ABSOLUTE MAXIMUM RATINGS

V _{DD} to AGND, DGND	0.3V to +6V
AGND to DGND	0.3V to +0.3V
REF, OUT to AGND	0.3V to $(V_{DD} + 0.3V)$
CLR, LDAC, SCLK, DIN, CS to DGND	$00.3V$ to $(V_{DD} + 0.3V)$
Maximum Current into Any Pin	50mA
Continuous Power Dissipation (TA = -	+70°C)
10-Pin µMAX (derate 5.6mW/°C ab	ove +70°C)444.4mW

Operating Temperature Ranges	
MAX520_CUB	0°C to +70°C
MAX520_EUB	40°C to +105°C
Junction Temperature	+150°C
Storage Temperature Range	60°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS—MAX5200/MAX5201

 $(V_{DD} = +4.75V \text{ to } +5.25V, f_{SCLK} = 10 \text{MHz} (50\% \text{ duty cycle}), \text{ output load} = 10 \text{k}\Omega \text{ in parallel with } 250 \text{pF}, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted}.$ Typical values are at $T_A = +25^{\circ}\text{C.})$

PARAMETER	SYMBOL	CONDITIONS		TYP	MAX	UNITS	
STATIC PERFORMANCE (Note 1)							
Resolution	N		16			Bits	
		MAX520_AEUB		±10	±20		
Integral Nonlinearity (Note 2)	INL	MAX520_ACUB		±10	±20	LSB	
		MAX520_BEUB		±20	±40		
Differential New York and the		MAX520_A_UB (Note 3)			±1		
Differential Nonlinearity (Note 2)	DNL	MAX520_BEUB (0°C to +105°C) (Note 3)			±1	LSB	
(14010-2)		MAX520_BEUB (-40°C to 0°C)			±2		
Offset Error		Inferred from measurement at 1C00 hex and FFFF hex		±3	±25	mV	
Gain Error	GE	Within DAC output range (Note 4)		±0.01	±1	%FSR	
Power-Supply Rejection	PSR	V _{DD} = 5V ±5%, midscale input		±0.06	±0.5	mV/V	
DYNAMIC PERFORMANCE							
DAC Output Range		(Note 2)		0 to V _{DD}		V	
Output-Voltage Slew Rate	SR			0.6		V/µs	
Output Settling Time		To ±1LSB of FS, VSTEP = 0.25 × VREF to 0.75 × VREF		25		μs	
Output Noise		DAC code = 8400 hex, 10kHz		175		nV/√ Hz	
DAC Glitch Impulse		Major carry transition (code 7FFF hex to code 8000 hex)		10		nV⋅s	
Digital Feedthrough				10		nV⋅s	
Wake-Up Time		From software shutdown to 90% of output code = FFFF hex, C _{REF} = 0.1µF		50		μs	
Power-Up Time		From power applied to 90% of output code = FFFF hex		10		ms	

ELECTRICAL CHARACTERISTICS—MAX5200/MAX5201 (continued)

