

ABSOLUTE MAXIMUM RATINGS

V_{IN} Voltage	10V
SW1 Voltage	20V
Sense Voltage	20V
Shutdown Voltage	10V
Select Voltage	10V
I_{LIM} Voltage	0.5V
Maximum Power Dissipation	500mW
Operating Temperature Range	
LT1301C	0°C to 70°C
LT1301I	40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec)	300°C

PACKAGE/ORDER INFORMATION

TOP VIEW N8 PACKAGE 8-LEAD PLASTIC DIP S8 PACKAGE 8-LEAD PLASTIC SOIC $T_{JMAX} = 100^{\circ}\text{C}$, $\theta_{JA} = 150^{\circ}\text{C/W}$	ORDER PART NUMBER
	LT1301CN8 LT1301CS8 LT1301IS8
	S8 PART MARKING
	1301 1301I

ELECTRICAL CHARACTERISTICS $T_A = 25^{\circ}\text{C}$, $V_{IN} = 2\text{V}$ unless otherwise noted.

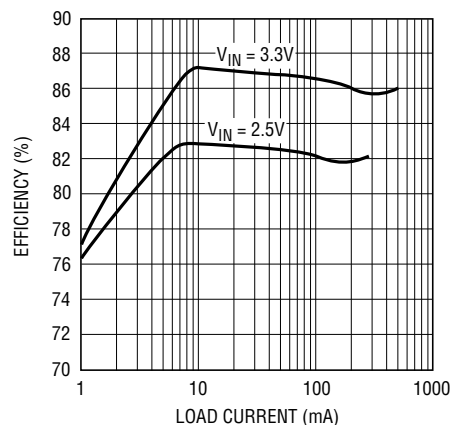
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
I_Q	Quiescent Current	$V_{SHDN} = 0.5\text{V}$, $V_{SEL} = 5\text{V}$, $V_{SENSE} = 5.5\text{V}$ $V_{SHDN} = 1.8\text{V}$	●		120	200	μA
			●		7	15	μA
V_{IN}	Input Voltage Range		●	1.8			V
			●	2.0			V
V_{OUT}	Output Sense Voltage	$V_{SEL} = 5\text{V}$ $V_{SEL} = 0\text{V}$	●	11.52	12.00	12.48	V
			●	4.75	5.00	5.25	V
	Output Referred Comparator Hysteresis	$V_{SEL} = 5\text{V}$ (Note 1) $V_{SEL} = 0\text{V}$ (Note 1)	●		50	100	mV
			●		22	50	mV
	Oscillator Frequency	Current Limit not Asserted.		120	155	185	kHz
	Oscillator TC				0.2		%/ $^{\circ}\text{C}$
DC	Maximum Duty Cycle			75	86	95	%
t_{ON}	Switch On-Time	Current Limit not Asserted.			5.6		μs
	Output Line Regulation	$1.8\text{V} < V_{IN} < 6\text{V}$	●		0.06	0.15	%/V
V_{CESAT}	Switch Saturation Voltage	$I_{SW} = 700\text{mA}$	●		130	200	mV
	Switch Leakage Current	$V_{SW} = 5\text{V}$, Switch Off	●		0.1	10	μA
	Peak Switch Current (Internal Trip Point)	I_{LIM} Floating (See Typical Application) I_{LIM} Grounded		0.75	1.0	1.25	A
					0.4		A
V_{SHDNH}	Shutdown Pin High		●	1.8			V
V_{SHDNL}	Shutdown Pin Low					0.5	V
V_{SELH}	Select Pin High		●	1.5			V
V_{SELL}	Select Pin Low		●			0.8	V
I_{SHDN}	Shutdown Pin Bias Current	$V_{SHDN} = 5\text{V}$ $V_{SHDN} = 2\text{V}$ $V_{SHDN} = 0\text{V}$	●		8	20	μA
			●		3		μA
			●		0.1	1	μA
I_{SEL}	Select Pin Bias Current	$0\text{V} < V_{SEL} < 5\text{V}$	●		1	3	μA

The ● denotes specifications which apply over the 0°C to 70°C temperature range.

