

#### **ABSOLUTE MAXIMUM RATINGS**

$\begin{tabular}{ll} Input Voltage, V_{IN} &$	Power Dissipation (Package) <sup>a</sup> 8-Pin TSSOP
Output Current, I <sub>OUT</sub> Short Circuit Protected	Thermal Impedance ( $\Theta_{JA}$ )
Output Voltage, V <sub>OUT</sub>	8-Pin TSSOPb
Maximum Junction Temperature, T <sub>J(max)</sub>	Notes
Storage Temperature, T <sub>STG</sub> 55°C to 150°C	<ul> <li>a. Device mounted with all leads soldered or welded to PC board.</li> <li>b. Derate 8.3 mW/°C above T<sub>A</sub> = 25°C</li> </ul>
ESD (Human Body Model)	b. Derate 0.5 mwy C above 1A = 25 C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **RECOMMENDED OPERATING RANGE**

Input Voltage, V <sub>IN</sub>	Operating Ambient Temperature, $T_A$ 40°C to 85°C
Output Voltage, V <sub>OUT</sub> (Adjustable Version) 1.5 V to 5 V	Operating Junction Temperature, T <sub>.1</sub> 40°C to 125°C
$\overline{\text{SD}}$ Input Voltage, $V_{\overline{\text{SD}}}$	
$C_{IN}$ = 2.2 $\mu$ F, $C_{OUT}$ = 2.2 $\mu$ F (ceramic, X5R or X7R type) , $C_{NOISE}$ = 0.1 $\mu$ F (ceramic)	
$C_{OUT}$ Range = 1 $\mu$ F to 10 $\mu$ F ( $\pm$ 10%, x5R or x7R type)	
$C_{IN} \geq C_{OUT}$	

SPECIFICATIONS								
		Test Conditions Unless Otherwise Specified		Limits -40 to 85°C				
Parameter	Symbol	$\begin{aligned} V_{IN} &= V_{OUT(nom)} + 1 \text{ V, } I_{OUT} = 1 \text{ mA} \\ C_{IN} &= 2.2 \text{ \muF, } C_{OUT} = 2.2 \text{ \muF, } V_{SD} = 1.5 \text{ V} \end{aligned}$	Temp <sup>a</sup>	Minb	Турс	Max <sup>b</sup>	Unit	
			_					
Output Voltage Range		Adjustable Version	Full	1.5		5	V	
Output Voltage Accuracy	V <sub>OUT</sub>	4 4 1 1 250 4	Room	-1.5		1.5	% V <sub>O(nom)</sub>	
(Fixed Versions)		$1 \text{ mA} \le I_{OUT} \le 350 \text{ mA}$	Full	-2.5		2.5		
F # 13/# (AD13/ : )	.,		Room	1.191	1.215	1.239		
Feedback Voltage (ADJ Version)	V <sub>ADJ</sub>		Full	1.179		1.251	- v	
Line Regulation (Except 5-V Version)		From $V_{IN} = V_{OUT(nom)} + 1 V$ to $V_{OUT(nom)} + 2 V$	Full	-0.18		0.18		
Line Regulation (5-V Version)	$\frac{\Delta V_{OUT} \times 100}{V_{IN} \times V_{OUT(nom)}}$	From $V_{IN} = 5.5 \text{ V to 6 V}$	Full	-0.18		0.18	%/V	
Line Regulation (ADJ Version)		$V_{OUT}$ = 1.5 V, From $V_{IN}$ = 2.5 V to 3.5 V	Full	-0.18		0.18	1	
		$V_{OUT} = 5 \text{ V}$ , From $V_{IN} = 5.5 \text{ V}$ to 6 V	Full	-0.18		0.18		
		I <sub>OUT</sub> = 10 mA	Room		5	20		
Dropout Voltage <sup>d</sup>		I <sub>OUT</sub> = 200 mA	Room		85	180		
(@V <sub>OUT</sub> ≥ 2 V)			Room		150	400	mV	
	V <sub>IN</sub> – V <sub>OUT</sub>	I <sub>OUT</sub> = 350 mA	Full			550		
		I <sub>OUT</sub> = 200 mA	Room		170	250		
Dropout Voltage <sup>d</sup> (@ $V_{OUT}$ < 2 V, $V_{IN} \ge 2$ V)			Room		290	425		
		I <sub>OUT</sub> = 350 mA	Full			575		
Ground Pin Current	I <sub>GND</sub>	I <sub>OUT</sub> = 0 mA	Room		150			
			Room		1000		1	
		I <sub>OUT</sub> = 200 mA	Full			1500	μΑ	
			Room		1500			
		I <sub>OUT</sub> = 350 mA	Full			2800		



