

General Description

The MAX907/MAX908/MAX909 are dual/guad/single. high-speed, ultra-low-power voltage comparators designed for use in systems powered from a single +5V supply; the MAX909 also accepts dual ±5V supplies. Their 40ns propagation delay (with 5mV input overdrive) is achieved with a power consumption of only 3.5mW per comparator. The wide input commonmode range extends from 200mV below ground (below the negative supply rail for the MAX909) to within 1.5V of the positive supply rail.

Because they are micropower, high-speed comparators that operate from a single +5V supply and include built-in hysteresis, these devices replace a variety of older comparators in a wide range of applications.

MAX907/MAX908/MAX909 outputs are TTL-compatible, requiring no external pullup circuitry. All inputs and outputs can be continuously shorted to either supply rail without damage. These easy-to-use comparators incorporate internal hysteresis to ensure clean output switching even when the devices are driven by a slow-moving input signal.

The MAX909 features complementary outputs and an output latch. A separate supply pin for extending the analog input range down to -5V is also provided.

The dual MAX907 and single MAX909 are available in 8-pin DIP and SO packages, and the guad MAX908 is available in 14-pin DIP and SO packages. These comparators are ideal for single +5V-supply applications that require the combination of high speed, precision, and ultra-low power dissipation.

Applications

Battery-Powered Systems

High-Speed A/D Converters

High-Speed V/F Converters

Line Receivers

Threshold Detectors/Discriminators

High-Speed Sampling Circuits

Zero-Crossing Detectors

Features

- ♦ 40ns Propagation Delay
- ♦ 700µA (3.5mW) Supply Current per Comparator
- ♦ Single 4.5V to 5.5V Supply Operation (or ±5V, MAX909 only)
- ♦ Wide Input Range Includes Ground (or -5V, MAX909 only)
- ♦ Low, 500µV Offset Voltage
- ♦ Internal Hysteresis Provides Clean Switching
- **♦ TTL-Compatible Outputs** (Complementary on MAX909)
- ♦ Input and Output Short-Circuit Protection
- ♦ Internal Latch (MAX909 only)

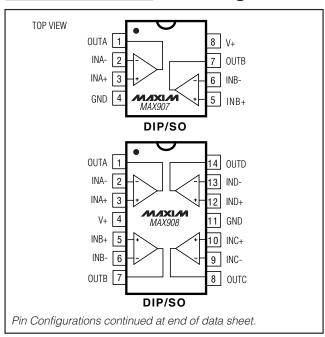
Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX907CPA	0°C to +70°C	8 Plastic DIP
MAX907CSA	0°C to +70°C	8 SO
MAX907EPA	-40°C to +85°C	8 Plastic DIP
MAX907ESA	-40°C to +85°C	8 SO
MAX907MSA/PR*	-55°C to +125°C	8 SO

Ordering Information continued at end of data sheet.

*Go to www.maxim-ic.com/PR-1 for details on high-reliability plastic processing.

Pin Configurations



NIXIN

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For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Positive Supply Voltage (V+ to GND)+6V
Negative Supply Voltage (V- to GND, MAX909 only)7V
Differential Input Voltage
MAX907/MAX9080.3V to (V+ + 0.3V)
MAX909(V 0.3V) to (V+ + 0.3V)
Common-Mode Input Voltage
MAX907/MAX9080.3V to (V+ + 0.3V)
MAX909(V 0.3V) to (V+ + 0.3V)
Latch Input Voltage (MAX909 only)0.3V to (V+ + 0.3V)
Input/Output Short-Circuit Duration to V+ or GND Continuous

