

LT3062 Series

ABSOLUTE MAXIMUM RATINGS

(Note 1)

IN Pin Voltage	±50V
OUT Pin Voltage	+40V, -50V
Input to Output Differential Voltage (Note 2)	±50V
ADJ Pin Voltage	±50V
SHDN Pin Voltage	±50V
REF/BYP Pin Voltage	-0.3V to 1V
Output Short-Circuit Duration	Indefinite

Operating Junction Temperature (Notes 3, 5, 12)	
LT3062E, LT3062I	-40°C to 125°C
LT3062MP	-55°C to 150°C
LT3062H	-40°C to 150°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec)	
MS8E Package Only	300°C

PIN CONFIGURATION

TOP VIEW DCB PACKAGE 8-LEAD (2mm × 3mm) PLASTIC DFN $T_{JMAX} = 150^{\circ}\text{C}$, $\theta_{JA} = 38^{\circ}\text{C/W}$ TO 45°C/W, $\theta_{JC} = 3.5^{\circ}\text{C/W}$ EXPOSED PAD (PIN 9) IS GND, MUST BE SOLDERED TO PCB	TOP VIEW MS8E PACKAGE 8-LEAD PLASTIC MSOP $T_{JMAX} = 150^{\circ}\text{C}$, $\theta_{JA} = 29^{\circ}\text{C/W}$ TO 45°C/W, $\theta_{JC} = 5^{\circ}\text{C/W}$ TO 10°C/W EXPOSED PAD (PIN 9) IS GND, MUST BE SOLDERED TO PCB
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ORDER INFORMATION <http://www.linear.com/product/LT3062#orderinfo>

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3062EDCB#PBF	LT3062EDCB#TRPBF	LGNC	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C
LT3062IDCB#PBF	LT3062IDCB#TRPBF	LGNC	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C
LT3062HDCB#PBF	LT3062HDCB#TRPBF	LGNC	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 150°C
LT3062MPDCB#PBF	LT3062MPDCB#TRPBF	LGNC	8-Lead (2mm × 3mm) Plastic DFN	-55°C to 150°C
LT3062EMS8E#PBF	LT3062EMS8E#TRPBF	LTGND	8-Lead Plastic MSOP	-40°C to 125°C
LT3062IMS8E#PBF	LT3062IMS8E#TRPBF	LTGND	8-Lead Plastic MSOP	-40°C to 125°C
LT3062HMS8E#PBF	LT3062HMS8E#TRPBF	LTGND	8-Lead Plastic MSOP	-40°C to 150°C
LT3062MPMS8E#PBF	LT3062MPMS8E#TRPBF	LTGND	8-Lead Plastic MSOP	-55°C to 150°C
LT3062EDCB-3.3#PBF	LT3062EDCB-3.3#TRPBF	LGYY	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C
LT3062IDCB-3.3#PBF	LT3062IDCB-3.3#TRPBF	LGYY	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C
LT3062EDCB-5#PBF	LT3062EDCB-5#TRPBF	LGYY	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C
LT3062IDCB-5#PBF	LT3062IDCB-5#TRPBF	LGYY	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C

3062fa

ORDER INFORMATION <http://www.linear.com/product/LT3062#orderinfo>

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3062EMS8E-3.3#PBF	LT3062EMS8E-3.3#TRPBF	LTGZG	8-Lead Plastic MSOP	-40°C to 125°C
LT3062IMS8E-3.3#PBF	LT3062IMS8E-3.3#TRPBF	LTGZG	8-Lead Plastic MSOP	-40°C to 125°C
LT3062EMS8E-5#PBF	LT3062EMS8E-5#TRPBF	LTGZF	8-Lead Plastic MSOP	-40°C to 125°C
LT3062IMS8E-5#PBF	LT3062IMS8E-5#TRPBF	LTGZF	8-Lead Plastic MSOP	-40°C to 125°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. *The temperature grade is identified by a label on the shipping container.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreel/>. Some packages are available in 500 unit reels through designated sales channels with #TRMPBF suffix.

ELECTRICAL CHARACTERISTICS The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. (Note 3)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Minimum Input Voltage (Note 4)	$I_{LOAD} = 200\text{mA}$ ●		1.6	2.1	V
Regulated Output Voltage (Note 5)	LT3062-3.3: $V_{IN} = 3.9$, $I_{LOAD} = 1\text{mA}$ ●	3.267	3.3	3.333	V
	LT3062-3.3: $3.9\text{V} < V_{IN} < 45\text{V}$, $1\text{mA} < I_{LOAD} < 200\text{mA}$ ●	3.234		3.366	V
	LT3062-5: $V_{IN} = 5.6$, $I_{LOAD} = 1\text{mA}$ ●	4.950	5	5.050	V
	LT3062-5: $5.6\text{V} < V_{IN} < 45\text{V}$, $1\text{mA} < I_{LOAD} < 200\text{mA}$ ●	4.900		5.100	V
ADJ Pin Voltage (Notes 4, 5)	$V_{IN} = 2.1\text{V}$, $I_{LOAD} = 1\text{mA}$ ●	594	600	606	mV
	$2.1\text{V} < V_{IN} < 45\text{V}$, $1\text{mA} < I_{LOAD} < 200\text{mA}$ (E-, I-Grades) ●	588	600	612	mV
	$2.1\text{V} < V_{IN} < 45\text{V}$, $1\text{mA} < I_{LOAD} < 200\text{mA}$ (MP-, H-Grades) ●	585	600	612	mV
Line Regulation (Note 4) $I_{LOAD} = 1\text{mA}$	LT3062-3.3: $\Delta V_{IN} = 3.9\text{V}$ to 45V (E-, I-Grades) ●		1.6	22	mV
	LT3062-5: $\Delta V_{IN} = 5.6\text{V}$ to 45V (E-, I-Grades) ●		3.1	33	mV
	LT3062: $\Delta V_{IN} = 2.1\text{V}$ to 45V (E-, I-Grades) ●		0.5	4	mV
	LT3062: $\Delta V_{IN} = 2.1\text{V}$ to 45V (MP-, H-Grades) ●			6	mV
Load Regulation (Note 4) $\Delta I_{LOAD} = 1\text{mA}$ to 200mA	LT3062-3.3: $V_{IN} = 3.9\text{V}$ (E-, I-Grades) ●		7.1	28	mV
	LT3062-5: $V_{IN} = 5.6\text{V}$ (E-, I-Grades) ●		7.6	39	mV
	LT3062: $V_{IN} = 2.1\text{V}$ (E-, I-Grades) ●		0.3	4	mV
	LT3062: $V_{IN} = 2.1\text{V}$ (MP-, H-Grades) ●			9	mV
Dropout Voltage $V_{IN} = V_{OUT(NOMINAL)}$ (Notes 6, 7)	$I_{LOAD} = 1\text{mA}$ ●		65	110	mV
	$I_{LOAD} = 1\text{mA}$ ●			180	mV
	$I_{LOAD} = 10\text{mA}$ ●		130	180	mV
	$I_{LOAD} = 10\text{mA}$ ●			270	mV
	$I_{LOAD} = 100\text{mA}$ ●		250	290	mV
GND Pin Current $V_{IN} = V_{OUT(NOMINAL)} + 0.6\text{V}$ (Notes 6, 8)	$I_{LOAD} = 100\text{mA}$ ●			430	mV
	$I_{LOAD} = 200\text{mA}$ ●		300	360	mV
	$I_{LOAD} = 200\text{mA}$ ●			530	mV
	$I_{LOAD} = 0$ ●		45	90	μA
	$I_{LOAD} = 1\text{mA}$ ●		70	120	μA
Output Voltage Noise	$I_{LOAD} = 10\text{mA}$ ●		225	500	μA
	$I_{LOAD} = 100\text{mA}$ ●		2	4	mA
	$I_{LOAD} = 200\text{mA}$ ●		5	10	mA
	$I_{LOAD} = 200\text{mA}$ ●				
Output Voltage Noise	$C_{OUT} = 10\mu\text{F}$, $I_{LOAD} = 200\text{mA}$, $C_{REF/BYP} = 0.01\mu\text{F}$		30		μV_{RMS}
	$V_{OUT} = 600\text{mV}$, $\text{BW} = 10\text{Hz}$ to 100kHz				

LT3062 Series

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. (Note 3)

PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
ADJ Pin Bias Current (Notes 4, 9)		●		15	60	nA
Shutdown Threshold	$V_{OUT} = \text{Off to On}$ $V_{OUT} = \text{On to Off}$	● ●	0.3	0.8 0.7	1.5	V V
SHDN Pin Current (Note 10)	$V_{SHDN} = 0\text{V}$ $V_{SHDN} = 45\text{V}$	● ●		1.2	<1 3	μA μA
Quiescent Current in Shutdown	$V_{IN} = 45\text{V}$, $V_{SHDN} = 0\text{V}$			0.3	1	μA
Ripple Rejection $V_{IN} - V_{OUT} = 1.5\text{V}$ (AVG), $V_{RIPPLE} = 0.5\text{V}_{P-P}$, $f_{RIPPLE} = 120\text{Hz}$, $I_{LOAD} = 200\text{mA}$	LT3062-3.3 LT3062-5 LT3062 (Note 4)		58 55 70	73 70 85		dB dB dB
Current Limit	$V_{IN} = 7\text{V}$, $V_{OUT} = 0$ $V_{IN} = V_{OUT(NOMINAL)} + 1\text{V}$ (Note 11), $\Delta V_{OUT} = -5\%$	●	220	320		mA mA
Input Reverse Leakage Current	$V_{IN} = -45\text{V}$, $V_{OUT} = 0$	●			1	mA
Reverse Output Current (Note 13)	LT3062-3.3: $V_{OUT} = 3.3\text{V}$, $V_{IN} = V_{SHDN} = 0\text{V}$ LT3062-5: $V_{OUT} = 5\text{V}$, $V_{IN} = V_{SHDN} = 0\text{V}$ LT3062: $V_{OUT} = 1.2\text{V}$, $V_{IN} = V_{SHDN} = 0\text{V}$			5.2 5.2 0.2	15 15 10	μA μA μA

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: Absolute maximum input to output differential voltage is not achievable with all combinations of rated IN pin and OUT pin voltages. With the IN pin at 50V, the OUT pin may not be pulled below 0V. The total measured voltage from IN to OUT must not exceed $\pm 50\text{V}$. If OUT is above ground, do not pull IN more than 40V below OUT.

