report that documents test setup, methodology, and results.

#### Human Body Model

Figure 9a shows the Human Body Model, and Figure 9b shows the current waveform it generates when discharged into a low impedance. This model consists of a 100pF capacitor charged to the ESD voltage of interest, which is then discharged into the device through a  $1.5k\Omega$  resistor.

#### **Driver Output Protection**

Two mechanisms prevent excessive output current and power dissipation caused by faults or bus contention. The first, a foldback current limit on the driver output stage, provides immediate protection against short circuits over the whole common-mode voltage range. The second, a thermal shutdown circuit, forces the driver outputs into a high-impedance state if the die temperature exceeds +160°C. Normal operation resumes when the die temperature cools to +140°C, resulting in a pulsed output during continuous short-circuit conditions.

### Hot-Swap Capability

#### Hot-Swap Input Circuitry

Hot-Swap Inputs Inserting circuit boards into a hot, or powered, backplane may cause voltage transients on DE, RE, and receiver inputs A and B that can lead to data errors. For example, upon initial circuit board insertion, the processor undergoes a power-up sequence. During this period, the high-impedance state of the output drivers makes them unable to drive the MAX13442E/MAX13443E/ MAX13444E enable inputs to a defined logic level. Meanwhile, leakage currents of up to 10µA from the high-impedance output, or capacitively coupled noise from VCC or GND, could cause an input to drift to an incorrect logic state. To prevent such a condition from occurring, the MAX13442E/MAX13443E/MAX13444E feature hot-swap input circuitry on DE, and RE to guard against unwanted driver activation during hot-swap situations. The MAX13444E has hot-swap input circuitry only on RE. When VCC rises, an internal pulldown (or pullup for RE) circuit holds DE low for at least 10µs, and until the current into DE exceeds 200µA. After the initial power-up sequence, the pulldown circuit becomes transparent, resetting the hot-swap tolerable input.

At the driver-enable input (DE), there are two NMOS devices, M1 and M2 (Figure 10). When V<sub>CC</sub> ramps from zero, an internal 15µs timer turns on M2 and sets the SR latch, which also turns on M1. Transistors M2, a 2mA current sink, and M1, a 100µA current sink, pull DE to GND through a 5.6k $\Omega$  resistor. M2 pulls DE to the disabled state against an external parasitic capacitance up to 100pF that may drive DE high. After 15µs, the timer deactivates M2 while M1 remains on, holding DE low against three-state leakage currents that may drive DE high. M1 remains on until an external current source overcomes the required input current. At this time, the SR latch resets M1 and turns off. When M1 turns off, DE reverts to a standard, high-impedance CMOS input. Whenever V<sub>CC</sub> drops below 1V, the input is reset.

A complementary circuit for  $\overline{\text{RE}}$  uses two PMOS devices to pull  $\overline{\text{RE}}$  to V<sub>CC</sub>.

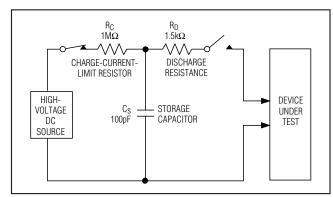


Figure 9a. Human Body ESD Test Model

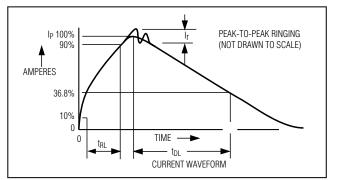


Figure 9b. Human Body Model Current Waveform

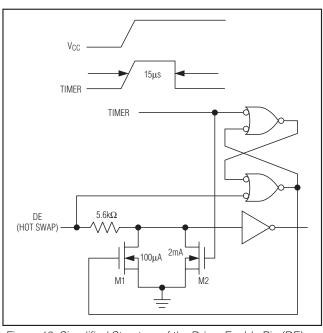


Figure 10. Simplified Structure of the Driver Enable Pin (DE)

## **Applications Information**

#### **128 Transceivers on the Bus**

The MAX13442E/MAX13443E/MAX13444E transceivers 1/4-unit-load receiver input impedance (48k $\Omega$ ) allows up to 128 transceivers connected in parallel on one communication line. Connect any combination of these devices, and/or other RS-485 devices, for a maximum of 32-unit loads to the line.

#### **Reduced EMI and Reflections**

The MAX13442E/MAX13444E are slew-rate limited, minimizing EMI and reducing reflections caused by improperly terminated cables. Figure 11 shows the driver output waveform and its Fourier analysis of a 125kHz signal transmitted by a MAX13443E. High-frequency harmonic components with large amplitudes are evident.

Figure 12 shows the same signal displayed for the MAX13442E transmitting under the same conditions. Figure 12's high-frequency harmonic components are much lower in amplitude, compared with Figure 11's, and the potential for EMI is significantly reduced.

In general, a transmitter's rise time relates directly to the length of an unterminated stub, which can be driven with only minor waveform reflections. The following equation expresses this relationship conservatively:

 $length = t_{RISE} / (10 \times 1.5 n_{s}/ft)$ 

where t<sub>RISE</sub> is the transmitter's rise time.