Continuous Power Dissipation ($T_A = +70^{\circ}C$)
8-Pin Plastic DIP (derate 9.09mW/°C above +70°C)727mW
8-Pin SO (derate 5.88mW/°C above +70°C)471mW
14-Pin Plastic DIP (derate 10.00mW/°C above +70°C)800mW
14-Pin SO (derate 8.33mW/°C above +70°C)667mW
Operating Temperature Ranges:
MAX90_C0°C to +70°C
MAX90_E40°C to +85°C
MAX907MSA/PR55°C to +125°C
Storage Temperature Range65°C to +150°C
Lead Temperature (soldering, 10s)+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V+ = 5V, T_A = +25°C; MAX909 only: V- = 0V, V_{LATCH} = 0V; unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
Positive Trip Point	V _{TRIP+}	(Note 1)			2	4	mV	
Negative Trip Point	V _{TRIP} -	(Note 1)			-2	-4	mV	
Input Offset Voltage	Vos	(Note 2)			0.5	2.0	mV	
Input Bias Current	Ι _Β	$V_{CM} = 0V, V$	$I_{IN} = V_{OS}$		100	300	nA	
Input Offset Current	Ios	$V_{CM} = 0V, V$	$I_{IN} = V_{OS}$		25	50	nA	
Input Voltage Dange	V	(Nloton 2, 4)	MAX907/MAX908/MAX909	-0.2		V+ - 1.5	V	
Input Voltage Range	V _{CMR}	(Notes 3, 4)	MAX909 only: V- = -5V	-5.2		V+ - 1.5]	
Common-Mode Rejection Ratio	CMRR	(Notes 4, 5)	(Notes 4, 5)		50	150	μV/V	
Power-Supply Rejection Ratio	PSRR	(Notes 4, 6)	(Notes 4, 6)		50	200	μV/V	
Output High Voltage	V _{OH}	I _{SOURCE} = 100µA		3.0	3.5		V	
Output Low Voltage	Va.	I _{SINK} = 3.2mA			0.36	0.475	V	
Output Low Voltage	V _{OL}	I _{SINK} = 8mA	I _{SINK} = 8mA		0.4		\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	
Positive Supply Current per	1+	(Note 7)	MAX907/MAX908		0.7	1.0	mA	
Comparator	I+	(Note 7)	MAX909		1.2	1.8	IIIA	
Negative Supply Current	ļ-	MAX909 only: V- = -5V			60	100	μΑ	
Power Dissipation per	DD	(Nata O)	MAX907/MAX908		3.5	5.5	, no \ \ \	
Comparator	PD	(Note 8)	MAX909		6	10	mW	
Output Rise Time	t _r	V _{OUT} = 0.4V to 2.4V, C _L = 10pF			12		ns	
Output Fall Time	t _f	$V_{OUT} = 2.4V \text{ to } 0.4V, C_L = 10pF$			6		ns	

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ELECTRICAL CHARACTERISTICS (continued)

 $(V + = 5V, T_A = +25^{\circ}C; MAX909 \text{ only: } V - = 0V, V_{LATCH} = 0V; unless otherwise noted.)$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Propagation Delay	t _{PD+,} t _{PD-}	V _{IN} = 100mV, V _{OD} = 5mV (Note 9)	40 50		ns	
Differential Propagation Delay	Δt_{PD}	V _{IN} = 100mV, V _{OD} = 5mV (Note 10)		1		ns
Propagation Delay Skew	t _{PD} skew	MAX909 only: $V_{IN} = 100$ mV, $V_{OD} = 5$ mV (Note 11)		2		ns
Latch Input Voltage High	V _{IH}	(Note 12)	2.0			V
Latch Input Voltage Low	V _{IL}	(Note 12)			0.8	V
Latch Input Current	I _{IH} , I _{IL}	(Note 12)			20	μΑ
Latch Setup Time	ts	(Note 12)		2		ns
Latch Hold Time	t _h	(Note 12)	2		ns	

