

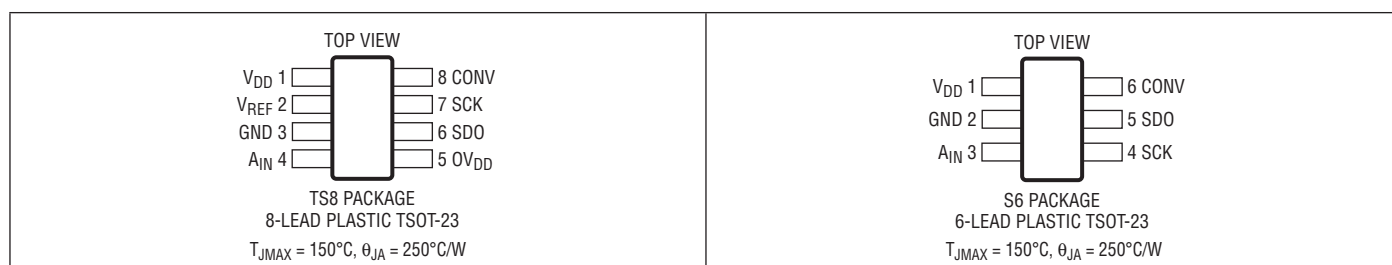
# LTC2360/LTC2361/LTC2362

## ABSOLUTE MAXIMUM RATINGS (Notes 1, 2)

Supply Voltage ( $V_{DD}$ ) ..... 4V  
Supply Voltage ( $OV_{DD}$ ) ..... Min ( $V_{DD} + 0.3V$ , 4.0V)  
 $V_{REF}$  and Analog Input Voltage  
(Note 3) .....  $-0.3V$  to ( $V_{DD} + 0.3V$ )  
Digital Input Voltage .....  $-0.3V$  to ( $V_{DD} + 0.3V$ )  
Digital Output Voltage .....  $-0.3V$  to ( $V_{DD} + 0.3V$ )  
Power Dissipation ..... 100mW

Operating Temperature Range  
LTC2360C/LTC2361C/LTC2362C .....  $0^{\circ}C$  to  $70^{\circ}C$   
LTC2360I/LTC2361I/LTC2362I .....  $-40^{\circ}C$  to  $85^{\circ}C$   
LTC2360H/LTC2361H/LTC2362H (Note 12) ..  $-40^{\circ}C$  to  $125^{\circ}C$   
Storage Temperature Range .....  $-65^{\circ}C$  to  $150^{\circ}C$   
Lead Temperature (Soldering, 10 sec) .....  $300^{\circ}C$

## PIN CONFIGURATION



## ORDER INFORMATION

### Lead Free Finish

TAPE AND REEL (MINI)	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LTC2362CTS8#TRMPBF	LTC2362CTS8#TRPBF	LTDBV	8-Lead Plastic TSOT23	$0^{\circ}C$ to $70^{\circ}C$
LTC2362ITS8#TRMPBF	LTC2362ITS8#TRPBF	LTDBV	8-Lead Plastic TSOT23	$-40^{\circ}C$ to $85^{\circ}C$
LTC2362HTS8#TRMPBF	LTC2362HTS8#TRPBF	LTDBV	8-Lead Plastic TSOT23	$-40^{\circ}C$ to $125^{\circ}C$
LTC2362CS6#TRMPBF	LTC2362CS6#TRPBF	LTDGP	6-Lead Plastic TSOT23	$0^{\circ}C$ to $70^{\circ}C$
LTC2362IS6#TRMPBF	LTC2362IS6#TRPBF	LTDGP	6-Lead Plastic TSOT23	$-40^{\circ}C$ to $85^{\circ}C$
LTC2362HS6#TRMPBF	LTC2362HS6#TRPBF	LTDGP	6-Lead Plastic TSOT23	$-40^{\circ}C$ to $125^{\circ}C$
LTC2361CTS8#TRMPBF	LTC2361CTS8#TRPBF	LTDGM	8-Lead Plastic TSOT23	$0^{\circ}C$ to $70^{\circ}C$
LTC2361ITS8#TRMPBF	LTC2361ITS8#TRPBF	LTDGM	8-Lead Plastic TSOT23	$-40^{\circ}C$ to $85^{\circ}C$
LTC2361HTS8#TRMPBF	LTC2361HTS8#TRPBF	LTDGM	8-Lead Plastic TSOT23	$-40^{\circ}C$ to $125^{\circ}C$
LTC2361CS6#TRMPBF	LTC2361CS6#TRPBF	LTDGN	6-Lead Plastic TSOT23	$0^{\circ}C$ to $70^{\circ}C$
LTC2361IS6#TRMPBF	LTC2361IS6#TRPBF	LTDGN	6-Lead Plastic TSOT23	$-40^{\circ}C$ to $85^{\circ}C$
LTC2361HS6#TRMPBF	LTC2361HS6#TRPBF	LTDGN	6-Lead Plastic TSOT23	$-40^{\circ}C$ to $125^{\circ}C$
LTC2360CTS8#TRMPBF	LTC2360CTS8#TRPBF	LTDGJ	8-Lead Plastic TSOT23	$0^{\circ}C$ to $70^{\circ}C$
LTC2360ITS8#TRMPBF	LTC2360ITS8#TRPBF	LTDGJ	8-Lead Plastic TSOT23	$-40^{\circ}C$ to $85^{\circ}C$
LTC2360HTS8#TRMPBF	LTC2360HTS8#TRPBF	LTDGJ	8-Lead Plastic TSOT23	$-40^{\circ}C$ to $125^{\circ}C$
LTC2360CS6#TRMPBF	LTC2360CS6#TRPBF	LTDGK	6-Lead Plastic TSOT23	$0^{\circ}C$ to $70^{\circ}C$
LTC2360IS6#TRMPBF	LTC2360IS6#TRPBF	LTDGK	6-Lead Plastic TSOT23	$-40^{\circ}C$ to $85^{\circ}C$
LTC2360HS6#TRMPBF	LTC2360HS6#TRPBF	LTDGK	6-Lead Plastic TSOT23	$-40^{\circ}C$ to $125^{\circ}C$

TRM = 500 pieces. \*Temperature grades are identified by a label on the shipping container.

Consult LTC Marketing for information on lead based finish parts.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreel/>

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## CONVERTER CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ . (Note 4)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Resolution (No Missing Codes)	●	12			Bits
Integral Linearity Error	(Notes 5, 6) ●		$\pm 0.25$	$\pm 1$	LSB
Differential Linearity Error	(Note 6) ●		$\pm 0.25$	$\pm 1$	LSB
Transition Noise	(Note 7)		0.25		$\text{LSB}_{\text{RMS}}$
Offset Error	(Note 6) ●		1	$\pm 3.5$	LSB
Gain Error	(Note 6) ●		0.1	$\pm 2$	LSB
Total Unadjusted Error	(Note 6) ●		1.1	$\pm 3.5$	LSB

## ANALOG INPUT

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ . (Note 4)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$V_{\text{IN}}$	Analog Input Voltage	S6 Package TS8 Package ●	-0.05 -0.05		$V_{\text{DD}} + 0.05$ $V_{\text{REF}} + 0.05$	V
$I_{\text{IN}}$	Analog Input Leakage Current	CONV = High ●			$\pm 1$	$\mu\text{A}$
$C_{\text{IN}}$	Analog Input Capacitance	Between Conversions During Conversions		20 4		pF pF
$V_{\text{REF}}$	Reference Input Voltage	TS8 Package ●	1.4		$V_{\text{DD}} + 0.05$	V
$I_{\text{REF}}$	Reference Input Leakage Current	TS8 Package ●			$\pm 1$	$\mu\text{A}$
$C_{\text{REF}}$	Reference Input Capacitance	TS8 Package		20		pF
$t_{\text{AP}}$	Sample-and-Hold Aperture Delay Time			1		ns
$t_{\text{JITTER}}$	Sample-and-Hold Aperture Delay Time Jitter			0.3		ns

## DYNAMIC ACCURACY

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ . (Note 4)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
SINAD	Signal-to-(Noise + Distortion) Ratio	$f_{\text{IN}} = 49\text{kHz}$ for LTC2360/LTC2361, $f_{\text{IN}} = 100\text{kHz}$ for LTC2362		72		dB
SNR	Signal-to-Noise Ratio	$f_{\text{IN}} = 49\text{kHz}$ for LTC2360/LTC2361, $f_{\text{IN}} = 100\text{kHz}$ for LTC2362		73		dB
THD	Total Harmonic Distortion	$f_{\text{IN}} = 49\text{kHz}$ for LTC2360/LTC2361, $f_{\text{IN}} = 100\text{kHz}$ for LTC2362		-85		dB
SFDR	Spurious Free Dynamic Range	$f_{\text{IN}} = 49\text{kHz}$ for LTC2360/LTC2361, $f_{\text{IN}} = 100\text{kHz}$ for LTC2362		86		dB
IMD	Intermodulation Distortion	$f_{\text{IN1}} = 97\text{kHz}$ , $f_{\text{IN2}} = 100\text{kHz}$ for LTC2362 $f_{\text{IN1}} = 47\text{kHz}$ , $f_{\text{IN2}} = 49\text{kHz}$ for LTC2360/LTC2361		-75		dB
	Full-Power Bandwidth	at 3dB at 0.1dB		10 2		MHz MHz
	Full-Linear Bandwidth	SINAD $\geq 68\text{dB}$		1		MHz

