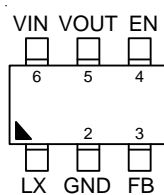


Pin Configurations

(TOP VIEW)

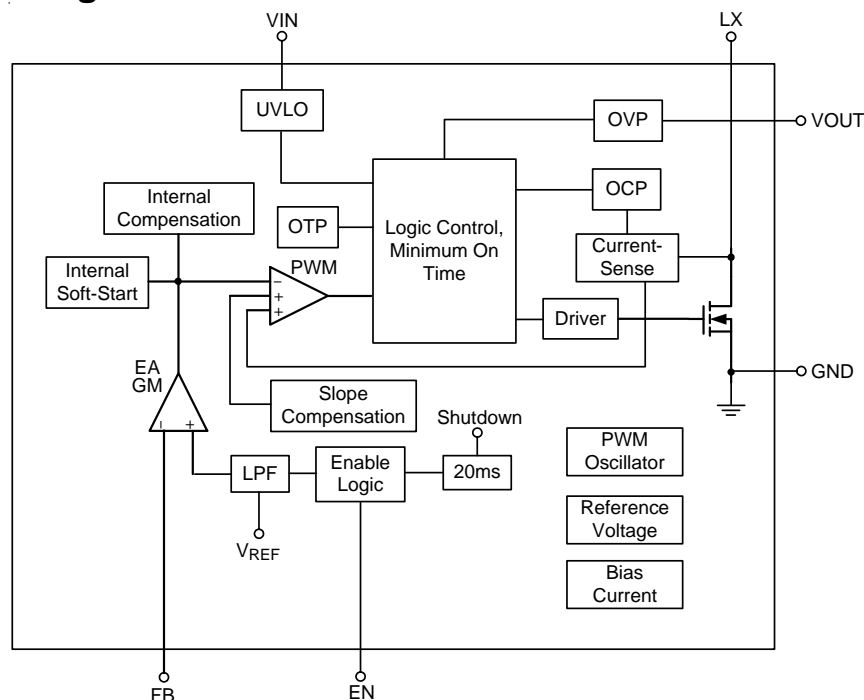


TSOT-23-6

Functional Pin Description

Pin No.	Pin Name	Pin Function
1	LX	Switch Node.
2	GND	Ground.
3	FB	Feedback Voltage Input. Put a resistor to GND to setting the current.
4	EN	Enable Control Input (PWM Dimming).
5	VOUT	Output Voltage Pin.
6	VIN	Supply Voltage Input.

Function Block Diagram



Operation

The RT4533/A is a constant frequency, current mode Boost LED driver. In normal operation, the N-MOSFET is turned on when the PWM Control circuit is set by the oscillator and is turned off when the current comparator resets the PWM Control circuit. While the N-MOSFET is turned off, the inductor current conducts through the external diode.

UVLO

When the input voltage is lower than the UVLO threshold (2.37V typ.), the driver will turn off. There is a 80mV for the UVLO hysteresis control.

Soft-Start

When the device is enabled, the Comp ramps up to the target voltage in a specific time. This ensures that the output voltage rises slowly to reduce the input inrush current.

EN Dimming

The EN pin is used for the control input for both PWM dimming mode and digital dimming mode. Shutdown delay when the EN voltage is logic low for more than specific time, the driver will be shut down.

OCP

The driver provides cycle-by-cycle current limit function to control the current on power switch. The boost switch turns off when the inductor current reaches this current threshold and it remains off until the beginning of the next switching cycle. This protects the RT4533/A and external component under overload conditions.

OVP

The Over Voltage Protection is detected by a resistor divider circuit from VOUT.

Once VOUT goes over the OVP voltage, LX pin stops switching and the power N-MOSFET will be turned off. Then, the VOUT will be discharged by external resistor. OVP will be released until EN goes from Lo to Hi again.

OTP

As the die temperature is higher than 160°C, the chip also will enter protection mode. The power MOSFET will be turned off during protection mode to prevent abnormal operation. The device is released from shutdown automatically when the junction temperature decreases by 15°C.

