

Figure 1. Block Diagram

# **Pin Descriptions**

Pin Name	Type	Pin Description
X1, X2	I/O	32.768 kHz crystal connection. When using an external oscillator, apply the clock to X1
		and a DC mid-level to X2 (see Crystal Type section for suggestions).
ACS	Output	Alarm/Calibration/SquareWave: This is an open-drain output that requires an external
		pullup resistor. The alarm, calibration, and square wave functions all share this output.
		In Alarm mode, this pin acts as the active-low alarm output. In Calibration mode, a 512
		Hz square-wave is driven out. In SquareWave mode, the user may select a frequency of
		1, 512, 4096, or 32768 Hz to be used as a continuous output. Refer to <i>Table 3. Control</i>
an i	7/0	Bit Settings for ACS Pin to determine the bit settings for each mode.
SDA	I/O	Serial Data & Address: This is a bi-directional line for the two-wire interface. It is
		open-drain and is intended to be wire-OR'd with other devices on the two-wire bus. The
		input buffer incorporates a Schmitt trigger for noise immunity and the output driver
		includes slope control for falling edges. A pull-up resistor is required.
SCL	Input	Serial Clock: The serial clock line for the two-wire interface. Data is clocked out of the
		part on the falling edge, and data into the device on the rising edge. The SCL input also
		incorporates a Schmitt trigger input for noise immunity.
VBAK	Supply	Backup supply voltage: A 3V battery or a large value capacitor. If no backup supply is
		used, this pin should be tied to V <sub>SS</sub> . The trickle charger is UL recognized and ensures no
		excessive current when using a lithium battery.
VDD	Supply	Supply Voltage.
VSS	Supply	Ground

#### Overview

The FM3130 device combines a serial nonvolatile RAM with a real-time clock (RTC) and alarm. These complementary but distinct functions share a common interface in a single package. Although monolithic, the product is organized as two logical devices, the F-RAM memory and the RTC/alarm. From the system perspective, they appear to be two separate devices with unique IDs on the serial bus.

The memory is organized as a stand-alone 2-wire nonvolatile memory with a standard device ID value. The real-time clock and alarm are accessed with a separate 2-wire device ID. This allows clock/calendar data to be read while maintaining the most recently used memory address. The clock and alarm are controlled by 15 special function registers. The registers are maintained by the power source on the VBAK pin, allowing them to operate from battery or backup capacitor power when  $V_{\rm DD}$  drops below a set threshold. Each functional block is described below.

## **Memory Operation**

The FM3130 integrates a 64Kb F-RAM. The memory is organized in bytes, 8192 addresses of 8 bits each. The memory is based on F-RAM technology. Therefore it can be treated as RAM and is read or written at the speed of the two-wire bus with no delays for write operations. It also offers effectively unlimited write endurance unlike other nonvolatile memory technologies. The two-wire interface protocol is described further on page 12.

The memory array can be write-protected by software. Two bits (WP0, WP1) in register 0Eh control the protection setting as shown in the following table. Based on the setting, the protected addresses cannot be written and the 2-wire interface will not acknowledge any data to protected addresses. The special function registers containing these bits are described in detail below.

**Table 1. F-RAM Write-Protect** 

Write-Protect Range	WP1	WP0
None	0	0
Bottom 1/4	0	1
Bottom 1/2	1	0
Full array	1	1

The WP bits are battery-backed. On a powerup without a backup source, the WP bits are cleared to a '0' state.

# **Real-Time Clock Operation**

The real-time clock (RTC) is a timekeeping device that can be battery or capacitor backed for permanently-powered operation. It offers a software calibration feature that allows high accuracy.

The RTC consists of an oscillator, clock divider, and a register system for user access. It divides down the 32.768 kHz time-base and provides a minimum resolution of seconds (1Hz). Static registers provide the user with read/write access to the time values. It includes registers for seconds, minutes, hours, day-of-the-week, date, months, and years. A block diagram (Figure 2) illustrates the RTC function.

The user registers are synchronized with the timekeeper core using R and W bits in register 00h described below. Changing the R bit from 0 to 1 transfers timekeeping information from the core into holding registers that can be read by the user. If a timekeeper update is pending when R is set, then the core will be updated prior to loading the user registers. The registers are frozen and will not be updated again until the R bit is cleared to '0'. R is used to read the time.

Setting the W bit to '1' locks the user registers. Clearing it to a '0' causes the values in the user registers to be loaded into the timekeeper core. W is used for writing new time values. Users should be certain not to load invalid values, such as FFh, to the timekeeping registers. Updates to the timekeeping core occur continuously except when locked. All timekeeping registers must be initialized at the first powerup or when the LB bit is set. See the description of the LB bit on page 11.

# **Backup Power**

The real-time clock/calendar is intended to be permanently powered. When the primary system power fails, the voltage on the  $V_{DD}$  pin will drop. When  $V_{DD}$  is less than  $V_{SW}$ , the RTC will switch to the backup power supply on  $V_{BAK}$ . The clock operates at extremely low current in order to maximize battery or capacitor life. However, an advantage of combining a clock function with F-RAM memory is that data is not lost regardless of the backup power source.

If a battery is applied without a  $V_{DD}$  power supply, the device has been designed to ensure the  $I_{BAK}$  current does not exceed the  $1\mu A$  maximum limit.

#### **Trickle Charger**

To facilitate capacitor backup the  $V_{\text{BAK}}$  pin can optionally provide a trickle charge current. When the

VBC bit (register 0Eh, bit 2) is set to a '1', the  $V_{BAK}$  pin will source approximately 80  $\mu A$  until  $V_{BAK}$  reaches  $V_{DD}$ . This charges the capacitor to  $V_{DD}$  without an external diode and resistor charger. There is a Fast Charge mode which is enabled by the FC bit (register 0Eh, bit 1). In this mode the trickle charger current is set to approximately 1 mA, allowing a large backup capacitor to charge more quickly.

 In the case where no battery is used, the V<sub>BAK</sub> pin should be tied to V<sub>SS</sub>.

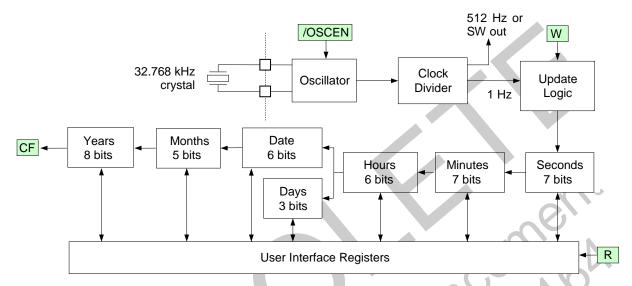


Figure 2. Real-Time Clock Core Block Diagram

#### Calibration

When the CAL bit in register 00h is set to '1', the clock enters calibration mode. In calibration mode, the ACS output pin is dedicated to the calibration function and the power fail output is temporarily unavailable. Calibration operates by applying a digital correction to the counter based on the frequency error. In this mode, the ACS pin is driven with a 512 Hz (nominal) square wave. Any measured deviation from 512 Hz translates into a timekeeping error. The user converts the measured error in ppm and writes the appropriate correction value to the calibration register. The correction factors are listed in the table below. Positive ppm errors require a negative adjustment that removes pulses. Negative ppm errors require a positive correction that adds pulses. Positive ppm adjustments have the CALS (sign) bit set to '1', whereas negative ppm adjustments have CALS = 0. After calibration, the clock will have a maximum error of  $\pm$  2.17 ppm or ± 0.09 minutes per month at the calibrated temperature.

