## Ordering Information

| Part Number | | Voltage | Junction Temp. Range | Package |
|---|---|---|---|---|
| **Standard** | **RoHS Compliant*** | | | |
| MIC2937A-3.3BU | MIC2937A-3.3WU | 3.3V | –40ºC to +125ºC | TO-263-3 |
| MIC2937A-3.3BT | MIC2937A-3.3WT | 3.3V | –40ºC to +125ºC | TO-220-3 |
| MIC2937A-5.0BU | MIC2937A-5.0WU | 5.0V | –40ºC to +125ºC | TO-263-3 |
| MIC2937A-5.0BT | MIC2937A-5.0WT | 5.0V | –40ºC to +125ºC | TO-220-3 |
| MIC2937A-12BU | MIC2937A-12WU | 12V | –40ºC to +125ºC | TO-263-3 |
| MIC2937A-12BT | MIC2937A-12WT | 12V | –40ºC to +125ºC | TO-220-3 |
| MIC29371-3.3BU | MIC29371-3.3WU | 3.3V | –40ºC to +125ºC | TO-263-5 |
| MIC29371-3.3BT | MIC29371-3.3WT | 3.3V | –40ºC to +125ºC | TO-220-5 |
| MIC29371-5.0BU | MIC29371-5.0WU | 5.0V | –40ºC to +125ºC | TO-263-5 |
| MIC29371-5.0BT | MIC29371-5.0WT | 5.0V | –40ºC to +125ºC | TO-220-5 |
| MIC29371-12BU | MIC29371-12WU | 12V | –40ºC to +125ºC | TO-263-5 |
| MIC29371-12BT | MIC29371-12WT | 12V | –40ºC to +125ºC | TO-220-5 |
| MIC29372BU | MIC29372WU | ADJ | –40ºC to +125ºC | TO-263-5 |
| MIC29372BT | MIC29372WT | ADJ | –40ºC to +125ºC | TO-220-5 |

* RoHS compliant with 'high-melting solder' exemption.

## Absolute Maximum Ratings

Power Dissipation (**Note 1**)..............................................Internally Limited
Lead Temperature (Soldering, 5 seconds)........................................260°C
Storage Temperature Range............................................–65°C to +150°C
Operating Junction Temperature Range ..........................–40°C to +125°C
TO-220 $\theta_{JC}$ .....................................................................................2.5°C/W
TO-263 $\theta_{JC}$ .....................................................................................2.5°C/W
Input Supply Voltage.............................................................–20V to +60V

Operating Input Supply Voltage.................................................2V† to 26V
Adjust Input Voltage (**Notes 9 and 10**)................................–1.5V to +26V
Shutdown Input Voltage ......................................................–0.3V to +30V
Error Comparator Output Voltage........................................–0.3V to +30V

† Across the full operating temperature, the minimum input voltage range for full output current is 4.3V to 26V. Output will remain in-regulation at lower output voltages and low current loads down to an input of 2V at 25°C.

## Electrical Characteristics

Limits in standard typeface are for $T_J$ = 25°C and limits in **boldface** apply over the full operating temperature range. Unless otherwise specified, $V_{IN} = V_{OUT}$ + 1V, $I_L$ = 5mA, $C_L$ = 10μF. The MIC29372 are programmed for a 5V output voltage, and $V_{SHUTDOWN}$ ≤ 0.6V (MIC29371-xx and MIC29372 only).

| Symbol | Parameter | Conditions | Min | Typical | Max | Units |
|---|---|---|---|---|---|---|
| $V_O$ | Output Voltage Accuracy | Variation from factory trimmed $V_{OUT}$ | –1 | | 1 | % |
| | | | **–2** | | **2** | |
| | | 5mA ≤ $I_L$ ≤ 500mA | **–2.5** | | **2.5** | |
| | | MIC2937A-12 and 29371-12 only: | –1.5 | | 1.5 | |
| | | | **–3** | | **3** | |
| | | 5mA ≤ $I_L$ ≤ 500mA | **–4** | | **4** | |
| $\frac{\Delta V_O}{\Delta T}$ | Output Voltage Temperature Coef. | (**Note 2**) | | 20 | **100** | ppm/°C |
| | | Output voltage > 10V | | 80 | **350** | |
| $\frac{\Delta V_O}{V_O}$ | Line Regulation | $V_{IN} = V_{OUT}$ + 1V to 26V | | 0.03 | 0.10 | % |
| | | | | | **0.40** | |
| $\frac{\Delta V_O}{V_O}$ | Load Regulation | $I_L$ = 5 to 500mA | | 0.04 | 0.16 | % |
| | | (**Note 3**) | | | **0.30** | |
| $V_{IN} - V_O$ | Dropout Voltage (**Note 4**) | $I_L$ = 5mA | | 80 | 150 | mV |
| | | | | | **180** | |
| | | $I_L$ = 100mA | | 200 | | |
| | | Output voltage > 10V | | 240 | | |
| | | $I_L$ = 500mA | | 300 | | |
| | | Output voltage > 10V | | 420 | | |
| | | $I_L$ = 750mA | | 370 | 600 | |
| | | | | | **750** | |
| $I_{GND}$ | Ground Pin Current (**Note 5**) | $I_L$ = 5mA | | 160 | 250 | μA |
| | | | | | **300** | |
| | | $I_L$ = 100mA | | 1 | 2.5 | mA |
| | | | | | **3** | |
| | | $I_L$ = 500mA | | 8 | 13 | |
| | | | | | **16** | |
| | | $I_L$ = 750mA | | 15 | 25 | |
| $I_{GNDDO}$ | Ground Pin Current at Dropout (**Note 5**) | $V_{IN}$ = 0.5V less than designed $V_{OUT}$ ($V_{OUT}$ ≥ 3.3V) $I_O$ = 5mA | | 200 | **500** | μA |
| $I_{LIMIT}$ | Current Limit | $V_{OUT}$ = 0V | | 1.1 | 1.5 | A |
| | | (**Note 6**) | | | **2** | |
| $\frac{\Delta V_O}{\Delta P_D}$ | Thermal Regulation | (**Note 7**) | | 0.05 | 0.2 | %/W |
| $e_n$ | Output Noise Voltage (10Hz to 100kHz) $I_L$ = 100mA | $C_L$ = 10μF | | 400 | | μV RMS |
| | | $C_L$ = 100μF | | 260 | | |