SPECIFICATIONS									
		Test Conditions Unless Otherwise Specified $V_{IN} = V_{OUT(nom)} + 1 \text{ V, } I_{OUT} = 1 \text{ mA}$ $C_{IN} = 2.2 \mu\text{F, } C_{OUT} = 2.2 \mu\text{F, } V_{\overline{SD}} = 1.5 \text{ V}$			Limits -40 to 85°C				
Parameter	Symbol			Temp <sup>a</sup>	Minb	Турс	Max <sup>b</sup>	Unit	
Shutdown Supply Current	I <sub>IN(off)</sub>	V <sub>SD</sub> = 0	V	Room		0.1	1	μΑ	
ADJ Pin Current	I <sub>ADJ</sub>	ADJ = 1.2	2 V	Room		5	100	nA	
Peak Output Current	I <sub>O(peak)</sub>	$V_{OUT} \ge 0.95 \times V_{OUT}$	<sub>nom)</sub> , t <sub>pw</sub> = 2 ms	Room	600			mA	
0 1 111 1 1/1		BW = 50 Hz to 100 kHz	w/o C <sub>NOISE</sub>	Room		200		μV (rms)	
Output Noise Voltage	e <sub>N</sub>	$I_{OUT} = 150 \text{ mA}$	C <sub>NOISE</sub> = 0.1 μF	Room		100			
			f = 1 kHz	Room		60			
Ripple Rejection	$\Delta V_{OUT}/\Delta V_{IN}$	I <sub>OUT</sub> = 150 mA	f = 10 kHz	Room		60		dB	
			f = 100 kHz	Room		40			
Dynamic Line Regulation	$\Delta V_{O(line)}$	V <sub>IN</sub> : V <sub>OUT(nom)</sub> + 1 V to V <sub>OUT(nom)</sub> + 2 V t <sub>R</sub> /t <sub>F</sub> = 5 μs, I <sub>OUT</sub> = 350 mA		Room		10		mV	
Dynamic Load Regulation	$\Delta V_{O(load)}$	I <sub>OUT</sub> : 1 mA to 150 m	nA, t <sub>R</sub> /t <sub>F</sub> = 2 μs	Room		30			
	· · · · · · · · · · · · · · · · · · ·	V <sub>IN</sub> = 4.3 V	w/o C <sub>NOISE</sub> Cap	Room		5		μs	
V <sub>OUT</sub> Turn-On-Time	t <sub>ON</sub>	V <sub>OUT</sub> = 3.3 V	C <sub>NOISE</sub> = 0.1 μF	Room		2		mS	
Thermal Shutdown		•	1	l	1				
Thermal Shutdown Junction Temp	t <sub>J(s/d)</sub>					165		_	
Thermal Hysteresis	t <sub>HYST</sub>			Room		20		°C	
Short Circuit Current	I <sub>SC</sub>	V <sub>OUT</sub> = 0	V	Room		800		mA	
Shutdown Input				l					
V <sub>IH</sub>		High = Regulator ON (Rising)		Full	1.5		V <sub>IN</sub>		
SD Input Voltage	V <sub>IL</sub>	Low = Regulator OFF (Falling)		Full			0.4	V	
_	I <sub>IH</sub>	V <sub>SD</sub> = 0 V, Regulator OFF		Room		0.01		μΑ	
SD Input Currente	I <sub>IL</sub>	V <sub>SD</sub> = 6 V, Regulator ON		Room		1.0			
Shutdown Hysteresis	V <sub>HYST</sub>	5		Full		100		mV	
Error Output					,				
Output High Leakage	I <sub>OFF</sub>	ERROR = V <sub>OUT(nom)</sub>		Full		0.01	2	μА	
Output Low Voltage <sup>g</sup>	V <sub>OL</sub>	I <sub>SINK</sub> = 2 mA		Full			0.4		
Power_Good Trip Threshold <sup>f, h</sup> (Rising)	$V_{TH}$			Full	0.93 x V <sub>OUT</sub>	0.95 x V <sub>OUT</sub>	0.97 x V <sub>OUT</sub>	٧	
Hysteresis <sup>f</sup>	V <sub>HYST</sub>			Room		2% x V <sub>OUT</sub>			
Delay Pin Current Source	I <sub>DELAY</sub>			Room	1.2	2.2	3.0	μΑ	