The calibration setting is battery-backed and must be reloaded should the backup source fail. It is accessed with bits CAL.4-0 in register 01h. This value only can be written when the CAL bit is set to a '1'. To exit the calibration mode, the user must clear the CAL bit to a '0'. When the CAL bit is '0', the ACS pin will revert to another function as defined in *Table 3. Control Bit Settings for ACS Pin*.

## **Crystal Type**

The crystal oscillator is designed to use a 12.5pF crystal without the need for external components, such as loading capacitors. The FM3130 device has built-in loading capacitors that match the crystal.

If a 32.768kHz crystal is not used, an external oscillator may be connected to the FM3130. Apply the oscillator to the X1 pin. Its high and low voltage levels can be driven rail-to-rail or amplitudes as low as approximately 500mV p-p. To ensure proper operation, a DC bias must be applied to the X2 pin. It should be centered between the high and low levels on the X1 pin. This can be accomplished with a voltage divider. See Figure 3.

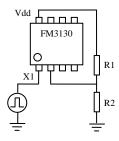
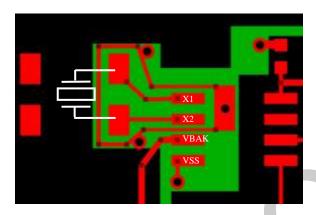


Figure 3. External Oscillator

In the example, R1 and R2 are chosen such that the X2 voltage is centered around the oscillator drive levels. If you wish to avoid the DC current, you may

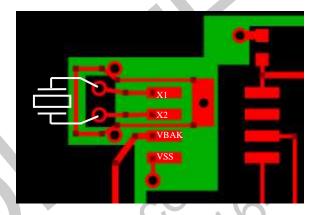


**Layout for Surface Mount Crystal** (red = top layer, green = bottom layer)

choose to drive X1 with an external clock and X2 with an inverted clock using a CMOS inverter.

## **Layout Recommendations**

The X1 and X2 crystal pins employ very high impedance circuits and the oscillator connected to these pins can be upset by noise or extra loading. To reduce RTC clock errors from signal switching noise, a guard ring should be placed around these pads and the guard ring grounded. SDA and SCL traces should be routed away from the X1/X2 pads. The X1 and X2 trace lengths should be less than 5 mm. The use of a ground plane on the backside or inner board layer is preferred. See layout example. Red is the top layer, green is the bottom layer.



**Layout for Through Hole Crystal** (red = top layer, green = bottom layer)

Table 2.	Calibratio	n Adjustments
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	Positive (	Calibration for slow	clocks: Calibrat	ion will achiev	e ± 2.17 PPM after calibration
	Measured Fre	quency Range	Error Range (F	PPM)	*
	Min	Max	Min	Max	Program Calibration Register to:
0	512.0000	511.9989	0	2.17	000000
1	511.9989	511.9967	2.18	6.51	100001
2	511.9967	511.9944	6.52	10.85	100010
3	511.9944	511.9922	10.86	15.19	100011
4	511.9922	511.9900	15.20	19.53	100100
5	511.9900	511.9878	19.54	23.87	100101
6	511.9878	511.9856	23.88	28.21	100110
7	511.9856	511.9833	28.22	32.55	100111
8	511.9833	511.9811	32.56	36.89	101000
9	511.9811	511.9789	36.90	41.23	101001
10	511.9789	511.9767	41.24	45.57	101010
11	511.9767	511.9744	45.58	49.91	101011
12	511.9744	511.9722	49.92	54.25	101100
13	511.9722	511.9700	54.26	58.59	101101
14	511.9700	511.9678	58.60	62.93	101110
15	511.9678	511.9656	62.94	67.27	101111
16	511.9656	511.9633	67.28	71.61	110000
17	511.9633	511.9611	71.62	75.95	110001
18	511.9611	511.9589	75.96	80.29	110010
19	511.9589	511.9567	80.30	84.63	110011
20	511.9567	511.9544	84.64	88.97	110100
21	511.9544	511.9522	88.98	93.31	110101
22	511.9522	511.9500	93.32	97.65	110110
23	511.9500	511.9478	97.66	101.99	110111

24	511.9478	511.9456	102.00	106.33	111000
25	511.9456	511.9433	106.34	110.67	111001
26	511.9433	511.9411	110.68	115.01	111010
27	511.9411	511.9389	115.02	119.35	111011
28	511.9389	511.9367	119.36	123.69	111100
29	511.9367	511.9344	123.70	128.03	111101
30	511.9344	511.9322	128.04	132.37	111110
31	511.9322	511.9300	132.38	136.71	111111

	Negative Calibration for fast clocks: Calibration will achieve $\pm$ 2.17 PPM after calibration										
	Measured Fre	quency Range	Error Rang	ge (PPM)							
	Min	Max	Min	Max	Program Calibration Register to:						
0	512.0000	512.0011	0	2.17	000000						
1	512.0011	512.0033	2.18	6.51	000001						
2	512.0033	512.0056	6.52	10.85	000010						
3	512.0056	512.0078	10.86	15.19	000011						
4	512.0078	512.0100	15.20	19.53	000100						
5	512.0100	512.0122	19.54	23.87	000101						
6	512.0122	512.0144	23.88	28.21	000110						
7	512.0144	512.0167	28.22	32.55	000111						
8	512.0167	512.0189	32.56	36.89	001000						
9	512.0189	512.0211	36.90	41.23	001001						
10	512.0211	512.0233	41.24	45.57	001010						
11	512.0233	512.0256	45.58	49.91	001011						
12	512.0256	512.0278	49.92	54.25	001100						
13	512.0278	512.0300	54.26	58.59	001101						
14	512.0300	512.0322	58.60	62.93	001110						
15	512.0322	512.0344	62.94	67.27	001111						
16	512.0344	512.0367	67.28	71.61	010000						
17	512.0367	512.0389	71.62	75.95	010001						
18	512.0389	512.0411	75.96	80.29	010010						
19	512.0411	512.0433	80.30	84.63	010011						
20	512.0433	512.0456	84.64	88.97	010100						
21	512.0456	512.0478	88.98	93.31	010101						
22	512.0478	512.0500	93.32	97.65	010110						
23	512.0500	512.0522	97.66	101.99	010111						
24	512.0522	512.0544	102.00	106.33	011000						
25	512.0544	512.0567	106.34	110.67	011001						
26	512.0567	512.0589	110.68	115.01	011010						
27	512.0589	512.0611	115.02	119.35	011011						
28	512.0611	512.0633	119.36	123.69	011100						
29	512.0633	512.0656	123.70	128.03	011101						
30	512.0656	512.0678	128.04	132.37	011110						
31	512.0678	512.0700	132.38	136.71	011111						

### Alarm

The alarm function compares user-programmed alarm values to the corresponding RTC time/date values. When a match occurs, an alarm event occurs. The alarm event sets an internal flag AF (register 00h, bit 6) and drives the ACS pin low, if the appropriate control bits are set in registers 00h and 0Eh. See Table 3. The alarm condition on the ACS pin and the AF bit are cleared by reading register 00h.