## Electrical Characteristics (Continued)

**MIC29372**

| Parameter | Conditions | Min | Typical | Max | Units |
|---|---|---|---|---|---|
| Reference Voltage | | 1.223<br>1.210 | 1.235 | 1.247<br>1.260 | V<br>V max |
| Reference Voltage | (**Note 8**) | **1.204** | | **1.266** | V |
| Adjust Pin Bias Current | | | 20 | 40<br>**60** | nA |
| Reference Voltage Temperature Coefficient | (**Note 7**) | | 20 | | ppm/°C |
| Adjust Pin Bias Current Temperature Coefficient | | | 0.1 | | nA/°C |
| **Error Comparator** | **MIC29371** | | | | |
| Output Leakage Current | $V_{OH}$ = 26V | | 0.01 | 1.00<br>**2.00** | µA |
| Output Low Voltage | $V_{IN}$ = 4.5V<br>$I_{OL}$ = 250µA | | 150 | 250<br>**400** | mV |
| Upper Threshold Voltage | (**Note 9**) | 40<br>**25** | 60 | | mV |
| Lower Threshold Voltage | (**Note 9**) | | 75 | 95<br>**140** | mV |
| Hysteresis | (**Note 9**) | | 15 | | mV |
| **Shutdown Input** | **MIC29371/MIC29372** | | | | |
| Input Logic Voltage Low (ON) | | | 1.3 | **0.7** | V |
| | High (OFF) | **2.0** | | | |
| Shutdown Pin Input Current | $V_{SHUTDOWN}$ = 2.4V | | 30 | 50<br>**100** | µA |
| | $V_{SHUTDOWN}$ = 26V | | 450 | 600<br>**750** | µA |
| Regulator Output Current in Shutdown | (**Note 10**) | | 3 | 10<br>**20** | µA |

**Note 1:** Absolute maximum ratings indicate limits beyond which damage to the component may occur. Electrical specifications do not apply when operating the device outside of its rated operating conditions. The maximum allowable power dissipation is a function of the maximum junction temperature, $T_{J\,(MAX)}$, the junction-to-ambient thermal resistance, $\theta_{JA}$, and the ambient temperature, $T_A$. The maximum allowable power dissipation at any ambient temperature is calculated using: $P_{(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$. Exceeding the maximum allowable power dissipation will result in excessive die temperature, and the regulator will go into thermal shutdown.

**Note 2:** Output voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.

**Note 3:** Regulation is measured at constant junction temperature using low duty cycle pulse testing. Changes in output voltage due to heating effects are covered by the thermal regulation specification.

**Note 4:** Dropout Voltage is defined as the input to output differential at which the output voltage drops 100 mV below its nominal value measured at 1V differential. At low values of programmed output voltage, the minimum input supply voltage of 4.3V over temperature must be taken into account. The MIC2937A operates down to 2V of input at reduced output current at 25°C.

**Note 5:** Ground pin current is the regulator quiescent current. The total current drawn from the source is the sum of the load current plus the ground pin current.

**Note 6:** The MIC2937A family features fold-back current limiting. The short circuit ($V_{OUT}$ = 0V) current limit is less than the maximum current with normal output voltage.

**Note 7:** Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 200mA load pulse at $V_{IN}$ = 20V (a 4W pulse) for T = 10ms.

**Note 8:** $V_{REF} \le V_{OUT} \le (V_{IN} - 1\text{ V})$, $4.3\text{V} \le V_{IN} \le 26\text{V}$, $5\text{mA} < I_L \le 750\text{ mA}$, $T_J \le T_{J\,MAX}$.