# LTC2360/LTC2361/LTC2362

## DIGITAL INPUTS AND DIGITAL OUTPUTS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ . (Note 4)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
$V_{IH}$	High Level Input Voltage	$2.7V < V_{DD} \leq 3.6V$ $2.35V \leq V_{DD} \leq 2.7V$	● ●	2 1.7			V V
$V_{IL}$	Low Level Input Voltage	$2.7V < V_{DD} \leq 3.6V$ $2.35V \leq V_{DD} \leq 2.7V$	● ●			0.8 0.7	V V
$I_{IH}$	High Level Input Current	$V_{IN} = V_{DD}$	●			2.5	$\mu\text{A}$
$I_{IL}$	Low Level Input Current	$V_{IN} = 0V$	●			-2.5	$\mu\text{A}$
$C_{IN}$	Digital Input Capacitance				2		pF
$V_{OH}$	High Level Output Voltage	$V_{DD} = 2.35V$ to $3.6V$ , $I_{SOURCE} = 200\mu\text{A}$	●	$V_{DD} - 0.2$			V
$V_{OL}$	Low Level Output Voltage	$V_{DD} = 2.35V$ to $3.6V$ , $I_{SINK} = 200\mu\text{A}$	●			0.2	V
$I_{OZ}$	Hi-Z Output Leakage	$CONV = V_{DD}$	●			$\pm 3$	$\mu\text{A}$
$C_{OZ}$	Hi-Z Output Capacitance	$CONV = V_{DD}$			4		pF
$I_{SOURCE}$	Output Source Current	$V_{OUT} = 0V$			-10		mA
$I_{SINK}$	Output Sink Current	$V_{OUT} = V_{DD}$			10		mA

## POWER REQUIREMENT

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ . (Note 4)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
$V_{DD}$	Supply Voltage		●	2.35	3.0	3.6	V
$OV_{DD}$	Digital Output Supply Voltage		●	1.0V		$V_{DD}$	V
$I_{DD}$	Supply Current						
	Operational Mode, LTC2362	$f_{SMPL} = 500\text{ksps}$	●		1.1	2	mA
	Operational Mode, LTC2361	$f_{SMPL} = 250\text{ksps}$	●		0.75	1.5	mA
	Operational Mode, LTC2360	$f_{SMPL} = 100\text{ksps}$	●		0.5	1	mA
	Sleep Mode	$0^\circ\text{C}$ to $70^\circ\text{C}$	●		0.1	2	$\mu\text{A}$
	Sleep Mode	$-40^\circ\text{C}$ to $85^\circ\text{C}$	●		0.1	2	$\mu\text{A}$
	Sleep Mode	$-40^\circ\text{C}$ to $125^\circ\text{C}$	●		0.1	5	$\mu\text{A}$
$P_D$	Power Dissipation						
	Operational Mode, LTC2362	$f_{SMPL} = 500\text{ksps}$	●		3.3	7.2	mW
	Operational Mode, LTC2361	$f_{SMPL} = 250\text{ksps}$	●		2.25	5.4	mW
	Operational Mode, LTC2360	$f_{SMPL} = 100\text{ksps}$	●		1.5	3.6	mW
	Sleep Mode	$0^\circ\text{C}$ to $70^\circ\text{C}$	●		0.3	7.2	$\mu\text{W}$
	Sleep Mode	$-40^\circ\text{C}$ to $85^\circ\text{C}$	●		0.3	7.2	$\mu\text{W}$
	Sleep Mode	$-40^\circ\text{C}$ to $125^\circ\text{C}$	●		0.3	18	$\mu\text{W}$

## TIMING CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^\circ\text{C}$ . (Note 4)

SYMBOL	PARAMETER	CONDITIONS		LTC2360			LTC2361			LTC2362			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$f_{\text{SMPL(MAX)}}$	Maximum Sampling Frequency	(Notes 8, 9)	●	100			250			500			kHz
$f_{\text{SCK}}$	Shift Clock Frequency	(Notes 8, 9)	●			10			25			50	MHz
$t_{\text{SCK}}$	Shift Clock Period		●	100			40			20			ns
$t_{\text{THROUGHPUT}}$	Minimum Throughput Time, $t_{\text{ACQ}} + t_{\text{CONV}}$		●			10			4			2	$\mu\text{s}$
$t_{\text{ACQ}}$	Acquisition Time		●	2			1			0.5			$\mu\text{s}$
$t_{\text{CONV}}$	Conversion Time		●	8			3			1.5			$\mu\text{s}$
$t_1$	Minimum Positive CONV Pulse Width	(Note 8)	●	8			3			1.5			$\mu\text{s}$
$t_2$	SCK↑ Setup Time After CONV↓	(Note 8)	●	16			16			16			ns
$t_3$	SDO Enabled Time After CONV↓	(Notes 8, 9)	●			16			16			16	ns
$t_4$	SDO Data Valid Access Time After SCK↓	(Notes 8, 9, 10)	●			8			8			8	ns
$t_5$	SCK Low Time	(Note 11)	●	40%			40%			40%			$t_{\text{SCK}}$
$t_6$	SCK High Time	(Note 11)	●	40%			40%			40%			$t_{\text{SCK}}$
$t_7$	SDO Data Valid Hold Time After SCK↓	(Notes 8, 9, 10)	●	4			4			4			ns
$t_8$	SDO Into Hi-Z State Time After CONV↑	(Notes 8, 9)			6			6			6		ns

**Note 1:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

**Note 2:** All voltage values are with respect to GND.

**Note 3:** When pins  $A_{\text{IN}}$  and  $V_{\text{REF}}$  are taken below GND or above  $V_{\text{DD}}$ , they will be clamped by internal diodes. These products can handle input currents greater than 100mA below GND or above  $V_{\text{DD}}$  without latch-up.

**Note 4:**  $V_{\text{DD}} = OV_{\text{DD}} = V_{\text{REF}} = 2.35\text{V}$  to  $3.6\text{V}$ ,  $f_{\text{SMPL}} = f_{\text{SMPL(MAX)}}$  and  $f_{\text{SCK}} = f_{\text{SCK(MAX)}}$  unless otherwise specified.

**Note 5:** Integral linearity is defined as the deviation of a code from a straight line passing through the actual endpoints of the transfer curve. The deviation is measured from the center of the quantization band.

**Note 6:** Linearity, offset and gain specifications apply for a single-ended  $A_{\text{IN}}$  input with respect to GND.

**Note 7:** Typical RMS noise at code transitions.

**Note 8:** Guaranteed by characterization. All input signals are specified with  $t_r = t_f = 2\text{ns}$  (10% to 90% of  $V_{\text{DD}}$ ) and timed from a voltage level of 1.6V.

**Note 9:** All timing specifications given are with a 10pF capacitance load. With a capacitance load greater than this value, a digital buffer or latch must be used.

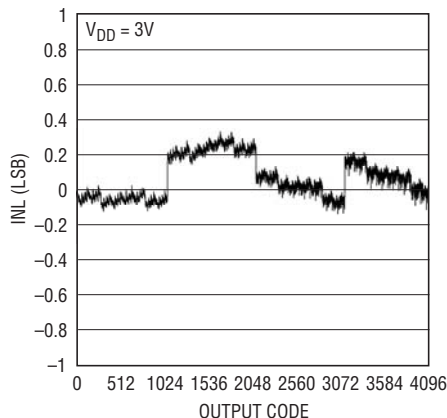
**Note 10:** The time required for the output to cross the  $V_{\text{IH}}$  or  $V_{\text{IL}}$  voltage.

**Note 11:** Guaranteed by design, not subject to test.

**Note 12:** High temperatures degrade operating lifetimes. Operating lifetime is derated at temperatures greater than  $105^\circ\text{C}$ .

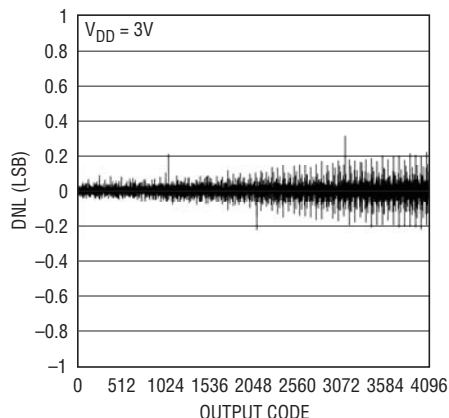
## TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$ , $V_{DD} = 0V_{DD} = V_{REF}$ (LTC2360, Note 4)