- b.
- Room =  $25^{\circ}$ C, Full = -40 to  $85^{\circ}$ C. The algebraic convention whereby the most negative value is a minimum and the most positive a maximum. Typical values are for DESIGN AID ONLY, not guaranteed nor subject to production testing. Typical values for dropout voltage at  $V_{OUT} \ge 2$  V are measured at  $V_{OUT} = 3.3$  V, while typical values for dropout voltage at  $V_{OUT} < 2$  V are measured at  $V_{OUT} = 1.8$  V. Dropout voltage is defined as the input to output differential voltage at which the output voltage drops 2% below the output voltage measured with a 1-V differential, provided that  $V_{IN}$  does not not drop below 2.0 V. The device's shutdown pin includes a typical  $6-M\Omega$  internal pull-down resistor connected to ground.  $V_{OUT}$  is defined as the output voltage of the DUT at 1 mA. The Error Output (Low) function is guaranteed from  $V_{OUT} = 2.0$  V to  $V_{OUT} = 5.0$  V. The Power\_Good trip threshold function is guaranteed from  $V_{OUT} = 1.5$  V to  $V_{OUT} = 5.0$  V and  $V_{IN} \ge 2.0$  V.



#### **TIMING WAVEFORMS**

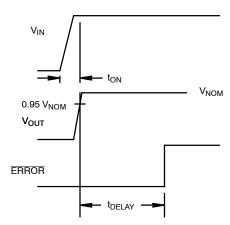
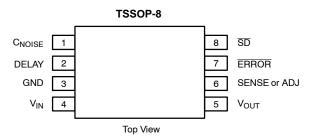


FIGURE 4. Timing Diagram for Power-Up

#### **PIN CONFIGURATION**



PIN DESCRIPTION				
Pin Number	Name	Function		
1	C <sub>NOISE</sub>	Noise bypass pin. For low noise applications, a $0.01$ - $\mu F$ or larger ceramic capacitor should be connected from this pin to ground.		
2	DELAY	Capacitor connected from this pin to ground will allow a delayed power-on-reset signal at the ERROR (Pin 7) output. Refer to Figure 4.		
3	GND	Ground pin. Local ground for C <sub>NOISE</sub> and C <sub>OUT</sub> .		
4	V <sub>IN</sub>	Input supply pin. Bypass this pin with a 2.2-μF ceramic or tantalum capacitor to ground.		
5	V <sub>OUT</sub>	Output voltage. Connect C <sub>OUT</sub> between this pin and ground.		
6	SENSE or ADJ	For fixed output voltage versions, this pin should be connected to V <sub>OUT</sub> (Pin 5). For adjustable output voltage version, this voltage feedback pin sets the output voltage via an external resistor divider.		
7	ERROR	This open drain output is an error flag output which goes low when V <sub>OUT</sub> drops 5% below its nominal voltage. This pin also provides a power-on-reset signal if a capacitor is connected to the DELAY pin.		
8	SD	By applying less than 0.4 V to this pin, the device will be turned off. Connect this pin to V <sub>IN</sub> if unused.		