Note 3: The LT3062 is tested and specified under pulse load conditions such that $T_J \approx T_A$. The LT3062E regulators are 100% tested at $T_A = 25^\circ\text{C}$ and performance is guaranteed from 0°C to 125°C . Performance at -40°C to 125°C is assured by design, characterization and correlation with statistical process controls. The LT3062I regulators are guaranteed over the full -40°C to 125°C operating junction temperature range. The LT3062MP regulators are 100% tested over the -55°C to 150°C operating junction temperature. The LT3062H regulators are 100% tested at the 150°C operating junction temperature. High junction temperatures degrade operating lifetimes. Operating lifetime is derated at junction temperature greater than 125°C .

Note 4: The LT3062 adjustable version is tested and specified for these conditions with the ADJ connected to the OUT pin.

Note 5: Maximum junction temperature limits operating conditions. The regulated output voltage specification does not apply for all possible combinations of input voltage and output current. Limit the output current range if operating at the maximum input voltage. Limit the input-to-output voltage differential if operating at maximum output current. Current limit foldback limits the maximum output current as a function of input-to-output voltage. See Current Limit vs $V_{IN} - V_{OUT}$ in the Typical Performance Characteristics section.

Note 6: To satisfy minimum input voltage requirements, the LT3062 adjustable version is tested and specified for these conditions with an external resistor divider (bottom 60k, top 230k) for an output voltage of 2.9V. The external resistor divider adds $10\mu\text{A}$ of DC load on the output. The external current is not factored into GND pin current.

Note 7: Dropout voltage is the minimum input-to-output voltage differential needed to maintain regulation at a specified output current. In dropout, the output voltage equals: $(V_{IN} - V_{DROPOUT})$.

Note 8: GND pin current is tested with $V_{IN} = V_{OUT(NOMINAL)} + 0.6\text{V}$ and a current source load. GND pin current will increase in dropout. See GND pin current curves in the Typical Performance Characteristics section. For fixed voltage options, an internal resistor divider will add $5\mu\text{A}$ to the GND pin current. See the GND Pin Current curves in the Typical Performance Characteristics section.

Note 9: ADJ pin bias current flows out of the ADJ pin.

Note 10: SHDN pin current flows into the SHDN pin.

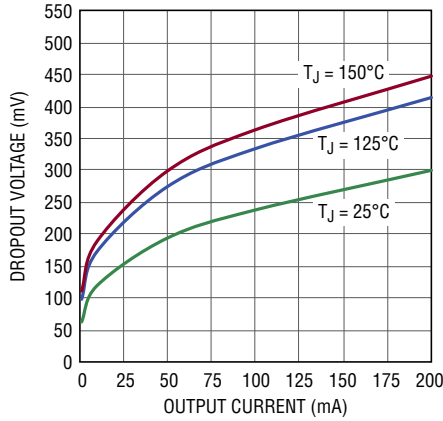
Note 11: To satisfy requirements for minimum input voltage, current limit is tested at $V_{IN} = V_{OUT(NOMINAL)} + 1\text{V}$ or $V_{IN} = 2.1\text{V}$, whichever is greater.

Note 12: This IC includes thermal limit which protects the device during momentary overload conditions. Junction temperature exceeds 125°C (E-Grade and I-Grade) or 150°C (MP-Grade and H-Grade) when thermal limit is active. Continuous operation above the specified maximum junction temperature may impair device reliability.

Note 13: Reverse output current is tested with the IN pin grounded and the OUT pin forced to the specified voltage. This current flows into the OUT pin and out of the GND pin.

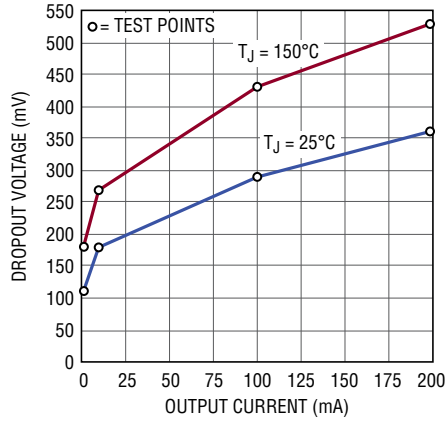
TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$, unless otherwise noted.

Typical Dropout Voltage



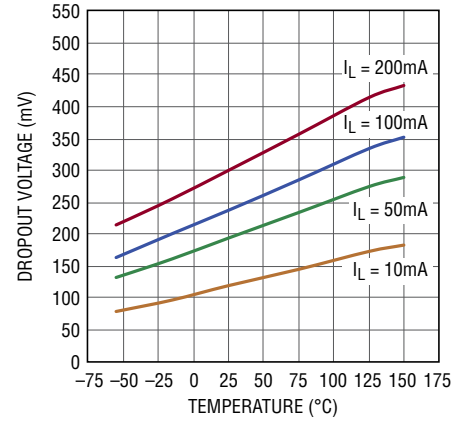
3062 G01

Guaranteed Dropout Voltage



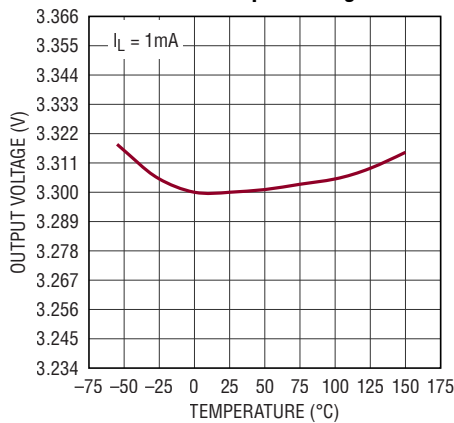
3062 G02

Dropout Voltage



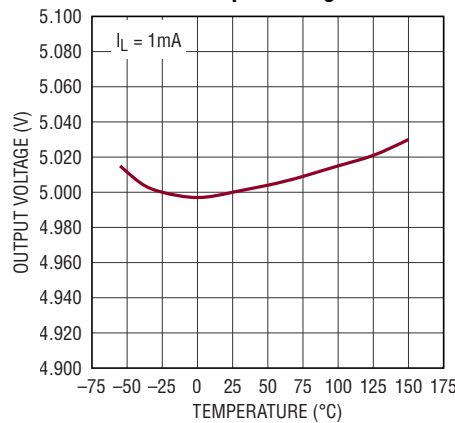
3062 G03

LT3062-3.3 Output Voltage



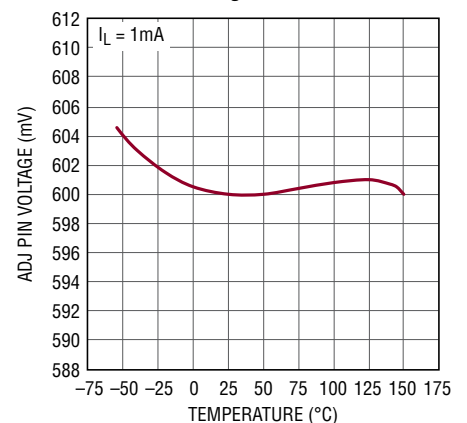
LT3062 G04

LT3062-5 Output Voltage



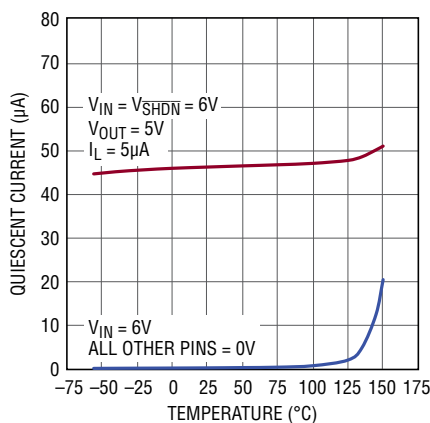
LT3062 G05

ADJ Pin Voltage



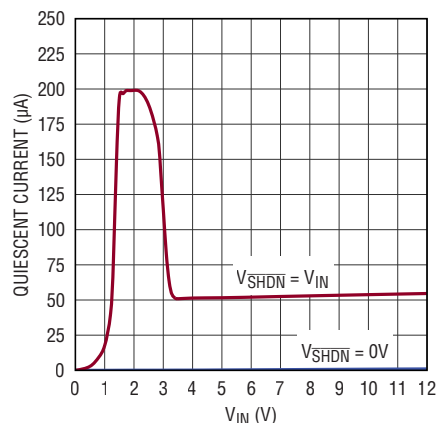
3062 G06

Quiescent Current



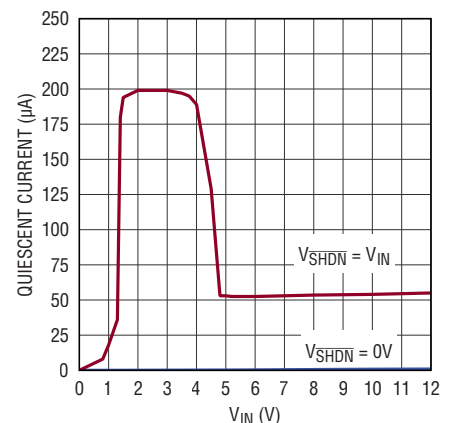
3062 G07

LT3062-3.3 Quiescent Current



3062 G08

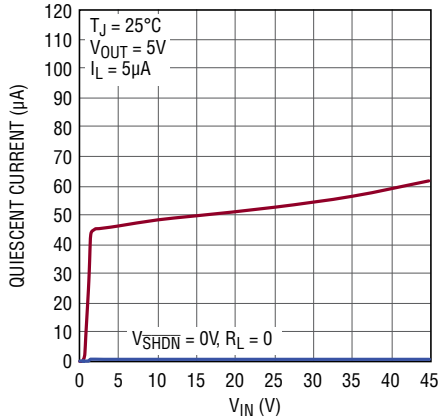
LT3062-5 Quiescent Current



3062 G09

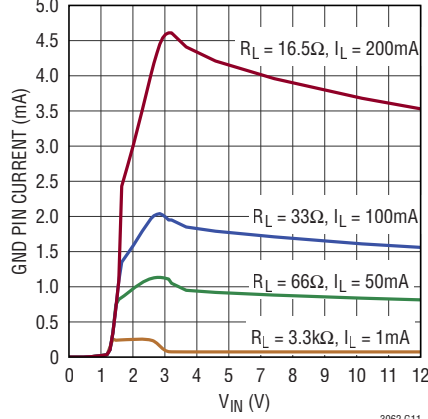
TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$, unless otherwise noted.