**Integral Nonlinearity vs Output Code**



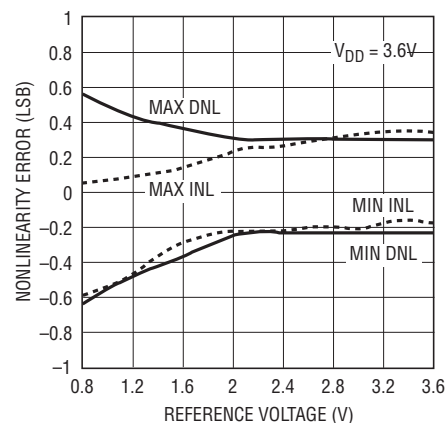
236012 G01

**Differential Nonlinearity vs Output Code**



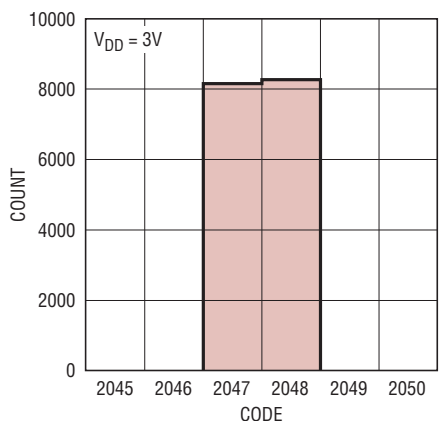
236012 G02

**Integral and Differential Nonlinearity vs Reference Voltage (TS8 Package)**



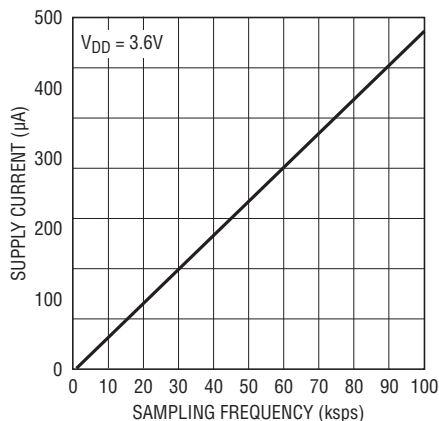
236012 G03

**Histogram for 16384 Conversions**



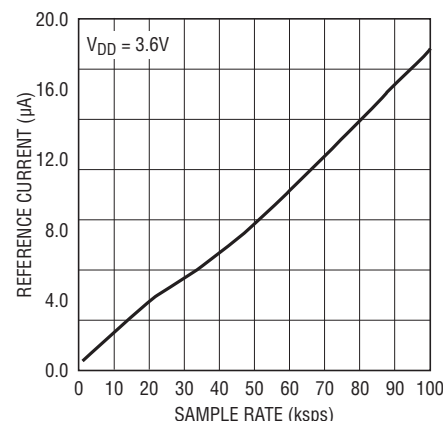
236012 G04

**Supply Current vs Sample Rate**



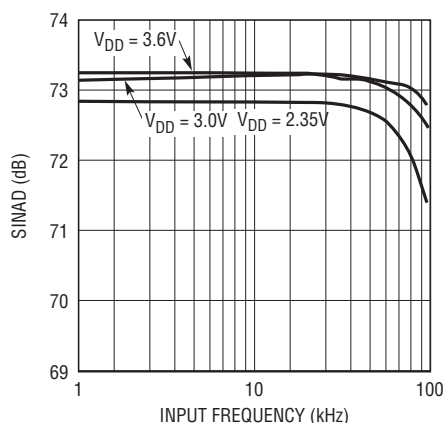
236012 G05

**Reference Current vs Sample Rate (TS8 Package)**



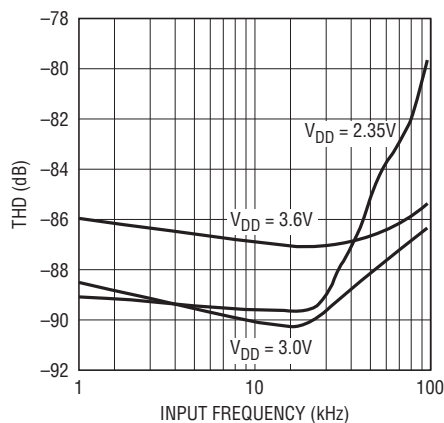
236012 G06

**SINAD vs Input Frequency**



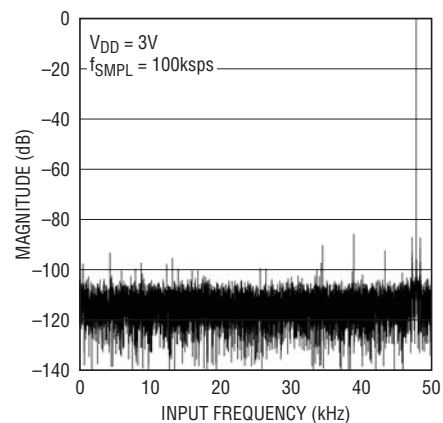
236012 G07

**THD vs Input Frequency**



236012 G08

**48kHz Sine Wave 8192 FFT Plot**

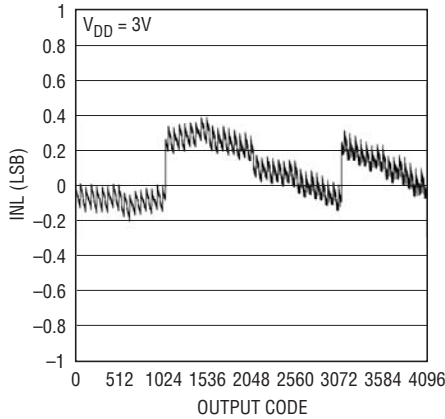


236012 G09

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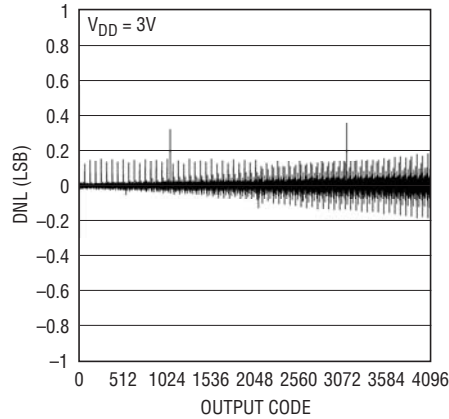
## TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$ , $V_{DD} = 0V_{DD} = V_{REF}$ (LTC2361, Note 4)