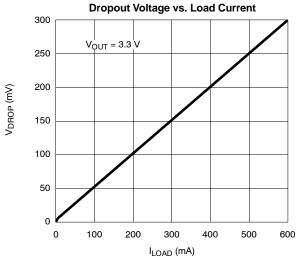


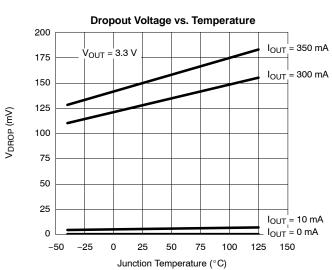
ORDERING INFORMATION						
Part Number	Lead (Pb)-Free Part Number	Marking	Voltage	Temperature Range	Package	
Si9181DQ-15-T1	Si9181DQ-15-T1—E3	115	1.5 V			
Si9181DQ-18-T1	Si9181DQ-18-T1—E3	118	1.8 V			
Si9181DQ-25-T1	Si9181DQ-25-T1—E3	125	2.5 V			
Si9181DQ-28-T1	Si9181DQ-28-T1—E3	128	2.8 V	–40 to 85°C	TSSOP-8	
Si9181DQ-30-T1	Si9181DQ-30-T1—E3	130	3.0 V	-40 to 65 C	13307-6	
Si9181DQ-33-T1	Si9181DQ-33-T1—E3	133	3.3 V			
Si9181DQ-50-T1	Si9181DQ-50-T1—E3	150	5.0 V			
Si9181DQ-AD-T1	Si9181DQ-AD-T1—E3	1AD	Adjustable			

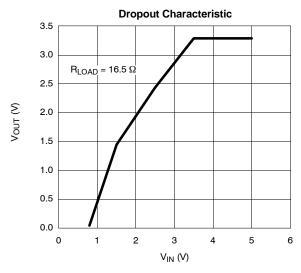
<sup>\*</sup> Additional voltage options are available.

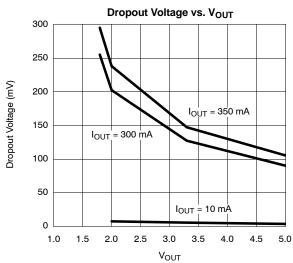
Eval Kit	Temperature Range	Board Type	
Si9181DB	−40 to 85°C	Surface Mount	

#### TYPICAL CHARACTERISTICS (INTERNALLY REGULATED, 25°C UNLESS NOTED)





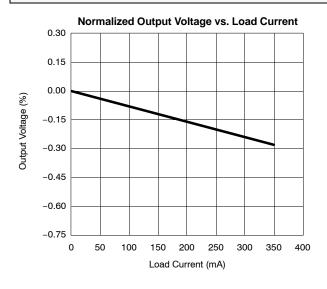


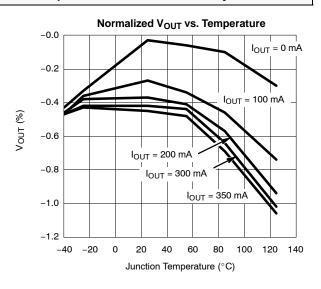


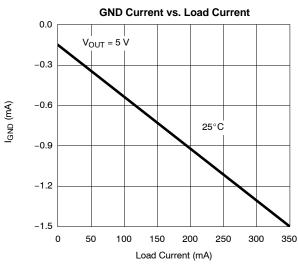
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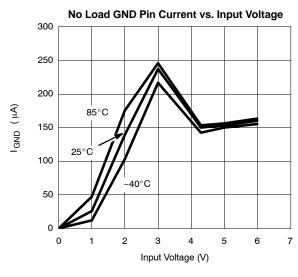


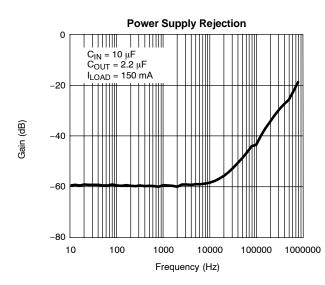
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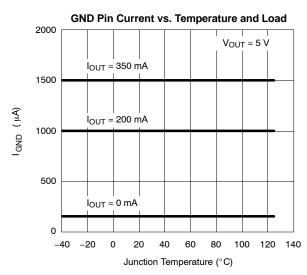