Absolute Maximum Ratings (Note 1)

• VIN	-----	-0.3V to 6V
• EN, FB to GND	-----	-0.3V to (VIN + 0.3V)
• LX to GND	-----	-0.3V to 38V
• VOUT to GND	-----	-0.3V to 40V
• Power Dissipation, PD @ TA = 25°C		
TSOT-23-6	-----	0.5W
• Package Thermal Resistance (Note 2)		
TSOT-23-6, θ_{JA}	-----	197.4°C/W
• Junction Temperature	-----	150°C
• Lead Temperature (Soldering, 10 sec.)	-----	260°C
• Storage Temperature Range	-----	-65°C to 150°C
• ESD Susceptibility (Note 3)		
HBM (Human Body Model)	-----	2kV
MM (Machine Model)	-----	200V

Recommended Operating Conditions (Note 4)

• Supply Input Voltage, VIN	-----	2.5V to 5.5V
• Junction Temperature Range	-----	-40°C to 125°C
• Ambient Temperature Range	-----	-40°C to 85°C

Electrical Characteristics

(VIN = 3.6V, CIN = 4.7μF, COUT = 1μF, L = 22μH, fSW = 1.1MHz, TA = 25°C, unless otherwise specified)

Parameter		Symbol	Test Conditions	Min	Typ	Max	Unit
Power Supply							
IC Operating Current (switching)		IQ_SW	VFB = 0V	0.25	0.8	1.5	mA
IC Quiescent Current (non-switching)		IQ	VFB = 0.4V	0.25	0.35	0.5	mA
VIN Pin Shutdown Current		ISD	EN = GND	--	0.3	1	μA
Under-Voltage Lockout Threshold,		UVLO	VIN Rising Edge	2.25	2.37	2.5	V
Under-Voltage Lockout Hysteresis		UVLO_hys	VIN Falling Edge	--	80	--	mV
Enable and Reference Control							
EN Input Voltage	Logic-High	VIH		1.4	--	--	V
	Logic-Low	VIL		--	--	0.4	
EN Pull Low Resistance		REN		--	300	--	kΩ
EN Minimum Shut down Pulse Width		Toff		20	--	--	ms
EN Minimum Logic High Pulse Width		THigh_Min		0.5	--	--	μs
EN Minimum logic low pulse width		TLow		1	--	10000	μs
Internal Comp Resistance		Rcomp		--	7	--	MΩ
PWM Minimum Duty		Dpwm min		--	3	--	%
PWM Frequency		fPWM		5	--	100	kHz

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Voltage And Current Control						
Feedback Regulation Voltage	RT4533	V_{FB}	196	200	204	mV
	RT4533A		294	300	306	
Feedback Pin Bias Current	I_{FB}		--	--	0.1	μA
Oscillator Frequency	f_{SW}		0.99	1.1	1.21	MHz
Maximum Duty Cycle	D_{MAX}		--	93	--	%
Power Switch						
N-MOSFET On-Resistance	$R_{DS(ON)}$	$V_{IN} = 3.6V$	--	0.4	0.7	Ω
OC and OLP						
Peak N-MOSFET Current Limit	I_{LIM}		1.0	1.2	1.4	A
Open LED Protection Threshold	OVP		35	36.5	38	V
Thermal Shutdown						
Thermal Shutdown Threshold	T_{SD}		--	160	--	$^{\circ}C$
Thermal Shutdown Hysteresis	ΔT_{SD}		--	15	--	$^{\circ}C$

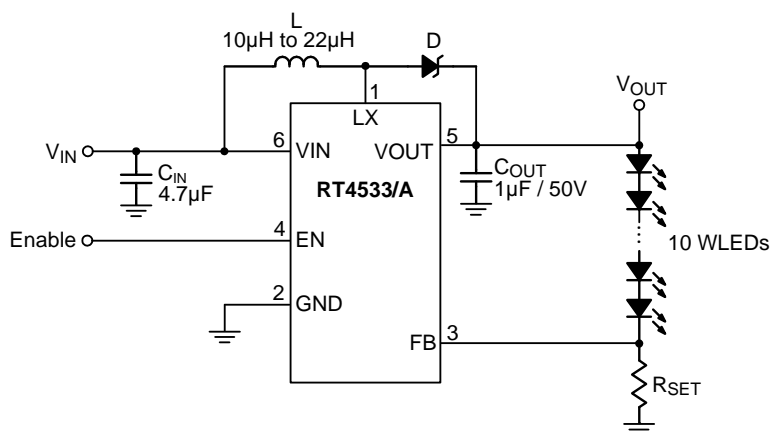
Note 1. Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Note 2. θ_{JA} is measured at $T_A = 25^{\circ}C$ on a high effective thermal conductivity four-layer test board per JEDEC 51-7.