**Integral Nonlinearity vs Output Code**



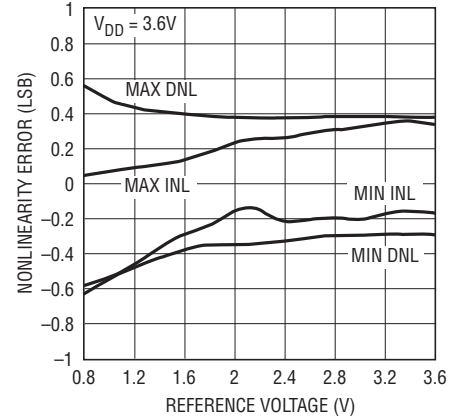
236012 G10

**Differential Nonlinearity vs Output Code**



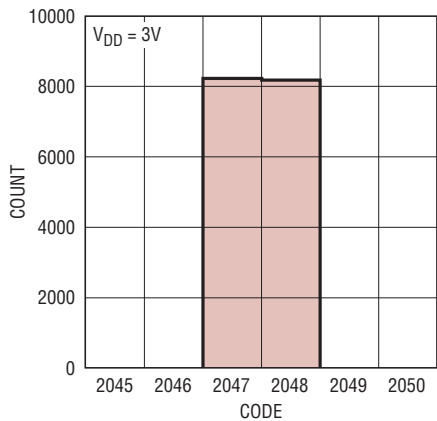
236012 G11

**Integral and Differential Nonlinearity vs Reference Voltage (TS8 Package)**



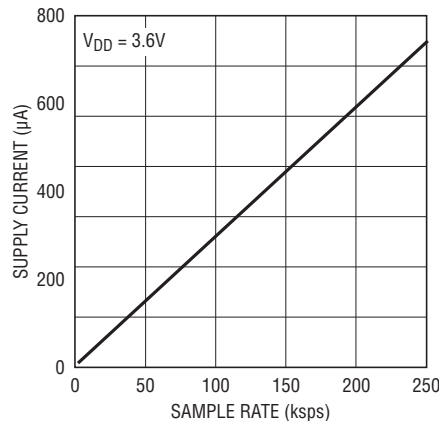
236012 G12

**Histogram for 16384 Conversions**



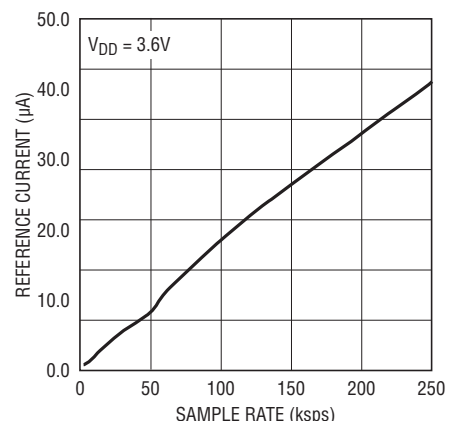
236012 G13

**Supply Current vs Sample Rate**



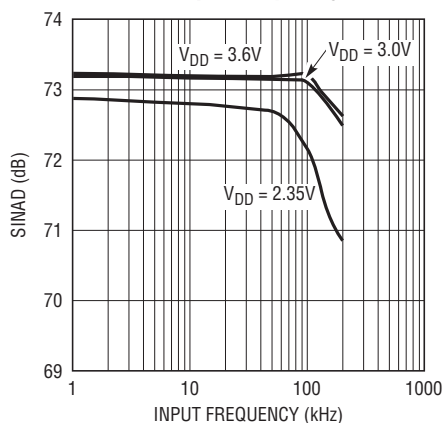
236012 G14

**Reference Current vs Sample Rate (TS8 Package)**



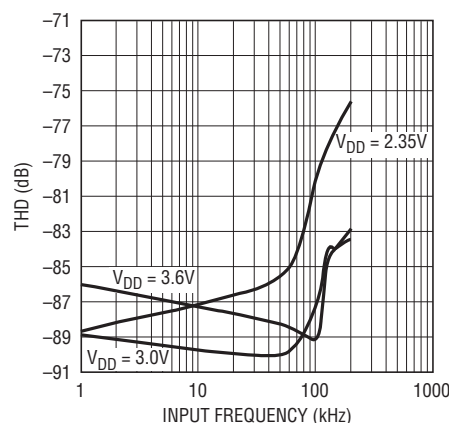
236012 G15

**SINAD vs Input Frequency**



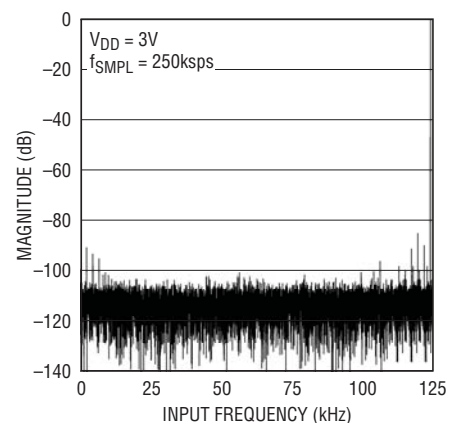
236012 G16

**THD vs Input Frequency**



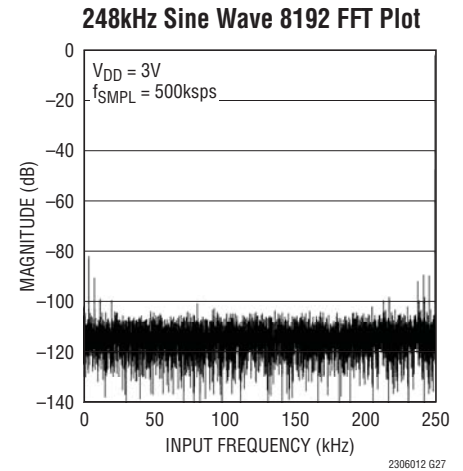
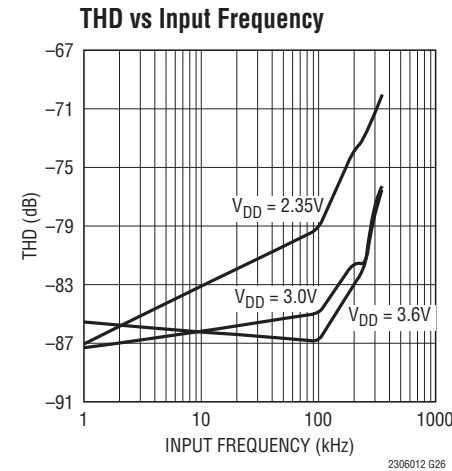
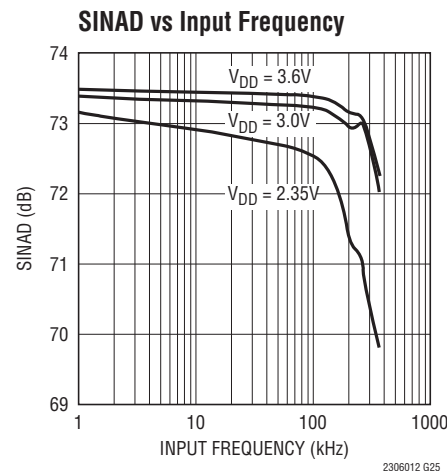
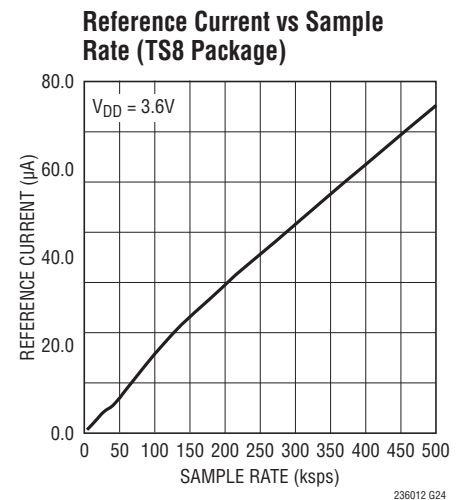
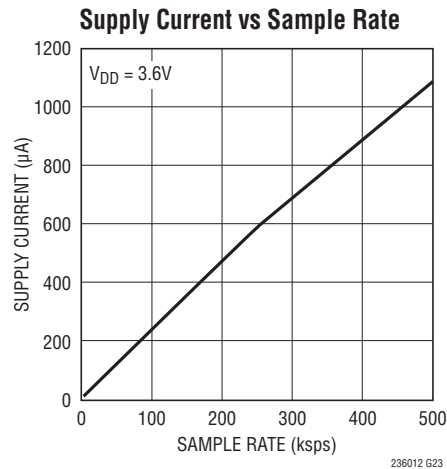
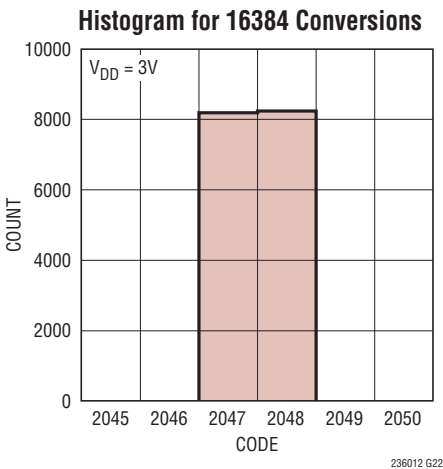
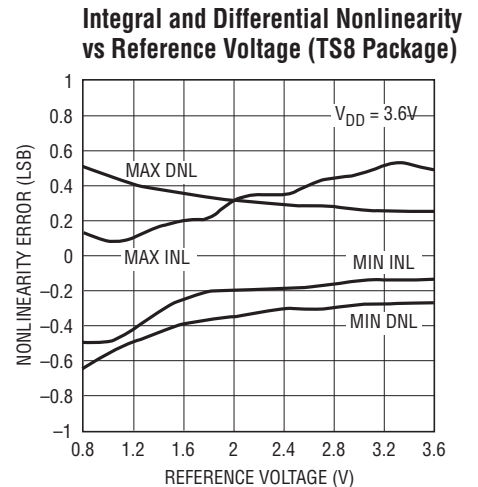
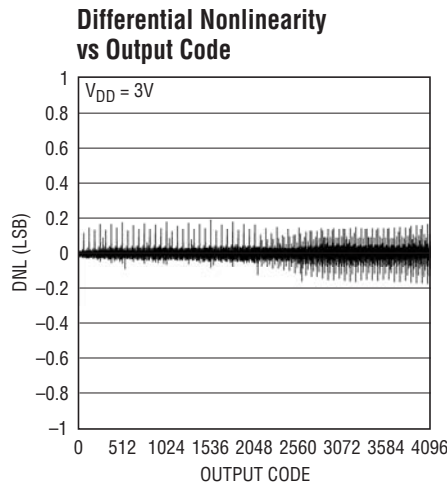
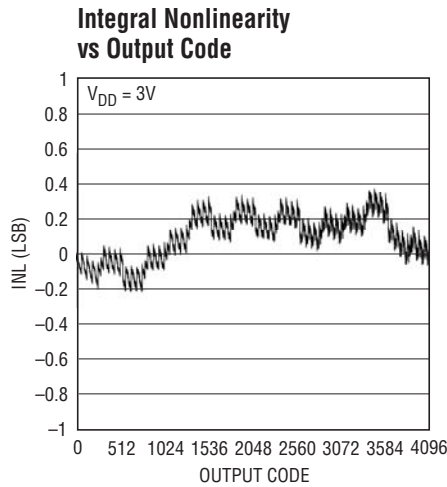
236012 G17

**124kHz Sine Wave 8192 FFT Plot**



236012 G18

## TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$ , $V_{DD} = 0V_{DD} = V_{REF}$ (LTC2362, Note 4)



## PIN FUNCTIONS

### S6 Package

**V<sub>DD</sub> (Pin 1):** Positive Supply. The V<sub>DD</sub> range is 2.35V to 3.6V. V<sub>DD</sub> also defines the input span of the ADC, 0V to V<sub>DD</sub>. Bypass to GND and to a solid ground plane with a 2.2μF ceramic capacitor (or 2.2μF tantalum in parallel with 0.1μF ceramic).

**GND (Pin 2):** Ground. The GND pin must be tied directly to a solid ground plane.

**A<sub>IN</sub> (Pin 3):** Analog Input. A<sub>IN</sub> is a single-ended input with respect to GND with a range from 0V to V<sub>DD</sub>.

**SCK (Pin 4):** Shift Clock Input. The SCK serial clock synchronizes the serial data transfer. SDO data transitions on the falling edge of SCK.

**SDO (Pin 5):** Three-State Serial Data Output. The A/D conversion result is shifted out on SDO as a serial data stream with MSB first. The data stream consists of 12 bits of conversion data followed by trailing zeros.

**CONV (Pin 6):** Convert Input. This active high signal starts a conversion on the rising edge. The device automatically powers down after conversion. A logic low on this input enables the SDO pin, allowing the data to be shifted out.

### TS8 Package

**V<sub>DD</sub> (Pin 1):** Positive Supply. The V<sub>DD</sub> range is 2.35V to 3.6V. Bypass to GND and to a solid ground plane with a 2.2μF ceramic capacitor (or 2.2μF tantalum in parallel with 0.1μF ceramic).

**V<sub>REF</sub> (Pin 2):** Reference Input. V<sub>REF</sub> defines the input span of the ADC, 0V to V<sub>REF</sub>. The V<sub>REF</sub> range is 1.4V to V<sub>DD</sub>. Bypass to GND and to a solid ground plane with a 2.2μF ceramic capacitor (or 2.2μF tantalum in parallel with 0.1μF ceramic).

**GND (Pin 3):** Ground. The GND pin must be tied directly to a solid ground plane.

**A<sub>IN</sub> (Pin 4):** Analog Input. A<sub>IN</sub> is a single-ended input with respect to GND with a range from 0V to V<sub>REF</sub>.

**OV<sub>DD</sub> (Pin 5):** Output Driver Supply for SDO. The OV<sub>DD</sub> range is 1V to V<sub>DD</sub>. Bypass to GND and to a solid ground plane with a 2.2μF ceramic capacitor (or 2.2μF tantalum in parallel with 0.1μF ceramic). OV<sub>DD</sub> can be driven separately from V<sub>DD</sub> and OV<sub>DD</sub> can be higher than V<sub>DD</sub>.

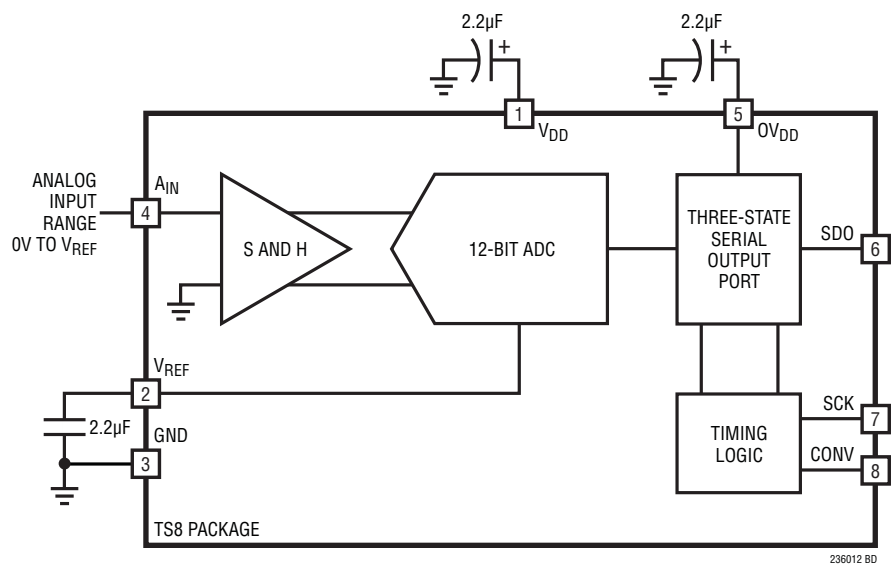
**SDO (Pin 6):** Three-State Serial Data Output. The A/D conversion result is shifted out on SDO as a serial data stream with MSB first. The data stream consists of 12 bits of conversion data followed by trailing zeros.