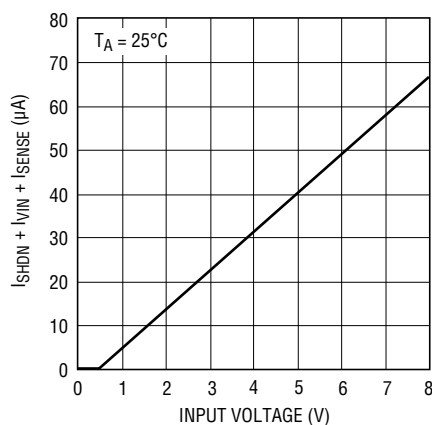
Note 1: Hysteresis specified is DC. Output ripple may be higher if output capacitance is insufficient or capacitor ESR is excessive. See operation section.

TYPICAL PERFORMANCE CHARACTERISTICS

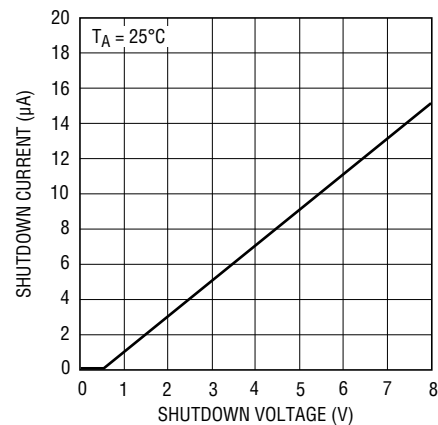
5V Output Efficiency



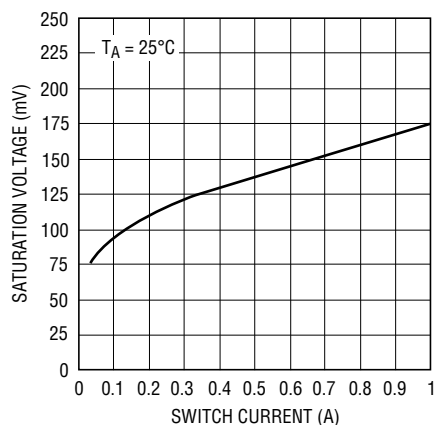
Total Quiescent Current in Shutdown



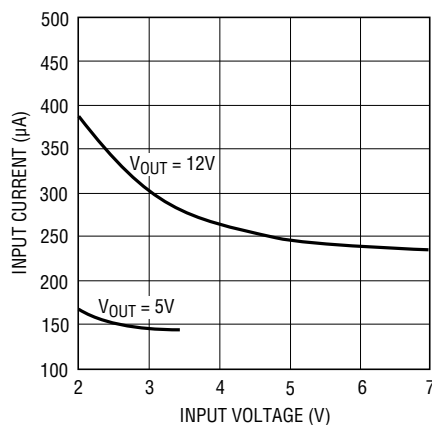
Shutdown Pin Bias Current



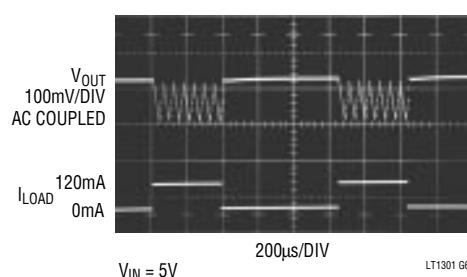
Saturation Voltage vs Switch Current



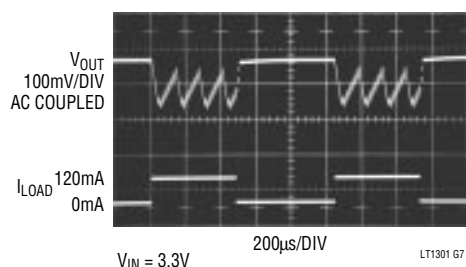
No-Load Input Current



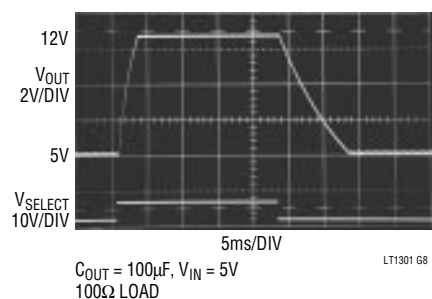
Load Transient Response of Figure 1 Circuit