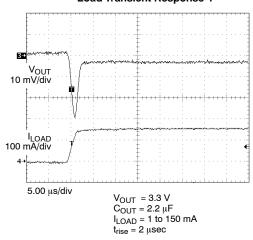




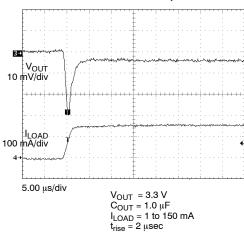


#### **TYPICAL WAVEFORMS**

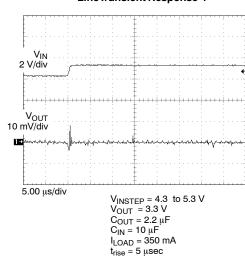
#### **Load Transient Response-1**



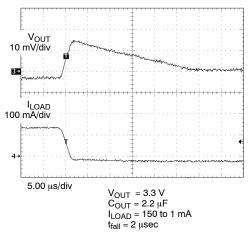
#### **Load Transient Response-3**



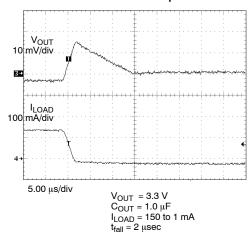
#### LineTransient Response-1



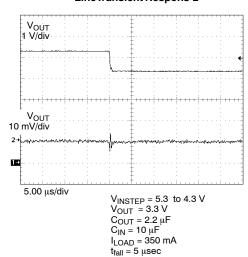
#### **Load Transient Response-2**



#### **Load Transient Response-4**

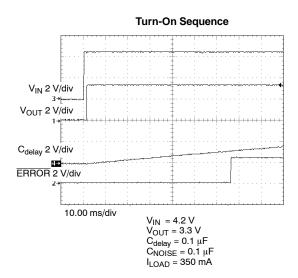


#### LineTransient Respons-2

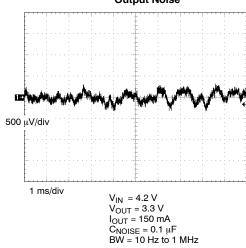




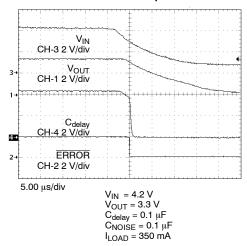
#### **TYPICAL WAVEFORMS**



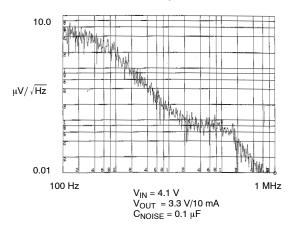
#### **Output Noise**



#### **Turn-Off Sequence**



#### Noise Spectrum





#### **BLOCK DIAGRAMS**

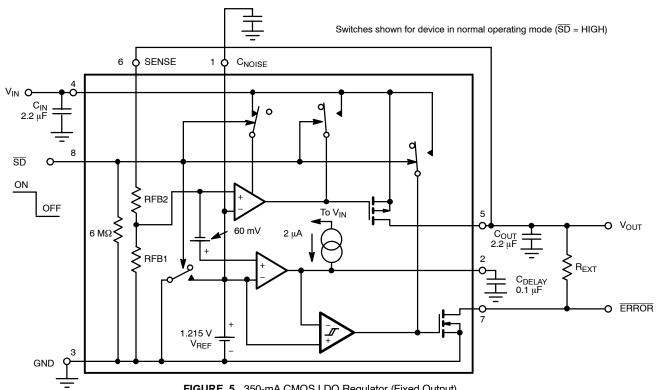
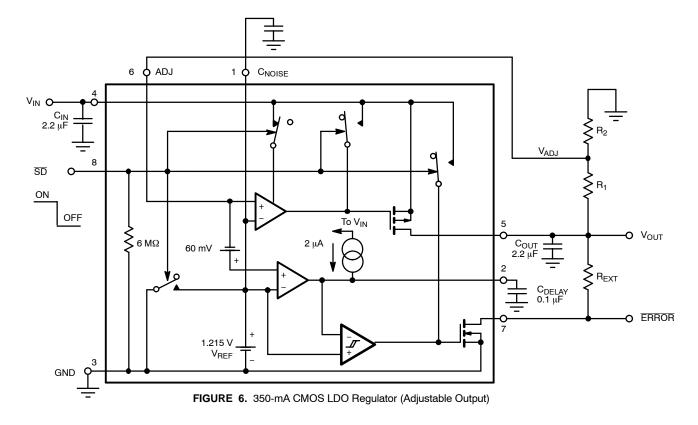


FIGURE 5. 350-mA CMOS LDO Regulator (Fixed Output)