Quiescent Current



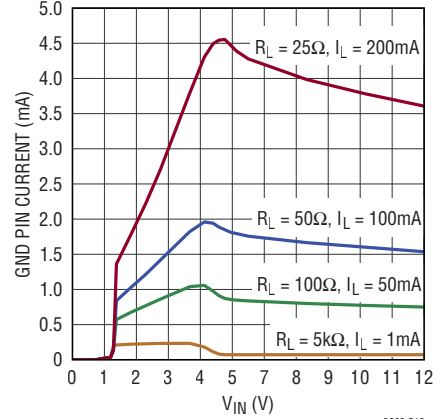
3062 G10

LT3062-3.3 GND Pin Current



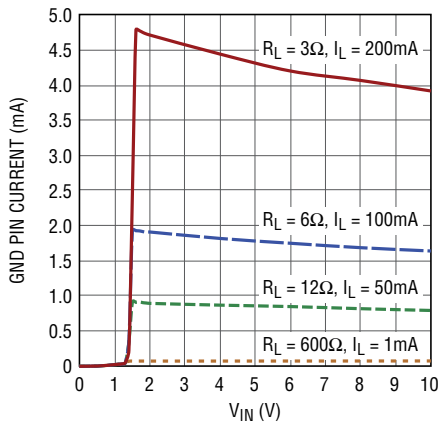
3062 G11

LT3062-5 GND Pin Current



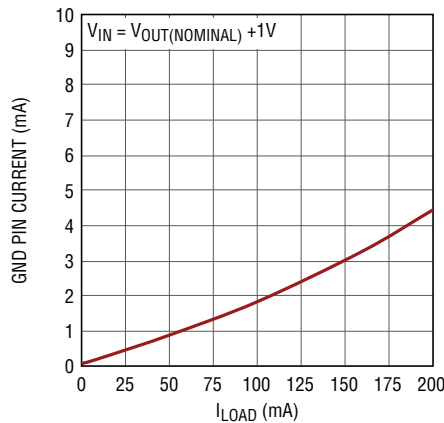
3062 G12

GND Pin Current, $V_{OUT} = 0.6\text{V}$



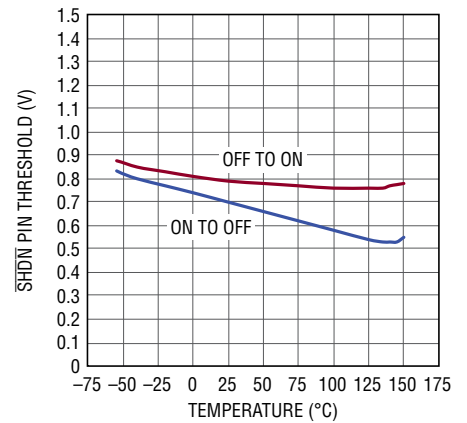
3063 G13

GND Pin Current vs I_{LOAD}



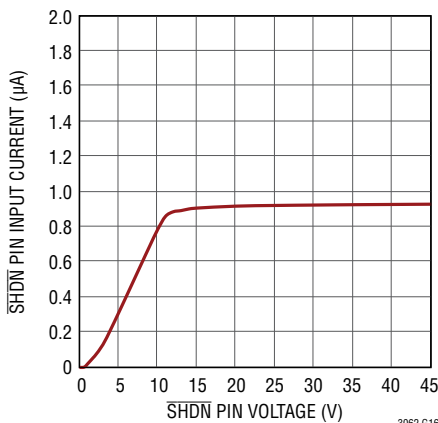
3062 G14

SHDN Pin Threshold



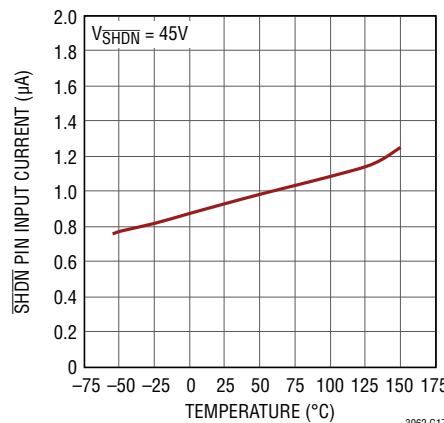
3062 G15

SHDN Pin Input Current



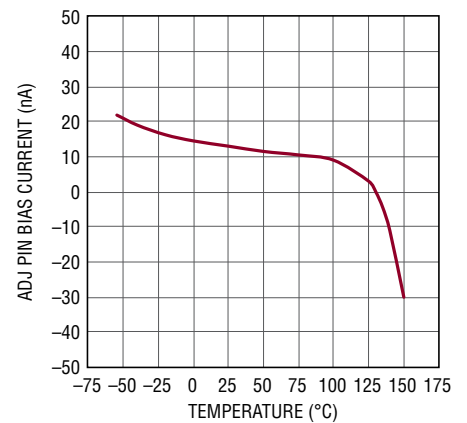
3062 G16

SHDN Pin Input Current



3062 G17

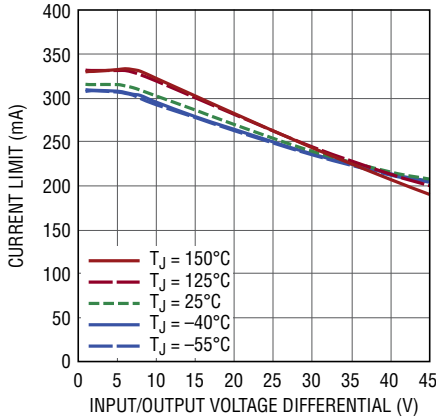
ADJ Pin Bias Current



3062 G18

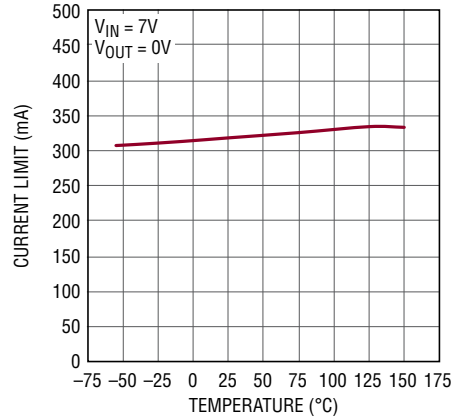
TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$, unless otherwise noted.

