MAX4063

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V _{CC} to GND)	0.3V to +6V
Any Other Pin to GND	0.3V to (V _{CC} + 0.3V)
Duration of Short Circuit to GND or V _{CC}	Continuous
Continuous Input Current (any pin)	±10mA
Continuous Power Dissipation ($T_A = +70^\circ$	
14-Pin TSSOP (derate 10.0mW/°C abov	/e +70°C)796.8mW

16-Pin TQFN (derate 25.0mW/°C above	+70°C)2000mW
Operating Temperature Range	40°C to +85°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{CC} = 3V, V_{GND} = 0V, V_{SHDN} = V_{CC}, V_{INT/AUX} = 0V, R_G = 11.11k\Omega, R_L = 100k\Omega$ to 1.5V, $R_{BIAS} = \infty$, V_{OUT} is measured between OUT and \overline{OUT} . T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Notes 1 and 2)

PARAMETER	SYMBOL	CON	NDITIONS	MIN	ТҮР	MAX	UNITS
Supply Voltage Range	V _{CC}	Inferred from PSRR test		2.4		5.5	V
Supply Current	ICC				0.75	1.1	mA
Output Common-Mode Voltage	VOCM			1.25	1.5	1.75	V
Slew Rate	SR	$A_V = 20V/V$			±1		V/µs
Supply Current in Shutdown	ISHDN	$V_{\overline{SHDN}} = 0V$			0.001	1	μA
Output Short-Circuit Current	I _{SC}	To GND			30		mA
		To V _{CC}			30		
DIFFERENTIAL INPUT (VINT/AUX = 0	1	1					
Input Offset Voltage	Vos	$T_A = +25^{\circ}C$			±1		mV
Common-Mode Input Voltage Range	VCM	Inferred from CMRF	R test	1		2	V
Maximum Differential Input Voltage	VDIFFMAX	$A_V = 2V/V$			1		V
Small-Signal Bandwidth	BW-3dB				600		kHz
Input Resistance	RIN	Either differential in	put		100		kΩ
Input Resistance Match	RMATCH				1		%
	e _n	$A_V = 20V/V$, f = 1kHz			70		nV/√Hz
Input Noise-Voltage Density		$A_V = 200V/V, f = 1kHz$		12			
RMS Output Noise Voltage	V _{NRMS}	$A_V = 20V/V$, BW = 22Hz to 22kHz			225		μV _{RMS}
Total Harmonic Distortion Plus Noise	THD+N	$A_V = 20V/V$, f = 1kHz, $V_{OUT} = 0.7V_{RMS}$, BW = 22Hz to 22kHz			0.05		%
	Avdiff	1V < V _{CM} < 2V, VOUT = 0.7V _{RMS}	RG = open		2		
Differential Gain			$R_G = 11.11 k\Omega$	19.2	20	20.8	V/V
			$R_G = 1.01 k\Omega$		200		
Common-Mode Rejection Ratio	CMRR	$V_{CM} = 500 \text{mV}_{P-P}, \text{ f} = 1 \text{kHz}$			70		dB
		$T_A = +25^{\circ}C$			95		
Power-Supply Rejection Ratio	PSRR	$T_A = T_{MIN} - T_{MAX}$		Ì	85		dB
		V _{CC} = 5V ±100mV, f = 1kHz			86		1