The alarm operates under  $V_{\rm DD}$  or  $V_{\rm BAK}$  power. If the system controller is being used to detect an alarm while the FM3130 is powered on  $V_{\rm BAK}$  only, the ACS pin may cause extra  $I_{\rm BAK}$  current when the alarm is activated. To avoid battery drain, the ACS pin can be

tri-stated by reading the AF flag, located in the RTC/Alarm Control register 00h.

There are five alarm match fields. They are Month, Date, Hours, Minutes, and Seconds. Each of these fields also has a Match bit that is used to determine if the field is used in the alarm match logic. Setting the Match bit to '0' indicates that the corresponding field will be used in the match process.

Depending on the Match bits, the alarm can occur as specifically as one particular second on one day of the month, or as frequently as once per second continuously. The MSB of each Alarm register is a Match bit. Examples of the Match bit settings are shown in *Table 4. Alarm Match Bit Examples*. Selecting none of the match bits (all '1's) indicates that no match is required. The alarm occurs every

second. Setting the match select bit for seconds to '0' causes the logic to match the seconds alarm value to the current time of day. Since a match will occur for only one value per minute, the alarm occurs once per minute. Likewise setting the seconds and minutes match select bits causes an exact match of these values. Thus, an alarm will occur once per hour. Setting seconds, minutes, and hours causes a match once per day. See Table 4 for other alarm setting examples.

#### **Function of the ACS Pin**

The ACS pin is a multifunction pin. The alarm, calibration, and square wave functions all share this output. There are two ways a user can detect an alarm event, by reading the AF flag or by monitoring the ACS pin. An interrupt pin on the host processor may be used to detect an alarm event. The AF flag in the register 00h (bit 6) will indicate that a time/date match has occurred. When a match occurs, the AF bit will be set to '1' and the ACS pin will drive low. The flag and ACS pin will remain in this state until the RTC/Alarm Control register is read which clears the AF bit.

Table 3 that shows the relationship between register control settings and the function of the ACS pin.

Table 3. Control Bit Settings for ACS Pin

State	e of Reg	Function of ACS pin	
CAL	AEN	AL/SW	
0	1	1	/Alarm
0	X	0	Sq Wave out
1	X	X	512 Hz out
0	0	1	Hi-Z

# Cal Output/SquareWave Output

When the RTC calibration mode is invoked by setting the CAL bit (register 00h, bit 2), the ACS output pin will be driven with a 512 Hz square wave and the alarm will continue to operate. Since most users only invoke the calibration mode during production, this should have no impact on the otherwise normal operation of the alarm.

The ACS output may also be used to drive the system with a continuous frequency. The AL/SW bit (register 0Eh, bit 7) must be a '0'. A user-selectable frequency is provided by F0 and F1 (register 0Eh, bits 5 and 6). The frequencies are 1, 512, 4096, and 32768 Hz. If a continuous frequency output is enabled by using the 512Hz or SquareWave out functions, the alarm function will not be available.

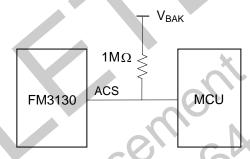


Figure 4. ACS Pin Requires Pullup

The ACS pin is an open-drain output that needs to be pulled up to a supply. The ACS pin and pullup resistor draws current only when the alarm is triggered.

**Table 4. Alarm Match Bit Examples** 

Seconds	Minutes	Hours	Date	Months	Alarm condition		
1	1	1	1	1	No match required = alarm 1/second		
0	1	1	1	1	Alarm when seconds match = alarm 1/minute		
0	0	1	1	1	Alarm when seconds, minutes match = alarm 1/hour		
0	0	0	1	1	Alarm when seconds, minutes, hours match = alarm 1/date		
0	0	0	0	1	Alarm when seconds, minutes, hours, date match = alarm 1/month		

# **Register Map**

The RTC, alarm, and other functions are accessed via 15 special function registers mapped to a separate 2-wire device ID. The interface protocol is described below. The registers contain timekeeping data, control bits, or information flags. A description of each register follows the summary table below.

# **Register Map Summary Table**

Address	D7	D6	D5	D4	D3	D2	D1	D0	Function	Range
0Eh	AL/SW	F1	F0	WP1	WP0	VBC	FC	TST	Alarm & WP Control	
0Dh	/Match	0	0	10 mo		Alarm	months		Alarm Month	01-12
0Ch	/Match	0	10	date		Alarm	n date		Alarm Date	01-31
0Bh	/Match	0	Alarm 1	0 hours		Alarm	hours		Alarm Hours	00-23
0Ah	/Match	Ala	rm 10 minu	ites		Alarm r	minutes		Alarm Minutes	00-59
09h	/Match	Ala	rm 10 seco	nds		Alarm	seconds		Alarm Seconds	00-59
08h		10 y	ears			ye	ars		RTC Years	00-99
07h	0	0	0	10 mo		moi	nths		RTC Month	1-12
06h	0	0	10	date		da	ate		RTC Date	1-31
05h	0	0	0	0	0		day		RTC Day	1-7
04h	0	0	10 h	ours		ho	urs		RTC Hours	0-23
03h	0		10 minutes			min	utes		RTC Minutes	0-59
02h	0		10 seconds		seconds			RTC Seconds	0-59	
01h	/OSCEN	•	CALS	CAL4	CAL3	CAL3 CAL2 CAL1 CAL0			CAL/Control	
00h	LB	AF	CF	POR	AEN	CAL	W	R	RTC/Alarm Control	

Note: When the device is first powered up, all registers should be treated as unknown and must be written. Otherwise, unpredictable behavior may result.