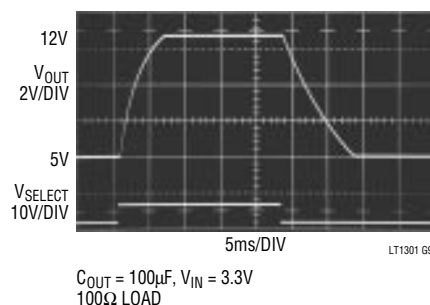
Load Transient Response of Figure 1 Circuit



Select Pin Transient Response



Select Pin Transient Response



PIN FUNCTIONS

GND (Pin 1): Signal Ground. Tie to PGND under the package.

Sel (Pin 2): Output Select. When tied to V_{IN} converter regulates at 12V. When grounded or floating converter regulates at 5V. May be driven under logic control.

SHDN (Pin 3): Shutdown. Pull high to shut down the LT1301. Ground for normal operation.

Sense (Pin 4): “Output” Pin. Goes to internal resistive divider. If operating at 5V output, a 0.1μF ceramic capacitor is required from Sense to Ground.

I_{LIM} (Pin 5): Float for 1A switch current limit. Tie to ground

for approximately 400mA. A resistor between I_{LIM} and ground sets peak current to some intermediate value .

V_{IN} (Pin 6): Supply Pin. Must be bypassed with a large value electrolytic to ground. Keep bypass within 0.2" of the device.

SW (Pin 7): Switch Pin. Connect inductor and diode here. Keep layout short and direct to minimize radio frequency interference.

PGND (Pin 8): Power Ground. Tie to signal ground (pin 1) under the package. Bypass capacitor from V_{IN} should be tied directly to PGND within 0.2" of the device.