For example, the MAX13442E's rise time is typically 800ns, which results in excellent waveforms with a stub length up to 53ft. A system can work well with longer unterminated stubs, even with severe reflections, if the waveform settles out before the UART samples them.

#### **RS-485 Applications**

The MAX13442E/MAX13443E/MAX13444E transceivers provide bidirectional data communications on multipoint bus transmission lines. Figure 13 shows a typical network application circuit. The RS-485 standard covers line lengths up to 4000ft. To minimize reflections and reduce data errors, terminate the signal line at both ends in its characteristic impedance, and keep stub lengths off the main line as short as possible.

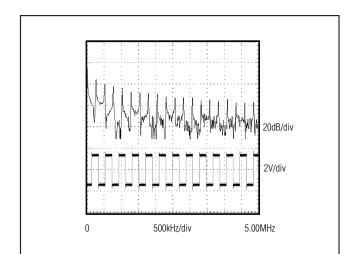


Figure 11. Driver Output Waveform and FFT Plot of the MAX13443E Transmitting a 125kHz Signal

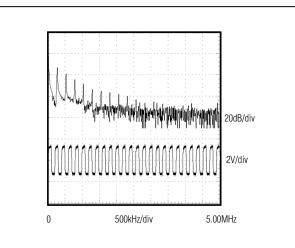


Figure 12. Driver Output Waveform and FFT Plot of the MAX13442E Transmitting a 125kHz Signal

#### **J1708** Applications

The MAX13444E is designed for J1708 applications. To configure the MAX13444E, connect  $\overline{\text{DE}}$  and  $\overline{\text{RE}}$  to GND. Connect the signal to be transmitted to TXD. Terminate the bus with the load circuit as shown in Figure 14. The drivers used by SAE J1708 are used in a dominant-mode application.  $\overline{\text{DE}}$  is active low; a high input on  $\overline{\text{DE}}$  places the outputs in high impedance. When the driver is disabled (TXD high or  $\overline{\text{DE}}$  high), the bus is pulled high by external bias resistors R1 and R2. Therefore, a logic-level high is encoded as recessive. When all transceivers are

idle in this configuration, all receivers output logic-high because of the pullup resistor on A and pulldown resistor on B. R1 and R2 provide the bias for the recessive state. C1 and C2 combine to form a lowpass filter, effective for reducing FM interference. R2, C1, R4, and C2 combine to form a 1.6MHz lowpass filter, effective for reducing AM interference. Because the bus is unterminated, at high frequencies, R3 and R4 perform a pseudotermination. This makes the implementation more flexible, as no specific termination nodes are required at the ends of the bus.

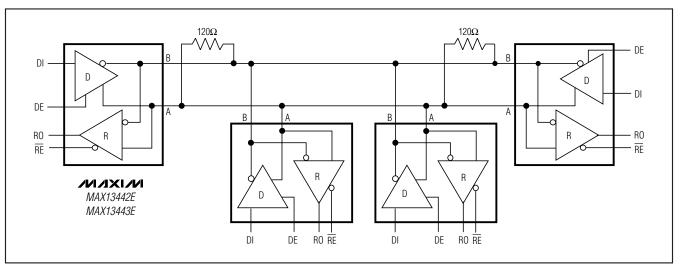


Figure 13. MAX13442E/MAX13443E Typical RS-485 Network

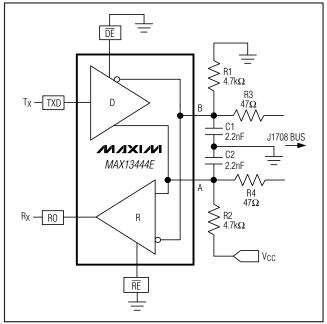


Figure 14. J1708 Application Circuit (See Tables 2 and 4)

R0 1

RE 2

DE 3

TXD 4

SO

8 V<sub>CC</sub>

7 B

6 А

5

SO

GND

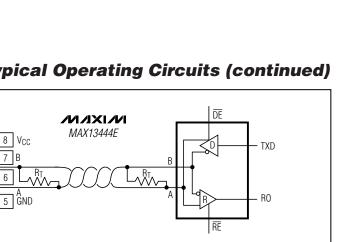
PROCESS: BICMOS

## Package Information

**Chip Information** 

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
8 SO	S8+4	<u>21-0041</u>	<u>90-0096</u>



# Pin Configurations and Typical Operating Circuits (continued)

R0 1

RE 2

DE 3

TXD 4

## **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	10/05	Initial release	—
1	3/06	Corrected the part numbers in the conditions for $\Delta V_{OC}$ in the <i>DC Electrical Characteristics</i> table; corrected the A, B current units from mA to $\mu$ A for the A, B Current vs. A, B Voltage (to Ground) graphs in the <i>Typical Operating Characteristics</i> section	2, 7
2	11/10	Added lead(Pb)-free parts to the Ordering Information table; added the soldering temperature to the Absolute Maximum Ratings section; updated Table 2 outputs	1, 2, 12

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18

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