**SCK (Pin 7):** Shift Clock Input. The SCK serial clock synchronizes the serial data transfer. SDO data transitions on the falling edge of SCK.

**CONV (Pin 8):** Convert Input. This active high signal starts a conversion on the rising edge. The device automatically powers down after conversion. A logic low on this input enables the SDO pin, allowing the data to be shifted out.



BLOCK DIAGRAM



TIMING DIAGRAMS

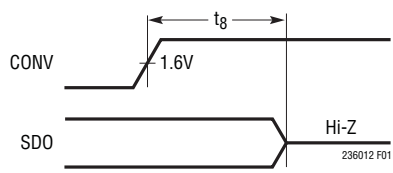


Figure 1. SDO Into Hi-Z State After CONV Rising Edge

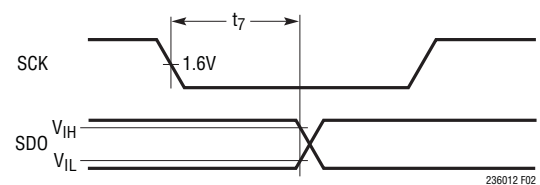


Figure 2. SDO Data Valid Hold Time After SCK Falling Edge

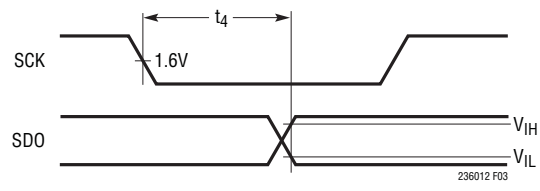


Figure 3. SDO Data Valid Access Time After SCK Falling Edge

## APPLICATIONS INFORMATION

### DC PERFORMANCE

The noise of an ADC can be evaluated in two ways: signal-to-noise ratio (SNR) in the frequency domain and histogram in the time domain. The LTC2360/LTC2361/LTC2362 excel in both. Figure 5 demonstrates that the LTC2360/LTC2361/LTC2362 have an SNR of over 73dB. The noise in the time domain histogram is the transition noise associated with a 12-bit resolution ADC which can be measured with a fixed DC signal applied to the input of the ADC. The resulting output codes are collected over a large number of conversions. The shape of the distribution of codes will give an indication of the magnitude of the transition noise. In Figure 4, the distribution of output codes is shown for a DC input that has been digitized 16384 times. The distribution is Gaussian and the RMS code transition is about 0.32LSB. This corresponds to a noise level of 73dB relative to a full scale of 3V.

### DYNAMIC PERFORMANCE

The LTC2360/LTC2361/LTC2362 have excellent high speed sampling capability. Fast fourier transform (FFT) test techniques are used to test the ADCs' frequency response, distortion and noise at the rated throughput. By applying a low distortion sine wave and analyzing the digital output using an FFT algorithm, the ADCs' spectral content can be examined for frequencies outside the fundamental. Figures 5 and 6 show typical LTC2361 and LTC2362 FFT plots respectively.

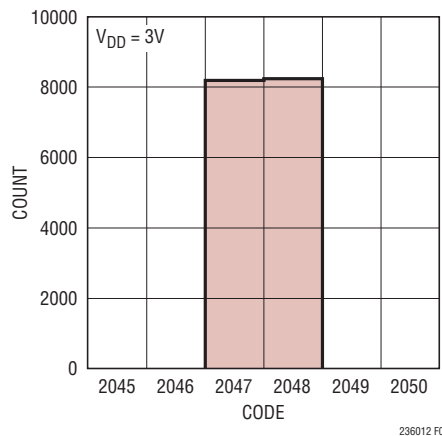


Figure 4. Histogram for 16384 Conversions

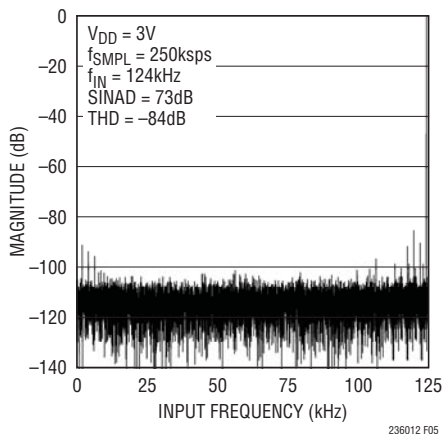


Figure 5. LTC2361 FFT Plot

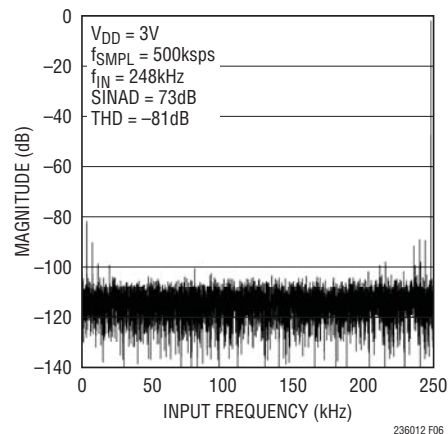


Figure 6. LTC2362 FFT Plot

## APPLICATIONS INFORMATION

### Signal-to-Noise plus Distortion Ratio

The signal-to-noise plus distortion ratio (SINAD) is the ratio between the RMS amplitude of the fundamental input frequency to the RMS amplitude of all other frequency components at the A/D output. The output is band limited to frequencies from above DC and below half the sampling frequency. Figure 6 shows a typical FFT with a 500kHz sampling rate and a 248kHz input. The dynamic performance is excellent for input frequencies up to and beyond the Nyquist frequency of 250kHz.

### Effective Number of Bits

The effective number of bits (ENOB) is a measurement of the resolution of an ADC and is directly related to SINAD by the equation:

$$\text{ENOB} = \frac{\text{SINAD} - 1.76}{6.02}$$

where ENOB is the effective number of bits of resolution and SINAD is expressed in dB. At the maximum sampling

rate of 500kHz, the LTC2362 maintains ENOB above 11 bits up to the Nyquist input frequency of 250kHz (refer to Figure 7).

### Total Harmonic Distortion

The total harmonic distortion (THD) is the ratio of the RMS sum of all harmonics of the input signal to the fundamental itself. The out-of-band harmonics alias into the frequency band between DC and half the sampling frequency. THD is expressed as:

$$\text{THD} = 20 \log \frac{\sqrt{V_2^2 + V_3^2 + V_4^2 + \dots + V_n^2}}{V_1}$$

where  $V_1$  is the RMS amplitude of the fundamental frequency and  $V_2$  through  $V_n$  are the amplitudes of the second through nth harmonics. THD vs Input Frequency is shown in Figure 8. The LTC2362 has excellent distortion performance up to the Nyquist frequency and beyond.

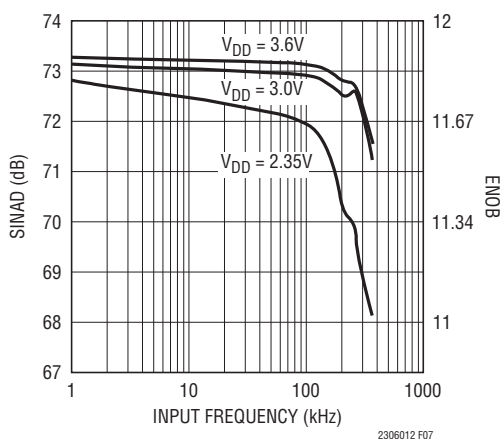


Figure 7. LTC2362 ENOB and SINAD vs Input Frequency

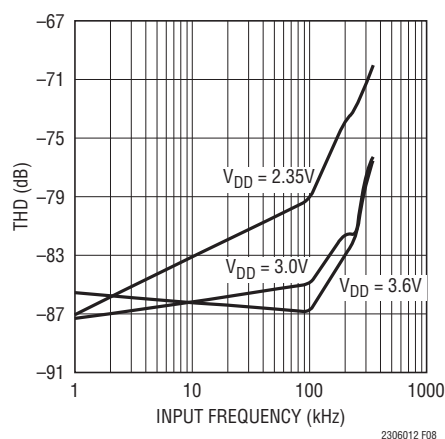


Figure 8. LTC2362 THD vs Input Frequency

## APPLICATIONS INFORMATION

### Intermodulation Distortion

If the ADC input signal consists of more than one spectral component, the ADC transfer function nonlinearity can produce intermodulation distortion (IMD) in addition to THD. IMD is the change in one sinusoidal input caused by the presence of another sinusoidal input at a different frequency.

If two pure sine waves of frequencies  $f_a$  and  $f_b$  are applied to the ADC input, nonlinearities in the ADC transfer function can create distortion products at the sum and difference frequencies of  $mf_a \pm nf_b$ , where  $m$  and  $n = 0, 1, 2, 3$ , etc. For example, the 2nd order IMD terms include  $(f_a \pm f_b)$ . If the two input sine waves are equal in magnitude, the value (in decibels) of the 2nd order IMD products can be expressed by the following formula:

$$\text{IMD}(f_a \pm f_b) = 20 \log \frac{\text{Amplitude at } (f_a \pm f_b)}{\text{Amplitude at } f_a}$$

### Peak Harmonic or Spurious Noise

The peak harmonic or spurious noise is the largest spectral component excluding the input signal and DC. This value is expressed in decibels relative to the RMS value of a full-scale input signal.