ELECTRICAL CHARACTERISTICS

(V+ = 5V, T_A = T_{MIN} to T_{MAX}; MAX909 only: V- = 0V, V_{LATCH} = 0V; unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
Positive Trip Point	V _{TRIP+}	(Note 1)			2	5	mV	
Negative Trip Point	V _{TRIP} -	(Note 1)			-2	-5	mV	
Input Offset Voltage	Vos	(Note 2)			1	3.5	mV	
Input Bias Current	IB	V _{CM} = 0V, V _{IN} = V	Vos		200	500	nA	
Input Offset Current	los	V _{CM} = 0V, V _{IN} = V	/os		50	100	nA	
		C/E temp.	MAX907/MAX908/MAX909	-0.2		V+ - 1.5		
Input Voltage Range	VCMR	(Notes 3, 4)	MAX909 only, V- = -5V	-5.2		V+ - 1.5	V	
		M temp. range (Notes 3, 4)	MAX907MSA/PR	-0.1		V+ - 1.5		
Common-Mode Rejection Ratio	CMRR	(Notes 4, 5)			75	300	μV/V	
Power-Supply Rejection Ratio	PSRR	(Notes 4, 6)			75	400	μV/V	
Output High Voltage	VoH	ISOURCE = 100µA	ISOURCE = 100µA		3.5		V	
			$T_A = T_{MAX}$			0.45		
Output Low Voltage	VoL	$I_{SINK} = 3.2 \text{mA}$	$T_{MIN} = 0^{\circ}C$			0.525	V	
			$T_{MIN} = -40^{\circ}C$			0.550		
Positive Supply Current per	1.	(Nlete 7)	MAX907/MAX908		0.8	1.2	A	
Comparator	I+	(Note 7)	MAX909		1.2	2.0	mA	
Negative Supply Current	I-	MAX909 only; V- = -5V			100	200	μV	
Power Dissipation per	DD	(1.1.1.2)	(N - 0)	MAX907/MAX908		4	7	14/
Comparator	PD	(Note 8)	MAX909		6	11	mW	

ELECTRICAL CHARACTERISTICS (continued)

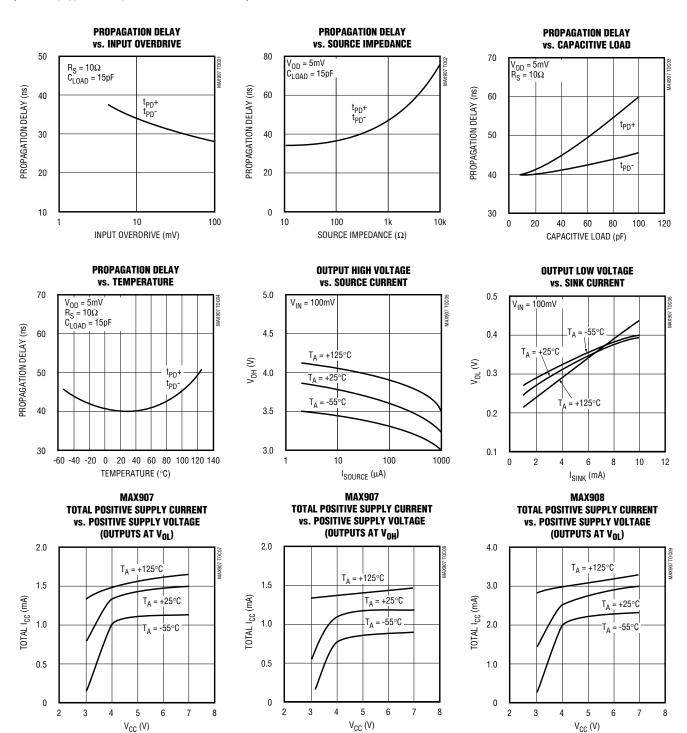
(V+ = 5V, TA = TMIN to TMAX; MAX909 only: V- = 0V, VLATCH = 0V; unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Propagation Delay	t _{PD+,} t _{PD-}	V _{IN} = 100mV, V _{OD} = 5mV (Note 9)		45	70	ns
Differential Propagation Delay	Δt_{PD}	V _{IN} = 100mV, V _{OD} = 5mV (Note 10)		2		ns
Propagation Delay Skew	t _{PD} skew	MAX909 only: $V_{IN} = 100$ mV, $V_{OD} = 5$ mV (Note 11)		4		ns
Latch Input Voltage High	V _{IH}	(Note 12)	2.0			V
Latch Input Voltage Low	VIL	(Note 12)			0.8	V
Latch Input Current	I _{IH} , I _{IL}	(Note 12)			20	μA
Latch Setup Time	ts	(Note 12)		4		ns
Latch Hold Time	t _h	(Note 12)		4		ns