BLOCK DIAGRAM

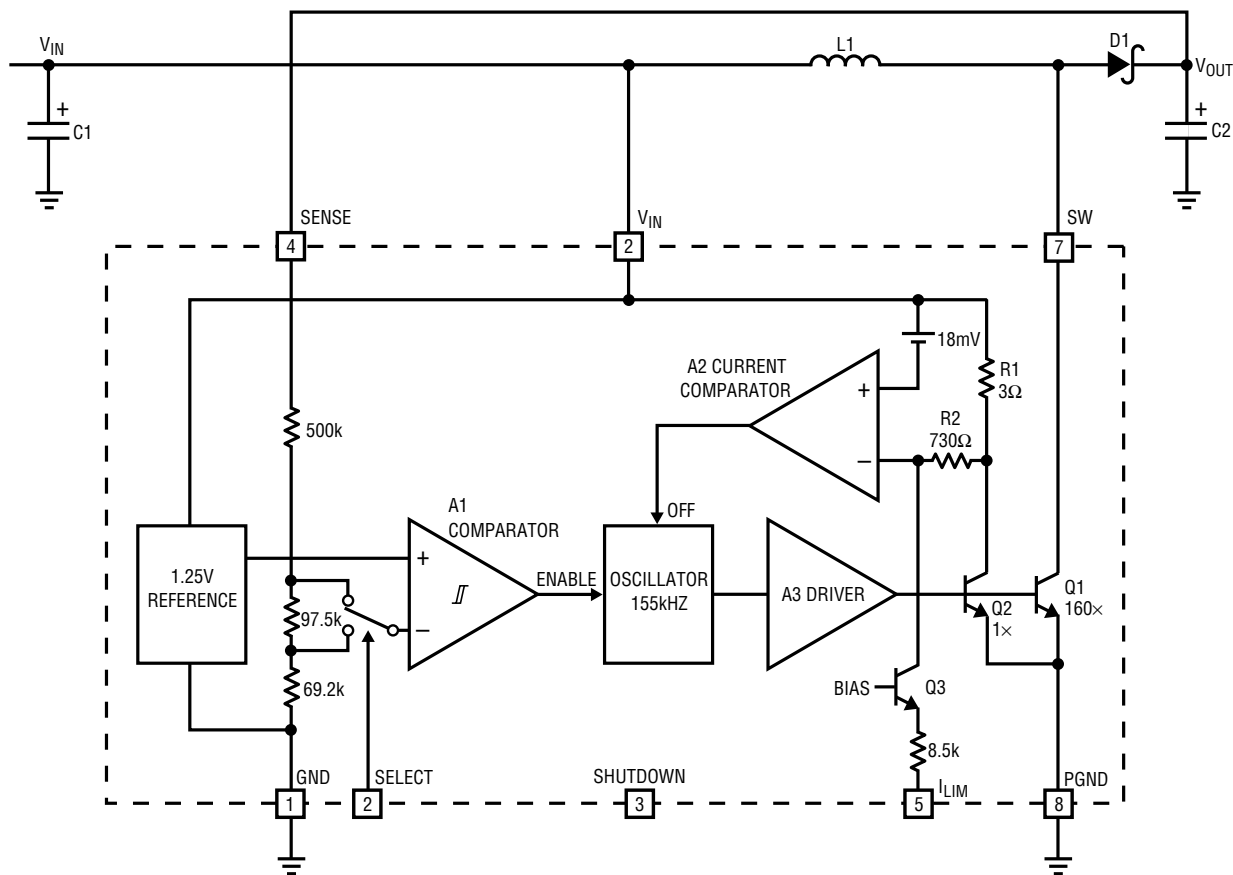
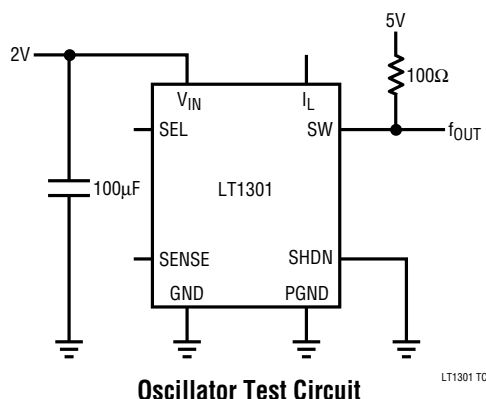


Figure 2.

LT1301 F2

TEST CIRCUIT



OPERATION

Operation of the LT1301 is best understood by referring to the Block Diagram in Figure 2. When A1's negative input, related to the Sense pin voltage by the appropriate resistor-divider ratio is higher than the 1.25V reference voltage, A1's output is low. A2, A3 and the oscillator are turned off, drawing no current. Only the reference and A1 consume current, typically 120μA. When A1's negative input drops below 1.25V, overcoming A1's 6mV hysteresis, A1's output goes high enabling the oscillator, current comparator A2, and driver A3. Quiescent current increases to 2mA as the device prepares for high current switching. Q1 then turns on in controlled saturation for (nominally) 5.3μs or until comparator A2 trips, whichever comes first. After a fixed off-time of (nominally) 1.2μs, Q1 turns on again. The LT1301's switching causes current to alternately build up in L1 and dump into output capacitor C2 via D1, increasing the output voltage. When the output is high enough to cause A1's output to go to low, switching action ceases. C2 is left to supply current to the load until V_{OUT} decreases enough to force A1's output high, and the entire cycle repeats. Figure 4 details relevant waveforms. A1's cycling causes low-to-mid-frequency ripple voltage on the output. Ripple can be reduced by making the output capacitor large. The 33μF unit specified results in ripple of 100mV to 200mV on the 12V output. A 100μF capacitor will decrease ripple to 50mV. If operating at 5V output a 0.1μF ceramic capacitor is required at the Sense pin in addition to the electrolytic.