# VISHAY.

#### **DETAILED DESCRIPTION**

The Si9181 is a low drop out, low quiescent current, and very linear regulator family with very fast transient response. It is primarily designed for battery powered applications where battery run time is at a premium. The low quiescent current allows extended standby time while low drop out voltage enables the system to fully utilize battery power before recharge. The Si9181 is a very fast regulator with bandwidth exceeding 50 kHz while maintaining low quiescent current at light load conditions. With this bandwidth, the Si9181 is the fastest LDO available today. The Si9181 is stable with any output capacitor type from 1  $\mu\text{F}$  to 10.0  $\mu\text{F}$ . However, X5R or X7R ceramic capacitors are recommended for best output noise and transient performance.

#### $V_{IN}$

 $V_{IN}$  is the input supply pin. The bypass capacitor for this pin is not critical as long as the input supply has low enough source impedance. For practical circuits, a 1.0- $\mu$ F or larger ceramic capacitor is recommended. When the source impedance is not low enough and/or the source is several inches from the Si9181, then a larger input bypass capacitor is needed. It is required that the equivalent impedance (source impedance, wire, and trace impedance in parallel with input bypass capacitor impedance) must be smaller than the input impedance of the Si9181 for stable operation. When the source impedance, wire, and trace impedance are unknown, it is recommended that an input bypass capacitor be used of a value that is equal to or greater than the output capacitor.

#### **V**OUT

 $V_{OUT}$  is the output voltage of the regulator. Connect a bypass capacitor from  $V_{OUT}$  to ground. The output capacitor can be any value from 1.0  $\mu F$  to 10.0  $\mu F$ . A ceramic capacitor with X5R or X7R dielectric type is recommended for best output noise, line transient, and load transient performance.

#### GND

Ground is the common ground connection for  $V_{IN}$  and  $V_{OUT}$ . It is also the local ground connection for  $C_{NOISE}$ , DELAY, SENSE or ADJ, and  $\overline{SD}$ .

#### **SENSE or ADJ**

SENSE is used to sense the output voltage. Connect SENSE to  $V_{OUT}$  for the fixed voltage version. For the adjustable output version, use a resistor divider R1 and R2, connect R1 from  $V_{OUT}$  to ADJ and R2 from ADJ to ground. R2 should be in the 25-k $\Omega$  to 150-k $\Omega$  range for low power consumption, while maintaining adequate noise immunity.

The formula below calculates the value of R1, given the desired output voltage and the R2 value,

$$R1 = \frac{\left(V_{OUT} - V_{ADJ}\right)R2}{V_{ADJ}}$$

$$V_{ADJ} \text{ is nominally 1.215 V.} \tag{1}$$

#### SHUTDOWN (SD)

 $\overline{SD}$  controls the turning on and off of the Si9181.  $V_{OUT}$  is guaranteed to be on when the  $\overline{SD}$  pin voltage equals or is greater than 1.5 V.  $V_{OUT}$  is guaranteed to be off when the  $\overline{SD}$  pin voltage equals or is less than 0.4 V. During shutdown mode, the Si9181 will draw less than 2- $\mu$ A current from the source. To automatically turn on  $V_{OUT}$  whenever the input is applied, tie the  $\overline{SD}$  pin to  $V_{IN}$ .

#### **ERROR**

 $\overline{\text{ERROR}}$  is an open drain output that goes low when  $V_{\text{OUT}}$  is less than 5% of its normal value. As with any open drain output, an external pull up resistor is needed. When a capacitor is connected from DELAY to GROUND, the error signal transition from low to high is delayed (see Delay section). This delayed error signal can be used as the power-on reset signal for the application system. (Refer to Figure 4.)

The ERROR pin is disconnected if not used.

#### **DELAY**

A capacitor from DELAY to GROUND sets the time delay for ERROR going from low to high state. The time delay can be calculated using the following formula:

$$T_{\text{delay}} = \frac{(V_{\text{ADJ}})C_{\text{delay}}}{I_{\text{delay}}}$$
 (2)

The DELAY pin should be an open circuit if not used.

#### CNOISE

For low noise application, connect a high frequency ceramic capacitor from  $C_{NOISE}$  to ground. A 0.01- $\mu F$  or a 0.1- $\mu F$  X5R or X7R is recommended.

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