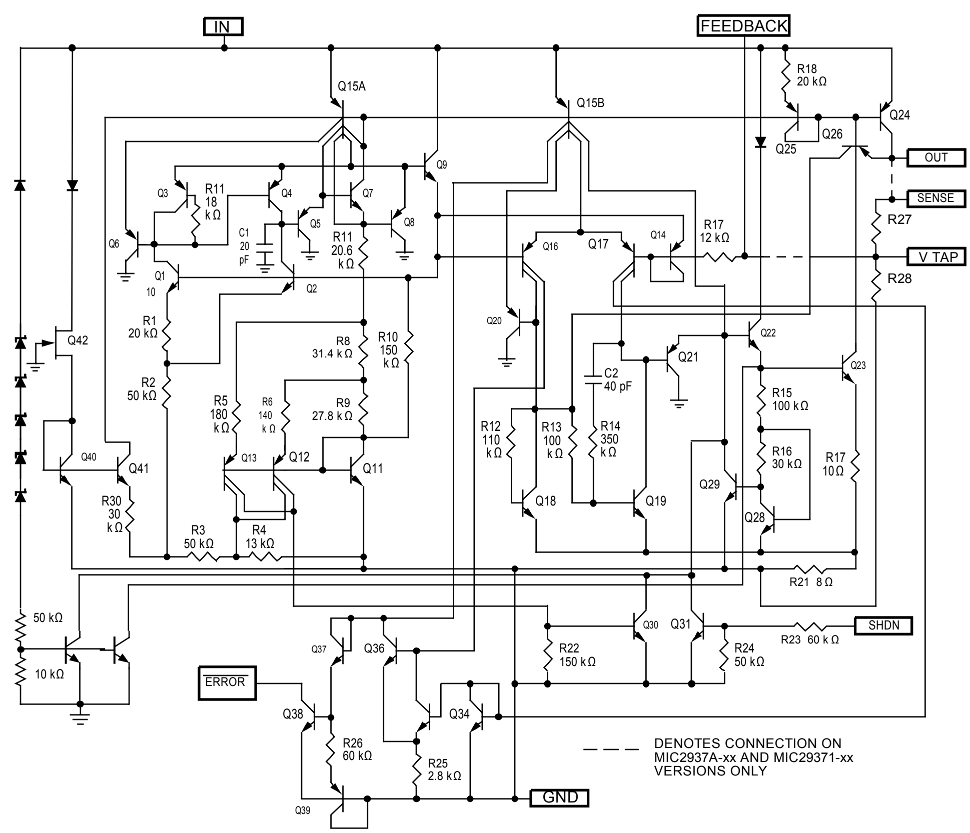
**Note 9:** Comparator thresholds are expressed in terms of a voltage differential at the Adjust terminal below the nominal reference voltage measured at 6V input (for a 5V regulator). To express these thresholds in terms of output voltage change, multiply by the error amplifier gain = $V_{OUT}/V_{REF}$ = (R1 + R2)/R2. For example, at a programmed output voltage of 5V, the Error output is guaranteed to go low when the output drops by 95 mV x 5V/1.235 V = 384 mV. Thresholds remain constant as a percent of $V_{OUT}$ as $V_{OUT}$ is varied, with the dropout warning occurring at typically 5% below nominal, 7.7% guaranteed.

**Note 10:** Circuit of Figure 3 with R1 ≥ 150kΩ. $V_{SHUTDOWN} \ge 2\text{V}$ and $V_{IN} \le 26\text{V}$, $V_{OUT} = 0$.

**Note 11:** When used in dual supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.

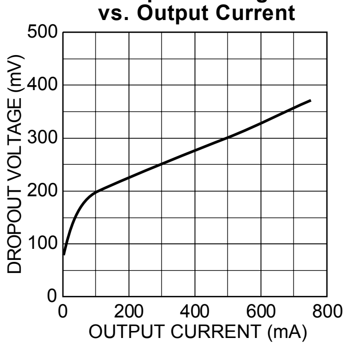
**Note 12:** Maximum positive supply voltage of 60V must be of limited duration (< 100ms) and duty cycle ( ≤ 1%). The maximum continuous supply voltage is 26V.