 $(V_{DD} = +4.75V \text{ to } +5.25V, f_{SCLK} = 10 \text{MHz} (50\% \text{ duty cycle}), \text{ output load} = 10 \text{k}\Omega \text{ in parallel with } 250 \text{pF}, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $T_A = +25^{\circ}\text{C.}$)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS			
INTERNAL REFERENCE	INTERNAL REFERENCE								
V _{REF} Output Voltage		T _A = +25°C	2.48	2.5	2.52	V			
V Tomoroso		$T_A = 0$ °C to +70°C		±15		10 C			
V _{REF} Tempco		$T_A = -40^{\circ}C \text{ to } +105^{\circ}C$		±20		ppm/°C			
DIGITAL INPUTS (DIN, SCLK, \overline{CS} ,	CLR, LDAC	5)							
Input High Voltage	VIH		2.4			V			
Input Low Voltage	VIL				0.8	V			
Input Hysteresis	VHYST			200		mV			
Input Leakage	I _{IN}	Digital inputs = 0 or V _{DD}			±1	μΑ			
Input Capacitance	CIN			15		рF			
POWER REQUIREMENTS									
Positive Power Supply	V_{DD}		4.75		5.25	V			
Positive Supply Current	I_{DD}	All digital inputs at 0 or V _{DD} (Note 5)		0.8	1.5	mA			
Shutdown Supply Current	ISHDN	All digital inputs at 0 or V _{DD}		1	10	μΑ			
TIMING CHARACTERISTICS									
SCLK Frequency	fsclk				10	MHz			
SCLK Clock Period	tcp		100			ns			
SCLK Pulse Width High	t _{CH}		40			ns			
SCLK Pulse Width Low	tCL		40			ns			
DIN Setup Time	t _{DS}		40			ns			
DIN Hold Time	tDH		0			ns			
CS Fall to SCLK Rise Setup Time	tcss		40			ns			
SCLK Rise to CS Rise Hold Time	tcsh		0			ns			
SCLK Rise to CS Fall Ignore	tcso		10			ns			
CS Rise to SCLK Rise Ignore	tcs1		40			ns			
LDAC Pulse Width	tLDAC		40			ns			
CS Rise to LDAC Low Setup	t _{LDACS}		40			ns			
SCLK Fall to CS Fall Ignore	tcsol		10			ns			
CS Pulse Width Low for Shutdown	tcswl		40			ns			
CS Pulse Width High	tcswh		100			ns			



ELECTRICAL CHARACTERISTICS—MAX5202/MAX5203

 $(V_{DD} = +2.7V \text{ to } +3.6V, \text{ f}_{SCLK} = 10 \text{MHz} (50\% \text{ duty cycle}), \text{ output load} = 10 \text{k}\Omega \text{ in parallel with } 250 \text{pF}, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25 ^{\circ}\text{C.})$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
STATIC PERFORMANCE (Note	1)					
Resolution	N		16			Bits
		MAX520_AEUB		±10	±20	
Integral Nonlinearity (Note 2)	INL	MAX520_ACUB		±10	±20	LSB
		MAX520_BEUB		±20	±40	
D''' I' IN I'		MAX520_A_UB (Note 3)			±1	
Differential Nonlinearity (Note 2)	DNL	MAX520_BEUB (0°C to +105°C) (Note 3)			±1	LSB
(14010-2)		MAX520_BEUB (-40°C to 0°C)			±2	
Offset Error		Inferred from measurement at 3800 hex and FFFF hex		±3	±25	mV
Gain Error	GE	Within DAC output range (Note 4)		±0.01	±1.0	%FSR
Power-Supply Rejection	PSR	V _{DD} = 3V ±10%, midscale input		±0.06	±0.5	mV/V
DYNAMIC PERFORMANCE						
DAC Output Range		(Note 2)		0 to V _{DD}		V
Voltage-Output Slew Rate	SR			0.6		V/µs
Output Settling Time		To ±1 LSB of FS, VSTEP = 0.25 × VREF to 0.75 × VREF		25		μs
Output Noise		Code = 8400 hex, 10kHz		175		nV/√Hz
Reference Feedthrough		Code = 0000 hex at 100kHz, V _{REF} = 1V _{P-P}		1		mV _{P-P}
DAC Glitch Impulse		Major carry transition (code 7FFF hex to code 8000 hex)		10		nV⋅s
Digital Feedthrough		Code = 0000 hex; $\overline{\text{CS}}$ = V _{DD} ; $\overline{\text{LDAC}}$ = 0; SCLK, DIN = 0 or V _{DD} levels		10		nV⋅s
Wake-Up Time		From software shutdown to 90% of output code = FFFF hex		50		μs
Power-Up Time		From power applied to 90% of output code = FFFF hex		10		ms

__ /N/IXI/M

ELECTRICAL CHARACTERISTICS—MAX5202/MAX5203 (continued)