- Note 1: Trip Point is defined as the input voltage required to make the comparator output change state. The difference between upper (V_{TRIP}+) and lower (V_{TRIP}-) trip points is equal to the width of the input-referred hysteresis zone (V_{HYST}). Specified for an input common-mode voltage (V_{CM}) of 0V (see Figure 1).
- Note 2: Input Offset Voltage is defined as the center of the input-referred hysteresis zone. Specified for V_{CM} = 0V (see Figure 1).
- **Note 3:** Inferred from the CMRR test. Note that a correct logic result is obtained at the output, provided that at least one input is within the V_{CMR} limits. Note also that either or both inputs can be driven to the upper or lower absolute maximum limit without damage to the part.
- **Note 4:** Tested with $V_{+} = 5.5V$ (and $V_{-} = 0V$ for MAX909). MAX909 also tested over the full analog input range (i.e., with $V_{-} = -5.5V$).
- Note 5: Tested over the full input voltage range (V_{CMR}).
- **Note 6:** Specified over the full tolerance of operating supply voltage: MAX907/MAX908 tested with 4.5V < V+ < 5.5V. MAX909 tested with 4.5V < V+ < 5.5V and with -5.5V < V- < 0V.
- Note 7: Positive Supply Current specified with the worst-case condition of all outputs at logic low (MAX907/MAX908), and with V+ = 5.5V.
- Note 8: Typical power specified with V+ = 5V: maximum with V+ = 5.5V (and with V- = -5.5V for MAX909).
- **Note 9:** Due to difficulties in measuring propagation delay with 5mV of overdrive in automatic test equipment, this parameter is guaranteed by design for the MAX907 and MAX908. Correlation tests show that the specification can be guaranteed if all other DC parameters are within the specified limits. V_{OS} must be added to the overdrive voltage for low values of overdrive. For the MAX909, propagation delay is typical only and there is no guaranteed maximum limit.
- **Note 10:** Differential Propagation Delay is specified as the difference between any two channels in the MAX907/MAX908 (both outputs making either a low-to-high or a high-to-low transition).
- **Note 11:** Propagation Delay Skew is specified as the difference between any single channel's output low-to-high transition (t_{PD}+) and high-to-low transition (t_{PD}-), and also between the QOUT and QOUT transition on the MAX909.
- Note 12: Latch specifications apply to MAX909 only (see Figure 2).

Typical Operating Characteristics

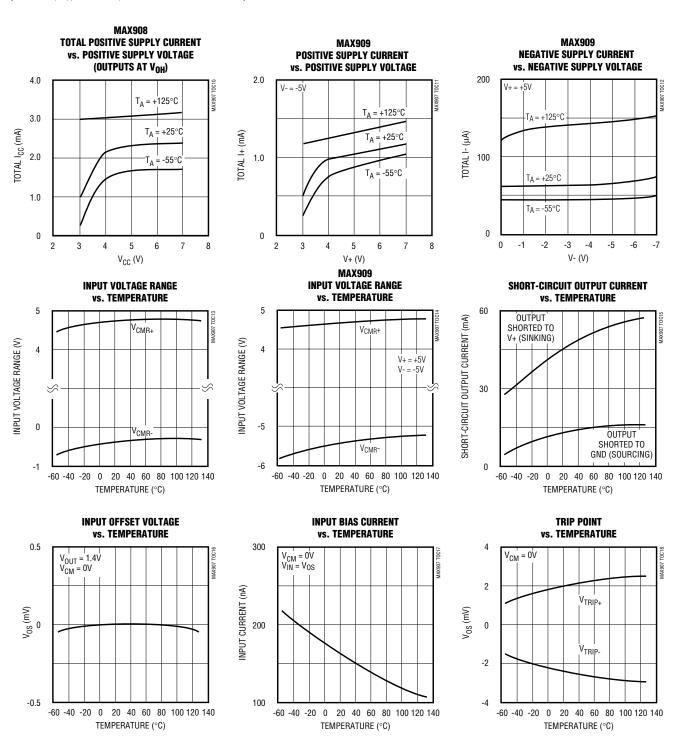
 $(V+ = 5V, T_A = +25$ °C, unless otherwise noted.)