Internal Current Limit



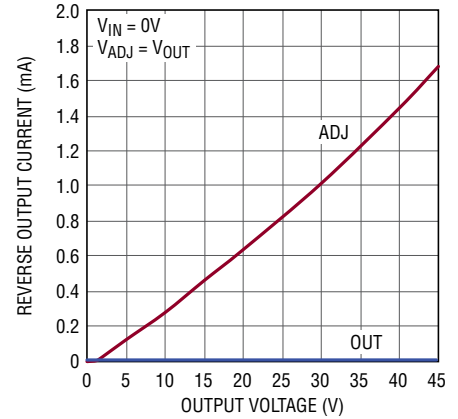
3062 G19

Internal Current Limit



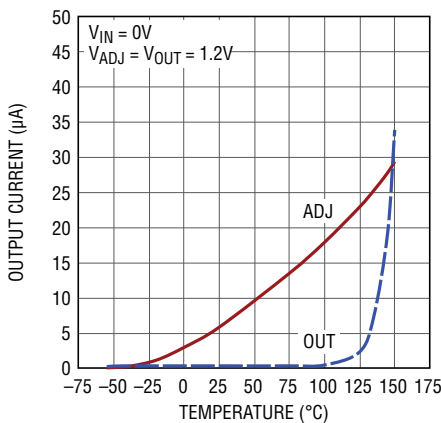
3062 G20

Reverse Output Current



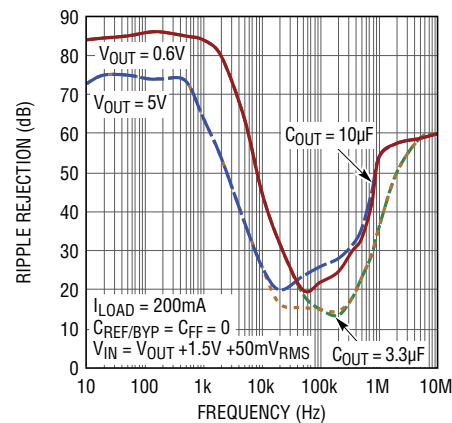
3062 G21

Reverse Output Current



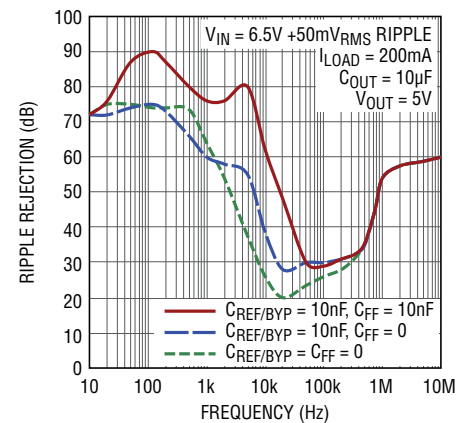
3062 G22

Input Ripple Rejection



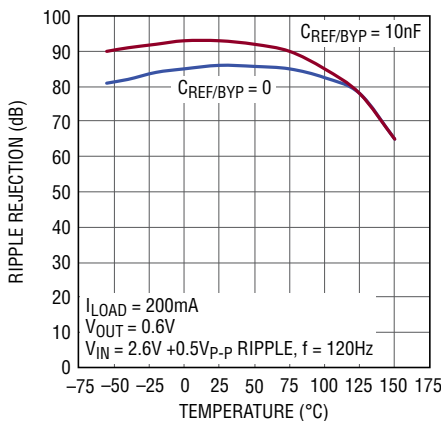
3062 G23

Input Ripple Rejection



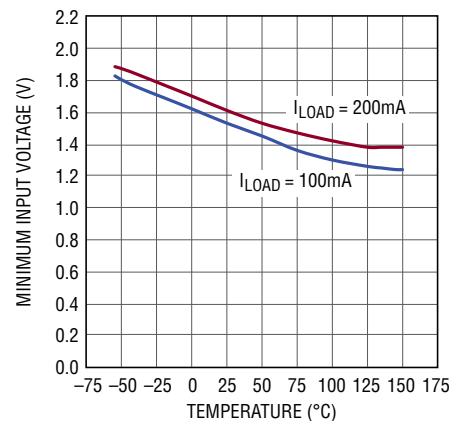
3062 G24

Input Ripple Rejection



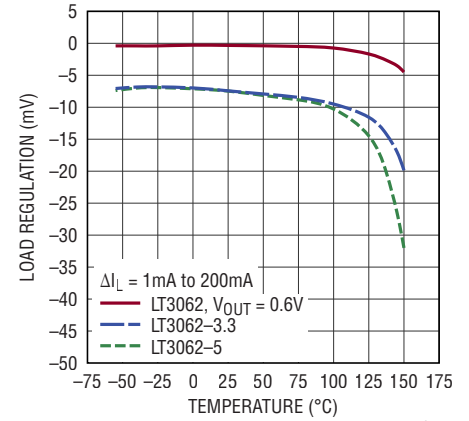
3062 G25

Minimum Input Voltage



3062 G26

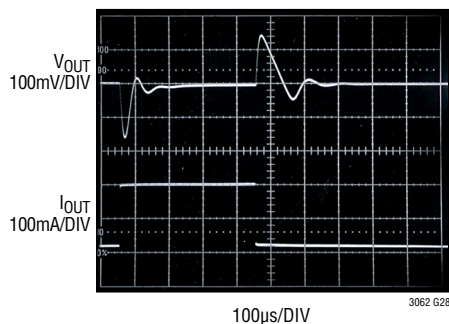
Load Regulation



3062 G27

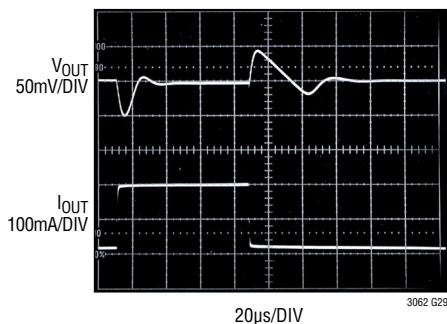
TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$, unless otherwise noted.

5V Transient Response
 $C_{FF} = 0$, $I_{OUT} = 20\text{mA}$ to 200mA



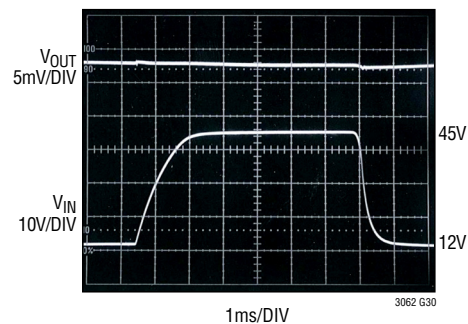
$V_{IN} = 6\text{V}$
 $C_{OUT} = 10\mu\text{F}$
 $I_{FB-DIVIDER} = 5\mu\text{A}$

5V Transient Response
 $C_{FF} = 10\text{nF}$, $I_{OUT} = 20\text{mA}$ to 200mA



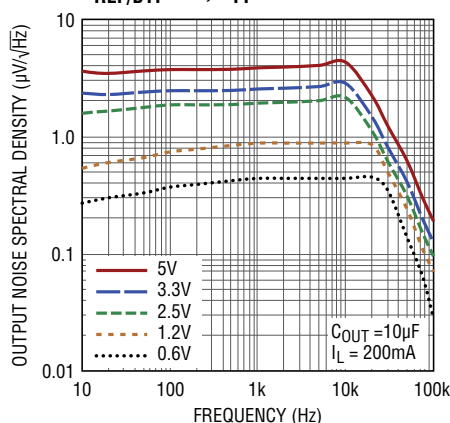
$V_{IN} = 6\text{V}$
 $C_{OUT} = 10\mu\text{F}$
 $I_{FB-DIVIDER} = 5\mu\text{A}$

5V Transient Response, $V_{OUT} = 5\text{V}$
Load Dump, $V_{IN} = 12\text{V}$ to 45V



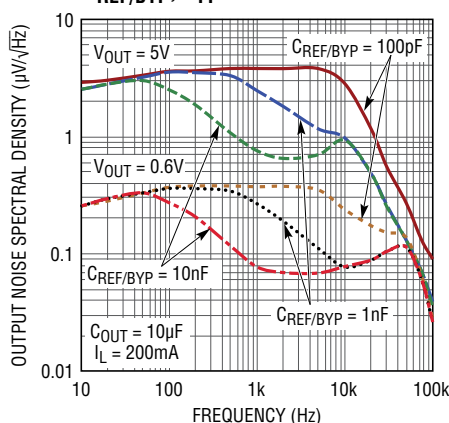
$C_{OUT} = 10\mu\text{F}$
 $C_{REF/BYP} = C_{FF} = 10\text{nF}$
 $I_{FB-DIVIDER} = 5\mu\text{A}$

Output Noise Spectral Density
 $C_{REF/BYP} = 0$, $C_{FF} = 0$



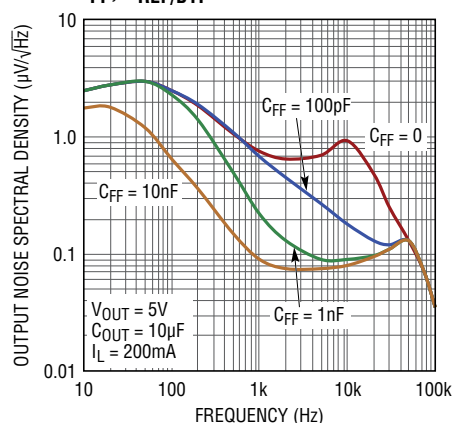
3062 G31

Output Noise Spectral Density vs
 $C_{REF/BYP}$, $C_{FF} = 0$



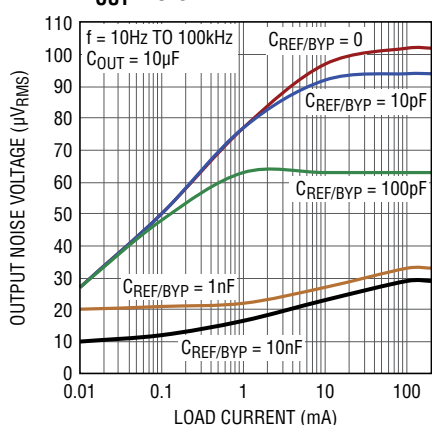
3062 G32

Output Noise Spectral Density vs
 C_{FF} , $C_{REF/BYP} = 10\text{nF}$



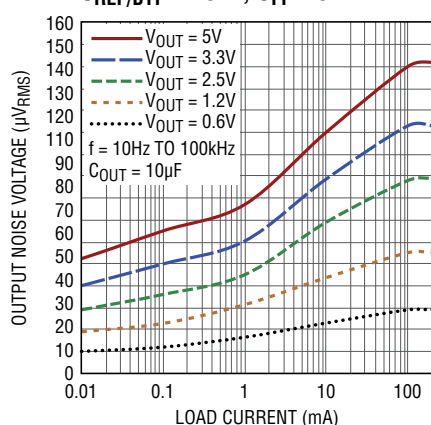
3062 G33

RMS Output Noise vs $C_{REF/BYP}$
 $V_{OUT} = 0.6\text{V}$



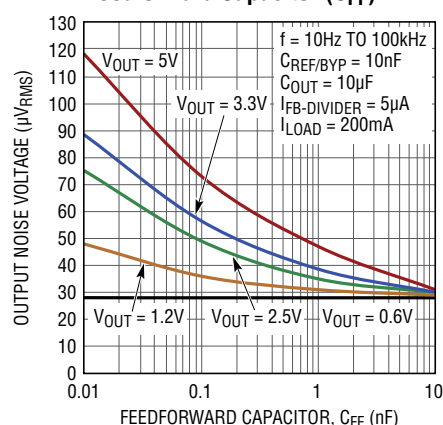
3062 G34

RMS Output Noise vs Load Current
 $C_{REF/BYP} = 10\text{nF}$, $C_{FF} = 0$



3062 G35

RMS Output Noise vs
Feedforward Capacitor (C_{FF})

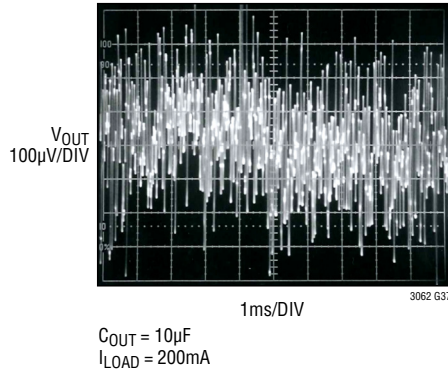


3062 G36

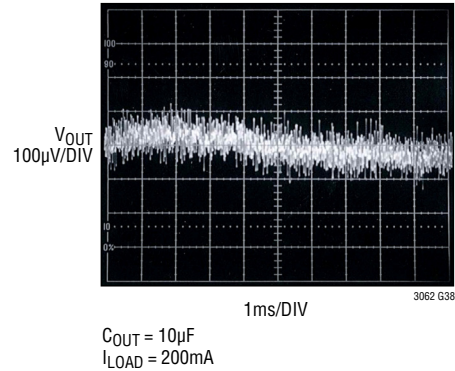
3062fa

TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$, unless otherwise noted.