## Schematic Diagram

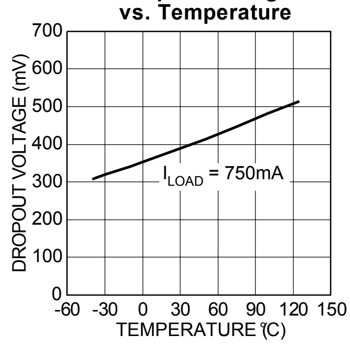


# Typical Characteristics

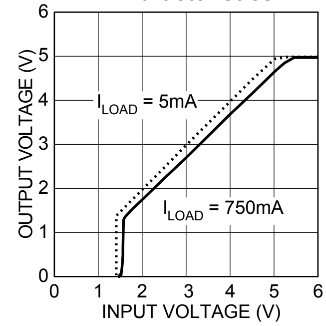
**Dropout Voltage vs. Output Current**



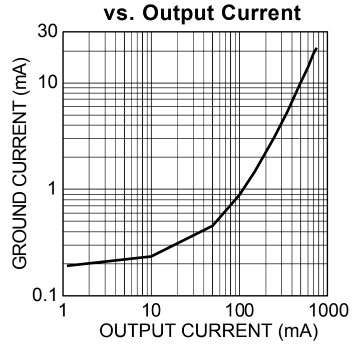

**Dropout Voltage vs. Temperature**



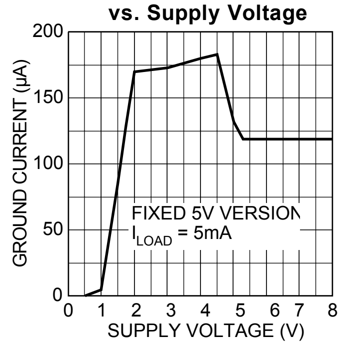

**Dropout Characteristics**



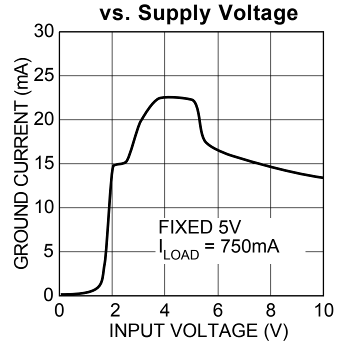

**Ground Current vs. Output Current**



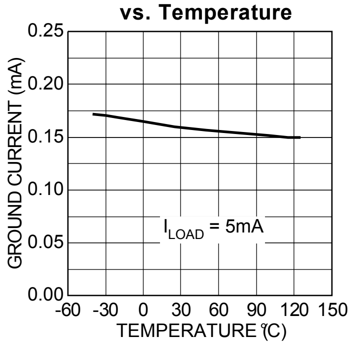

**Ground Current vs. Supply Voltage**



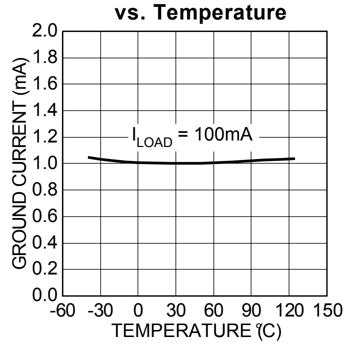

**Ground Current vs. Supply Voltage**



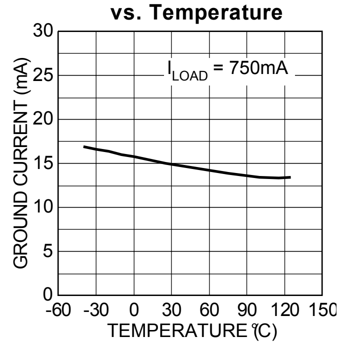

**Ground Current vs. Temperature**



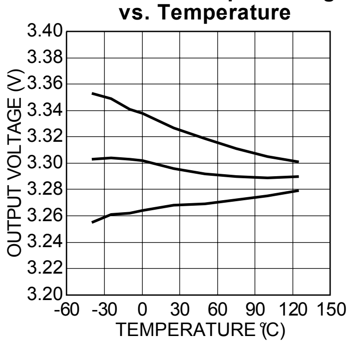

**Ground Current vs. Temperature**



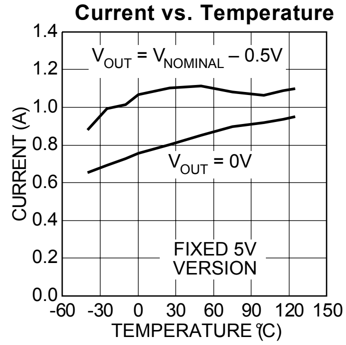

**Ground Current vs. Temperature**



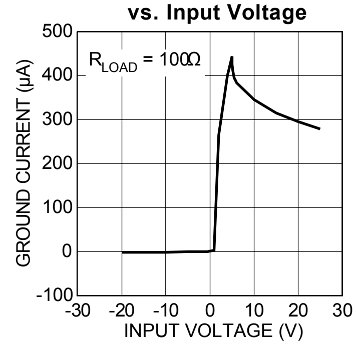

**Fixed 3.3V Output Voltage vs. Temperature**

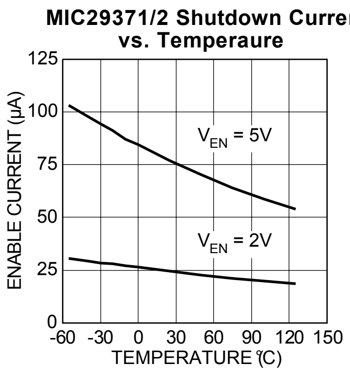


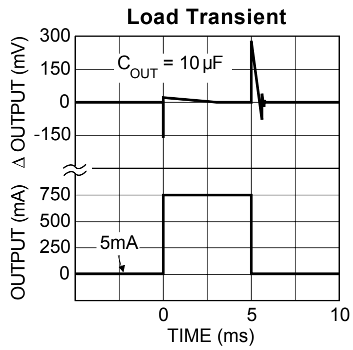

**Short Circuit and Maximum Current vs. Temperature**

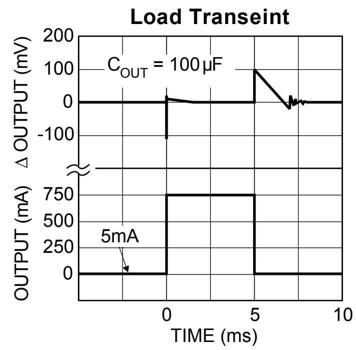


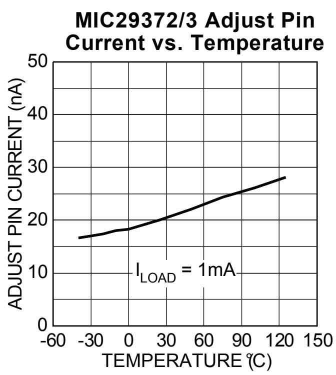

**Ground Current vs. Input Voltage**

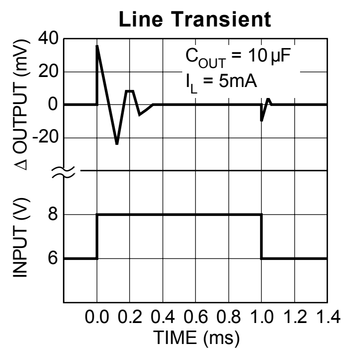


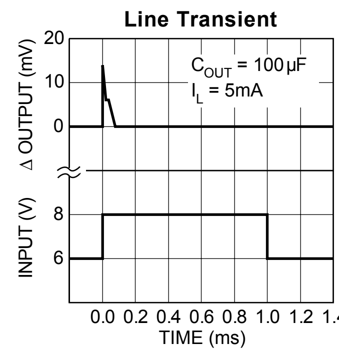

MIC29371/2 Shutdown Current
vs. Temperaure
ENABLE CURRENT (µA)
TEMPERATURE (°C)
V_EN = 5V
V_EN = 2V
125
100
75
50
25
0
-60 -30 0 30 60 90 120 150