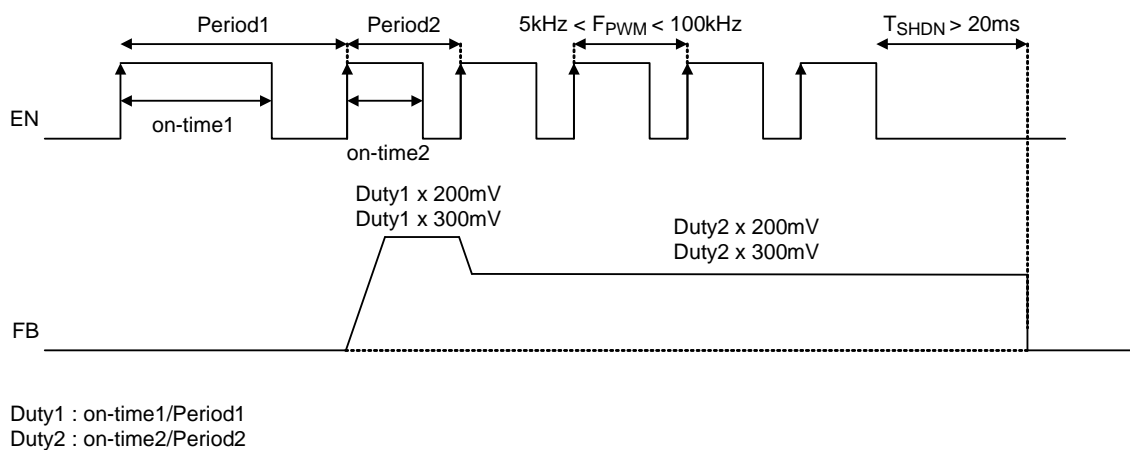
Note 3. Devices are ESD sensitive. Handling precaution is recommended.

Note 4. The device is not guaranteed to function outside its operating conditions.