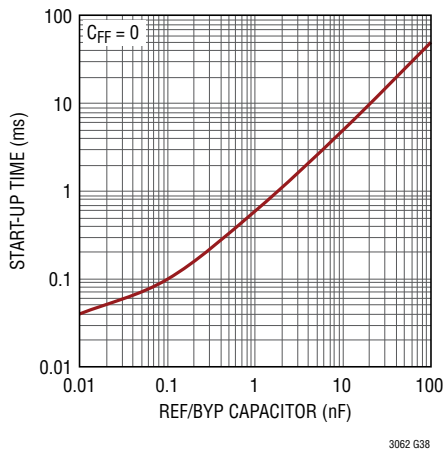
5V 10Hz to 100kHz Output Noise
 $C_{\text{REF/BYP}} = 10\text{nF}$, $C_{\text{FF}} = 0$



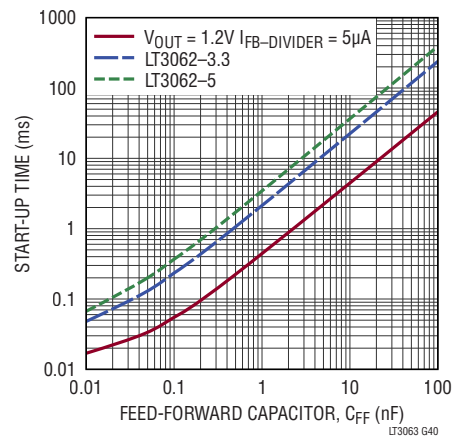
5V 10Hz to 100kHz Output Noise
 $C_{\text{REF/BYP}} = 10\text{nF}$, $C_{\text{FF}} = 10\text{nF}$



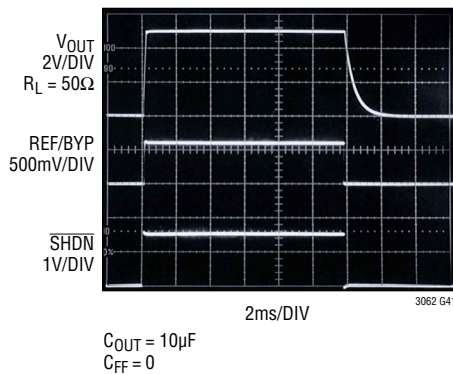
Start-Up Time vs REF/BYP Capacitor



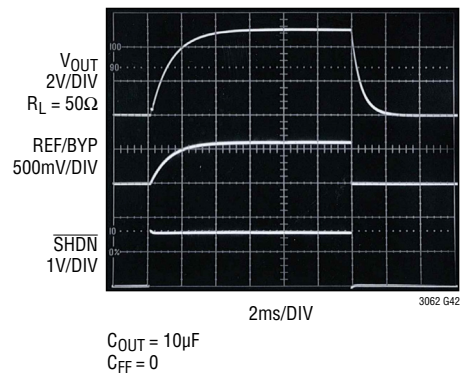
Start-Up Time vs C_{FF}



SHDN Transient Response
 $C_{\text{REF/BYP}} = 0$



SHDN Transient Response
 $C_{\text{REF/BYP}} = 10\text{nF}$



PIN FUNCTIONS

REF/BYP (Pin 1): Reference/ Bypass. Connecting a single capacitor from this pin to GND bypasses the LT3062's reference noise and soft-starts the reference. A 10nF bypass capacitor typically reduces output voltage noise to $30\mu\text{V}_{\text{RMS}}$ in a 10Hz to 100kHz bandwidth. Soft-start time is directly proportional to the REF/BYP capacitor value. If the LT3062 is placed in shutdown, REF/BYP is actively pulled low by an internal device to reset soft-start. If low noise or soft-start performance is not required, this pin must be left floating (unconnected). Do not drive this pin with any active circuitry.

ADJ (Pin 2): Adjust. This pin is the error amplifier's inverting terminal. Its typical bias current of 15nA flows out of the pin (see curve of ADJ Pin Bias Current vs Temperature in the Typical performance Characteristics section). The ADJ pin voltage is 600mV referenced to GND.

OUT (Pins 3, 4): Output. These pins supply power to the load. A minimum output capacitor of $3.3\mu\text{F}$ is required to prevent oscillations. Large load transient applications require larger output capacitors to limit peak voltage transients. See the Applications Information section for more information on transient response and reverse output characteristics. The output voltage range is 600mV to 40V.

IN (Pins 5, 6): Input. These pins supply power to the device. The LT3062 requires a bypass capacitor at IN if the device is located more than six inches from the main input filter capacitor. In general, the output impedance of a battery rises with frequency, so it is advisable to include a bypass capacitor in battery-powered circuits. A bypass capacitor in the range of $1\mu\text{F}$ to $10\mu\text{F}$ suffices. See Input Capacitance and Stability in the Application Information section for more information.

The LT3062 withstands reverse voltages on the IN pin with respect to the GND and OUT pins. In a reversed input situation, such as the battery plugged in backwards, the LT3062 behaves as if a large value resistor is in series with its input. Limited reverse current flows into the LT3062 and no reverse voltage appears at the load. The device protects itself and the load.

$\overline{\text{SHDN}}$ (Pin 7): Shutdown. Pulling the $\overline{\text{SHDN}}$ pin low puts the LT3062 into a low power state and turns the output off. Drive the SHDN pin with either logic or an open collector/drain with a pull-up resistor. The resistor supplies the pull-up current to the open collector/drain logic, normally several microamperes, and the $\overline{\text{SHDN}}$ pin current, typically less than $3\mu\text{A}$. If unused, connect the $\overline{\text{SHDN}}$ pin to V_{IN} . The LT3062 does not function if the $\overline{\text{SHDN}}$ pin is not connected. The $\overline{\text{SHDN}}$ pin cannot be driven below GND unless tied to the IN pin. If the $\overline{\text{SHDN}}$ pin is driven below GND while IN is powered, the output will turn on. $\overline{\text{SHDN}}$ pin logic cannot be referenced to a negative rail.

GND (Pin 8, Exposed Pad Pin 9): Ground. Connect the bottom of the external resistor divider that sets the output voltage directly to GND for optimum regulation. Tie the exposed pad Pin 9 directly to Pin 8 and the PCB ground. This exposed pad provides enhanced thermal performance with its connection to the PCB ground. See the Applications Information section for thermal considerations and calculating junction temperature.

APPLICATIONS INFORMATION

The LT3062 is a 200mA low dropout regulator with shutdown that is available in fixed output and adjustable versions. The device is capable of supplying 200mA at a typical dropout voltage of 300mV and operates over a 1.6V to 45V input range.

A single external capacitor provides programmable low noise reference performance and output soft-start functionality. For example, connecting a 10nF capacitor from the REF/BYP pin to GND lowers output noise to 30μV_{RMS} over a 10Hz to 100kHz bandwidth. This capacitor also soft-starts the reference and prevents output voltage overshoot at turn-on.

The LT3062's quiescent current is merely 45μA, while providing fast transient response with a 3.3μF minimum low ESR ceramic output capacitor. In shutdown, quiescent current is less than 1μA and the reference soft-start capacitor and output are reset.

The LT3062 optimizes stability and transient response with low ESR, ceramic output capacitors. The LT3062 does not require the addition of ESR as is common with other regulators. The LT3062 has an adjustable output and typically provides 0.1% line regulation and 0.1% load regulation. A curve of load regulation appears in the Typical Performance Characteristics section.

Internal protection circuitry includes reverse-battery protection, reverse-current protection, current limit with foldback and thermal shutdown.

Adjustable Operation

The adjustable LT3062 has an output voltage range of 0.6V to 40V. Output voltage is set by the ratio of two external resistors, as shown in Figure 1. The device regulates the output to maintain the ADJ pin voltage at 0.6V referenced to ground. The current in R1 equals 0.6V/R1, and R2's current is R1's current minus the ADJ pin bias current.

The ADJ pin bias current, 15nA at 25°C, flows from the ADJ pin through R1 to GND. Calculate the output voltage using the formula in Figure 1. R1's value should not be greater than 124k to provide a minimum ~5μA load cur-

rent so that output voltage errors, caused by the ADJ pin bias current, are minimized. Note that in shutdown, the output is turned off and the divider current is zero. Curves of ADJ Pin Voltage vs Temperature and ADJ Pin Bias Current vs Temperature appear in the Typical Performance Characteristics section.

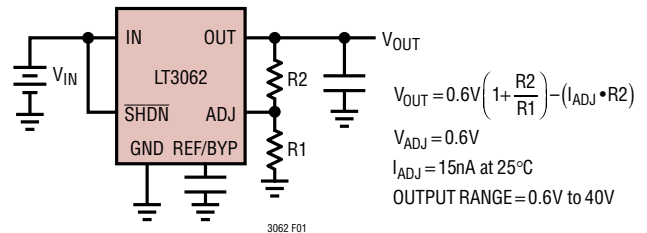


Figure 1. Adjustable Operation

The LT3062 is tested and specified with the ADJ pin tied to the OUT pin for an output voltage of 0.6V. Specifications for output voltages greater than 0.6V are proportional to the ratio of the desired output voltage to 0.6V: $V_{OUT}/0.6V$. For example, load regulation for an output current change of 1mA to 200mA is -0.3mV typical at $V_{OUT} = 0.6V$. At $V_{OUT} = 12V$, load regulation is:

$$\frac{12V}{0.6V} \cdot (-0.3mV) = -6mV$$

Table 1 shows 1% resistor divider values for some common output voltages with a resistor divider current of 5μA.