# **Register Description**

# **Address Description**

0Eh	Alarm & WP Control										
	D7	D6	D5	D4	D3	D2	D1	D0			
	AL/SW	F1	F0	WP1	WP0	VBC	FC	TST			
AL/SW	Alarm/Square	e Wave Select:						hen set to 0,			
		Square Wave Fr	eq will be driv	ven on the ACS	S pin, and an a	larm match onl	y sets the AF f	lag. Battery-			
F(1:0)	backed, read/	Freq Select: T	hasa bita salas	t the frequency	y on the ACS n	in when the C	AI and AI/SW	I hita ara			
F(1.0)	both 0. Batter		nese ons selec	t the frequency	y on the ACS p	iii when the Ca	AL and AL/S w	ons are			
	0041 0. 2410										
	Setting $F(1:0)$ Setting $F(1:0)$										
			0 (default)			0					
	-	512 Hz 0	l	327	68 Hz 1	1					
WP1,WP0	Write Protect	. These bits cor	ntrol the write	protection of t	he memory arr	ay. Battery-bac	cked, read/writ	e.			
		ite-Protect addı									
	No: Bot	ne ttom 1/4		$egin{pmatrix} 0 & 0 & 0 \\ 0 & 1 & 1 \end{pmatrix}$							
		ttom 1/2		1 0				A.,			
	Ful	l array		1 1							
MDC	LIDAU CI	G . 1 G	" UDC	1 ( 1 FC 0)		00 4 (1 4 )	TEG 1) . : 11				
VBC		ger Control: Se supplied on V <sub>B</sub>									
FC		Setting FC to									
		C to 0 disables						, DAK			
TST	Invokes facto	ory test mode. U	Jsers should al	ways set this b	oit to 0.						
ADI.	A1 N/I	41-				$ \sim$ $\cup$		V)			
0Dh	Alarm – M D7	D6	D5	D4	D3	D2	D1	D0			
	_										
	Contains the	0 alarm value for	the month on	10 Month	Month.3	Month.2	Month.1	Month.0			
/M		ig this bit to a '						is bit to a			
/11/1		e match circuit					ogiei setting til	15 010 00 0			
0Ch	Alarm – Da	ate				0.					
	D7	D6	D5	D4	D3	D2	D1	D0			
	M	0	10 date.1	10 date.0	Date.3	Date.2	Date.1	Date.0			
		alarm value for									
/M		g this bit to a '					ic. Setting this	bit to a '1'			
0Bh	Alarm – Ho	atch circuit to ig	gnore the Date	value. Battery	/-backed, read/	write.					
UDII	D7	D6	D5	D4	D3	D2	D1	<b>D</b> 0			
	_										
	Contains the	0 alarm value for	10 hours.1	10 hours.0	Hours.3	Hours2	Hours.1	Hours.0			
/M								s bit to a '1'			
		Match: Setting this bit to a '0' causes the Hours value to be used in the alarm match logic. Setting this bit to a '1' causes the match circuit to ignore the Hours value. Battery-backed, read/write.									
0Ah	Alarm – M	inutes				<del>_</del>	<del>_</del>				
	D7	D6	D5	D4	D3	D2	D1	D0			
	<u></u>	10 min.2	10 min.1	10 min.0	Min.3	Min.2	Min.1	Min.0			
		alarm value for									
/M		g this bit to a '					logic. Setting t	this bit to a			
i	1 '1' causes the	e match circuit	to 1gnore the N	/Inutes value.	Battery-backed	i, read/write.					

09h	Alarm – Seconds												
	D7	D7 D6 D5 D4 D3 D2 D1 D0											
	M	10 sec.2	10 sec.1	10 sec.0	Seconds.3	Seconds.2	Seconds.1	Seconds.0					
	Contains the	alarm value fo	r the seconds a	and the mask b	it to select or d	leselect the Sec	conds value.						
/M	Match: Setting this bit to a '0' causes the Seconds value to be used in the alarm match logic. Setting this bit to a												
	'1' causes the	e match circuit	to ignore the S	Seconds value.	Battery-backe	d, read/write.							

08h	Timekeep	Timekeeping – Years													
	D7	D6	D5	D4	D3	D2	D1	D0							
	10 year.3	10 year.2	10 year.1	10 year.0	Year.3	Year.2	Year.1	Year.0							
				ne year. Lower i											
				perates from 0											
	read/write.														
07h	Timekeep	oing – Mont													
	<b>D7</b>	D6	D5	D4	D3	D2	D1	D0							
	0	0	0	10 Month	Month.3	Month.2	Month.1	Month.0							
	Contains th	ne BCD digits	for the month.	Lower nibble co	ontains the lowe	er digit and ope	rates from 0 to	9; upper							
	nibble (one bit) contains the upper digit and operates from 0 to 1. The range for the register is 1-12. Battery-														
	backed, rea														
06h			of the month	1				1							
	D7	D6	D5	D4	D3	D2	D1	D0							
	0	0	10 date.1	10 date.0	Date.3	Date.2	Date.1	Date.0							
		Contains the BCD digits for the date of the month. Lower nibble contains the lower digit and operates from 0 to 9 upper nibble contains the upper digit and operates from 0 to 3. The range for the register is 1-31. Battery-backed,													
	read/write.														
05h		oing – Day o				, 									
	D7	D6	D5	D4	D3	D2	D1	D0							
	0	0	0	0	0	Day.2	Day.1	Day.0							
	Lower nibble contains a value that correlates to day of the week. Day of the week is a ring counter that counts														
				nust assign mea	ning to the day	value, as the da	y is not integra	ited with the							
0.41		ry-backed, rea				V ·									
04h		oing – Hours		D4	72	D2	D1	D.O.							
	D7	D6	D5	D4	D3	D2	D1	D0							
	0	0	10 hours.1	10 hours.0	Hours.3	Hours2	Hours.1	Hours.0							
		Contains the BCD value of hours in 24-hour format. Lower nibble contains the lower digit and operates from 0 to													
	9; upper nibble (two bits) contains the upper digit and operates from 0 to 2. The range for the register is 0-23. Battery-backed, read/write.														
03h		oing – Minu		. 40		4									
USII	D7	Ding – Miniu D6	D5	D4	D3	D2	D1	<b>D</b> 0							
	D/	Du	DS	D4	D3		DI	DU							
	0	10 min.2	10 min.1	10 min.0	Min.3	Min.2	Min.1	Min.0							
				wer nibble cont											
	read/write.	e upper minut	es digit and ope	erates from 0 to	5. The range to	or the register is	0-59. Battery-	васкеа,							
02h		oing – Secon	de	-											
V211	D7	Ding – Secon	D5	D4	D3	D2	D1	D0							
	0	10 sec.2	10 sec.1	10 sec.0	Seconds.3	Seconds.2	Seconds.1	Seconds.0							
				wer nibble cont											
				om $0$ to $5$ . The ra											

01h	CAL/Control								
	D7	D7 D6 D5 D4 D3 D2 D1 D0							
	OSCEN	ı	CALS	CAL.4	CAL.3	CAL.2	CAL.1	CAL.0	
/OSCEN	Oscillator Enable. When set to 1, the oscillator is halted. When set to 0, the oscillator runs. Disabling the								
	oscillator can save battery power during storage. On an initial power-up of V <sub>DD</sub> with or without V <sub>BAK</sub> , this bit								
	is internally s	et to 1, which t	urns off the osc	cillator. Battery	-backed, read	/write.			