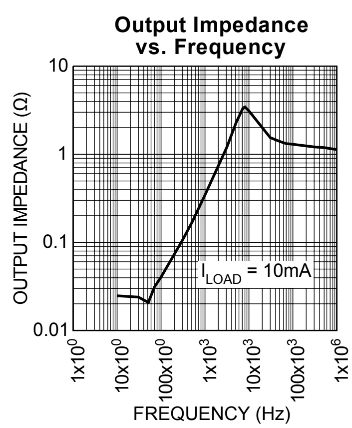

Load Transient
C_OUT = 10µF
Δ OUTPUT (mV)
OUTPUT (mA)
5mA
TIME (ms)
300
150
0
-150
750
500
250
0
0 5 10


Load Transeint
C_OUT = 100µF
Δ OUTPUT (mV)
OUTPUT (mA)
5mA
TIME (ms)
200
100
0
-100
750
500
250
0
0 5 10


MIC29372/3 Adjust Pin
Current vs. Temperature
ADJUST PIN CURRENT (nA)
TEMPERATURE (°C)
I_LOAD = 1mA
50
40
30
20
10
0
-60 -30 0 30 60 90 120 150


Line Transient
C_OUT = 10µF
I_L = 5mA
Δ OUTPUT (mV)
INPUT (V)
TIME (ms)
40
20
0
-20
8
6
0.0 0.2 0.4 0.6 0.8 1.0 1.2 1.4


Line Transient
C_OUT = 100µF
I_L = 5mA
Δ OUTPUT (mV)
INPUT (V)
TIME (ms)
20
10
0
8
6
0.0 0.2 0.4 0.6 0.8 1.0 1.2 1.4


Output Impedance
vs. Frequency
OUTPUT IMPEDANCE (Ω)
FREQUENCY (Hz)
I_LOAD = 10mA
10
1
0.1
0.01
1x10^0
10x10^0
100x10^0
1x10^3
10x10^3
100x10^3
1x10^6

# Applications Information

## External Capacitors

A 10µF (or greater) capacitor is required between the MIC2937A output and ground to prevent oscillations due to instability. Most types of tantalum or aluminum electrolytics will be adequate; film types will work, but are costly and therefore not recommended. Many aluminum electrolytics have electrolytes that freeze at about –30°C, so solid tantalums are recommended for operation below –25°C. The important parameters of the capacitor are an effective series resistance of about 5Ω or less and a resonant frequency above 500kHz. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.5µF for current below 10mA or 0.15µF for currents below 1 mA. Adjusting the MIC29372 to voltages below 5V runs the error amplifier at lower gains so that more output capacitance is needed. For the worst-case situation of a 750mA load at 1.23V output (Output shorted to Adjust) a 22µF (or greater) capacitor should be used.

The MIC2937A/29371 will remain in regulation with a minimum load of 5mA. When setting the output voltage of the MIC29372 version with external resistors, the current through these resistors may be included as a portion of the minimum load.

A 0.1µF capacitor should be placed from the input to ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

## Error Detection Comparator Output (MIC29371)

A logic low output will be produced by the comparator whenever the MIC29371 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 75mV divided by the 1.235V reference voltage. (Refer to the block diagram on Page 1). This trip level remains "5% below normal" regardless of the programmed output voltage of the MIC29371. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage,extremely high input voltage, current limiting, or thermal limiting.

Figure 1 is a timing diagram depicting the $\overline{\text{ERROR}}$ signal and the regulated output voltage as the MIC29371 input is ramped up and down. The $\overline{\text{ERROR}}$ signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which $V_{OUT}$ = 4.75). Since the MIC29371's dropout voltage is load-dependent (see curve in Typical Performance Characteristics), the input voltage trip point (about 5V) will vary with the load current. The output voltage trip point (approximately 4.75V) does not vary with load.

The error comparator has an NPN open-collector output which requires an external pull-up resistor. Depending on system requirements, this resistor may be returned to the 5V output or some other supply voltage. In determining a value for this resistor, note that while the output is rated to sink 250µA, this sink current adds to battery drain in a low battery condition. Suggested values range from 100k to 1MΩ. The resistor is not required if this output is unused.

## Programming the Output Voltage (MIC29372)

The MIC29372 may programmed for any output voltage between its 1.235V reference and its 26V maximum rating. An external pair of resistors is required, as shown in Figure 3.

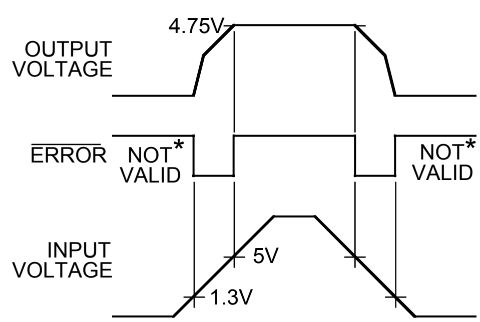
The complete equation for the output voltage is

$$V_{OUT} = V_{REF} \times \{ 1 + R_1/R_2 \} - |I_{FB}| R_1$$

where $V_{REF}$ is the nominal 1.235 reference voltage and $I_{FB}$ is the Adjust pin bias current, nominally 20nA. The minimum recommended load current of 1µA forces an upper limit of 1.2MΩ on the value of $R_2$, if the regulator must work with no load (a condition often found in CMOS in standby), $I_{FB}$ will produce a –2% typical error in $V_{OUT}$ which may be eliminated at room temperature by trimming $R_1$. For better accuracy, choosing $R_2$ = 100k reduces this error to 0.17% while increasing the resistor program current to 12µA. Since the MIC29372 typically draws 100µA at no load with SHUTDOWN open-circuited, this is a negligible addition.