Table 1. Output Voltage Resistor Divider Values

V _{OUT} (V)	R1 (kΩ)	R2 (kΩ)
1.2	118	118
1.5	121	182
1.8	124	249
2.5	115	365
3	124	499
3.3	124	562
5	115	845
12	124	2370
15	124	3010

APPLICATIONS INFORMATION

Bypass Capacitance, Output Voltage Noise and Transient Response

The LT3062 regulator provides low output voltage noise over the 10Hz to 100kHz bandwidth while operating at full load with the addition of a bypass capacitor ($C_{REF/BYP}$) from the REF/BYP pin to GND. A good quality low leakage capacitor is recommended. This capacitor bypasses the reference of the regulator, providing a low frequency noise pole for the internal reference. With the use of 10nF for $C_{REF/BYP}$, the output voltage noise decreases to as low as 30 μ V_{RMS} when the output voltage is set for 0.6V. For higher output voltages (generated by using a resistor divider), the output voltage noise gains up accordingly when using $C_{REF/BYP}$ by itself.

To lower the output voltage noise for higher output voltages, include a feedforward capacitor (C_{FF}) from V_{OUT} to the ADJ pin. A good quality, low leakage capacitor is recommended. This capacitor bypasses the error amplifier of the regulator, providing a low frequency noise pole. With the use of 10nF for both C_{FF} and $C_{REF/BYP}$, output voltage noise decreases to 30 μ V_{RMS} when the output voltage is set to 5V by a 5 μ A feedback resistor divider. If the current in the feedback resistor divider is doubled, C_{FF} must also be doubled to achieve equivalent noise performance.

Higher values of output voltage noise may be measured if care is not exercised with regard to circuit layout and testing. Crosstalk from nearby traces can induce unwanted noise onto the LT3062's output. Power supply ripple rejection must also be considered. The LT3062 regulator does not have unlimited power supply rejection and will pass a small portion of the input noise through to the output.

Using a feedforward capacitor (C_{FF}) from V_{OUT} to the ADJ pin has the added benefit of improving transient response for output voltages greater than 0.6V. With no feedforward

capacitor, the settling time will increase as the output voltage is raised above 0.6V. Use the equation in Figure 2 to determine the minimum value of C_{FF} to achieve a transient response that is similar to 0.6V output voltage performance regardless of the chosen output voltage (see Figure 3 and Transient Response in the Typical Performance Characteristics section).

Feedforward capacitance can also be used in fixed-voltage parts; the feedforward capacitor is connected from OUT to ADJ in the same manner. In this case, the current in the internal feedback resistor divider is 5 μ A.

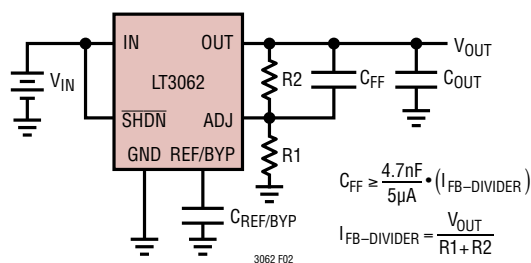


Figure 2. Feedforward Capacitor for Fast Transient Response

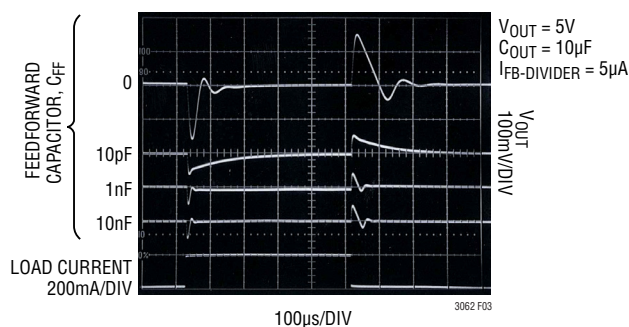


Figure 3. Transient Response vs Feedforward Capacitor

APPLICATIONS INFORMATION

During start-up, the internal reference soft-starts if a reference bypass capacitor is present. Regulator startup time is directly proportional to the size of the bypass capacitor, slowing to 6ms with a 10nF bypass capacitor (See SHDN Transient Response vs REF/BYP Capacitor in the Typical Performance Characteristics section). The reference bypass capacitor is actively pulled low during shutdown to reset the internal reference.

Start-up time is also affected by the use of a feedforward capacitor. Start-up time is directly proportional to the size of the feedforward capacitor and output voltage, and is inversely proportional to the feedback resistor divider current, slowing to 15ms with a 4.7nF feedforward capacitor and a 10 μ F output capacitor for an output voltage set to 5V by a 5 μ A feedback resistor divider.

Output Capacitance

The LT3062 regulator is stable with a wide range of output capacitors. The ESR of the output capacitor affects stability, most notably with small capacitors. Use a minimum output capacitor of 3.3 μ F with an ESR of 3 Ω or less to prevent oscillations. If a feedforward capacitor is used with output voltages set for greater than 24V, use a minimum output capacitor of 10 μ F. The LT3062 is a micropower device and output load transient response is a function of output capacitance. Larger values of output capacitance decrease the peak deviations and provide improved transient response for larger load current changes. Bypass capacitors, used to decouple individual components powered by the LT3062, will increase the effective output capacitor value. For applications with large load current transients, a low ESR ceramic capacitor in parallel with a bulk tantalum capacitor often provides an optimally damped response.

Give extra consideration to the use of ceramic capacitors. Manufacturers make ceramic capacitors with a variety of dielectrics, each with different behavior across temperature and applied voltage. The most common dielectrics are specified with EIA temperature characteristic codes of Z5U, Y5V, X5R and X7R. The Z5U and Y5V dielectrics provide high C-V products in a small package at low cost, but exhibit strong voltage and temperature coefficients as shown in Figures 4 and 5. When used with a 5V regulator, a 16V 10 μ F Y5V capacitor can exhibit an effective value as low as 1 μ F to 2 μ F for the DC bias voltage applied and over the operating temperature range. The X5R and X7R dielectrics yield much more stable characteristics and are more suitable for use as the output capacitor. The X7R type works over a wider temperature range and has better temperature stability, while the X5R is less expensive and is available in higher values. Care still must be exercised when using X5R and X7R capacitors; the X5R and X7R codes only specify operating temperature range and maximum capacitance change over temperature. Capacitance change due to DC bias with X5R and X7R capacitors is better than Y5V and Z5U capacitors, but can still be significant enough to drop capacitor values below appropriate levels. Capacitor DC bias characteristics tend to improve as component case size increases, but expected capacitance at operating voltage should be verified.

Voltage and temperature coefficients are not the only sources of problems. Some ceramic capacitors have a piezoelectric response. A piezoelectric device generates voltage across its terminals due to mechanical stress, similar to the way a piezoelectric accelerometer or microphone works. For a ceramic capacitor, the stress can be induced by vibrations in the system or thermal transients.

APPLICATIONS INFORMATION

The resulting voltages produced can cause appreciable amounts of noise. A ceramic capacitor produced the trace in Figure 6 in response to light tapping from a pencil. Similar vibration induced behavior can masquerade as increased output voltage noise.

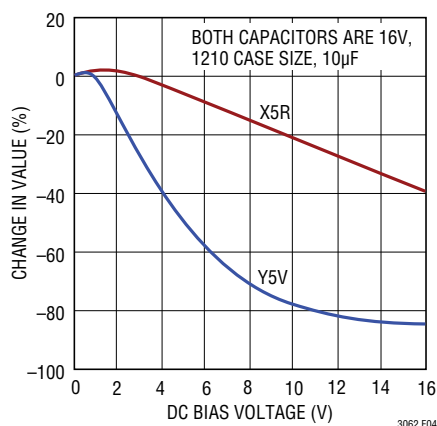


Figure 4. Ceramic Capacitor DC Bias Characteristics

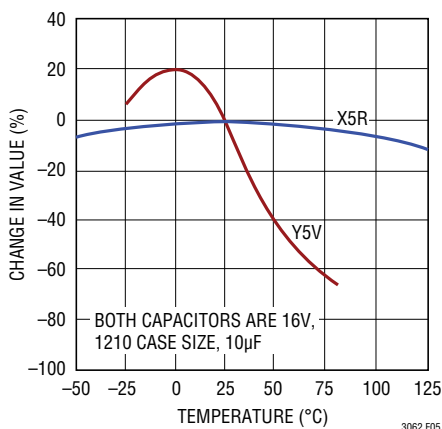


Figure 5. Ceramic Capacitor Temperature Characteristics

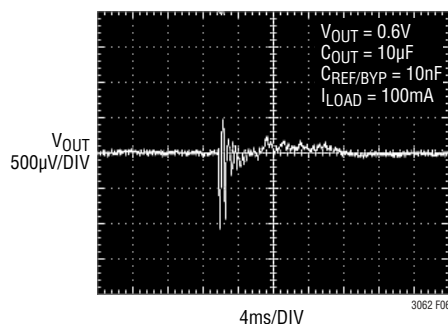


Figure 6. Noise Resulting from Tapping on a Ceramic Capacitor

Input Capacitance and Stability

Low ESR, ceramic input bypass capacitors are acceptable for applications without long input leads. However, applications connecting a power supply to an LT3062 circuit's IN and GND pins with long input wires combined with a low ESR, ceramic input capacitor are prone to voltage spikes, reliability concerns and application-specific board oscillations.