 $(V_{DD} = +2.7V \text{ to } +3.6V, f_{SCLK} = 10 \text{MHz} (50\% \text{ duty cycle}), \text{ output load} = 10 \text{k}\Omega \text{ in parallel with } 250 \text{pF}, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25 ^{\circ}\text{C.})$

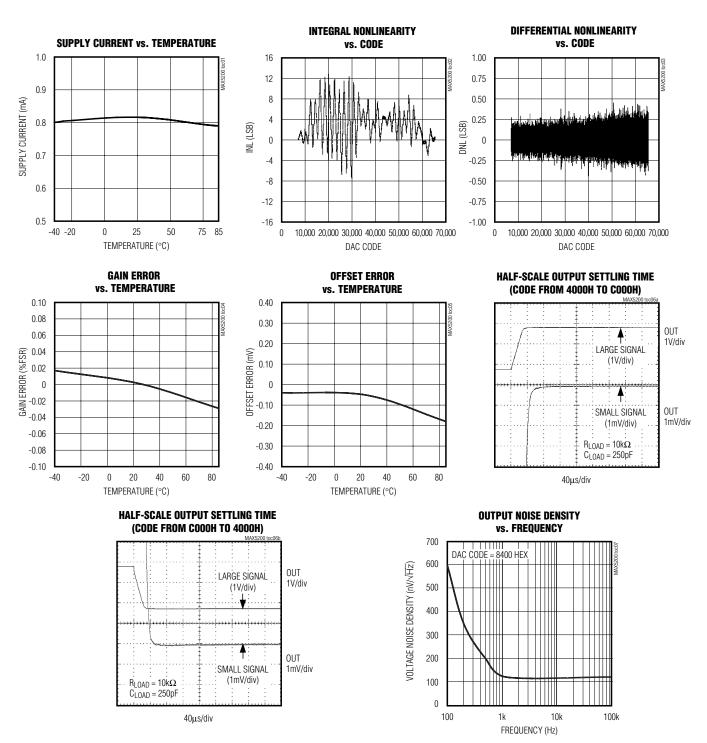
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
INTERNAL REFERENCE						
V _{REF} Output Voltage		$T_A = +25^{\circ}C$	1.46	1.5	1.54	V
		$T_A = 0$ °C to +70°C		±15		20m/0C
V _{REF} Tempco		$T_A = -40^{\circ}C$ to $+105^{\circ}C$		±20		ppm/°C
DIGITAL INPUTS (DIN, SCLK, CS	, CLR, LDAC	5)				
Input High Voltage	V _{IH}		2.1			V
Input Low Voltage	VIL				0.6	V
Input Hysteresis	V _{HYST}			200		mV
Input Leakage	I _{IN}	Digital inputs = 0 or V _{DD}			±1	μΑ
Input Capacitance	CIN			15		рF
POWER REQUIREMENTS						
Positive Power Supply	V_{DD}		2.7		3.6	V
Positive Supply Current	I _{DD}	All digital inputs at 0 or V _{DD} (Note 5)		0.8	1.5	mA
Shutdown Supply Current	ISHDN	All digital inputs at 0 or V _{DD}		1	10	μΑ
TIMING CHARACTERISTICS						
SCLK Frequency	fsclk				10	MHz
SCLK Clock Period	tCP		100			ns
SCLK Pulse Width High	tch		40			ns
SCLK Pulse Width Low	tcL		40			ns
DIN Setup Time	tDS		40			ns
DIN Hold Time	tDH		0			ns
CS Fall to SCLK Rise Setup Time	tcss		40			ns
SCLK Rise to CS Rise Hold Time	tcsh		0			ns
SCLK Rise to CS Fall Ignore	tcso		10			ns
CS Rise to SCLK Rise Ignore	tcs1		40			ns
LDAC Pulse Width	tLDAC		40			ns
CS Rise to LDAC Low Setup	tLDACS		40			ns
SCLK Fall to CS Fall Ignore	tcsol		10			ns
CS Pulse Width Low for Shutdown	tcswl		40			ns
CS Pulse Width High	tcswh		100			ns

- Note 1: Static performance tested at $V_{DD} = +5.0V$ (MAX5200/MAX5201) and at $V_{DD} = +3.0V$ (MAX5202/MAX5203).
- Note 2: INL and DNL are guaranteed for outputs between 0.5V to (V_{DD} 0.5V).
- Note 3: Guaranteed monotonic.
- **Note 4:** $V_{REF} = 2.5V$ (MAX5200/MAX5201) and $V_{REF} = 1.5V$ (MAX5202/MAX5203).
- **Note 5:** $R_L = \infty$, digital inputs are at V_{DD} or DGND.