# RAMTRON

CALS	Calibration Sign: Determines if the calibration adjustment is applied as an addition to or as a subtraction from
	the time-base. This bit can be written only when CAL=1. Battery-backed, read/write.
CAL.4-0	Calibration Code: These five bits control the calibration of the clock. These bits can be written only when
	CAL=1. Battery-backed, read/write.

00h	RTC/Alarm Control							
	D7	D6	D5	D4	D3	D2	D1	D0
	LB	AF	CF	POR	AEN	CAL	W	R
LB		Flag: If the V <sub>B</sub>						
		'1'. All registe						
	treated as unknown. The user should clear it to '0' when initializing the system. Battery-backed. Rea							ead/Write
	(internally set, user can clear bit by writing to a '0').							
AF		Alarm Flag: This read-only bit is set to 1 when the time/date match the values stored in the alarm registers with the Match bit(s) = 0. It is cleared when the RTC/Alarm Control register is read. Battery-backed.						
CF		rflow Flag: This						
		ndicates a new c						
	Battery-back	ury information	as needed. In	is bit is cleared	when the KTC	Alarm Cont	roi register is	read.
POR		eset Flag: When	V drong ho	low V the D	OD bit will be	got to '1' D	ottory booked	
IOK		internally set, u				set to 1. B	attery-backed.	
AEN		e: This bit enab				nd CAL cleare	ed) the ACS r	nin operates
7 LLI		low alarm and the						
		s cleared, no ne						
		ed, read/write.				1		<b>X</b>
CAL	Calibration M	Mode: When CA	L is set to '1',	the clock enter	s calibration r	node. When C	CAL is set to	0', the cloc
		nally, and the A						
W		Setting the W b						
		dated values. S			the contents of	f the time regi	sters to be train	nsferred to
		ing counters. Ba						
R		Setting the R bi						
		e user can then a 0' to '1' causes						
		ed, read/write.	ше ишекеери	ig capture, so ii	ie dit must be	returned to 0	prior to read	ing agam.
	Buttery buck	su, redu/ write.				10	<b>K</b> 5	<u>P</u>
					- 0.4			
					ישכ			
				*		*		
					. 6	> +		
4				0	1.5			
			+, •(		7/1			
			( ) '		<b>O</b> -			
				3/1/				
			\ \ \ \ \	(O)				

## **Two-wire Interface**

The FM3130 employs an industry standard two-wire bus that is familiar to many users. This product is unique since it incorporates two logical devices in one chip. Each logical device can be accessed individually. Although monolithic, it appears to the system software to be two separate products. One is a memory device. It has a Slave Address (Slave ID = 1010b) that operates the same as a stand-alone memory device. The second device is a real-time clock and alarm which have a unique Slave Address (Slave ID = 1101b).

By convention, any device that is sending data onto the bus is the transmitter while the target device for this data is the receiver. The device that is controlling the bus is the master. The master is responsible for generating the clock signal for all operations. Any device on the bus that is being controlled is a slave. The FM3130 is always a slave device.

The bus protocol is controlled by transition states in the SDA and SCL signals. There are four conditions: Start, Stop, Data bit, and Acknowledge. The figure below illustrates the signal conditions that specify the four states. Detailed timing diagrams are shown in the Electrical Specifications section.

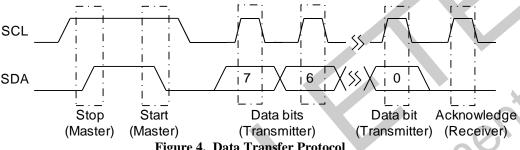


Figure 4. Data Transfer Protocol

#### **Start Condition**

A Start condition is indicated when the bus master drives SDA from high to low while the SCL signal is high. All read and write transactions begin with a Start condition. An operation in progress can be aborted by asserting a Start condition at any time. Aborting an operation using the Start condition will ready the FM3130 for a new operation.

## **Stop Condition**

A Stop condition is indicated when the bus master drives SDA from low to high while the SCL signal is high. All operations must end with a Stop condition. If an operation is pending when a stop is asserted, the operation will be aborted. The master must have control of SDA (not a memory read) in order to assert a Stop condition.

## Data/Address Transfer

All data transfers (including addresses) take place while the SCL signal is high. Except under the two conditions described above, the SDA signal should not change while SCL is high.

## Acknowledge

The Acknowledge (ACK) takes place after the 8<sup>th</sup> data bit has been transferred in any transaction. During this state the transmitter must release the SDA bus to allow the receiver to drive it. The receiver drives the SDA signal low to acknowledge receipt of the byte. If the receiver does not drive SDA low, the condition is a No-Acknowledge (NACK) and the operation is aborted.

The receiver might NACK for two distinct reasons. First is that a byte transfer fails. In this case, the NACK ends the current operation so that the part can be addressed again. This allows the last byte to be recovered in the event of a communication error.

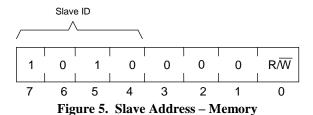
Second and most common, the receiver does not send an ACK to deliberately terminate an operation. For example, during a read operation, the FM3130 will continue to place data onto the bus as long as the receiver sends ACKs (and clocks). When a read operation is complete and no more data is needed, the receiver must NACK the last byte. If the receiver ACKs the last byte, this will cause the FM3130 to attempt to drive the bus on the next clock while the master is sending a new command such as a Stop.

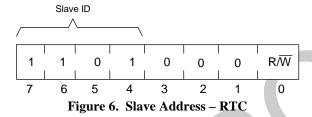
#### **Slave Address**

The first byte that the FM3130 expects after a Start condition is the slave address. As shown in figures below, the slave address contains the Slave ID and a bit that specifies if the transaction is a read or a write.

The FM3130 has two Slave Addresses (Slave IDs) associated with two logical devices. To access the memory device, bits 7-4 should be set to 1010b. The other logical device within the FM3130 is the real-time clock and alarm. To access this device, bits 7-4 of the slave address should be set to 1101b. A bus transaction with this slave address will not affect the memory in any way. The figures below illustrate the two Slave Addresses.

Bits 3 through 1 of the Slave Address must be logic 0. Bit 0 is the read/write bit. A '1' indicates a read operation, and a '0' indicates a write operation.





# Addressing Overview - Memory

After the FM3130 acknowledges the Slave Address, the master can place the memory address on the bus for a write operation. The address requires two bytes. The first is the MSB (upper byte). The first 3 unused address bits are don't cares, but should be set to '0' to maintain upward compatibility. Following the MSB is the LSB (lower byte) which contains the remaining eight address bits. The address is latched internally. Each access causes the latched address to be incremented automatically. The current address is the value that is held in the latch, either a newly written value or the address following the last access. The current address will be held as long as  $V_{DD}$  is greater than V<sub>SW</sub> or until a new value is written. Accesses to the clock do not affect the current memory address. Reads always use the current

address. A random read address can be loaded by beginning a write operation as explained below.