## Reducing Output Noise

In reference applications it may be advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is relatively inefficient, as increasing the capacitor from 1µF to 220µF only decreases the noise from 430µV to 160µ$V_{RMS}$ for a 100kHz bandwidth at 5V output. Noise can be reduced by a factor of four with the adjustable regulators




* SEE APPLICATIONS INFORMATION

**Figure 1. $\overline{\text{ERROR}}$ Output Timing**

with a bypass capacitor across $R_1$, since it reduces the high frequency gain from 4 to unity. Pick

$$C_{BYPASS} \cong \frac{1}{2\pi R_1 \bullet 200\text{ Hz}}$$

or about 0.01µF. When doing this, the output capacitor must be increased to 10µF to maintain stability. These changes reduce the output noise from 430µV to $100\mu V_{RMS}$ for a 100 kHz bandwidth at 5V output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

### Automotive Applications

The MIC2937A is ideally suited for automotive applications for a variety of reasons. It will operate over a wide range of input voltages with very low dropout voltages (40mV at light loads), and very low quiescent currents (100µA typical). These features are necessary for use in battery powered systems, such as automobiles. It is a "bulletproof" device with the ability to survive both reverse battery (negative transients up to 20V below ground), and load dump (positive transients up to 60V) conditions. A wide operating temperature range with low temperature coefficients is yet another reason to use these versatile regulators in automotive designs.

## Typical Applications

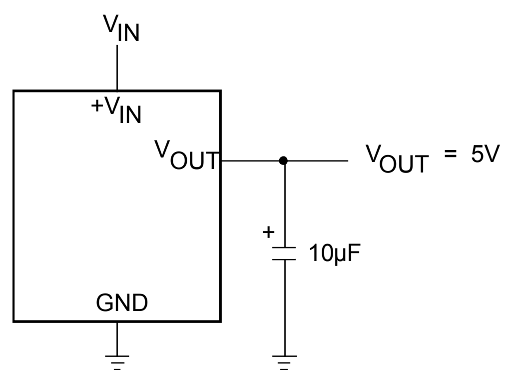



Figure 2. MIC2937A-5.0 Fixed +5V Regulator

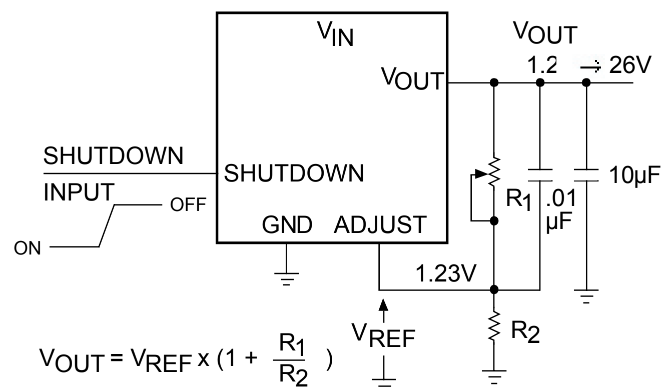
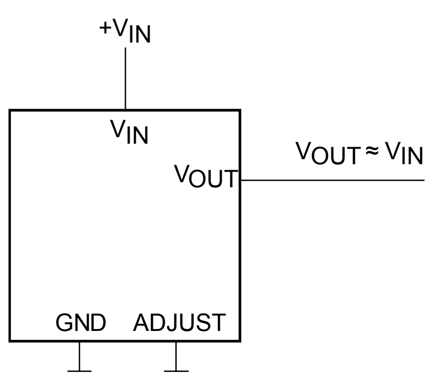
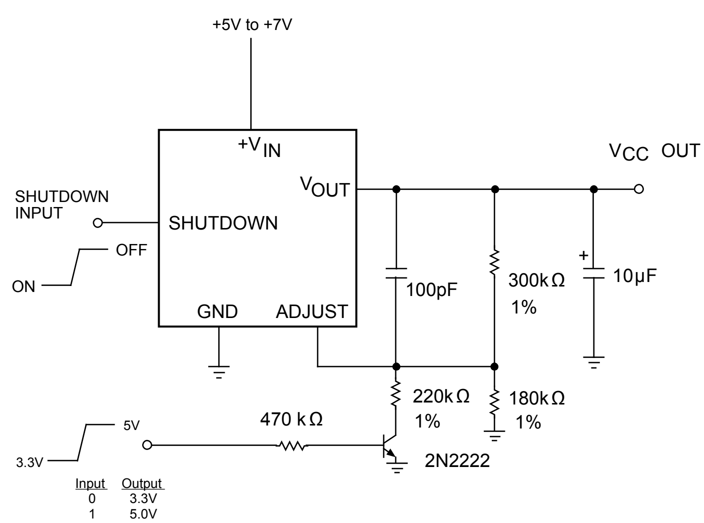



Figure 3. MIC29372 Adjustable Regulator




*MINIMUM INPUT-OUTPUT VOLTAGE RANGES FROM 40mV TO 400mV, DEPENDING ON LOAD CURRENT.