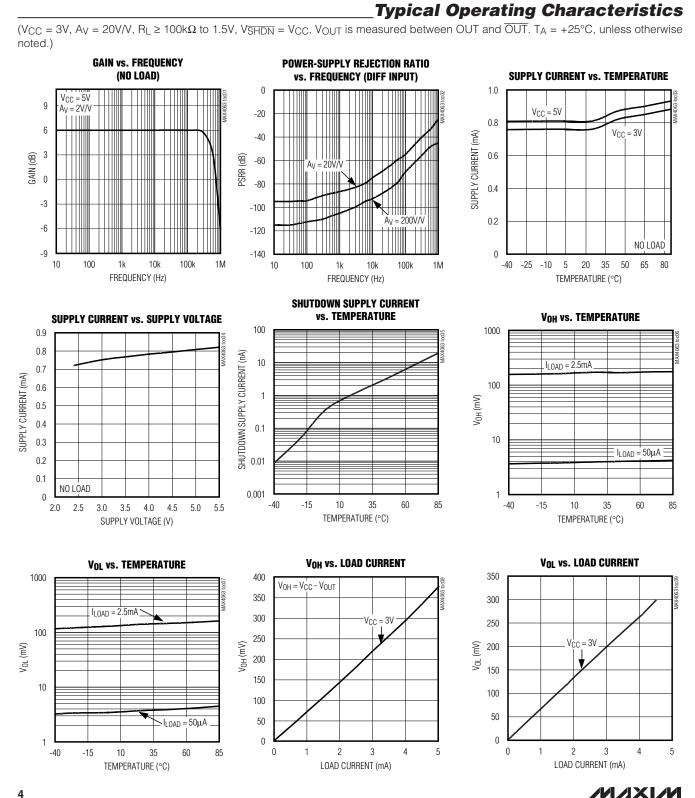
ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = 3V, V_{GND} = 0V, V_{\overline{SHDN}} = V_{CC}, V_{\overline{INT}/AUX} = 0V, R_G = 11.11k\Omega, R_L = 100k\Omega$ to 1.5V, $R_{BIAS} = \infty$, V_{OUT} is measured between OUT and \overline{OUT} . T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Notes 1 and 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
AUXILIARY INPUT (INT/AUX = V _{CC})			•			
Small-Signal Bandwidth	BW-3dB			230		kHz
Input Resistance	R _{IN}			100		kΩ
Input Noise-Voltage Density	en	f = 1kHz		200		nV/√Hz
RMS Output Noise Voltage	V _{NRMS}	BW = 22Hz to $22kHz$		620		μV _{RMS}
Total Harmonic Distortion Plus Noise	THD+N	f = 1kHz, BW = 22Hz to 22kHz		0.007		%
Power-Supply Rejection Ratio	PSRR	$T_A = +25^{\circ}C$	80	100		dB
Fower-Supply Rejection Ratio	ronn	T _A = T _{MIN} - T _{MAX}	72			
Voltage Gain	AVAUX	$V_{OUT} = 0.7 V_{RMS}$	-19.5	-20	-20.5	V/V
BIAS OUTPUT						
Output Voltage	Vout	$I_{BIAS} = 0.5 mA$ to GND	2	2.2		V
Output Resistance	Rout	$I_{BIAS} = 0.5 mA$ to GND		16	30	Ω
Output Noise Voltage	V _{NRMS}	$I_{BIAS} = 0.5$ mA to GND, BW = 22Hz to 22kHz		20		μV _{RMS}
		$I_{BIAS} = 0.5$ mA to GND, $V_{CC} = 2.4$ V to 5.5V	60	74		
Power-Supply Rejection Ratio	PSRR	$I_{BIAS} = 0.5$ mA, $V_{CC} = 3V + 100$ mV _{P-P} , f = 1kHz		71		dB
DIGITAL INPUTS (SHDN and INT/AU	X)					
Input Leakage Current	l _{IN}	$V_{IN} = 0V \text{ or } V_{CC}$			±1	μΑ
Input Voltage High	VINH		0.7 × V	СС		V
Input Voltage Low	V _{INL}				$0.3 \times V_{CC}$	V
Shutdown Enable Time	ton			10		μs
Shutdown Disable Time	toff			10		μs

Note 1: All specifications are 100% tested at $T_A = +25$ °C. Specification limits over temperature ($T_A = T_{MIN}$ to T_{MAX}) are guaranteed by design, not production tested.

Note 2: MAX4063 requires a 1µF capacitor from BIAS to ground and a 10pF capacitor from ADJ to OUT.



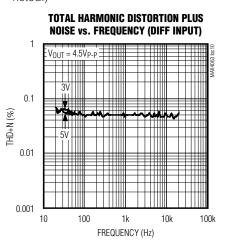
MAX4063

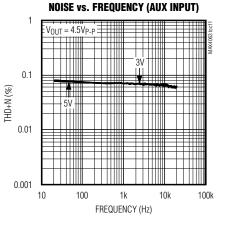
Downloaded from Arrow.com.

_Typical Operating Characteristics (continued)

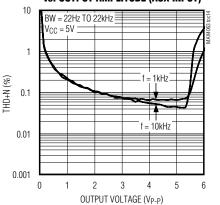
 $(V_{CC} = 3V, A_V = 20V/V, R_L \ge 100k\Omega$ to 1.5V, $V_{\overline{SHDN}} = V_{CC}$, V_{OUT} is measured between OUT and \overline{OUT} . $T_A = +25^{\circ}C$, unless otherwise noted.)