### Full-Power and Full-Linear Bandwidth

The full-power bandwidth is that input frequency at which the amplitude of reconstructed fundamental is reduced by 3dB for full-scale input signal.

The full-linear bandwidth is the input frequency at which the SINAD has dropped to 68dB (11 effective bits). The LTC2362 has been designed to optimize input bandwidth, allowing the ADC to undersample input signals with frequencies above the converter's Nyquist frequency. The noise floor stays very low at high frequencies; SINAD becomes dominated by distortion at frequencies far beyond Nyquist.

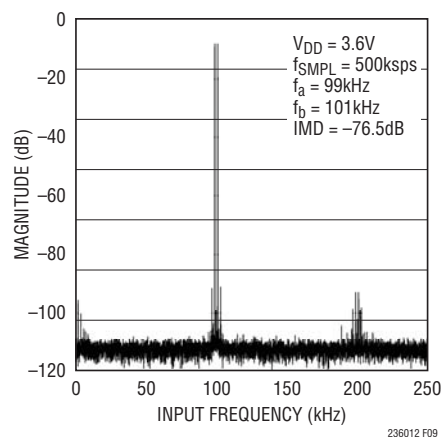


Figure 9. LTC2362 Intermodulation Distortion Plot

## APPLICATIONS INFORMATION

### OVERVIEW

The LTC2360/LTC2361/LTC2362 use a successive approximation algorithm and internal sample-and-hold circuit to convert an analog signal to a 12-bit serial output. All devices operate from a single 2.35V to 3.6V supply. The conversion time of the devices is controlled by an internal oscillator, which allows the LTC2360/LTC2361/LTC2362 to sample at a rate of 100ksps, 250ksps and 500ksps respectively.

The LTC2360/LTC2361/LTC2362 contain a 12-bit, switched-capacitor ADC, a sample-and-hold, a serial interface (see Block Diagram) and are available in tiny 6- or 8-lead TSOT-23 packages.

The S6 package of the LTC2360/LTC2361/LTC2362 uses  $V_{DD}$  as the reference and has an analog input range of 0V to  $V_{DD}$ . The ADC samples the analog input with respect to GND and outputs the result through the serial interface.

The TS8 package provides two additional pins: a reference pin,  $V_{REF}$ , and an output supply pin,  $OV_{DD}$ . The ADC can operate with reduced spans down to 1.4V and achieve 342 $\mu$ V resolution.  $OV_{DD}$  controls the output swing of the digital output pin, SDO, and allows the device to communicate with 1.8V, 2.5V or 3V digital systems.

### SERIAL INTERFACE

The LTC2360/LTC2361/LTC2362 communicate with micro-controllers, DSPs and other external circuitry via a 3-wire interface. Figure 10 shows the operating sequence of the serial interface.

### Data Transfer

A rising CONV edge starts a conversion and disables SDO. After the conversion, the ADC automatically goes into sleep mode, drawing only leakage current.

CONV going low enables SDO and clocks out the MSB bit, B11. SCK then synchronizes the data transfer with each bit being transmitted on the falling SCK edge and can be captured on the rising SCK edge. After completing the data transfer, if further SCK clocks are applied with CONV low, SDO will output zeros indefinitely (see Figure 10). For example, 16-clocks at SCK will produce the 12-bit data and four trailing zeros on SDO.

### SLEEP MODE

The LTC2360/LTC2361/LTC2362 enter sleep mode to save power after each conversion if CONV remains high. In sleep mode, all bias currents are shut down and only leakage currents remain (about 0.1 $\mu$ A). The sample-and-hold is in hold mode while the ADC is in sleep mode. The ADC returns to sample mode after the falling edge of CONV during power-up (see Figure 10).

### Exiting Sleep Mode and Power-Up Time

By taking CONV low, the ADC powers up and acquires an input signal completely after the acquisition time ( $t_{ACQ}$ ). After  $t_{ACQ}$ , the ADC can perform a conversion as described in the Serial Interface section (see Figure 10).

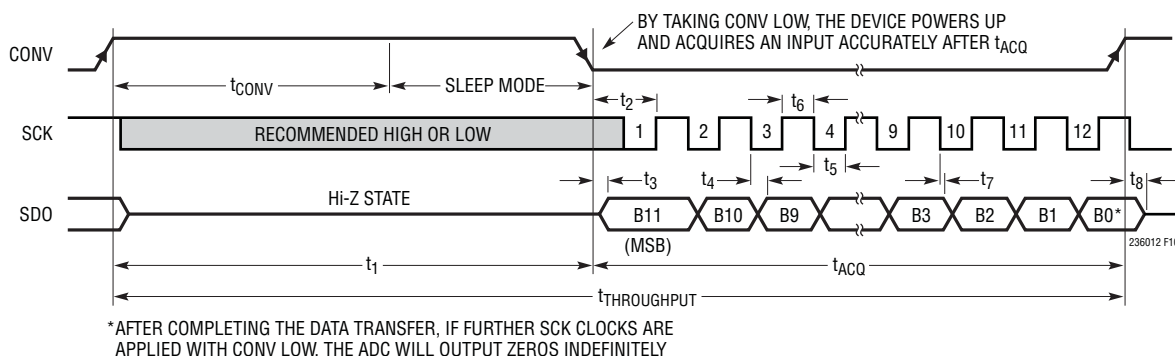


Figure 10. LTC2360/LTC2361/LTC2362 Serial Interface Timing Diagram

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## APPLICATIONS INFORMATION

### ACHIEVING MICROPOWER PERFORMANCE

With typical operating currents of 0.5mA, 0.75mA and 1.1mA for the LTC2360/LTC2361/LTC2362 and automatically entering sleep mode right after a conversion, these devices achieve extremely low power consumption over a wide range of sample rates (see Figure 11). The sleep mode allows the supply current to drop with reduced sample rate. Several things must be taken into account to achieve such low power consumption.

#### Minimize Power Consumption in Sleep Mode

The LTC2360/LTC2361/LTC2362 enter sleep mode after each conversion if CONV remains high and draw only leakage current (see Figure 10). If the CONV input is not running rail-to-rail, the input logic buffer will draw current. This current may be large compared to the typical supply

current. To obtain the lowest supply current, bring the CONV pin to GND when it is low and to  $V_{DD}$  when it is high.

After the conversion with CONV staying high, the converter is in sleep mode and draws only leakage current. The status of the SCK input has no effect on supply current during this time. For the best performance, hold SCK either high or low while the ADC is converting.

#### Minimize the Device Active Time

In systems that have significant time between conversions, the ADC draws a minimal amount of power. Figures 12 and 13 show two ways to minimize the amount of time the ADC draws power. In Figure 12, the ADC draws power during  $t_{ACQ}$  and  $t_{CONV}$  and is in sleep mode for the rest of the time. The conversion results are available at the next CONV falling edge. In Figure 13, the ADC draws twice the power than that in Figure 12, but the conversion results are available during  $t_{DATA}$ . The user can use the fastest SCK available in the system to shorten data transfer time,  $t_{DATA}$  as long as  $t_4$  and  $t_7$  are not violated.

#### SDO Loading

Capacitive loading on the digital output can increase power consumption. A 100pF capacitor on the SDO pin can add more than 50 $\mu$ A to the supply current at a 200kHz clock frequency. An extra 50 $\mu$ A or so of current goes into charging and discharging the load capacitor. The same goes for digital lines driven at a high frequency by any logic. The  $C \cdot V \cdot f$  currents must be evaluated with the troublesome ones minimized.

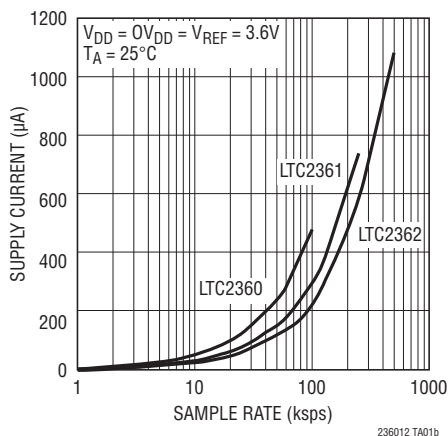


Figure 11. Supply Current vs Sample Rate

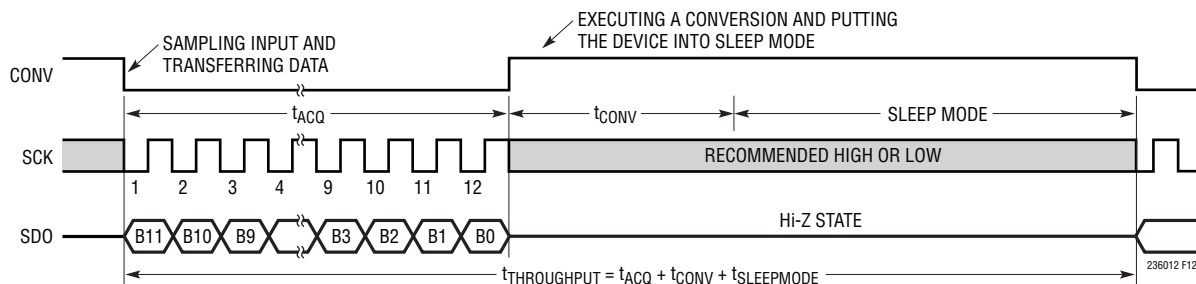


Figure 12. Minimize the Time When the Device Draws Power, While the Conversion Results are Available After the Device Wakes Up