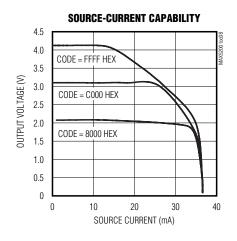
Typical Operating Characteristics

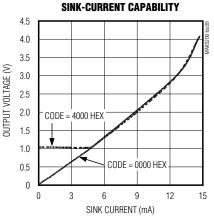
 $(V_{DD} = +5V, T_A = +25^{\circ}C, unless otherwise noted.)$

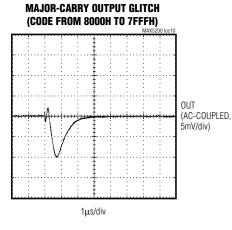


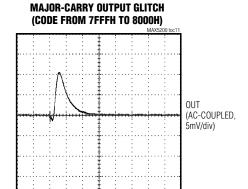
_Typical Operating Characteristics (continued)

 $(V_{DD} = +5V, T_A = +25^{\circ}C, unless otherwise noted.)$

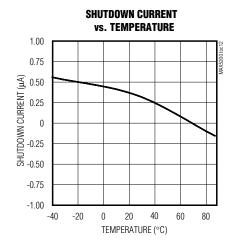








1μs/div



Pin Description

PIN	NAME	FUNCTION
1	CLR	Reset DAC Active-Low Input. Pull $\overline{\text{CLR}}$ low to reset the DAC output to midscale output (8000 hex) for MAX5200/MAX5202 and to zero-scale output (0000 hex) for MAX5201/MAX5203. For normal operation, connect $\overline{\text{CLR}}$ to V_{DD} .
2	REF	Reference Voltage Output. Provides a +2.5V (MAX5200/MAX5201) or +1.5V (MAX5202/MAX5203) nominal output. For improved noise performance, bypass with a minimum 0.1µF capacitor to AGND.
3	AGND	Analog Ground
4	V _{DD}	Positive Supply Voltage. Bypass V _{DD} to AGND with a 10µF capacitor in parallel with a 0.1µF capacitor.
5	OUT	DAC Output Voltage
6	CS	Active-Low Chip-Select Input
7	LDAC	Load DAC Input
8	DIN	Serial Data Input
9	SCLK	Serial Clock Input. Duty cycle must be 40% to 60%.
10	DGND	Digital Ground

Detailed Description

The MAX5200–MAX5203 serial 16-bit, voltage-output DACs are easily configured with a 3-wire serial interface. These devices offer full 16-bit performance with less than ± 20 LSB integral linearity error and less than ± 1 LSB differential linearity error, thus ensuring monotonic performance. Serial data transfer minimizes the number of package pins required. The MAX5200–MAX5203 include control-logic circuitry, a 16-bit data-in shift register, and a DAC register. In addition, these devices employ a precision-bandgap reference and trimmed internal resistors to produce a gain of 2V/V, maximizing the output voltage swing. The MAX5200–MAX5203 output is buffered and the full-scale output voltage is 2 × VREF.

The MAX5200–MAX5203 feature a hardware reset input (CLR) that, when pulled low, clears the DAC output to zero code 0000H (MAX5201/MAX5203) or resets the DAC output to midscale code 8000 hex (MAX5200/MAX5202). For normal operation, connect CLR to VDD.