TOTAL HARMONIC DISTORTION PLUS

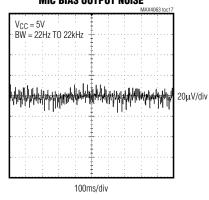




TOTAL HARMONIC DISTORTION PLUS NOISE vs. Output Amplitude (AUX Input)

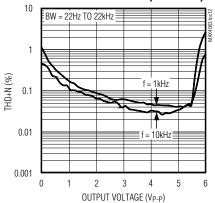


MIC BIAS OUTPUT NOISE

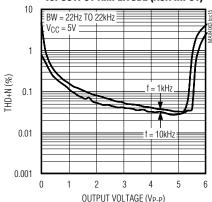


TOTAL HARMONIC DISTORTION PLUS NOISE vs. Output Amplitude (DIFF Input)

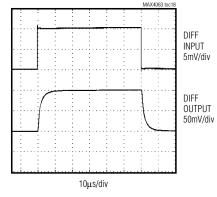
MAX4063



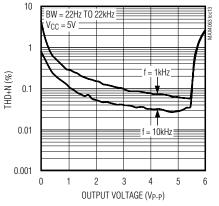
TOTAL HARMONIC DISTORTION PLUS NOISE vs. Output Amplitude (AUX Input)



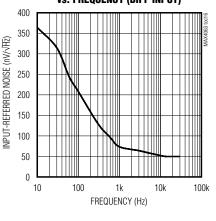
SMALL-SIGNAL TRANSIENT RESPONSE FOR DIFF INPUT



TOTAL HARMONIC DISTORTION PLUS NOISE vs. Output Amplitude (DIFF Input)



INPUT-REFERRED NOISE vs. FREQUENCY (DIFF INPUT)



<u>///XI///</u>

Downloaded from Arrow.com.

Typical Operating Characteristics (continued)

MAX4063 toc20

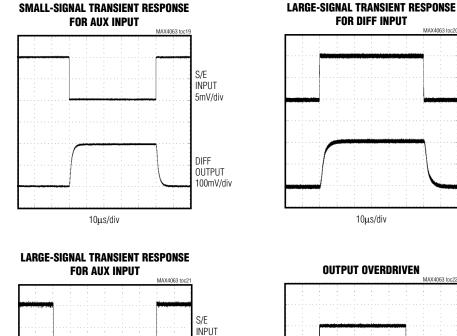
DIFF

INPUT 100mV/div

DIFF OUTPUT

2V/div

 $(V_{CC} = 3V, A_V = 20V/V, R_L \ge 100k\Omega$ to 1.5V, $V_{SHDN} = V_{CC}, V_{OUT}$ is measured between OUT and \overline{OUT} . $T_A = +25^{\circ}C$, unless otherwise noted.)



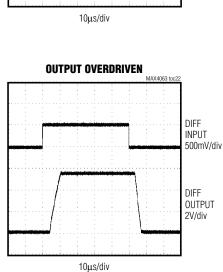
100mV/div

DIFF

OUTPUT

2V/div

10µs/div



MAX4063

6

Pin Description

PIN			FUNCTION
TSSOP	TQFN	NAME	FUNCTION
1	15	G2	Gain-Selectable Input. Connect an external resistor between G1 and G2 to set the gain for the differential amplifier (see <i>Adjustable Differential Gain-Setting</i> section).
2	16	ĪNT/AUX	Internal (Differential) or Auxiliary (Single-Ended) Input Select. Drive INT/AUX low to select differential in or high to select auxiliary in.
3	1	SHDN	Shutdown Input. Drive SHDN high for normal operation. Drive SHDN low for shutdown mode.
4	2	OUT	Amplifier Output. OUT is high impedance when in shutdown mode.
5	3	BIAS	External Electret Microphone Capsule Bias Output. Bypass BIAS with a $1\mu\text{F}$ capacitor to ground.
6	4	VCC	Power Supply. Bypass the V_{CC} to GND with a 0.1µF capacitor.
7	5	ADJ	Adjustable Gain Select for AUX_IN (see <i>Auxiliary Input-Gain Adjustment</i> section). Connect a 10pF capacitor between OUT and ADJ.
8	7	OUT	Complementary Amplifier Output. OUT is high impedance in shutdown mode.
9	9	IN+	Noninverting Differential Amplifier Input. AC-couple the audio signal into IN+.
10	10	IN-	Inverting Differential Amplifier Input. AC-couple the audio signal into IN
11	11	GND	Ground
12	12	AUX_IN	Single-Ended Input for Auxiliary Microphone. AC-couple the audio signal into AUX_IN.
13	6, 8, 14	N.C.	No Connection. Not internally connected.
14	13	G1	Gain-Selectable Input. Connect an external resistor between G1 and G2 to set the gain for the differential amplifier.
		EP	Exposed Pad (TQFN Only). Internally connected to GND. Connect to a large ground plane to maximize thermal performance. Not intended as an electrical connection point.