The input wire inductance found in many battery-powered applications, combined with the low ESR ceramic input capacitor, forms a high Q LC resonant tank circuit. In some instances this resonant frequency beats against the output current dependent LDO bandwidth and interferes with proper operation. Simple circuit modifications/solutions are then required. This behavior is not indicative of LT3062 instability, but is a common ceramic input bypass capacitor application issue.

The self-inductance, or isolated inductance, of a wire is directly proportional to its length. Wire diameter is not a major factor on its self-inductance. For example, the self-inductance of a 2-AWG isolated wire (diameter = 0.26") is about half the self-inductance of a 30-AWG wire (diameter = 0.01"). One foot of 30-AWG wire has approximately 465nH of self-inductance.

APPLICATIONS INFORMATION

Two methods can reduce wire self-inductance. One method divides the current flowing towards the LT3062 between two parallel conductors. In this case, the farther apart the wires are from each other, the more the self-inductance is reduced; up to a 50% reduction when placed a few inches apart. Splitting the wires connects two equal inductors in parallel, but placing them in close proximity creates mutual inductance adding to the self-inductance. The second and most effective way to reduce overall inductance is to place both forward and return current conductors (the input and GND wires) in very close proximity. Two 30-AWG wires separated by only 0.02", used as forward and return current conductors, reduce the overall self-inductance to approximately one-fifth that of a single isolated wire.

If a battery, mounted in close proximity, powers the LT3062, a 1 μ F input capacitor suffices for stability. However, if a distant supply powers the LT3062, use a larger value input capacitor. Use a rough guideline of 1 μ F (in addition to the 1 μ F minimum) per 8 inches of wire length. The minimum input capacitance needed to stabilize the application also varies with power supply output impedance variations. Placing additional capacitance on the LT3062's output also helps. However, this requires an order of magnitude more capacitance in comparison with additional LT3062 input bypassing. Series resistance between the supply and the LT3062 input also helps stabilize the application; as little as 0.1 Ω to 0.5 Ω suffices. This impedance dampens the LC tank circuit at the expense of dropout voltage. A better alternative is to use higher ESR tantalum or electrolytic capacitors at the LT3062 input in place of ceramic capacitors.

Overload Recovery

Like many IC power regulators, the LT3062 has safe operating area protection. The safe area protection decreases current limit as input-to-output voltage increases and keeps the power transistor inside a safe operating region for all values of input-to-output voltage. The protective design provides some output current at all values of input-to-output voltage up to the device breakdown.

When power is first applied, as input voltage rises, the output follows the input, allowing the regulator to start up into very heavy loads. During start-up, as the input voltage is rising, the input-to-output voltage differential is small, allowing the regulator to supply large output currents. With a high input voltage, a problem can occur wherein removal of an output short will not allow the output to recover.

The problem occurs with a heavy output load when the input voltage is high and the output voltage is low. Common situations include immediately after the removal of a short-circuit or if the shutdown pin is pulled high after the input voltage has already been turned on. The load line for such a load may intersect the output current curve at two points. If this happens, there are two stable output operating points for the regulator. With this double intersection, the input power supply may need to be cycled down to zero and brought up again to make the output recover.

APPLICATIONS INFORMATION

Thermal Considerations

The power handling capability of the device will be limited by the maximum rated junction temperature (125°C for LT3062E, LT3062I or 150°C for LT3062MP, LT3062H). Two components comprise the power dissipated by the device:

1. Output current multiplied by the input/output voltage differential: $I_{OUT} \cdot (V_{IN} - V_{OUT})$, and
2. GND pin current multiplied by the input voltage: $I_{GND} \cdot V_{IN}$

GND pin current is determined using the GND Pin Current curves in the Typical Performance Characteristics section. Power dissipation will be equal to the sum of the two components listed above.

The LT3062 regulator has internal thermal limiting that protects the device during overload conditions. For continuous normal conditions, the maximum junction temperature of 125°C (E-grade, I-grade) or 150°C (MP-grade, H-grade) must not be exceeded. Carefully consider all sources of thermal resistance from junction to ambient including other heat sources mounted in proximity to the LT3062.

The undersides of the LT3062 packages have exposed metal from the lead frame to the die attachment. The package allows heat to directly transfer from the die junction to the printed circuit board metal to control maximum operating junction temperature. The dual-in-line pin arrangement allows metal to extend beyond the ends of the package on the topside (component side) of a PCB. Connect this metal to GND on the PCB. The multiple IN and OUT pins of the LT3062 also assist in spreading heat to the PCB.

For surface mount devices, heat sinking is accomplished by using the heat spreading capabilities of the PC board and its copper traces. Copper board stiffeners and plated through-holes can also be used to spread the heat generated by power devices.

The following tables list thermal resistance for several different board sizes and copper areas. All measurements were taken in still air on a 4 layer FR-4 board with 1oz solid internal planes and 2oz top/bottom external trace planes with a total board thickness of 1.6mm. The four layers were electrically isolated with no thermal vias present. PCB layers, copper weight, board layout and thermal vias will affect the resultant thermal resistance. For more information on thermal resistance and high thermal conductivity test boards, refer to JEDEC standard JESD51, notably JESD51-12 and JESD51-7. Achieving low thermal resistance necessitates attention to detail and careful PCB layout.

Table 2. Measured Thermal Resistance for DFN Package

COPPER AREA		BOARD AREA (mm ²)	THERMAL RESISTANCE (JUNCTION-TO-AMBIENT)
TOPSIDE* (mm ²)	BACKSIDE (mm ²)		
2500	2500	2500	38°C/W
1000	2500	2500	38°C/W
225	2500	2500	40°C/W
100	2500	2500	45°C/W

*Device is mounted on topside

Table 3. Measured Thermal Resistance for MSOP Package

COPPER AREA		BOARD AREA (mm ²)	THERMAL RESISTANCE (JUNCTION-TO-AMBIENT)
TOPSIDE* (mm ²)	BACKSIDE (mm ²)		
2500	2500	2500	29°C/W
1000	2500	2500	30°C/W
225	2500	2500	32°C/W
100	2500	2500	45°C/W

*Device is mounted on topside

APPLICATIONS INFORMATION

Calculating Junction Temperature

Example: Given an output voltage of 2.5V, an input voltage range of 12V $\pm$ 5%, an output current range of 0mA to 50mA and a maximum ambient temperature of 85°C, what will the maximum junction temperature be?

The power dissipated by the device equals:

$$I_{OUT(MAX)} \cdot (V_{IN(MAX)} - V_{OUT}) + I_{GND} \cdot V_{IN(MAX)}$$

where,

$$I_{OUT(MAX)} = 50\text{mA}$$

$$V_{IN(MAX)} = 12.6\text{V}$$

$$I_{GND} \text{ at } (I_{OUT} = 50\text{mA}, V_{IN} = 12\text{V}) = 1\text{mA}$$

So,

$$P = 50\text{mA} \cdot (12.6\text{V} - 2.5\text{V}) + 1\text{mA} \cdot 12.6\text{V} = 0.518\text{W}$$

Using a DFN package, the thermal resistance will be in the range of 38°C/W to 45°C/W depending on the copper area. So the junction temperature rise above ambient will be approximately equal to:

$$0.518\text{W} \cdot 45^\circ\text{C/W} = 23.3^\circ\text{C}$$

The maximum junction temperature equals the maximum ambient temperature plus the maximum junction temperature rise above ambient or:

$$T_{JMAX} = 85^\circ\text{C} + 23.3^\circ\text{C} = 108.3^\circ\text{C}$$

Protection Features

The LT3062 incorporates several protection features that make it ideal for use in battery-powered circuits. In addition to the normal protection features associated with monolithic regulators, such as current limiting and thermal limiting, the device also protects against reverse-input voltages, reverse-output voltages and reverse output-to-input voltages.

Current limit protection and thermal overload protection protect the device against current overload conditions at the output of the device. The typical thermal limit temperature is 165°C. For normal operation, do not exceed a junction temperature of 125°C (LT3062E, LT3062I) or 150°C (LT3062MP, LT3062H).

The LT3062 IN pin withstands reverse voltages of 50V. The device limits current flow to less than 1mA (typically less than 250μA) and no negative voltage appears at OUT. The device protects both itself and the load against batteries that are plugged in backwards.

The $\overline{\text{SHDN}}$ pin cannot be driven below GND unless tied to the IN pin. If the $\overline{\text{SHDN}}$ pin is driven below GND while IN is powered, the output will turn on. SHDN pin logic cannot be referenced to a negative rail.

The LT3062 incurs no damage if its output is pulled below ground. If the input is left open-circuit or grounded, the output can be pulled below ground by 50V. No current flows through the pass transistor from the output. However, current flows in (but is limited by) the resistor divider that sets the output voltage. Current flows from the bottom resistor in the divider and from the ADJ pin's internal clamp through the top resistor in the divider to the external circuitry pulling OUT below ground. If the input is powered by a voltage source, the output sources current equal to its current limit capability and the LT3062 protects itself by thermal limiting. In this case, grounding the $\overline{\text{SHDN}}$ pin turns off the device and stops the output from sourcing current.

APPLICATIONS INFORMATION

The LT3062 incurs no damage if the ADJ pin is pulled above or below ground by 50V. If the input is left open circuit or grounded, the ADJ pin performs like a large resistor (typically 30k) in series with a diode when pulled above or below ground.

In circuits where a backup battery is required, several different input/output conditions can occur. The output voltage may be held up temporarily or otherwise while the input is pulled to ground, pulled to some intermediate voltage or left open-circuit. Current flow back into the OUT and ADJ pins follows the curve shown in Figure 7. If the LT3062's IN pin is forced below the OUT pin or the OUT pin is pulled above the IN pin, regardless of the state of the SHDN pin, input current typically drops to less than 1μA.