Internal Reference

The MAX5200/MAX5201 (+5V supply) include an internal reference of 2.5V while the MAX5202/MAX5203 (+3V supply) include an internal reference of 1.5V. The DAC output range is from 0 to $2 \times V_{REF}$. Do not drive external circuitry from this reference. To improve DAC output noise performance, bypass with a low leakage 0.1µF minimum capacitor to AGND.

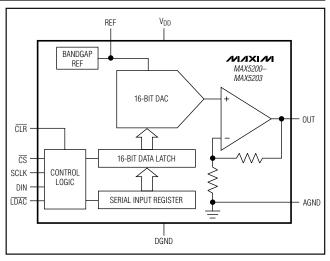


Figure 1. MAX5200-MAX5203 Simplified Functional Diagram

Digital Interface

The MAX5200-MAX5203 digital interface is a standard 3-wire connection compatible with SPI/QSPI/MICROWIRE and most DSP interfaces. All of the digital input pins (\overline{CS} , SCLK, DIN, \overline{CLR} , and \overline{LDAC}) are TTL compatible. SCLK can accept clock frequencies as high as 10MHz for a +5V supply and 10MHz for a +3V or +3.3V supply.

One of two methods can be used when interfacing and updating the MAX5200–MAX5203. The first requires three digital inputs: $\overline{\text{CS}}$, DIN, and SCLK (Figure 2). The active-low chip-select input ($\overline{\text{CS}}$) enables the serial

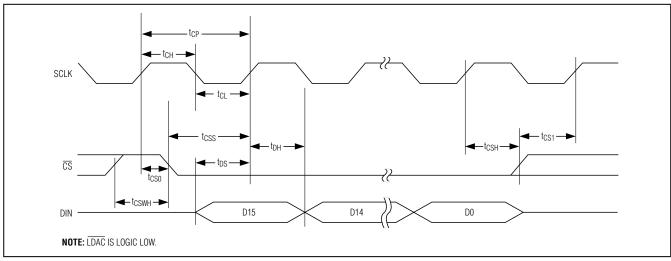


Figure 2. 3-Wire Interface Timing Diagram

data loading at the data input (DIN). Pull \overline{CS} low and clock in each bit of the 16-bit digital word on the rising edge of the serial clock (SCLK). Two 8-bit bytes can be used, and do not require any additional time between them. Pulling \overline{CS} high after loading the 16-bit word transfers that code into the DAC register and then updates the output. If \overline{CS} is not kept low during the entire loading of the 16-bit word, data is corrupted. In this case, a new 16-bit word must be loaded. \overline{LDAC} must be kept low at all times for the above instructions.

An alternate method of interfacing and updating the MAX5200–MAX5203 can be done with a fourth digital input, the active-low load DAC ($\overline{\text{LDAC}}$). $\overline{\text{LDAC}}$ allows the output to update asynchronously after $\overline{\text{CS}}$ goes high. It is useful when updating multiple MAX5200–MAX5203s synchronously when sharing a single $\overline{\text{LDAC}}$ and $\overline{\text{CS}}$ line. $\overline{\text{LDAC}}$ must be kept high at all times during the data-loading sequence and must only be asserted when $\overline{\text{CS}}$ is high. Asserting $\overline{\text{LDAC}}$ when $\overline{\text{CS}}$ is low can cause corrupted data. To operate the MAX5200–MAX5203 using $\overline{\text{LDAC}}$, pull $\overline{\text{LDAC}}$ high, pull $\overline{\text{CS}}$ low, load the 16-bit word as described in the previous paragraph, and pull $\overline{\text{CS}}$ high again. Following these commands, the DAC output only updates when $\overline{\text{LDAC}}$ is asserted low (Figure 3).

Shutdown Mode

The low-power shutdown mode reduces supply current to typically $1\mu A$ and a maximum of $10\mu A$. Shutdown mode is not activated through command words, as is common among D/A converters. These devices require careful manipulation of \overline{CS} and SCLK (Figure 4).

Shutting Down

To shut down the MAX5200–MAX5203, change the state of SCLK (either a high to low or low to high transition can be used) and pulse two falling $\overline{\text{CS}}$ edges. In order to keep the device in shutdown mode, SCLK **must** not change state. SCLK must remain in the state it is in after the two $\overline{\text{CS}}$ pulses.