Detailed Description

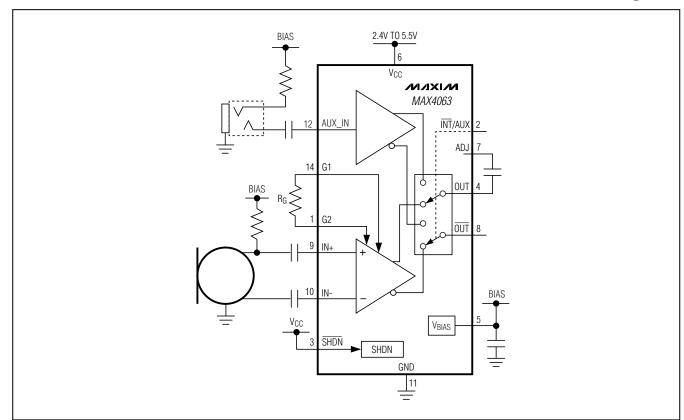
The MAX4063 is a differential microphone preamplifier providing high-quality amplification, optimized for use in computer and mobile applications. This device features adjustable gain, very high power-supply rejection (95dB), and common-mode rejection (79dB), making it ideal for low-noise applications. The MAX4063 provides a differential input stage, making the device particularly effective when layout constraints force the microphone amplifier to be physically remote from the ECM microphone.

The MAX4063 is capable of switching its output between the differential input and an inverting singleended input. INT/AUX selects either the differential input or single-ended auxiliary input. In addition, the MAX4063 has an integrated microphone bias source, simplifying system design and eliminating the need for external components. The MAX4063 has a complementary output allowing CODECs and other devices with differential inputs to be optimally driven (see *Functional Diagram*). The MAX4063 includes a 0.3µA shutdown mode for ultimate power savings. The differential gain of the MAX4063 is set with a single resistor connected between the G1 and G2 pins. The MAX4063 has an internal default gain of 20V/V on the AUX_IN input. The AUX_IN gain can be increased with a single external resistor (see the *Differential-Gain Adjustment and Auxiliary Input-Gain Adjustment* sections).

Differential Input

The main microphone amplifier is a low-noise, differential input structure. This is an almost essential element when faced with amplification of low-amplitude analog signals in digitally intense environments such as note-

Functional Diagram



book PCs or PDAs. Used correctly, the advantages over a single-ended solution are:

- Better power-supply noise rejection.
- Less degradation from noise in PC board ground planes.
- The microphone and preamplifier may be placed physically further apart, easing PC board layout restrictions.

Differential-Gain Adjustment

The MAX4063 allows the user to alter the differential gain to optimize the signal-to-noise ratio (SNR) of their system. The gain is set by a single external resistor (R_G) connected between the G1 and G2 pins:

$$R_{G} = \frac{200k\Omega}{A_{VD} - 2}$$

where A_V is the required voltage gain.

Hence, an 11.11k Ω resistor yields a gain of 20V/V, or 26dB. Leaving the pins unconnected results in a gain of 2V/V. Gain is defined as:

For differential out:

$$A_{VD} = \frac{V_{OUT} - V_{OUT}}{V_{IN+} - V_{IN-}}$$

The resistor can be either fixed or variable, allowing the use of a digitally controlled potentiometer to alter the gain under software control.

Auxiliary Input-Gain Adjustment

The MAX4063 provides an option to increase the AUX_IN (see Figure 3). To increase the gain, connect resistor R_{ADJ} between the ADJ and AUX_IN pins. R_{ADJ} is calculated from the following formula:

$$R_{ADJ} = \frac{2M\Omega}{AV_{AUX} - 20}$$
 (to increase the gain)