Typical Application Circuit

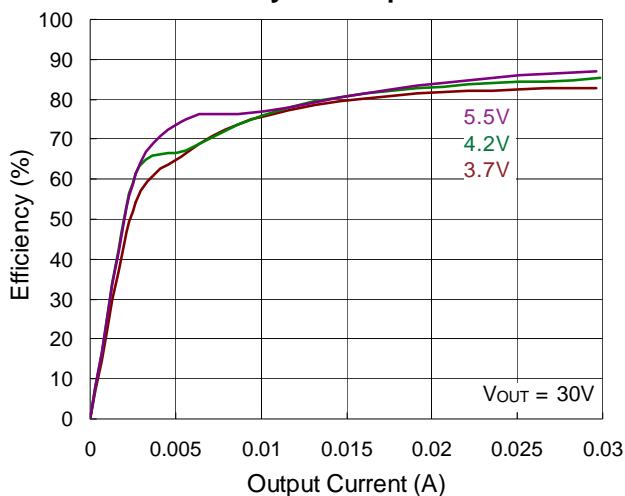


Timing Diagram

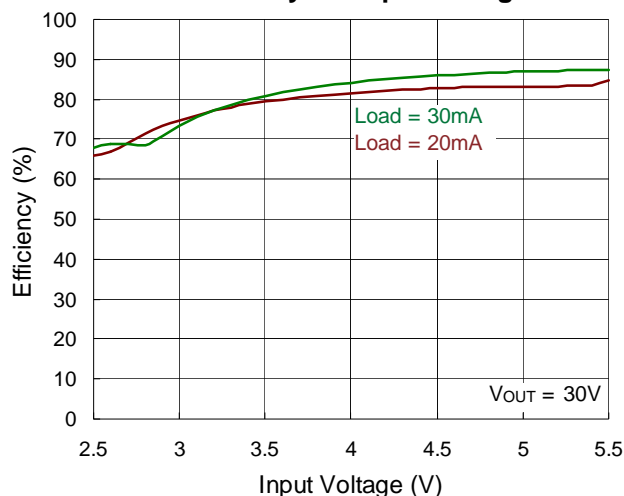


Typical Operating Characteristics

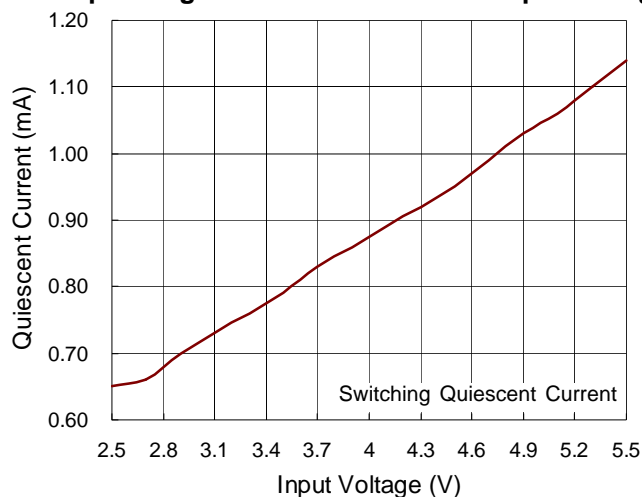
Efficiency vs. Output Current



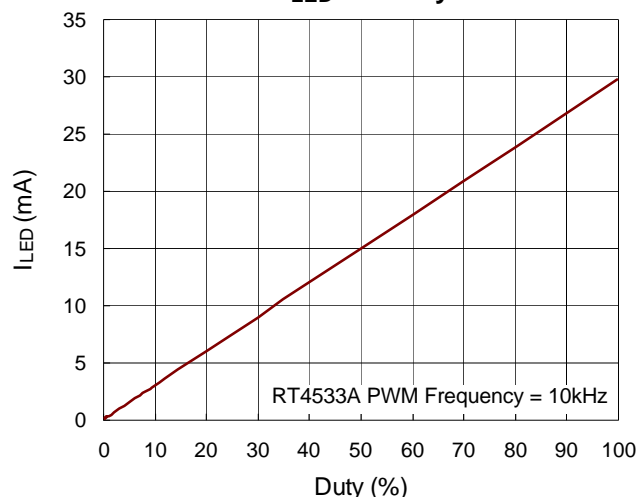
Efficiency vs. Input Voltage



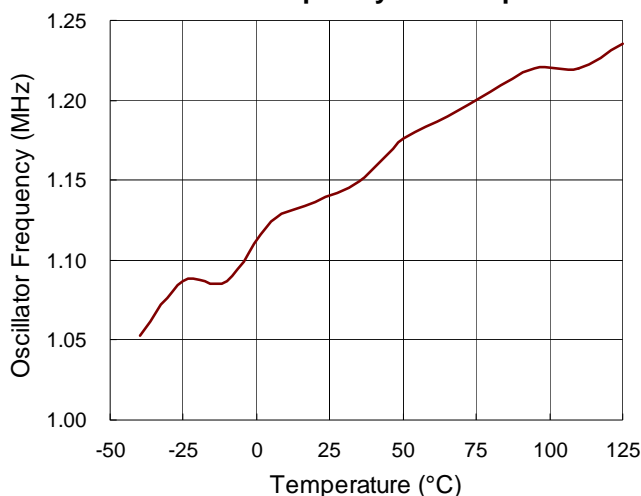
Operating Quiescent Current vs. Input Voltage



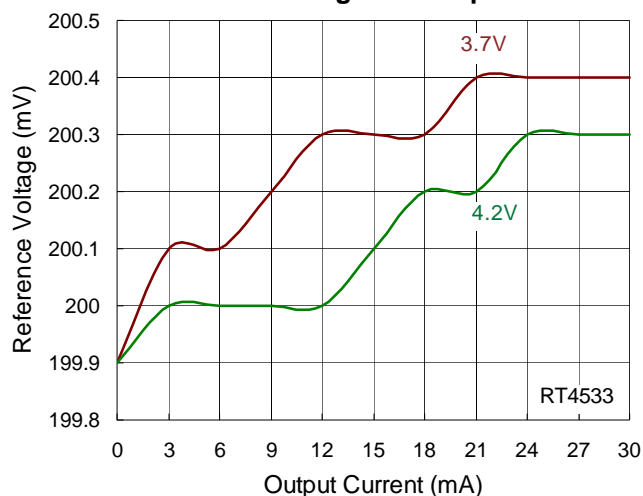
I_{LED} vs. Duty



Oscillator Frequency vs. Temperature



Reference Voltage vs. Output Current



Application Information

Soft-Start

The RT4533/A includes a soft-start function to avoid high inrush current during start-up. The soft-start function is achieved by clamping the output voltage of the error amplifier with another voltage source that is increased slowly from zero to near V_{IN} .