Waking Up

There are two methods to wake up the MAX5200–MAX5203. Pulse one falling \overline{CS} edge or transition SCLK. It takes 50µs typically from the \overline{CS} falling edge or SCLK transition for the DAC to return to normal operation.

Power-On Reset

The MAX5200–MAX5203 have a power-on reset circuit to set the DAC's output to a known state when V_{DD} is first applied. The MAX5200/MAX5202 reset to midscale (code 8000 hex) upon power-up. The MAX5201/MAX5203 reset to zero scale (code 0000 hex) upon power-up. This ensures that unwanted output voltages do not occur immediately following a system power-up, such as a loss of power. It is required to apply V_{DD} first before any other inputs (DIN, SCLK, CLR, LDAC, and CS).

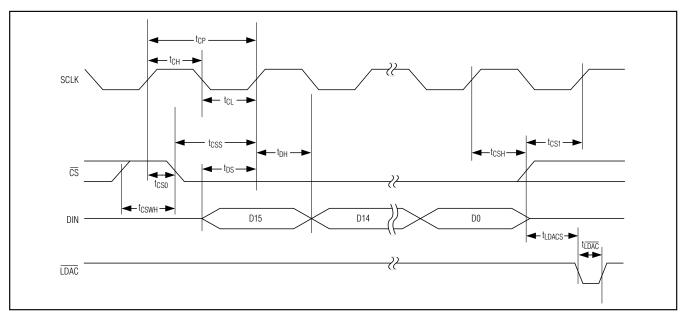


Figure 3. 4-Wire Interface Timing Diagram

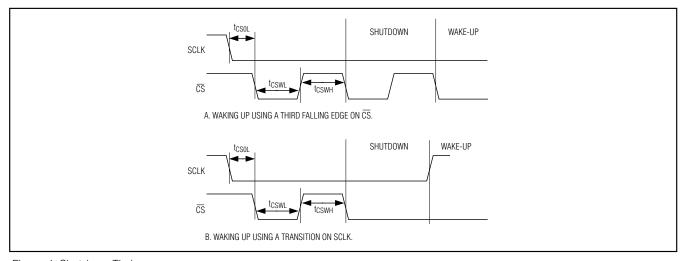


Figure 4. Shutdown Timing

Applications Information

Power-Supply and Bypassing Considerations

Bypass the power supply with a $10\mu F$ capacitor in parallel with a $0.1\mu F$ capacitor to AGND. Minimize lead lengths to reduce lead inductance. If noise becomes an issue, use shielding and/or ferrite beads to increase isolation.

Output Buffer

The MAX5200-MAX5203 include low-offset, low-noise buffers enabling the output to source 15mA or sink 5mA. The output buffer operates at a slew rate of 0.6V/µs. With a 1/4 FS to 3/4 FS output transition, the buffer output typically settles to 1 LSB in less than 25µs. The MAX5200-MAX5203 output buffers provide a low 0.2 Ω typical output impedance. The MAX5200-MAX5203 buffer amplifiers typically produce 175nV/ $\sqrt{\rm Hz}$ noise at 10kHz.