8

MAX4063

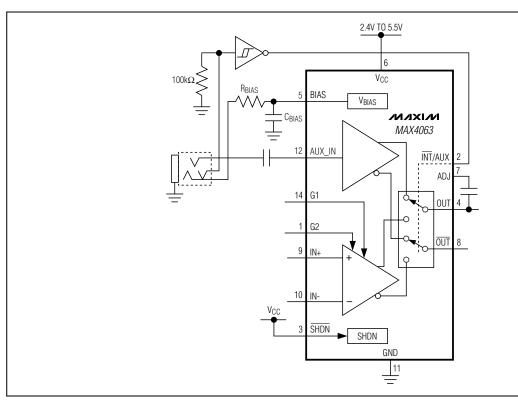


Figure 1. MAX4063 with Auxiliary Input Configuration

where:

$$AV_{AUX} = \frac{V_{OUT} - \overline{V_{OUT}}}{V_{AUX \ IN}}$$

RADJ is placed between AUX_IN and ADJ.

Input Capacitors

The two differential microphone inputs and the singleended auxiliary input of the MAX4063 have on-chip bias components, allowing the user to AC-couple any signals into the input. The input resistance is $100k\Omega$ (typ), so the capacitor size may be chosen accordingly to define the LF rolloff desired. This can be calculated as:

$C_{IN} = 1 / (2\pi f_{CUT} R_{IN})$

This assumes a low source impedance is driving the inputs.

A further consideration for the differential input is the effect of these series input capacitors on low-frequency, common-mode rejection. Any mismatch in the values of these two capacitors degrades the CMRR at frequencies where the impedance of the capacitor is significant

compared to the input resistance of the amplifier—this is usually most noticeable at low frequencies. One way to avoid the need for matched or tight tolerance capacitors is to deliberately oversize the values on the differential inputs and to set the lower 3dB point (f_{CUT}) of the amplifier by sizing the output capacitor appropriately.

The input impedance matching on the differential input is typically 1%, allowing input capacitor matching to be effective at improving low-frequency PSRR.

Common-Mode Rejection Ratio

The common-mode rejection ratio (CMRR) refers to the amount of rejection that the amplifier is capable of providing to any signal applied equally to the IN+ and IN-inputs. In the case of amplifying low-level microphone signals in noisy digital environments, it is a key figure of merit. In audio circuits, this is generally measured for V_{IN} as an AC signal:

$CMRR(dB) = A_{DM} / A_{CM}$

where A_{DM} is the differential gain, A_{CM} is the common-mode gain.

Input voltages should be sufficiently small such that the output is not clipped in either differential or common-

mode application. The topology used in the MAX4063 means that the CMRR actually improves at higher differential gains—another advantage of using differential sensing.

Auxiliary Input

The auxiliary input is a single-ended input intended to be used with a jack-socket type microphone input (Figure 1). Internal DC-bias components (as on the main inputs) allow the input signal to be AC-coupled. Mechanically switched jack sockets can be used in conjunction with the INT/AUX select pin, allowing the auxiliary microphone input to be automatically selected when a jack socket is inserted.

Microphone Bias Voltage

On the MAX4063 thin QFN package, connect the exposed paddle (backside of PRS) to the ground plane. The MAX4063 has an integrated low-noise, low-output impedance bias voltage. An optimum electret bias resistor can be set externally. This gives a low-noise, flexible solution that can run from 2.4V to 5.5V, which is suitable for hand-held devices such as PDAs that typically have audio power supplies in the 3V region (Figure 2).

Output

DC Bias

In shutdown mode, the bias voltage is disabled. OUT and OUT have a low-noise, DC-bias voltage independent of the power supplies, resulting in superior PSRR performance. The MAX4063 outputs are high impedance when the part is in shutdown mode.

Differential Output

The MAX4063 features a differential output stage (OUT and OUT), allowing optimum performance when connected to ADCs and CODECs with differential inputs. This differential output is particularly useful in designs where the microphone preamplifier is mounted some distance away from the CODEC/ADC, as the low-impedance, differential line provides excellent noise rejection and immunity (Figure 4).

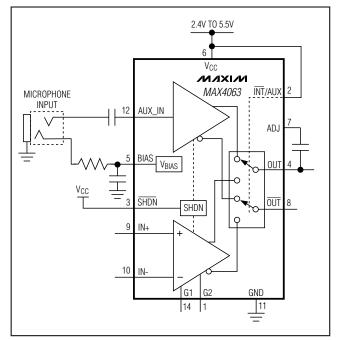


Figure 2. MAX4063 Used for Biasing a Microphone

Applications Information

Shutdown Mode

The MAX4063 features a low-power, complete shutdown mode. When SHDN goes low, the supply current drops to 0.3μ A, the output enters a high-impedance state, and the bias current to the microphone is switched off. Driving SHDN high enables the amplifier. SHDN should not be left unconnected.