## APPLICATIONS INFORMATION

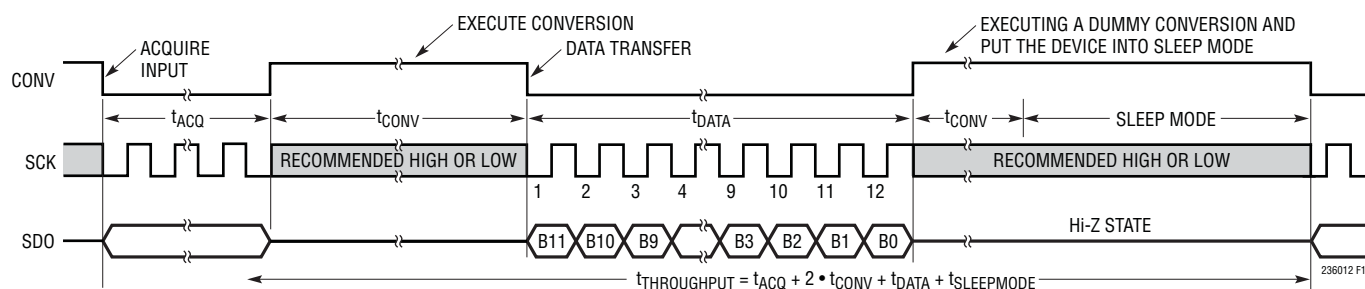


Figure 13. Minimize the Time When the Device Draws Power, While the Conversion Results are Available Right After Conversion

## SINGLE-ENDED ANALOG INPUT

## Driving the Analog Input

The analog input of the LTC2360/LTC2361/LTC2362 is easy to drive. The input draws only one small current spike while charging the sample-and-hold capacitor with the ADC going into track mode. During the conversion, the analog input draws only a small leakage current. If the source impedance of the driving circuit is low, then the input of the LTC2360/LTC2361/LTC2362 can be driven directly. As source impedance increases, so will acquisition time. For minimum acquisition time with high source impedance, a buffer amplifier should be used. The main requirement is that the amplifier driving the analog input must settle after the small current spike before the next conversion starts (settling time must be less than  $t_{ACQ}$  for full throughput rate). While choosing an input amplifier, also keep in mind the amount of noise and harmonic distortion the amplifier contributes.

## Choosing an Input Amplifier

Choosing an input amplifier is easy if a few requirements are taken into consideration. First, to limit the magnitude of the voltage spike seen by the amplifier from charging the sampling capacitor, choose an amplifier that has a low output impedance ( $<100\Omega$ ) at the closed-loop bandwidth frequency. For example, if an amplifier is used in a gain of 1 and has a unity-gain bandwidth of 10MHz, then the output impedance at 10MHz must be less than  $100\Omega$ . The second requirement is that the closed-loop bandwidth must be greater than 8MHz to ensure adequate small-signal settling for full throughput rate. If slower op amps are

used, more time for settling can be provided by increasing the time between conversions. The best choice for an op amp to drive the LTC2360/LTC2361/LTC2362 will depend on the application. Generally, applications fall into two categories: AC applications where dynamic specifications are most critical and time domain applications where DC accuracy and settling time are most critical. The following list is a summary of the op amps that are suitable for driving the LTC2360/LTC2361/LTC2362. (More detailed information is available on the Linear Technology website at [www.linear.com](http://www.linear.com).)

**LTC1566-1:** Low Noise 2.3MHz Continuous Time Low-pass Filter.

**LT<sup>®</sup>1630:** Dual 30MHz Rail-to-Rail Voltage FB Amplifier. 2.7V to  $\pm 15V$  supplies. Very high  $A_{VOL}$ , 500 $\mu V$  offset and 520ns settling to 0.5LSB for a 4V swing. THD and noise are  $-93dB$  to 40kHz and below 1LSB to 320kHz ( $A_V = 1$ ,  $2V_{P-P}$  into 1k,  $V_S = 5V$ ), making the part excellent for AC applications (to 1/3 Nyquist) where rail-to-rail performance is desired. Quad version is available as LT1631.

**LTC6241:** Dual 18MHz, Low Noise, Rail-to-Rail, CMOS Voltage FB Amplifier. 2.8V to 6V supplies. Very high  $A_{VOL}$  and 125 $\mu V$  offset. It is suitable for applications with a single 5V supply. Quad version is available as LTC6242.

**LT1797:** Unity-Gain Stable 10MHz, Rail-to-Rail Voltage Feedback Amplifier.

**LT1801:** 180MHz GBWP,  $-75dBc$  at 500kHz, 2mA/Amplifier, 8.5nV/ $\sqrt{Hz}$ .

**LT6203:** 100MHz GBWP,  $-80dBc$  Distortion at 1MHz, Unity-Gain Stable, R-R In and Out, 3mA/Amplifier, 1.9nV/ $\sqrt{Hz}$ .



## APPLICATIONS INFORMATION

### Input Filtering and Source Impedance

The noise and the distortion of the input amplifier and other circuitry must be considered since they will add to the LTC2360/LTC2361/LTC2362 noise and distortion. The small-signal bandwidth of the sample-and-hold circuit is 10MHz. Any noise or distortion products that are present at the analog inputs will be summed over this entire bandwidth. Noisy input circuitry should be filtered prior to the analog inputs to minimize noise. A simple 1-pole RC filter is sufficient for many applications. For example, Figure 14 shows a 220pF capacitor from  $A_{IN}$  to ground and a 51Ω source resistor to limit the input bandwidth to 10MHz. The 220pF capacitor also acts as a charge reservoir for the input sample-and-hold and isolates the ADC input from sampling-glitch sensitive circuitry. High quality capacitors and resistors should be used since these components can add distortion. NPO and silvermica type dielectric capacitors have excellent linearity. Carbon surface mount resistors can generate distortion from self heating and from damage that may occur during soldering. Metal film surface mount resistors are much less susceptible to both problems. When high amplitude unwanted signals are close in frequency to the desired signal frequency, a multiple pole filter is required. High external source resistance, combined with the 20pF of input capacitance, will reduce the rated 10MHz bandwidth and increase acquisition time beyond 500ns.

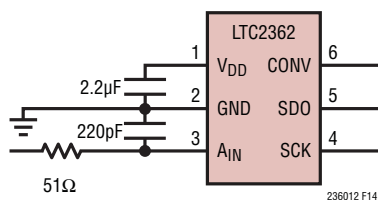


Figure 14. RC Input Filter

### Reference Input

On the TS8 package of the LTC2360/LTC2361/LTC2362, the voltage on the  $V_{REF}$  pin defines the full-scale range of the ADC. The reference voltage can range from  $V_{DD}$  down to 1.4V.

### Input Range

The analog input of the LTC2360/LTC2361/LTC2362 is driven single-ended with respect to GND from a single supply. The input may swing up to  $V_{DD}$  for the S6 package and to  $V_{REF}$  for the TS8 package. The 0V to 2.5V range is also ideally suited for single-ended input use with  $V_{DD}$  or  $V_{REF} = 2.5V$  for single supply applications. If the difference between the  $A_{IN}$  input and GND exceeds  $V_{DD}$  for the S6 package or  $V_{REF}$  for the TS8 package, the output code will stay fixed at all ones, and if this difference goes below 0V, the output code will stay fixed at all zeros.

Figure 15 shows the ideal input/output characteristics for the LTC2360/LTC2361/LTC2362. The code transitions occur midway between successive integer LSB values (i.e., 0.5LSB, 1.5LSB, 2.5LSB, ...,  $FS - 1.5LSB$ ). The output code is straight binary with  $1LSB = V_{DD}/4096$  for the S6 package and  $1LSB = V_{REF}/4096$  for the TS8 package.

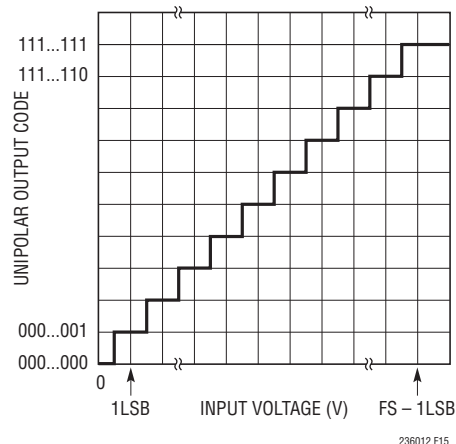


Figure 15. Transfer Characteristics



## APPLICATIONS INFORMATION

### BOARD LAYOUT AND BYPASSING

Wire wrap boards are not recommended for high resolution and/or high speed A/D converters. To obtain the best performance from the LTC2360/LTC2361/LTC2362, a printed circuit board with ground plane is required. Layout for the printed circuit board should ensure that digital and analog signal lines are separated as much as possible. In particular, care should be taken not to run any digital track alongside an analog signal track or underneath the ADC. The analog input should be screened by the ground plane.

High quality tantalum and ceramic bypass capacitors should be used at the  $V_{DD}$  pin as shown in the Block Diagram on the first page of this data sheet. For optimum performance, a  $2.2\mu\text{F}$  surface mount AVX capacitor with a  $0.1\mu\text{F}$  ceramic is recommended for the  $V_{DD}$  and  $V_{REF}$  pins. Alternatively,  $2.2\mu\text{F}$  ceramic chip capacitors such as Murata GRM235Y5V106Z016 may be used. The capacitors must be located as close to the pins as possible. The traces

connecting the pins and the bypass capacitors must be kept short and should be made as wide as possible.

Figure 16 shows the recommended system ground connections. All analog circuitry grounds should be terminated at the LTC2360/LTC2361/LTC2362. The ground return from the LTC2360/LTC2361/LTC2362 to the power supply should be low impedance for noise free operation. Digital circuitry grounds must be connected to the digital supply common.

In applications where the ADC data outputs and control signals are connected to a continuously active microprocessor bus, it is possible to get errors in the conversion results. These errors are due to feedthrough from the microprocessor to the successive approximation comparator. The problem can be eliminated by forcing the microprocessor into a wait state during conversion or by using three-state buffers to isolate the ADC data bus.