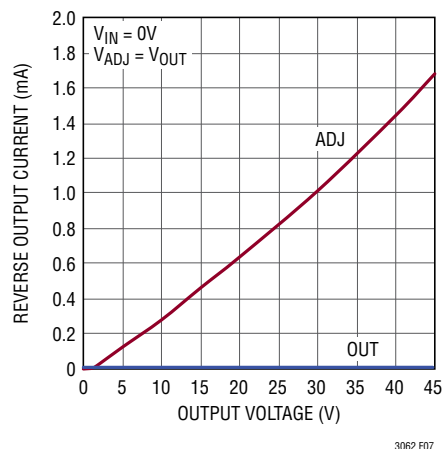
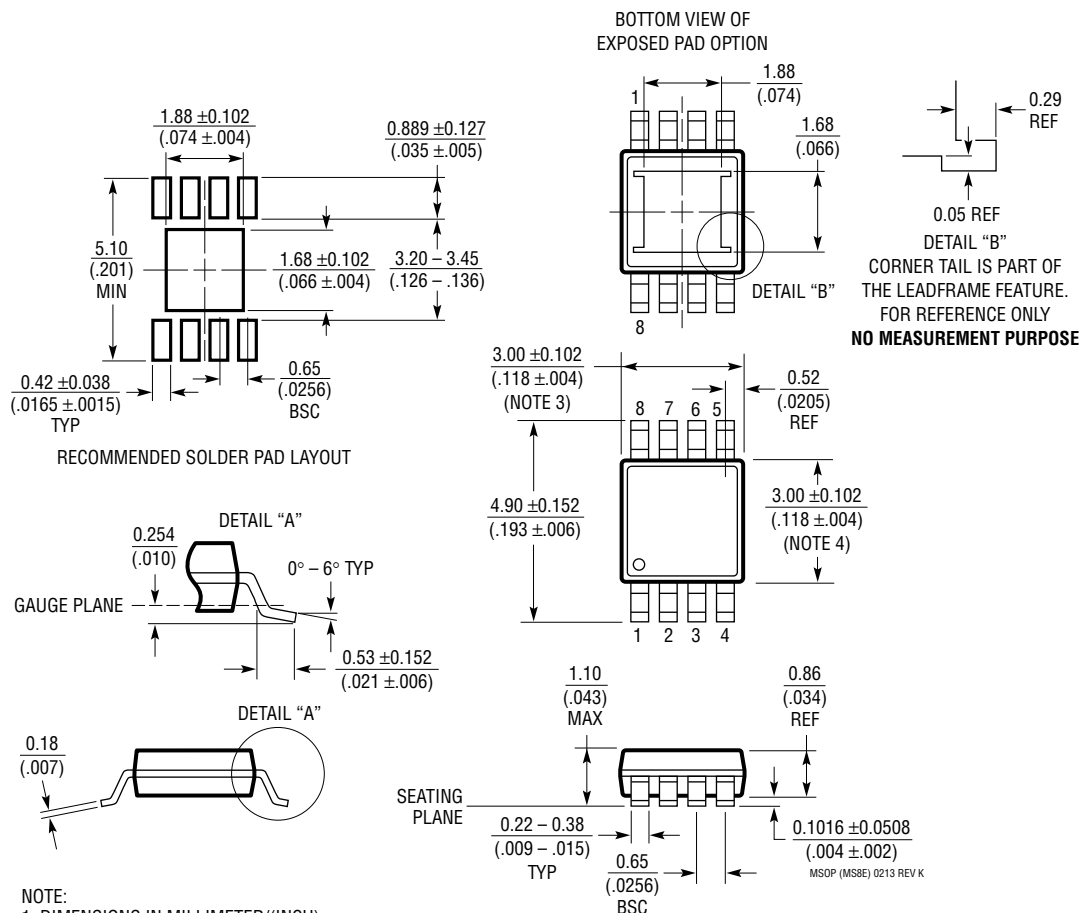


Figure 7. Reverse Output Current

PACKAGE DESCRIPTION

Please refer to <http://www.linear.com/product/LT3062#packaging> for the most recent package drawings.

MS8E Package 8-Lead Plastic MSOP, Exposed Die Pad (Reference LTC DWG # 05-08-1662 Rev K)



NOTE:

1. DIMENSIONS IN MILLIMETER/(INCH)
2. DRAWING NOT TO SCALE
3. DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.152mm ($.006$) PER SIDE
4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm ($.006$) PER SIDE
5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm ($.004$) MAX
6. EXPOSED PAD DIMENSION DOES INCLUDE MOLD FLASH. MOLD FLASH ON E-PAD SHALL NOT EXCEED 0.254mm ($.010$) PER SIDE.

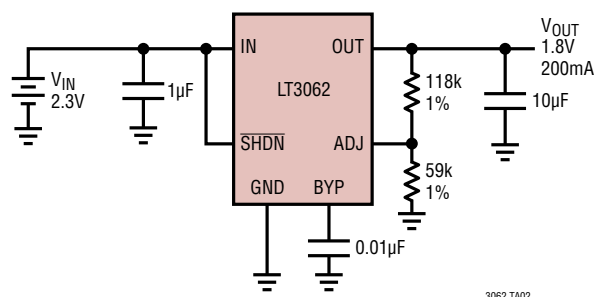
REVISION HISTORY

REV	DATE	DESCRIPTION	PAGE NUMBER
A	09/16	Add 3.3V/5V fixed voltage options	1-22

LT3062 Series

TYPICAL APPLICATION

1.8V Low Noise Regulator



3062 TA02

RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1761	100mA, Low Noise LDO	300mV Dropout Voltage, Low Noise: 20µVRMS, VIN = 1.8V to 20V, ThinSOT™ Package
LT1762	150mA, Low Noise LDO	300mV Dropout Voltage, Low Noise: 20µVRMS, VIN = 1.8V to 20V, MS8 Package
LT1763	500mA, Low Noise LDO	300mV Dropout Voltage, Low Noise: 20µVRMS, VIN = 1.8V to 20V, S08 Package
LT1962	300mA, Low Noise LDO	270mV Dropout Voltage, Low Noise: 20µVRMS, VIN = 1.8V to 20V, MS8 Package
LT3008	20mA, 45V, 3µA IQ Micropower LDO	300mV Dropout Voltage, Low IQ: 3µA, VIN = 2V to 45V, VOUT = 0.6V to 44.5V; ThinSOT and 2mm × 2mm DFN-6 Packages
LT3009	20mA, 3µA IQ Micropower LDO	280mV Dropout Voltage, Low IQ: 3µA, VIN = 1.6V to 20V, 2mm × 2mm DFN-6 and SC70 Packages
LT3042	20V, 200mA Ultralow Noise, Ultrahigh PSRR LDO	0.8µVRMS Noise and 79dB PSRR at 1MHz, VIN = 1.8V to 20V, 350mV Dropout Voltage, Programmable ILIM and Power Good, 3mm × 3mm DFN and MSOP Packages
LT3050	100mA, Low Noise Linear Regulator with Precision Current Limit and Diagnostic Functions.	340mV Dropout Voltage, Low Noise: 30µVRMS, VIN: 1.6V to 45V, VOUT: 0.6V to 44.5V, Programmable Precision Current Limit: ±5%, Programmable Minimum IOUT Monitor, Output Current Monitor, Fault Indicator, Reverse Protection; 12-Lead 3mm × 2mm DFN and MSOP Packages.
LT3060	45V VIN, Micropower, Low Noise, 100mA Low Dropout Linear Regulator	Input Voltage Range: 1.6V to 45V, Quiescent Current: 40µA, Dropout Voltage: 300mV, Low Noise: 30µVRMS (10Hz to 100kHz), Adjustable Output: VREF = 600mV, 8-Lead 2mm × 2mm DFN and 8-Lead ThinSOT
LT3061	45V VIN, Micropower, Low Noise, 100mA Low Dropout Linear Regulator with Active Output Discharge	Active Output Discharge, Input Voltage Range: 1.6V to 45V, Quiescent Current: 45µA, Dropout Voltage: 250mV, Low Noise: 30µVRMS (10Hz to 100kHz), Adjustable Output: 600mV to 19V, 8-Lead MSOP and 2mm × 3mm DFN
LT3063	45V VIN, Micropower, Low Noise, 200mA Low Dropout Linear Regulator with Active Output Discharge	Active Output Discharge, Input Voltage Range: 1.6V to 45V, Quiescent Current: 45µA, Dropout Voltage: 300mV, Low Noise: 30µVRMS (10Hz to 100kHz), Adjustable Output: 600mV to 19V, 8-Lead MSOP and 2mm × 3mm DFN
LT3082	200mA, Parallelable, Single Resistor, Low Dropout Linear Regulator	Outputs May Be Paralleled for Higher Output, Current or Heat Spreading, Wide Input Voltage Range: 1.2V to 40V Low Value Input/Output Capacitors Required: 0.22µF, Single Resistor Sets Output Voltage Initial Set Pin Current Accuracy: 1%, Low Output Noise: 40µVRMS (10Hz to 100kHz) Reverse-Battery Protection, Reverse-Current Protection; 8-Lead SOT-23, 3-Lead SOT-223 and 8-Lead 3mm × 3mm DFN Packages
LT3085	500mA, Parallelable, Low Noise, Low Dropout Linear Regulator	275mV Dropout Voltage (2-Supply Operation), Low Noise: 40µVRMS, VIN: 1.2V to 36V, VOUT: 0V to 35.7V, Current-Based Reference with 1-Resistor VOUT Set, Directly Parallelable (No Op Amp Required), Stable with Ceramic Capacitors; MS8E and 2mm × 3mm DFN-6 Packages
LT3092	200mA 2-Terminal Programmable Current Source	Programmable 2-Terminal Current Source, Maximum Output Current: 200mA Wide Input Voltage Range: 1.2V to 40V, Resistor Ratio Sets Output Current Initial Set Pin Current Accuracy: 1%, Current Limit and Thermal Shutdown Protection Reverse-Voltage Protection, Reverse-Current Protection; 8-Lead SOT-23, 3-Lead SOT-223 and 8-Lead 3mm × 3mm DFN Packages

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