If switch current reaches 1A, causing A2 to trip, switch on-time is reduced and off-time increases slightly. This allows continuous mode operation during bursts. A2 monitors

the voltage across 3Ω resistor R1 which is directly related to the switch current. Q2's collector current is set by the emitter-area ratio to 0.6% of Q1's collector current. When R1's voltage drop exceeds 18mV, corresponding to 1A switch current, A2's output goes high, truncating the on-time portion of the oscillator cycle and increasing off-time to about 2μs as shown in Figure 3, trace A. This programmed peak current can be reduced by tying the I_{LIM} pin to ground, causing 15μA to flow through R2 into Q3's collector. Q3's current causes a 10.4mV drop in R2 so that only an additional 7.6mV is required across R1 to turn off the switch. This corresponds to a 400mA switch current as shown in Figure 3, trace B. The reduced peak switch current reduces I^2R losses in Q1, L1, C1 and D1. Efficiency can be increased by doing this provided that the accompanying reduction in full load current is acceptable. Lower peak currents also extend alkaline battery life due to the alkaline cell's high internal impedance.

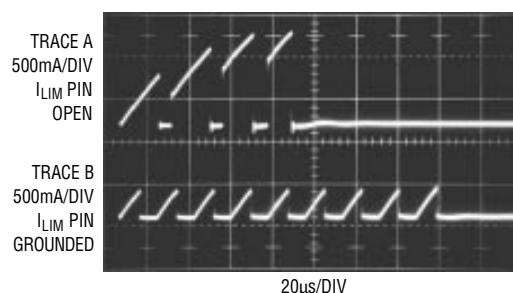


Figure 3. Switch Pin Current With I_{LIM} Floating or Grounded

APPLICATIONS INFORMATION

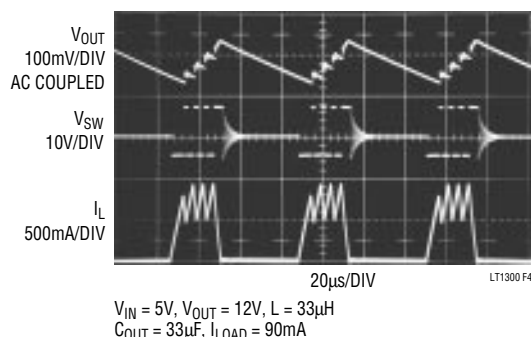


Figure 4. Burst Mode Operation in Action

Output Voltage Selection

The LT1301 can be selected to 5V or 12V under logic control or fixed at either by tying Select to ground or V_{IN} respectively. It is permissible to tie Select to a voltage higher than V_{IN} as long as it does not exceed 10V. Efficiency in 5V mode will be slightly less than in 12V mode due to the fact that the diode drop is a greater percentage of 5V than 12V. Since the bipolar switch in the LT1301 gets its base drive from V_{IN} , no reduction in switch efficiency occurs when in 5V mode. When V_{IN} exceeds the programmed output voltage the output will follow the input. This is characteristic of the simple step-up or “boost” converter topology. A circuit example that provides a regulated output with an input voltage above or below the output (known as a buck-boost or SEPIC) is shown in the Typical Applications section.

Shutdown

The converter can be turned off by pulling SHDN (pin 3) high. Quiescent current drops to 10μA in this condition. Bias current of 8μA to 10μA flows into the pin (at 5V input). It is recommended that SHDN not be left floating. Tie the pin to ground if the feature is not used. SHDN can be driven high even if V_{IN} is floating.

I_{LIM} Function

The LT1301's current limit (I_{LIM}) pin can be used for soft start. Upon start-up, the LT1301 will draw maximum current from the supply (about 1A) from the supply to charge the output capacitor. Figure 5 shows V_{OUT} and I_{IN} waveforms as the device is turned on. The high current flow can create IR drops along supply and ground lines or cause the input supply to drop out momentarily. By