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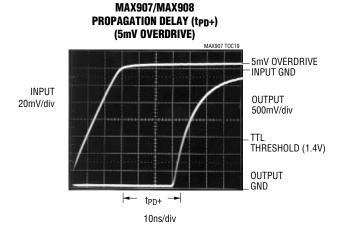
Typical Operating Characteristics (continued)

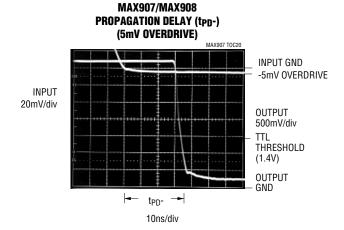
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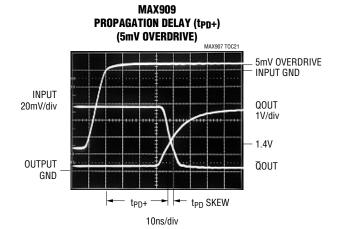


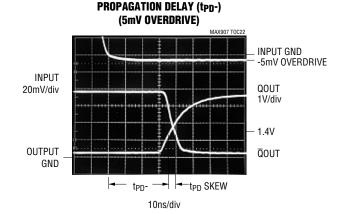
Typical Operating Characteristics (continued)

 $(V + = 5V, T_A = +25$ °C, unless otherwise noted.)



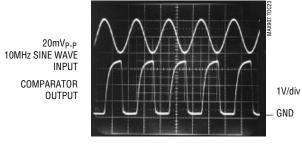






MAX909

RESPONSE TO 10MHz SINE WAVE



50ns/div

Pin Description

PIN		NAME	FUNCTION		
MAX907	MAX908	MAX909	NAIVIE	FUNCTION	
1	1	_	OUTA	Comparator A Output	
2	2	_	INA-	Comparator A Inverting Input	
3	3	_	INA+	Comparator A Noninverting Input	
4	11	6	GND	Ground	
5	5	_	INB+	Comparator B Noninverting Input	
6	6	_	INB-	Comparator B Inverting Input	
7	7	_	OUTB	Comparator B Output	
8	4	1	V+	Positive Supply	
_	8		OUTC	Comparator C Output	
_	9	_	INC-	Comparator C Inverting Input	
_	10	_	INC+	Comparator C Noninverting Input	
_	12	_	IND+	Comparator D Noninverting Input	
_	13	_	IND-	Comparator D Inverting Input	
_	14	_	OUTD	Comparator D Output	
_	_	2	IN+	Noninverting Input	
_	_	3	IN-	Inverting Input	
_	_	4	V-	Negative Supply or Ground	
_	_	5	LE	The latch is transparent when LE is low. The comparator output is stored when LE is high.	
_	_	7	QOUT	Comparator Output	
_	_	8	QOUT	Inverted Comparator Output	

Detailed Description

Timing

Noise or undesired parasitic AC feedback cause most high-speed comparators to oscillate in the linear region (i.e., when the voltage on one input is at or near the voltage on the other input). The MAX907/MAX908/MAX909 eliminate this problem by incorporating internal hysteresis. When the two comparator input voltages are equal, hysteresis effectively causes one comparator input voltage to move quickly past the other, thus taking the input out of the region where oscillation occurs. Standard comparators require that hysteresis be added through the use of external resistors. The MAX907/MAX908/MAX909's fixed internal hysteresis eliminates these resistors (and the equations required to determine appropriate values).

Adding hysteresis to a comparator creates two trip points: one for the input voltage rising and one for the input voltage falling (Figure 1). The difference between these two input-referred trip points is the hysteresis.

Figure 1 illustrates the case where IN- is fixed and IN+ is varied. If the inputs were reversed, the figure would look the same, except the output would be inverted.

The MAX909 includes an internal latch, allowing the result of a comparison to be stored. If LE is low, the latch is transparent (i.e., the comparator operates as though the latch is not present). The state of the comparator output is stored when LE is high (Figure 2).