After transmission of each data byte, just prior to the Acknowledge, the FM3130 increments the internal address. This allows the next sequential byte to be accessed with no additional addressing externally. After the last address is reached, the address latch will roll over to 0000h. There is no limit to the number of bytes that can be accessed with a single read or write operation.

## Addressing Overview - RTC/Alarm

The RTC/Alarm operates in a similar manner to the memory, except that it uses only one byte of address. Addresses 00h to 0Eh correspond to the RTC/Alarm and control registers. Attempting to load addresses above 0Eh is an illegal condition; the FM3130 will return a NACK and abort the 2-wire transaction.

#### **Data Transfer**

After the address information has been transmitted, data transfer between the bus master and the FM3130 begins. For a read, the FM3130 will place 8 data bits on the bus then wait for an ACK from the master. If the ACK occurs, the FM3130 will transfer the next byte. If the ACK is not sent, the FM3130 will end the read operation. For a write operation, the FM3130 will accept 8 data bits from the master then send an Acknowledge. All data transfer occurs MSB (most significant bit) first.

## **Memory Write Operation**

All memory writes begin with a Slave Address, then a memory address. The bus master indicates a write operation by setting the slave address LSB to a '0'. After addressing, the bus master sends each byte of data to the memory and the memory generates an Acknowledge condition. Any number of sequential bytes may be written. If the end of the address range is reached internally, the address counter will wrap to 0000h. Internally, the actual memory write occurs after the 8<sup>th</sup> data bit is transferred. It will be complete before the Acknowledge is sent. Therefore, if the user desires to abort a write without altering the memory contents, this should be done using a Start or Stop condition prior to the 8<sup>th</sup> data bit. The figures below illustrate a single- and multiple-writes to memory.

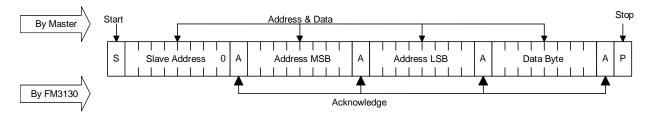


Figure 7. Single Byte Memory Write

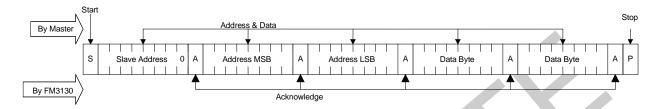


Figure 8. Multiple Byte Memory Write

## **Memory Read Operation**

There are two types of memory read operations. They are current address read and selective address read. In a current address read, the FM3130 uses the internal address latch to supply the address. In a selective read, the user performs a procedure to first set the address to a specific value.

## Current Address & Sequential Read

As mentioned above the FM3130 uses an internal latch to supply the address for a read operation. A current address read uses the existing value in the address latch as a starting place for the read operation. The system reads from the address immediately following that of the last operation.

To perform a current address read, the bus master supplies a slave address with the LSB set to 1. This indicates that a read operation is requested. After receiving the complete device address, the FM3130 will begin shifting data out from the current address on the next clock. The current address is the value held in the internal address latch.

Beginning with the current address, the bus master can read any number of bytes. Thus, a sequential read is simply a current address read with multiple byte transfers. After each byte the internal address counter will be incremented.

Each time the bus master acknowledges a byte, this indicates that the FM3130 should read out the next sequential byte.

There are four ways to terminate a read operation. Failing to properly terminate the read will most likely create a bus contention as the FM3130 attempts to read out additional data onto the bus. The four valid methods follow.

- 1. The bus master issues a NACK in the 9<sup>th</sup> clock cycle and a Stop in the 10<sup>th</sup> clock cycle. This is illustrated in the diagrams below and is preferred.
- 2. The bus master issues a NACK in the 9<sup>th</sup> clock cycle and a Start in the 10<sup>th</sup>.
- 3. The bus master issues a Stop in the 9<sup>th</sup> clock cycle.
- 4. The bus master issues a Start in the 9<sup>th</sup> clock cycle.

If the internal address reaches the top of memory, it will wrap around to 0000h on the next read cycle. The figures below show the proper operation for current address reads.

## Selective (Random) Read

There is a simple technique that allows a user to select a random address location as the starting point for a read operation. This involves using the first three bytes of a write operation to set the internal address followed by subsequent read operations.

To perform a selective read, the bus master sends out the slave address with the LSB set to 0. This specifies a write operation. According to the write protocol, the bus master then sends the address bytes that are loaded into the internal address latch. After the FM3130 acknowledges the address, the bus master

issues a Start condition. This simultaneously aborts the write operation and allows the read command to be issued with the slave address LSB set to a '1'. The operation is now a read from the current address. Read operations are illustrated below.

## **RTC/Alarm Write Operation**

All RTC/Alarm writes operate in a similar manner to memory writes. The distinction is that a different device ID is used and only one byte address is needed instead of two. Figure 12 illustrates a single byte write to the RTC/Alarm.

## **RTC/Alarm Read Operation**

As with writes, a read operation begins with the Slave Address. To perform a register read, the bus master supplies a Slave Address with the LSB set to a '1'. This indicates that a read operation is requested. After receiving the complete Slave Address, the

FM3130 will begin shifting data out from the current register address on the next clock. Auto-increment operates for the special function registers as with the memory address. A current address read for the registers look exactly like the memory except that the device ID is different.

The FM3130 contains two separate address registers, one for the memory address and the other for the register address. This allows the contents of one address register to be modified without affecting the current address of the other register. For example, this would allow an interrupted read to the memory while still providing fast access to an RTC register. A subsequent memory read will then continue from the memory address where it previously left off, without requiring the load of a new memory address. However, a write sequence always requires an address to be supplied.

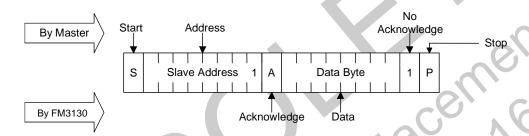


Figure 9. Current Address Memory Read

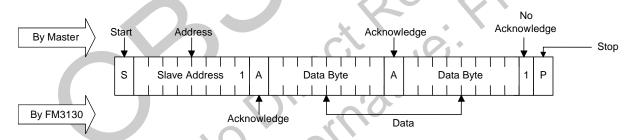


Figure 10. Sequential Memory Read

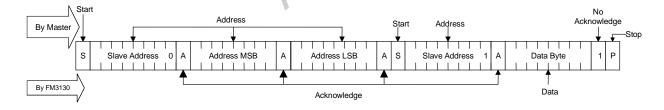


Figure 11. Selective (Random) Memory Read

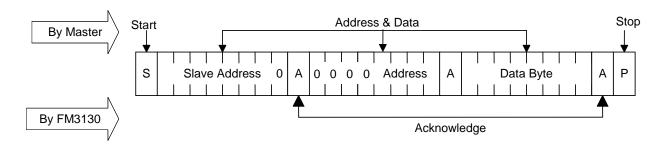


Figure 12. Register Byte Write

\* Although not required, it is recommended that A7-A4 in the Register Address byte are zeros in order to preserve compatibility with future devices.