LED Current

The loop control of the Boost converter keeps V_{FB} equal to a reference voltage, V_{REF} . Therefore, when R_{SET} is connected between the FB pin and GND, the LED current will be determined by the current through R_{SET} , which is equal to V_{FB} / R_{SET} .

Current Limit

The current flowing through the inductor during a charging period is detected by a current sensing circuit. If the value exceeds the current limit, the N-MOSFET will be turned off. The inductor will then be forced to leave charging stage and enter discharging stage. Therefore, the inductor current will not increase to reach current limit.

Shutdown Delay

When the EN voltage is in logic low for 20ms during PWM dimming, the system will enter shutdown.

PWM Dimming

When the EN pin is constantly high, the FB voltage is regulated to 200mV / 300mV typically. The RT4533/A allows a PWM signal from EN pin to reduce the regulation voltage; and achieve LED brightness dimming. The relationship between the duty cycle and FB voltage is given by

$$V_{FB} = \text{Duty} \times V_{REF}$$

Where

Duty = duty cycle of the PWM signal V_{REF} = internal reference voltage (200mV / 300mV typ.)

The RT4533/A choose the internal 200mV reference voltage via the duty cycle of the PWM signal. Therefore, although a PWM signal is used for brightness dimming, only the LED DC current is modulated, which is often referred as

analog dimming. This eliminates the audible noise which often occurs when the LED current is pulsed in replica of the frequency and duty cycle of PWM control. For optimum performance, use the PWM dimming frequency in the range of 5kHz to 100kHz, and the PWM dimming frequency is strongly suggested to be over than 20kHz to avoid audio noise.

Inductor Selection

The recommended value of inductor for 10 LEDs or high brightness LED applications is 10μH to 22μH. Smaller size and better efficiency are the major concerns for portable devices. The inductor should have low core loss at 1MHz and low DCR for better efficiency. The inductor saturation current rating should be considered to cover the inductor peak current. Table 1 lists the recommended inductor for the RT4533/A.

Table 1

Part Number	L (μH)	DCR MAX (mΩ)	Saturation Current (A)	Size (L x W x H mm)	Vendor
NR4018T100M	10	216	1.2	4 x 4 x 1.8	TAIYO
NR4018T220M	22	432	0.8	4 x 4 x 1.8	TAIYO

Capacitor Selection

For low ripple voltage, ceramic capacitors with low ESR are recommended. X5R and X7R types are suitable because of their wide voltage range and good operating temperature characteristics. For the application of the RT4533/A to drive 10 LEDs in series, a 4.7μF for input capacitor, an 1μF / 50V for output capacitor.

Thermal Considerations

For continuous operation, do not exceed absolute maximum operation junction temperature. The maximum power dissipation depends on the thermal resistance of IC package, PCB layout, the rate of surroundings airflow and temperature difference between junction to ambient. The maximum power dissipation can be calculated by following formula :

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

Where $T_{J(MAX)}$ is the maximum operation junction temperature, T_A is the ambient temperature and the θ_{JA} is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance, θ_{JA} , is layout dependent. For TSOT-23-6 package, the thermal resistance, θ_{JA} , is 197.4°C/W on the standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at $T_A = 25^\circ\text{C}$ can be calculated by following formula :

$$P_{D(MAX)} = (125^\circ\text{C} - 25^\circ\text{C}) / (197.4^\circ\text{C/W}) = 0.5\text{W for TSOT-23-6 package}$$

The maximum power dissipation depends on operating ambient temperature for fixed $T_{J(MAX)}$ and thermal resistance θ_{JA} . The derating curve in Figure 1 of derating curves allows the designer to see the effect of rising ambient temperature on the maximum power allowed.

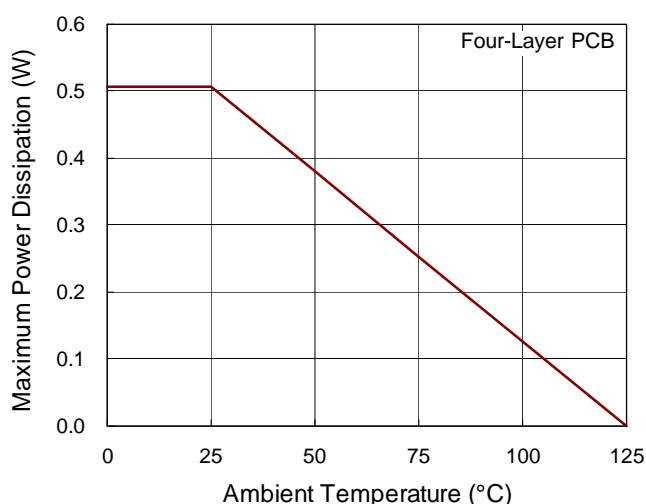


Figure 1. Derating Curve of Maximum Power Dissipation

Layout Consideration

For best performance of the RT4533/A, the following guide lines must be strictly followed.