Note that the MAX909 can be operated with V- connected to ground or to a negative supply voltage. The MAX909's input range extends from (V- - 0.2V) to (V+ - 1.5V).

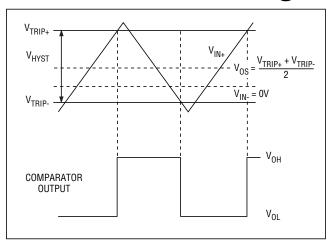


Figure 1. Input and Output Waveforms, Noninverting Input Varied

Applications Information

Circuit Layout

Because of the MAX907/MAX908/MAX909's high gain bandwidth, special precautions must be taken to realize the full high-speed capability. A printed circuit board with a good, low-inductance ground plane is mandatory. Place the decoupling capacitor (a 0.1µF ceramic capacitor is a good choice) as close to V+ as possible. Pay close attention to the decoupling capacitor's bandwidth, keeping leads short. Short lead lengths on the inputs and outputs are also essential to avoid unwanted parasitic feedback around the comparators. Solder the device directly to the printed circuit board instead of using a socket.

Overdriving the Inputs

The inputs to the MAX907/MAX908/MAX909 may be driven beyond the voltage limits given in the *Absolute Maximum Ratings*, as long as the current flowing into the device is limited to 25mA. However, if the inputs are overdriven, the output may be inverted. The addition of an external diode prevents this inversion by limiting the input voltage to 200mV to 300mV below ground (Figure 3).

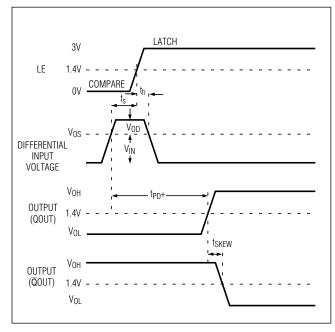


Figure 2. MAX909 Timing Diagram

Battery-Operated Infrared Data Link

Figure 4's circuit allows reception of infrared data. The MAX403 converts the photodiode current to a voltage, and the MAX907 determines whether the amplifier output is high enough to be called a "1". The current consumption of this circuit is minimal: The MAX403 and MAX907 require typically 250µA and 700µA, respectively.

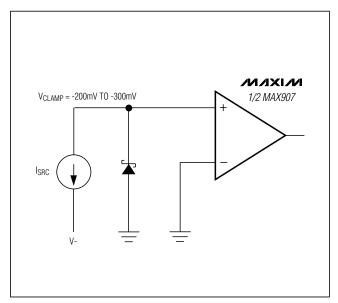


Figure 3. Schottky Clamp for Input Driven Below Ground

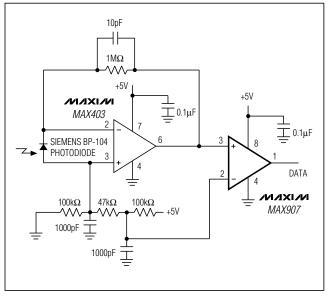
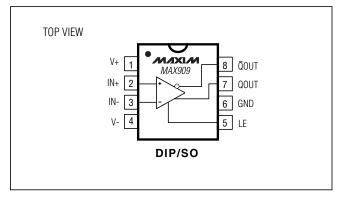


Figure 4. Battery-Operated Infrared Data Link Consumes Only 1mA

_Pin Configurations (continued)



_Ordering Information (continued)

PART	TEMP RANGE	PIN-PACKAGE
MAX908CPD	0°C to +70°C	14 Plastic DIP
MAX908CSD	0°C to +70°C	14 SO
MAX908EPD	-40°C to +85°C	14 Plastic DIP
MAX908ESD	-40°C to +85°C	14 SO
MAX909CPA	0°C to +70°C	8 Plastic DIP
MAX909CSA	0°C to +70°C	8 SO
MAX909EPA	-40°C to +85°C	8 Plastic DIP
MAX909ESA	-40°C to +85°C	8 SO

Chip Information

MAX907 TRANSISTOR COUNT: 262 MAX908 TRANSISTOR COUNT: 536 MAX909 TRANSISTOR COUNT: 140

PROCESS: Bipolar

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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