Power Supplies and Layout

The MAX4063 operates from a 2.4V to 5.5V single supply. Bypass the power supply with a 0.1μ F capacitor to ground. In systems where analog and digital grounds are available, the MAX4063 should be connected to the analog ground.

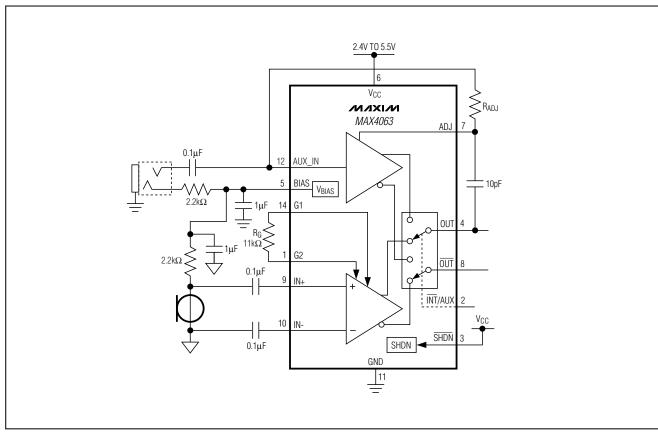


Figure 3. MAX4063 Used to Bias a Microphone Connected to the Auxiliary Input and the Differential Input

MAX4063



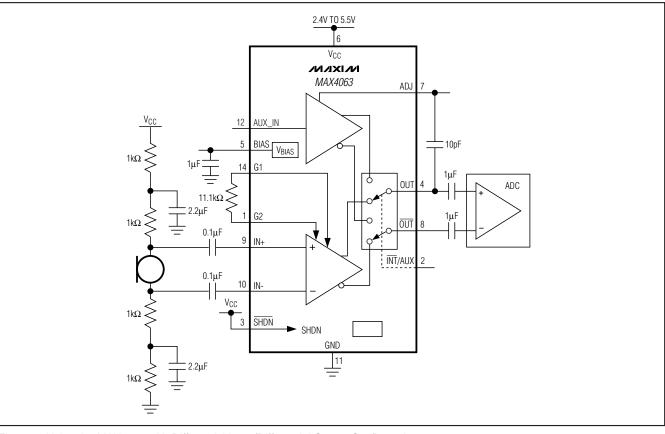
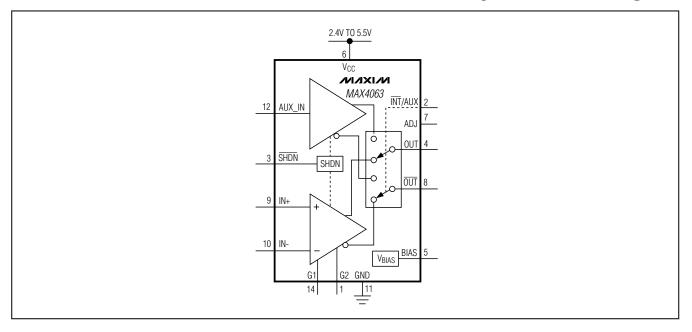


Figure 4. Using the MAX4063 with Differential Input/Differential Output Configuration

_Simplified Block Diagram

MAX4063



Chip Information

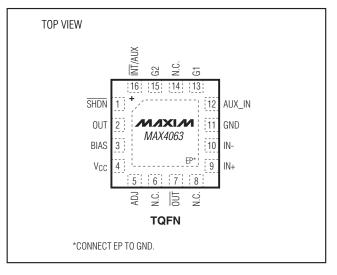
PROCESS: BICMOS

Package Information

For the latest package outline information and land patterns (footprints), go to **www.maxim-ic.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
14 SSOP	U14+1	<u>21-0066</u>	<u>90-0113</u>
16 TQFN	T1644+4	<u>21-0139</u>	<u>90-0070</u>

_Pin Configurations (continued)



Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	
0	1/03	Initial release	—
1	6/11	Added EP information to <i>Pin Description</i> , updated power dissipation ratings, updated <i>Ordering Information</i> and <i>Pin Configuration</i> for lead-free parts	1, 2, 7, 13

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

14

MAX4063

___Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 408-737-7600

© 2011 Maxim Integrated Products

Maxim is a registered trademark of Maxim Integrated Products, Inc.