- ▶ Input and Output capacitors should be placed close to the IC and connected to ground plane to reduce noise coupling.
- ▶ The GND and Exposed Pad should be connected to a strong ground plane for heat sinking and noise protection.
- ▶ Keep the main current traces as possible as short and wide.
- ▶ LX node of DC/DC converter is with high frequency voltage swing. It should be kept at a small area.
- ▶ Place the feedback components as close as possible to the IC and keep away from the noisy devices.

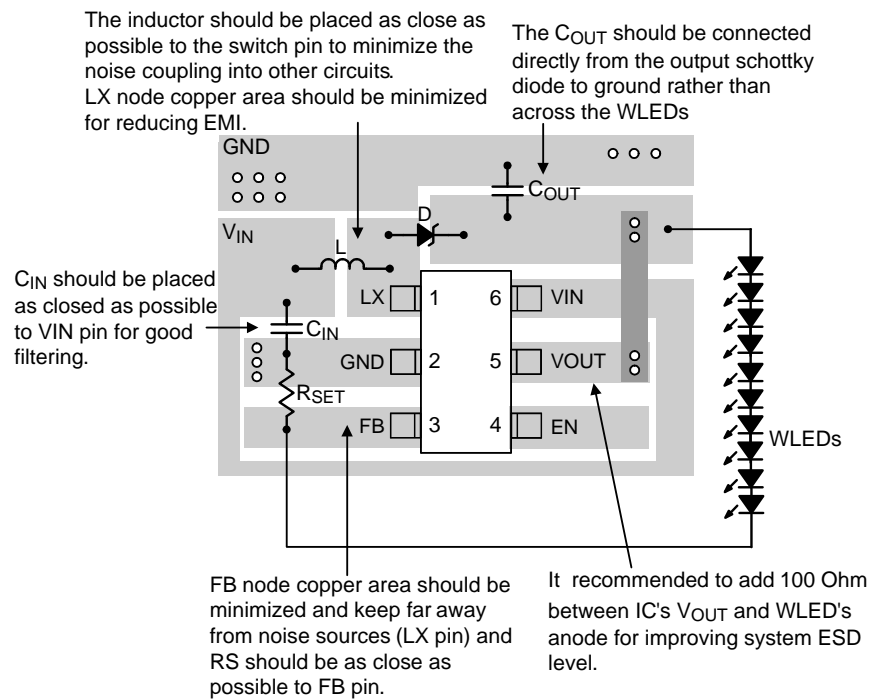
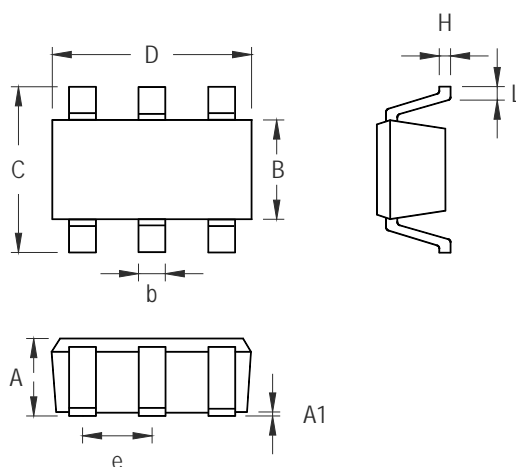


Figure 2. PCB Layout Guide

Outline Dimension



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.700	1.000	0.028	0.039
A1	0.000	0.100	0.000	0.004
B	1.397	1.803	0.055	0.071
b	0.300	0.559	0.012	0.022
C	2.591	3.000	0.102	0.118
D	2.692	3.099	0.106	0.122
e	0.838	1.041	0.033	0.041
H	0.080	0.254	0.003	0.010
L	0.300	0.610	0.012	0.024

TSOT-23-6 Surface Mount Package

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