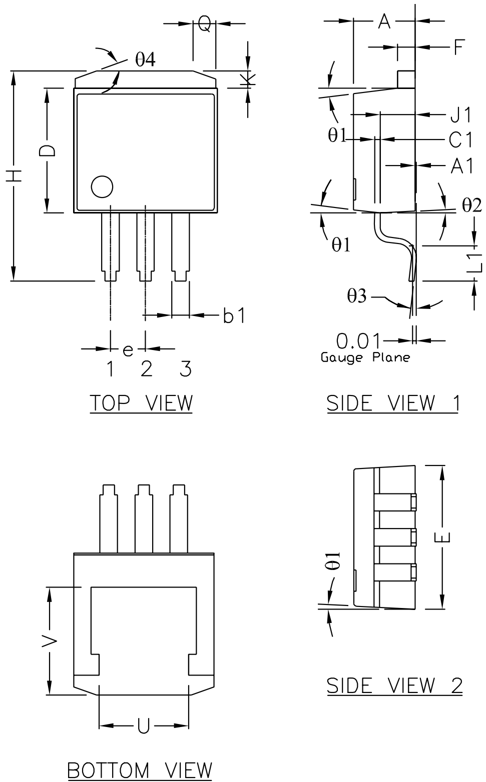
Figure 4. MIC29372 Wide Input Voltage Range Current Limiter




SHUTDOWN PIN LOW= ENABLE OUTPUT. Q1 ON = 3.3V, Q1 OFF = 5.0V.

Figure 5. MIC29372 5.0V or 3.3V Selectable Regulator with Shutdown.

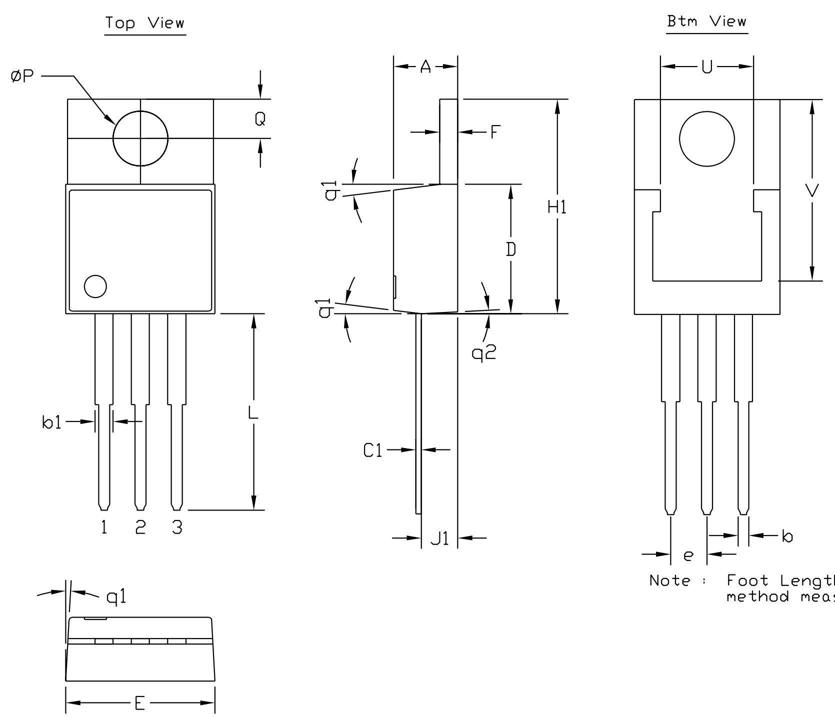
# Package Information




| POS | INCH MIN | INCH MAX | MM MIN | MM MAX |
|---|---|---|---|---|
| A | 0.171 | 0.181 | 4.343 | 4.597 |
| A1 | 0.000 | 0.012 | 0.000 | 0.305 |
| b1 | 0.047 | 0.053 | 1.194 | 1.346 |
| C1 | 0.012 | 0.018 | 0.305 | 0.457 |
| D | 0.351 | 0.361 | 8.915 | 9.169 |
| E | 0.400 | 0.420 | 10.160 | 10.668 |
| e | 0.095 | 0.105 | 2.413 | 2.667 |
| F | 0.045 | 0.055 | 1.143 | 1.397 |
| H | 0.575 | 0.625 | 14.605 | 15.875 |
| J1 | 0.080 | 0.120 | 2.032 | 3.048 |
| L1 | 0.090 | 0.110 | 2.286 | 2.794 |
| K | 0.045 | 0.055 | 1.143 | 1.397 |
| θ1 | 3° | 10° | 3° | 10° |
| θ2 | 1° | 7° | 1° | 7° |
| θ3 | 0° | 8° | 0° | 8° |
| θ4 | 18° | 22° | 18° | 22° |
| Q | 0.055 | 0.075 | 1.397 | 1.905 |
| U | 0.256 Ref. | | 6.502 Ref. | |
| V | 0.303 Ref. | | 7.696 Ref. | |

NOTE:
FOOT LENGTH USING GAUGE PLANE METHOD MEASUREMENT 0.010".

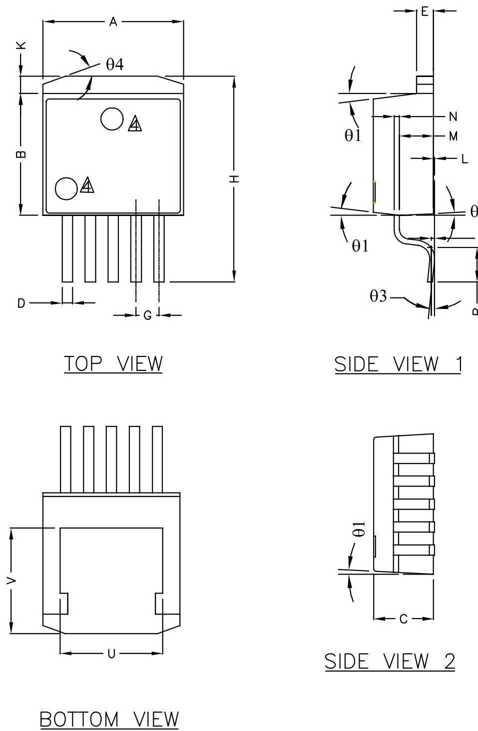
**3-Pin TO-263 (U)**




| POS | INCH MIN | INCH MAX |
|---|---|---|
| A | 0.160 | 0.190 |
| b | 0.025 | 0.040 |
| b1 | 0.045 | 0.060 |
| C1 | 0.012 | 0.022 |
| D | 0.351 | 0.361 |
| E | 0.385 | 0.420 |
| e | 0.090 | 0.110 |
| F | 0.045 | 0.055 |
| H1 | 0.560 | 0.595 |
| J1 | 0.080 | 0.120 |
| øP | 0.146 | 0.156 |
| Q | 0.103 | 0.113 |
| L | 0.500 | 0.580 |
| q1 | 3° | 10° |
| q2 | 1° | 7° |
| U | 0.256 Ref. | |
| V | 0.486 Ref. | |