10 _______/N/XI/VI

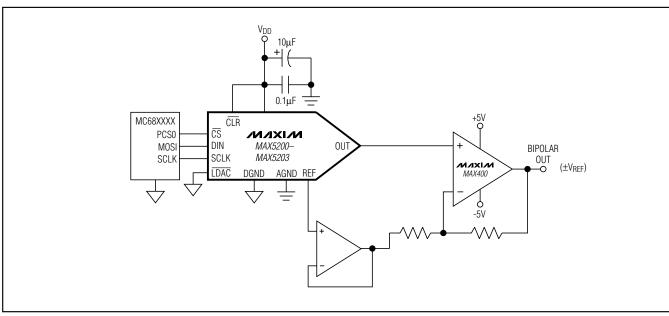


Figure 5. MAX5200-MAX5203 Typical Operating Circuit—Bipolar Output

Table 1. Bipolar Code Table

DAC LATCH CO	NTENTS	ANALOG OUTPUT, VOUT
MSB	LSB	ANALOG OUTPUT, VOUT
1111 1111 11	11 1111	+V _{REF} × (32,767 / 32,768)
1000 0000 000	00 0001	+V _{REF} × (1 / 32,768)
1000 0000 000	0000 0000	ov
0111 1111 11	11 1111	-V _{REF} × (1 / 32,768)
0000 0000 000	0000 0000	-V _{REF} × (32,768 / 32,768)

Bipolar Configuration

The MAX5200–MAX5203 are designed for unipolar operation, but can be used in bipolar applications with an external amplifier and resistors. Figure 5 shows the MAX5200–MAX5203 configured for bipolar operation. The op amp is set for unity gain. Table 1 lists the offset binary code for this circuit. The output voltage range is $\pm V_{REF}$.

Layout Considerations

Digital and AC transient signals coupling to AGND can create noise at the output. Connect AGND to the highest quality ground available. Use proper grounding techniques, such as a multilayer board with a lowinductance ground plane. Wire-wrapped boards and sockets are not recommended. For optimum system performance, use printed circuit (PC) boards with separate analog and digital ground planes. Connect the two ground planes together at the low-impedance power-supply source. Connect DGND and AGND pins together at the IC. The best ground connection is achieved by connecting the DAC's DGND and AGND together, and then connecting that point to the system analog ground plane. If the DAC's DGND is connected to the system digital ground, digital noise can get through the DAC's analog portion.

Chip Information

TRANSISTOR COUNT: 8764

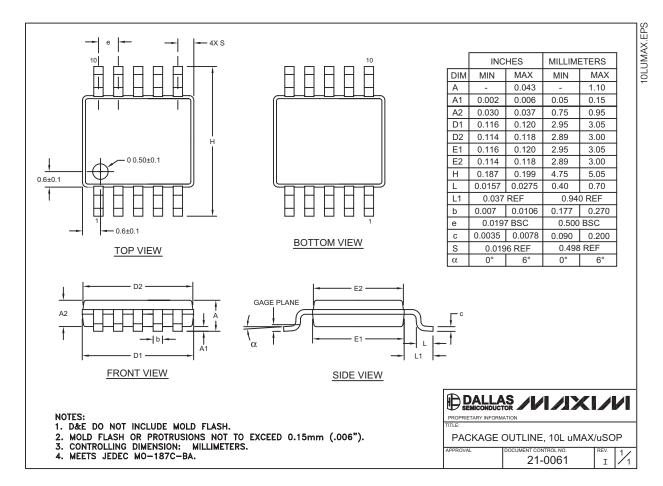
PROCESS: BiCMOS

Selector Guide

PART	INTEGRAL NONLINEARITY (LSB, MAX)	SUPPLY VOLTAGE RANGE (V)	REFERENCE INPUT RANGE (V)	POWER-ON-RESET VALUE
MAX5200AEUB	20	4.75 to 5.25	2.5	Midscale
MAX5200ACUB	20	4.75 to 5.25	2.5	Midscale
MAX5200BEUB	40	4.75 to 5.25	2.5	Midscale
MAX5201AEUB	20	4.75 to 5.25	2.5	Zero
MAX5201ACUB	20	4.75 to 5.25	2.5	Zero
MAX5201BEUB	40	4.75 to 5.25	2.5	Zero
MAX5202AEUB	20	2.7 to 3.6	1.5	Midscale
MAX5202ACUB	20	2.7 to 3.6	1.5	Midscale
MAX5202BEUB	40	2.7 to 3.6	1.5	Midscale
MAX5203AEUB	20	2.7 to 3.6	1.5	Zero
MAX5203ACUB	20	2.7 to 3.6	1.5	Zero
MAX5203BEUB	40	2.7 to 3.6	1.5	Zero

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)



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