# **Electrical Specifications**

## **Absolute Maximum Ratings**

Symbol	Description	Ratings
$V_{ m DD}$	Power Supply Voltage with respect to V <sub>SS</sub>	-1.0V to +5.0V
$V_{IN}$	Voltage on any signal pin with respect to V <sub>SS</sub>	-1.0V to +5.0V * and
		$V_{IN} \le V_{DD} + 1.0V **$
$V_{BAK}$	Backup Supply Voltage	-1.0V to +4.5V
$T_{STG}$	Storage Temperature	-55°C to + 125°C
$T_{LEAD}$	Lead Temperature (Soldering, 10 seconds)	300° C
$V_{ESD}$	Electrostatic Discharge Voltage	
	- Human Body Model (JEDEC Std JESD22-A114-B)	4kV
	- Charged Device Model (JEDEC Std JESD22-C101-A)	1kV
	- Machine Model (JEDEC Std JESD22-A115-A)	200V
	Package Moisture Sensitivity Level	MSL-1

<sup>\*\*</sup> The " $V_{\rm IN} < V_{\rm DD} + 1.0V$ " restriction does not apply to the SCL, SDA, and ACS pins which do not employ a diode to  $V_{\rm DD}$ . Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only, and the functional operation of the device at these or any other conditions above those listed in the operational section of this specification is not implied. Exposure to absolute maximum ratings conditions for extended periods may affect device reliability.

# **DC Operating Conditions** ( $T_A = -40^{\circ} \text{ C to} + 85^{\circ} \text{ C}$ , $V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$ unless otherwise specified)

Symbol	Parameter	Min	Тур	Max	Units	Notes
$V_{DD}$	Main Power Supply	2.7	-	3.6	V	1
$I_{DD}$	V <sub>DD</sub> Supply Current @ SCL = 100 kHz @ SCL = 1 MHz			150 500	μΑ μΑ	2
$I_{SB}$	Standby Current Trickle Charger Off (VBC=0) Trickle Chrg On, Fast Chrg Off (VBC=1, FC=0) Trickle Chrg On, Fast Chrg On (VBC=1, FC=1)			50 190 2600	μΑ μΑ μΑ	3
$V_{BAK}$	RTC Backup Supply Voltage	2.0	3.0	3.6	V	4
$I_{BAK}$	RTC Backup Supply Current			1	μΑ	5
$I_{BAKTC}$	Trickle Charge Current with V <sub>BAK</sub> =0V	08				6
	Fast Charge Off (FC = $0$ )	25		120	μΑ	
	Fast Charge On (FC = 1)	200	•	2500	μΑ	
$I_{LI}$	Input Leakage Current		0.	±1	μΑ	7
$I_{LO}$	Output Leakage Current	J • •		±1	μΑ	7
V <sub>IH</sub>	Input High Voltage	$0.7 V_{DD}$	7	$V_{DD} + 0.3$	V	
$V_{\rm IL}$	Input Low Voltage	-0.3		$0.3~\mathrm{V_{DD}}$	V	
$V_{OL1}$	Output Low Voltage ( $I_{OL} = 3 \text{ mA}$ )	70		0.4	V	
`	- Applies to SDA and ACS pin					
	$-V_{DD} > V_{SW}$					
$V_{OL2}$	Output Low Voltage ( $I_{OL} = 80 \mu A$ )					
	- Applies only to ACS pin	-		0.4	V	
	- $V_{BAK}$ applied, $V_{DD} < V_{SW}$					
$V_{SW}$	Battery Switchover Voltage	2.0		2.7	V	

#### Notes

- 1. Full complete operation. RTC operates to lower voltages as specified.
- 2. SCL toggling between  $V_{DD}$ -0.3V and  $V_{SS}$ , other inputs  $V_{SS}$  or  $V_{DD}$ -0.3V. VBC=0.  $I_{DD}$  is linear vs frequency.
- 3. All inputs at  $V_{SS}$  or  $V_{DD}$  static. Stop command issued.
- 4. The VBAK trickle charger automatically regulates the maximum voltage on this pin for capacitor backup applications.
- 5.  $V_{BAK} = 3.0V$ ,  $V_{DD} < V_{SW}$ , oscillator running.
- 6.  $V_{BAK}$  will source current when the trickle charger is enabled (VBC=1),  $V_{DD} > V_{BAK}$  and  $V_{DD} > V_{SW}$ .
- 7.  $V_{IN}$  or  $V_{OUT} = V_{SS}$  to  $V_{DD}$ .

AC Parameters ( $T_A = -40^{\circ} \text{ C to} + 85^{\circ} \text{ C}$ ,  $V_{DD} = 2.7 \text{ V to} 3.6 \text{ V}$ ,  $C_L = 100 \text{ pF}$  unless otherwise specified)

Symbol	Parameter	Min	Max	Min	Max	Min	Max	Units	Notes
$f_{SCL}$	SCL Clock Frequency		100	0	400	0	1000	kHz	
$t_{LOW}$	Clock Low Period	4.7		1.3		0.6		μs	
t <sub>HIGH</sub>	Clock High Period	4.0		0.6		0.4		μs	
$t_{AA}$	SCL Low to SDA Data Out Valid		3		0.9		0.55	μs	
t <sub>BUF</sub>	Bus Free Before New Transmission	4.7		1.3		0.5		μs	
t <sub>HD:STA</sub>	Start Condition Hold Time	4.0		0.6		0.25		μs	
t <sub>SU:STA</sub>	Start Condition Setup for Repeated	4.7		0.6		0.25		μs	
	Start							-	
$t_{HD:DAT}$	Data In Hold Time	0		0		0		ns	
$t_{SU:DAT}$	Data In Setup Time	250		100		100		ns	
$t_R$	Input Rise Time		1000		300		300	ns	1
$t_{\rm F}$	Input Fall Time		300		300		100	ns	1
$t_{SU:STO}$	Stop Condition Setup Time	4.0		0.6		0.25		μs	
$t_{\mathrm{DH}}$	Data Output Hold (from SCL @ VIL)	0		0		0		ns	
$t_{SP}$	Noise Suppression Time Constant		50		50		50	ns	
	on SCL, SDA								

All SCL specifications as well as start and stop conditions apply to both read and write operations.

**Supervisor Timing** ( $T_A = -40^{\circ} \text{ C to} + 85^{\circ} \text{ C}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$ )

Symbol	Parameter	Min	Max	Units	Notes
t <sub>VR</sub>	V <sub>DD</sub> Rise Time	50	-	μs/V	1,2
$t_{VF}$	V <sub>DD</sub> Fall Time	100		μs/V	1,2
$t_{RPU}$	Device active after V <sub>DD</sub> >2.7V	-	20	ms	

Capacitance  $(T_A = 25^{\circ} C, f=1.0 MHz, V_{DD} = 3.0V)$ 

Symbol	Parameter	Тур	Max	Units	Notes
$C_{IO}$	Input/Output Capacitance	-	8	pF	1
$C_{XTL}$	X1, X2 Crystal pin Capacitance	25	A	pF	1, 3

# Notes

- 1 This parameter is characterized but not tested.
- 2 Slope measured at any point on  $V_{DD}$  waveform.
- 3 The crystal attached to the X1/X2 pins must be rated as 12.5pF.