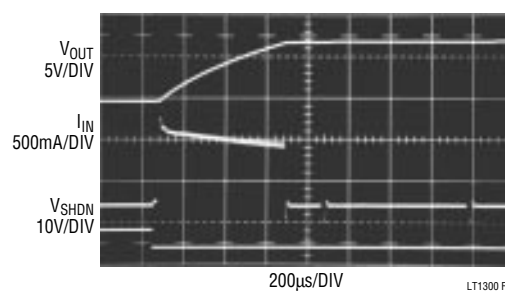


Figure 5. Start-Up Response

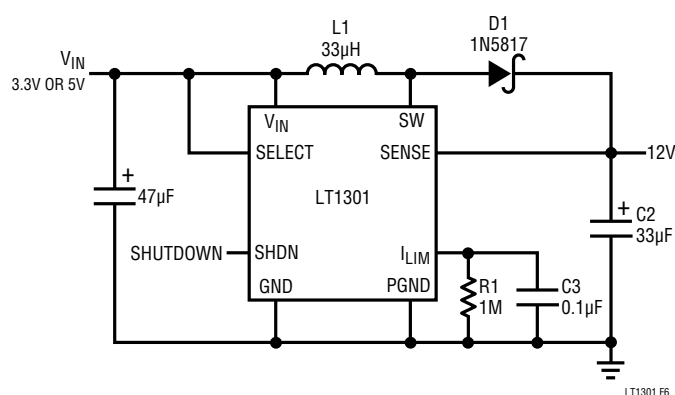


Figure 6.

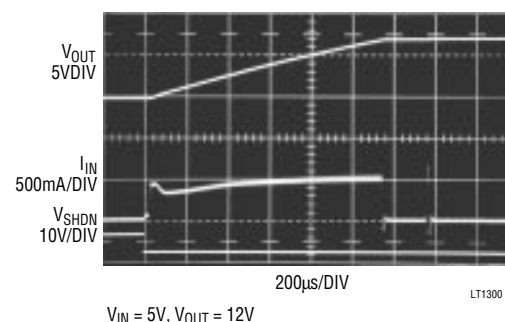


Figure 7. Startup Response Soft-Start Circuitry Added

adding R1 and C3 as shown in Figure 6, the switch current in the LT1301 is initially limited to 400mA until the 15μA flowing out of the I_{LIM} pin charges up C3. Input current is held to under 500mA while the output voltage ramps up to 12V as shown in Figure 7. R1 provides a discharge path for the capacitor without appreciably decreasing peak switch current. When using the I_{LIM} pin soft-start mode a minimum load of a few hundred microamperes is recommended to prevent C3 from discharging, as no current flows out of I_{LIM} when the LT1301 is not

APPLICATIONS INFORMATION

Table 1. Recommended Inductors

PART NUMBER	VENDOR	L (μ H)	DCR (Ω)	V _{IN} (V)	I _{LIM} PIN	EFFICIENCY (%)			COMPONENT HEIGHT (mm)	PHONE NUMBER
						30mA	60mA	120mA		
D03316-333	Coilcraft	33	0.088	3.3	Open	84	84	85	5.5	(708) 639-6400
				5	Open	89	89	90		
D01608-223	Coilcraft	22	.31	3.3	Open	82	82	—	3.5	
				3.3	Ground	85	—	—		
				5	10k	86	87	—		
				5	Ground	88	—	—		
D01608-103	Coilcraft	10	.11	2	Open	78	—	—	3.5	
CTX20-1	Coiltronics	20	.175	3.3	Open	84	84	—	4.2	(407) 241-7876
				5	Open	88	88	89		
GA10-332	Gowanda	33	.077	3.3	Open	86	86	87	Through-Hole	(716) 532-2234
				5	Open	89	89	90		
LQH3G220K04M00	Murata-Erie	22	0.7	3.3	Ground	81	—	—	2.0	(404) 436-1300
				5	Ground	85	—	—		
CD73-330KC	Sumida	33	0.131	3.3	Open	84	85	86	3.5	(708) 956-0666
				5	Open	88	88	89		
CDRH62-330MC	Sumida	33	0.48	3.3	Open	80	80	81	3.0	
					Ground	85	—	—		
				5	Open	84	84	85		
					Ground	83	—	—		

switching. Zero load current causes the LT1301 to switch so infrequently that C3 can completely discharge reducing subsequent peak switch current to 400mA. If a load is suddenly applied, output voltage will sag until C3 can be recharged and peak switch current returns to 1A.