Note : Foot Length using gauge plane method measurement 0.010".

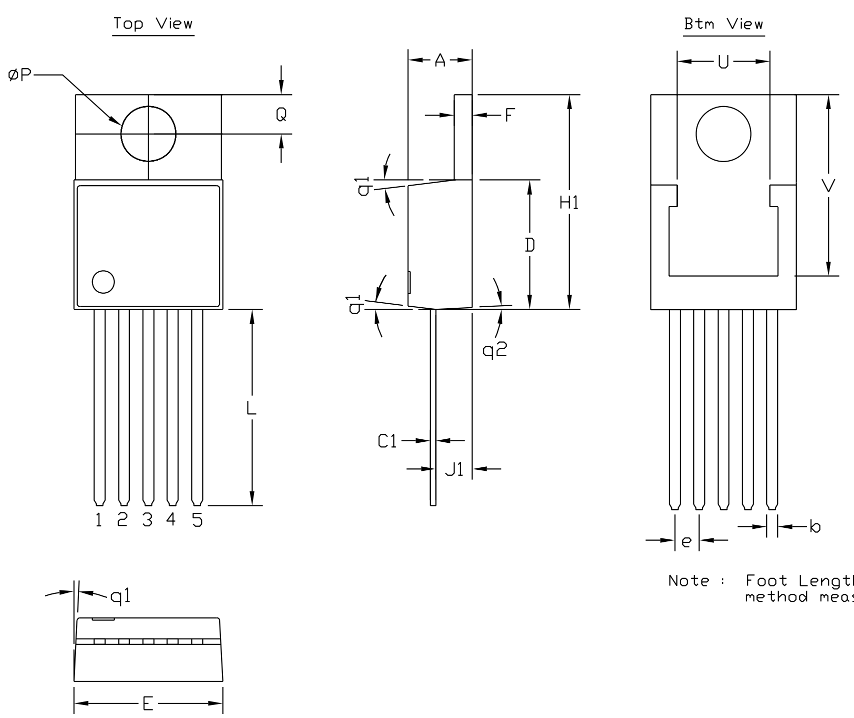
**3-Pin TO-220 (T)**




| POS | INCH MIN | INCH MAX | MM MIN | MM MAX |
|---|---|---|---|---|
| A | 0.396 | 0.420 | 10.058 | 10.668 |
| B | 0.330 | 0.361 | 8.382 | 9.169 |
| C | 0.170 | 0.181 | 4.318 | 4.597 |
| D | 0.026 | 0.036 | 0.660 | 0.914 |
| E | 0.045 | 0.055 | 1.143 | 1.397 |
| G | 0.067 Ref. | | 1.70 Ref. | |
| H | 0.575 | 0.625 | 14.605 | 15.875 |
| K | 0.045 | 0.066 | 1.143 | 1.676 |
| L | 0 | 0.012 | 0 | 0.305 |
| M | 0.080 | 0.120 | 2.032 | 3.048 |
| N | 0.012 | 0.023 | 0.305 | 0.584 |
| P | 0.090 | .0110 | 2.286 | 2.794 |
| θ1 | 3° | 10° | 3° | 10° |
| θ2 | 1° | 7° | 1° | 7° |
| θ3 | 0° | 8° | 0° | 8° |
| θ4 | 18° | 22° | 18° | 22° |
| U | 0.300 Ref. | | 7.620 Ref. | |
| V | 0.305 Ref. | | 7.747 Ref. | |
| W | 0.010 Ref. | | 0.254 Ref. | |

NOTE:
1. PACKAGE OUTLINE EXCLUSIVE OF MOLD FLASH & METAL BURR.
2. PACKAGE OUTLINE INCLUSIVE OF PLATING THICKNESS.
3. FOOT LENGTH MEASURED AT INTERCEPT POINT BETWEEN DATUM A & LEAD SURFACE
4. PACKAGE TOP MARK MAY BE IN TOP CENTER OR LOWER LEFT CORNER

**5-Pin TO-263 (U)**




| POS | INCH MIN | INCH MAX |
|---|---|---|
| A | 0.160 | 0.190 |
| b | 0.025 | 0.040 |
| C1 | 0.012 | 0.022 |
| D | 0.351 | 0.361 |
| E | 0.385 | 0.420 |
| e | 0.062 | 0.072 |
| F | 0.045 | 0.055 |
| H1 | 0.560 | 0.595 |
| J1 | 0.080 | 0.120 |
| øP | 0.146 | 0.156 |
| Q | 0.103 | 0.113 |
| L | 0.500 | 0.580 |
| q1 | 3° | 10° |
| q2 | 1° | 7° |
| U | 0.256 Ref. | |
| V | 0.486 Ref. | |

Note : Foot Length using gauge plane method measurement 0.010".

**5-Pin TO-220 (T)**

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