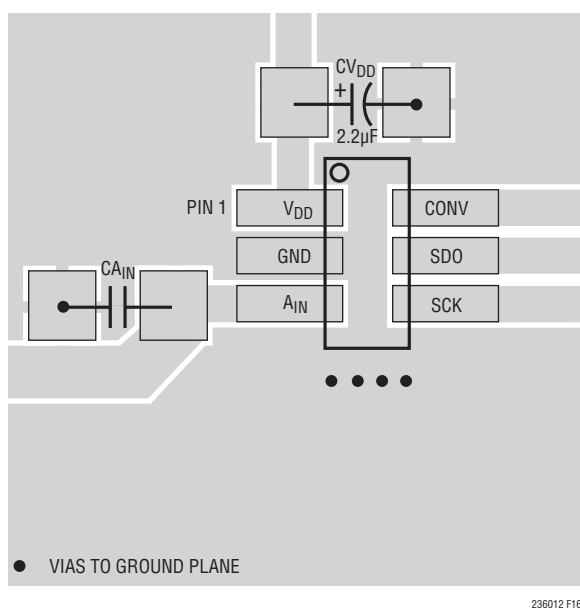
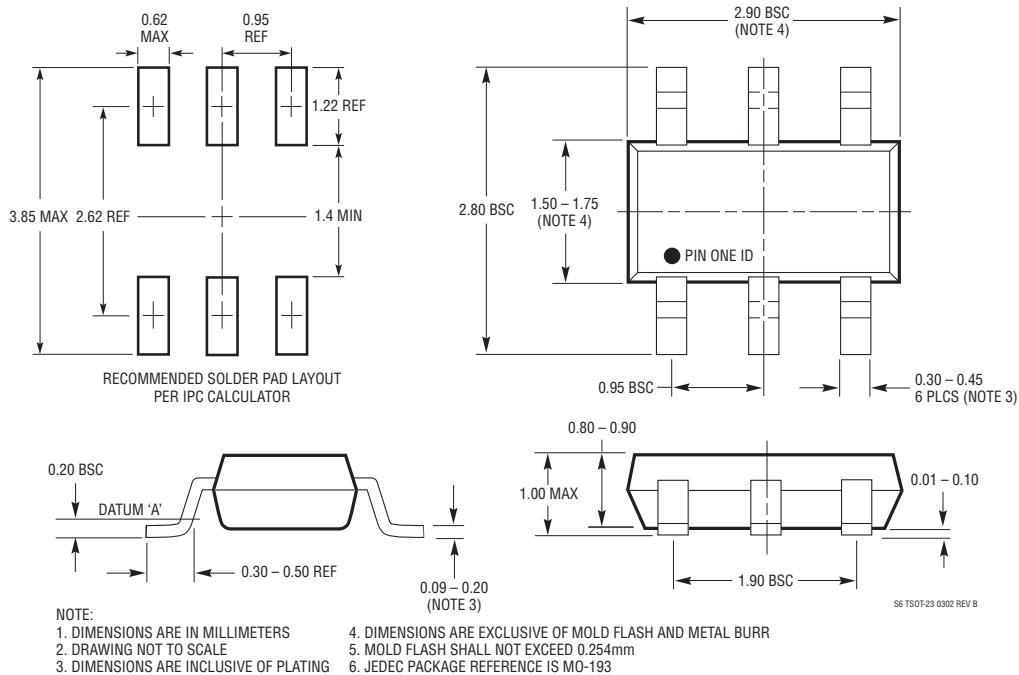


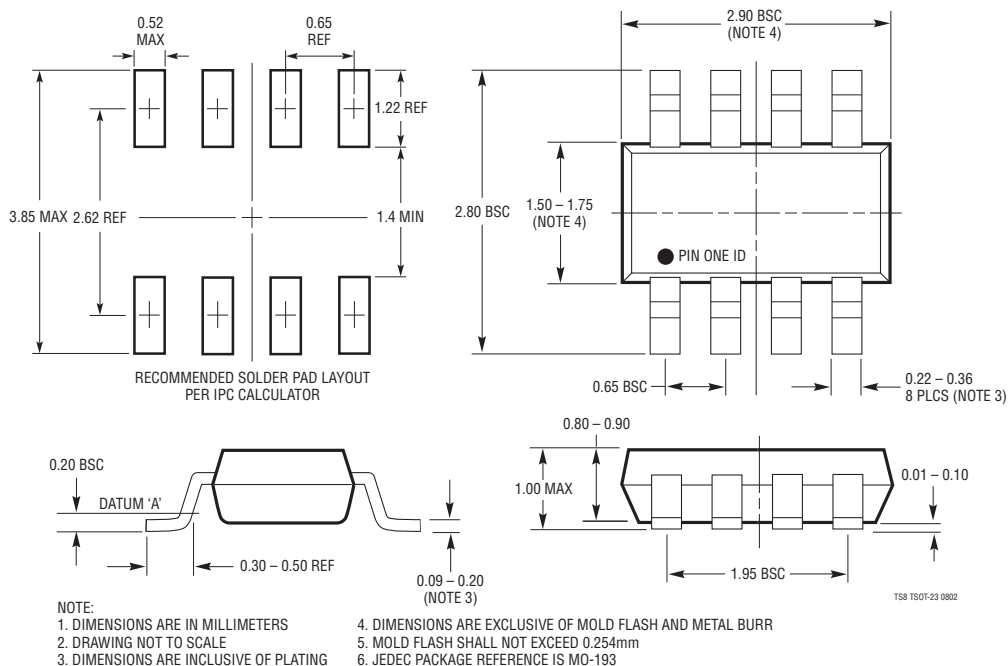
Figure 16. Power Supply Ground Practice

# PACKAGE DESCRIPTION

## S6 Package 6-Lead Plastic TSOT-23 (Reference LTC DWG # 05-08-1636)



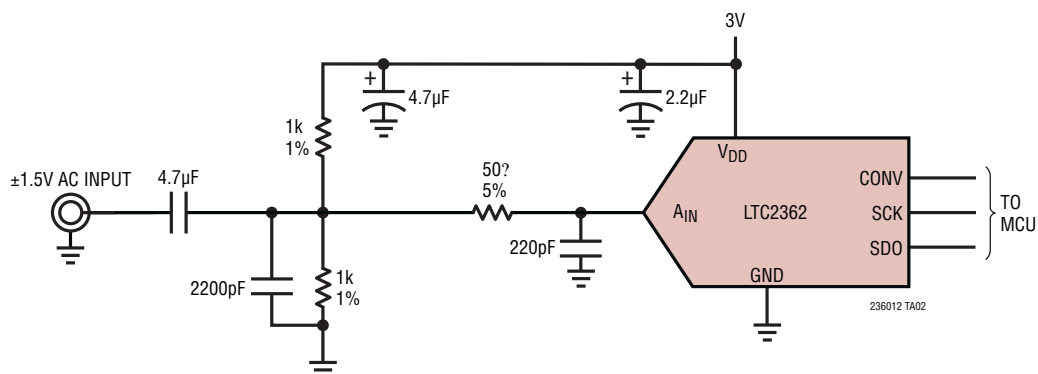
## TS8 Package 8-Lead Plastic TSOT-23 (Reference LTC DWG # 05-08-1637)



# LTC2360/LTC2361/LTC2362

## TYPICAL APPLICATION

Recommended AC Test Circuitry for the LTC2362



## RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
<b>ADCs</b>		
LTC1403/LTC1403A	12-/14-Bit, 2.8Msps Serial Sampling ADC	3V, Differential Input, 12mW, MSOP Package
LTC1407/LTC1407A	12-/14-Bit, 3Msps Simultaneous Sampling ADC	3V, 2-Channel Differential, 14mW, MSOP Package
LTC1860	12-Bit, 250ksps Serial ADC	5V Supply, 1-Channel, 4.3mW, MSOP-8 Package
LTC1860L	12-Bit, 150ksps Serial ADC	3V Supply, 1-Channel, 1.3mW, MSOP-8 Package
LTC1861	12-Bit, 250ksps Serial ADC	5V Supply, 2-Channel, 4.3mW, MSOP-8 Package
LTC1861L	12-Bit, 150ksps Serial ADC	3V Supply, 2-Channel, 1.3mW, MSOP-8 Package
LTC1863	12-Bit, 200ksps Serial ADC 8-Channel ADC	5V Supply, 6.5mW, SSOP-16 Package, Pin Compatible to LTC1863L, LTC1867
LTC1863L	12-Bit, 250ksps Serial ADC 8-Channel ADC	5V Supply, 2.2mW, SSOP-16 Package, Pin Compatible to LTC1863, LTC1867L
LTC1864/LTC1865	16-Bit, 250ksps Serial ADC	5V Supply, 1 and 2 Channel, 4.3mW, MSOP Package
LTC1867	16-Bit, 200ksps Serial ADC 8-Channel ADC	5V Supply, 6.5mW, SSOP-16 Package, Pin Compatible to LTC1863, LTC1867L
LTC1867L	16-Bit, 175ksps Serial ADC 8-Channel ADC	3V Supply, 2.2mW, SSOP-16 Package, Pin Compatible to LTC1863L, LTC1867
LTC2355/LTC2356	12-/14-Bit, 3.5Msps Serial ADCs	3.3V Supply, Differential Input, 18mW, MSOP Package
LTC2365/LTC2366	12-Bit, 1/3 Msps Serial ADCs in TSOT23	2.35V to 3.6V Supply, Pin and Software Compatible to LTC2360/LTC2361/LTC2362
<b>DACs</b>		
LTC1592	16-Bit, Serial SoftSpan™ I <sub>OUT</sub> DAC	±1LSB INL/DNL, Software Selectable Spans
LTC1666/LTC1667/LTC1668	12-/14-/16-Bit, 50Msps DACs	87dB SFDR, 20ns Settling Time
LTC2630	12-/10-/8-Bit Single V <sub>OUT</sub> DACs	SC70 6-Pin Package, Internal Reference, ±1LSB INL (12 Bits)
<b>References</b>		
LT1460-2.5	Micropower Series Voltage Reference	0.1% Initial Accuracy, 10ppm Drift
LT1461-2.5	Precision Voltage Reference	0.05% Initial Accuracy, 3ppm Drift
LT1790-2.5	Micropower Series Reference in SOT-23	0.05% Initial Accuracy, 10ppm Drift
LT6660	Ultra-Tiny Micropower Series Reference	2mm × 2mm DFN Package, 0.2% Initial Accuracy, 10ppm Drift

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