# **Data Retention** ( $V_{DD} = 2.7V \text{ to } 3.6V$ )

Symbol	Parameter	Min	Units	Notes
$T_{DR}$	Data Retention			
	@ +75°C	45	Years	
	@ +80°C	20	Years	
	@ +85°C	10	Years	

## **AC Test Conditions**

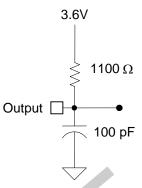
Input Pulse Levels  $0.1 V_{DD}$  to  $0.9 V_{DD}$ 

Input rise and fall times 10 ns Input and output timing levels  $0.5 V_{DD}$ 

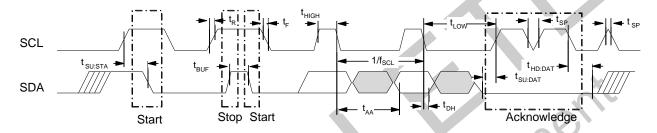
## **Diagram Notes**

All start and stop timing parameters apply to both read and write cycles. Clock specifications are identical for read and write cycles. Write timing parameters apply to slave address, word address, and write data bits. Functional relationships are illustrated in the relevant data sheet sections. These diagrams illustrate the timing parameters only.

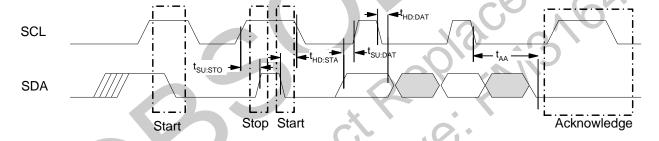
## **Equivalent AC Test Load Circuit**



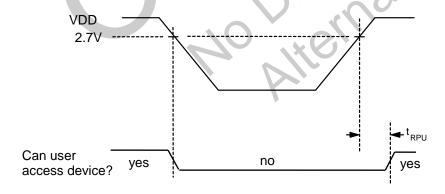
## **Read Bus Timing**



## Write Bus Timing

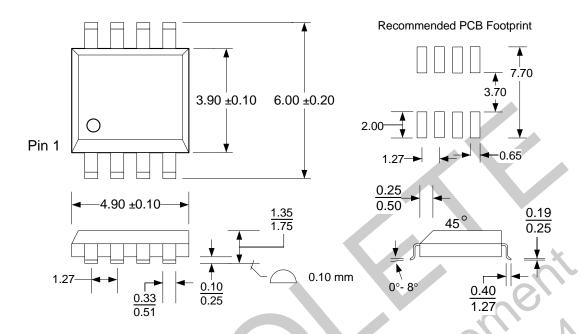


# **Power Cycle Timing**

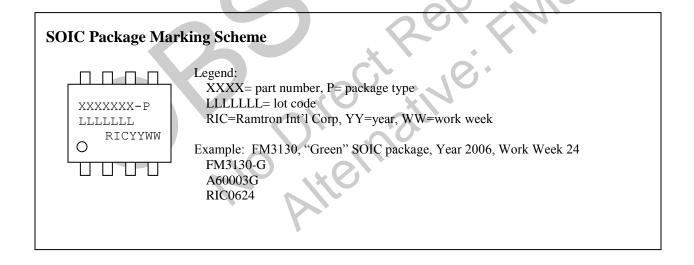


# **Mechanical Drawing**

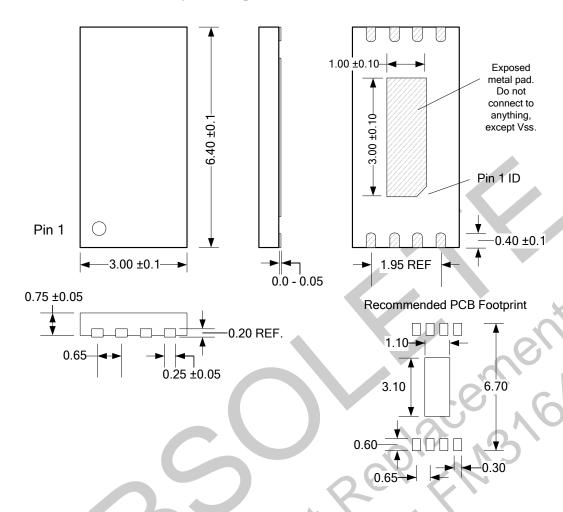
# 8-pin SOIC (JEDEC Standard MS-012 variation AA)



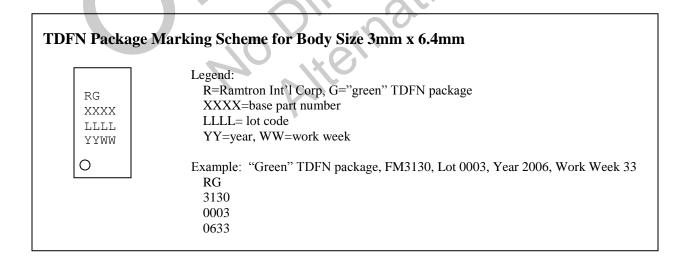
Refer to JEDEC MS-012 for complete dimensions and notes. All dimensions in <u>millimeters</u>.



8-pin TDFN (3.0 mm x 6.4 mm body, 0.65mm pitch)



Note: All dimensions in <u>millimeters</u>. This package is footprint compatible with the 8-pin TSSOP, however care must be taken to ensure PCB traces and vias are not placed within the exposed metal pad area.



# **Revision History**

Revision	Date	Summary
		υ ·
0.0	12/14/05	Initial release.
0.1	2/28/06	All register space is battery-backed. Moved location of some bits in regs 00
		and 01h. Removed serial number. Added LB and POR flags, and fast charge
		mode to trickle charger. Industrial temp grade.
0.2	5/10/06	Updated AEN, AF, and ACS pin descriptions and Backup Power section.
		Changed trickle charger limits and added V <sub>SW</sub> parameter. Added TDFN
		package.
1.0	9/18/06	Changed to Preliminary status. Changed I <sub>BAKTC</sub> (FC=0) from 50 to 25µA.
1.1	6/26/07	Added ESD and package MSL ratings.
3.0	2/29/2008	Changed to Production status. Updated ESD ratings.
3.1	2/9/2009	Added tape and reel ordering information. Expanded data retention ratings.
		Added UL Recognition of trickle charger. Added exposed pad dimensions
		and pcb footprint to TDFN drawing.
3.2	9/7/2011	End of Life. No direct replacement. Alternative device: FM3164.