If the full capacity of the LT1301 is not required peak current can be reduced by changing the value of R3 as shown in Figure 8. With R3 = 0 switch current is limited to approximately 400mA. Smaller, less expensive inductors with lower saturation ratings can then be used.

Inductor Selection

For full output power, the inductor should have a saturation current rating of 1.25A for worst-case current limit, although it is acceptable to bias an inductor 20% or more into saturation. Smaller inductors can be used in conjunction with the I_{LIM} pin. Efficiency is significantly affected by inductor DCR. For best efficiency limit the DCR to 0.03 Ω or less. Toroidal types are preferred in some cases due to their inherent flux containment and EMI/RFI superiority. Recommended inductors are listed in Table 1.

Table 2. Recommended Capacitors

VENDOR	SERIES	TYPE	PHONE#
AVX	TPS	Surface Mount	(803)448-9411
Sanyo	OS-CON	Through-Hole	(619) 661-6835
Panasonic	HFQ	Through-Hole	(201) 348-5200

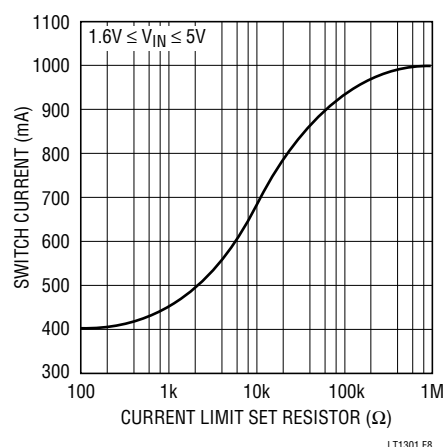


Figure 8. Peak Switch Current vs. Current Limit Set Resistor

APPLICATIONS INFORMATION

Capacitor Selection

Low ESR capacitors are required for both input and output of the LT1301. ESR directly affects ripple voltage and efficiency. For surface mount applications AVX TPS series tantalum capacitors are recommended. These have been specially designed for SMPS and have low ESR along with high surge current ratings. For through-hole applications Sanyo OS-CON capacitors offer extremely low ESR in a small size. Again, if peak switch current is reduced using the I_{LIM} pin, capacitor requirements can be relaxed and smaller, higher ESR units can be used. Suggested capacitor sources are listed in Table 2.

Diode Selection

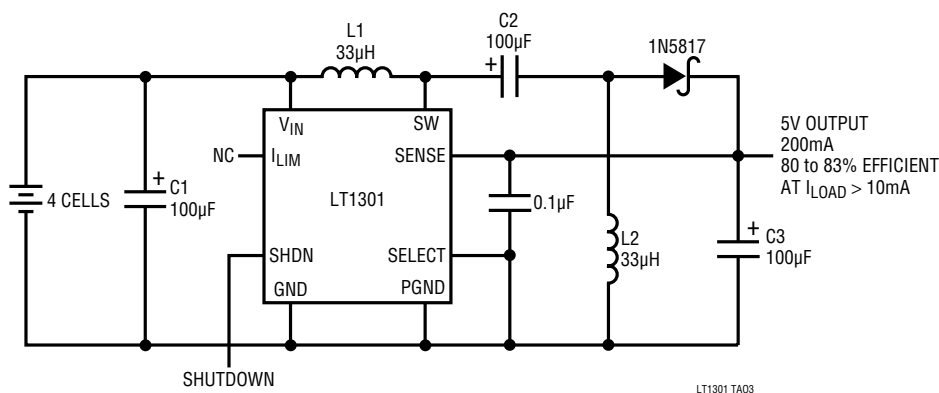
Best performance is obtained with a Schottky rectifier diode such as the 1N5817. Phillips Components makes this in surface mount as the PRL5817. Motorola makes the MBRS130LT3 which is slightly better and also in surface mount. For lower output power a 1N4148 can be used although efficiency will suffer substantially.

Layout Considerations

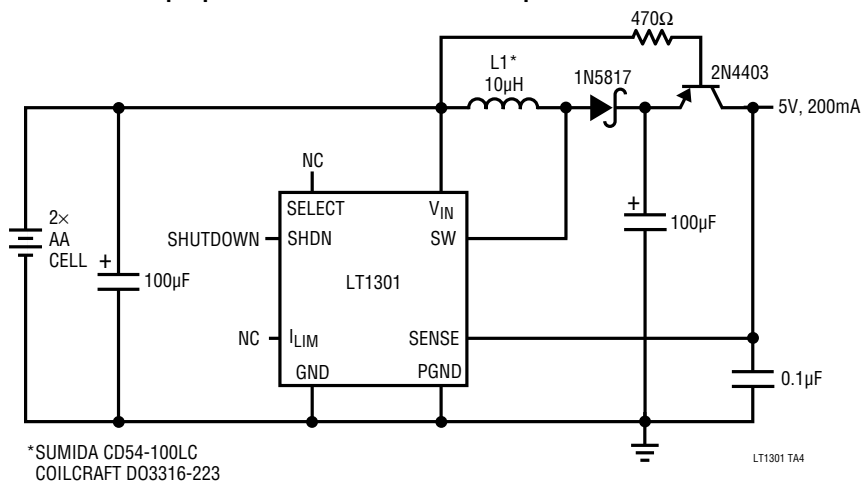
The LT1301 is a high speed, high current device. The input capacitor must be no more than 0.2" from V_{IN} (pin 6) and ground. Connect the PGND and GND (pins 8 and 1) together under the package. Place the inductor adjacent to SW (pin 7) and make the switch pin trace as short as possible. This keeps radiated noise to a minimum.

TYPICAL APPLICATIONS

Four-Cell to 5V Converter



Step-Up Converter with Automatic Output Disconnect



CONTRACT
V_{OUT} -4V TO -29V 12mA
MAXIMUM FROM 1.8V SUPPLY
(77% EFFICIENT)
20mA MAXIMUM FROM
3V SUPPLY (83% EFFICIENT)

T1 = DALE LPE-5047-A045 (605) 665-9301

PWM IN
0% TO 100%
CMOS DRIVE 0V TO 5V

LT1300 TA5

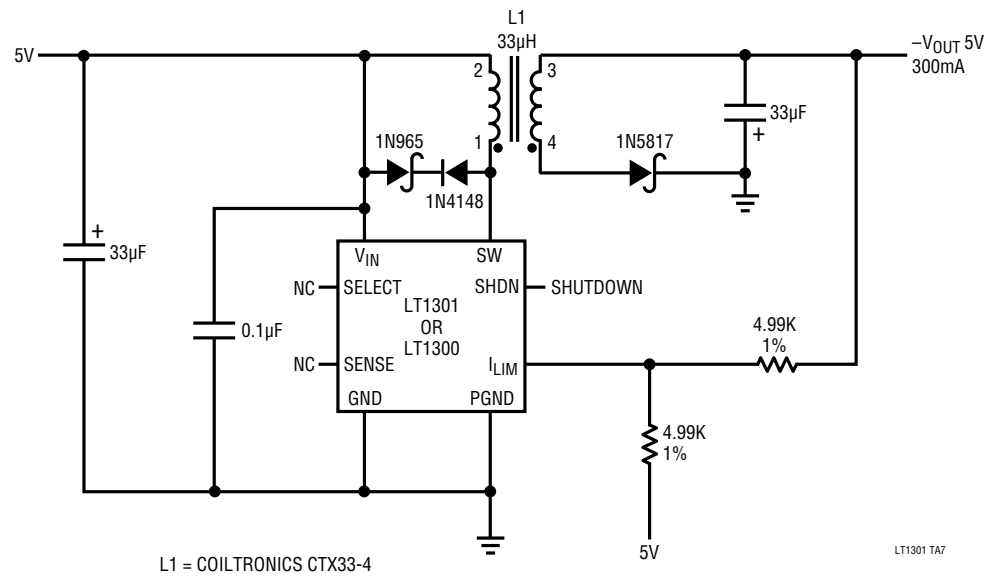
SHUTDOWN

0 - 5VDC IN
INTENSITY ADJUST
100μA TO 2mA BULB CURRENT

LT1300 TAg

TYPICAL APPLICATIONS

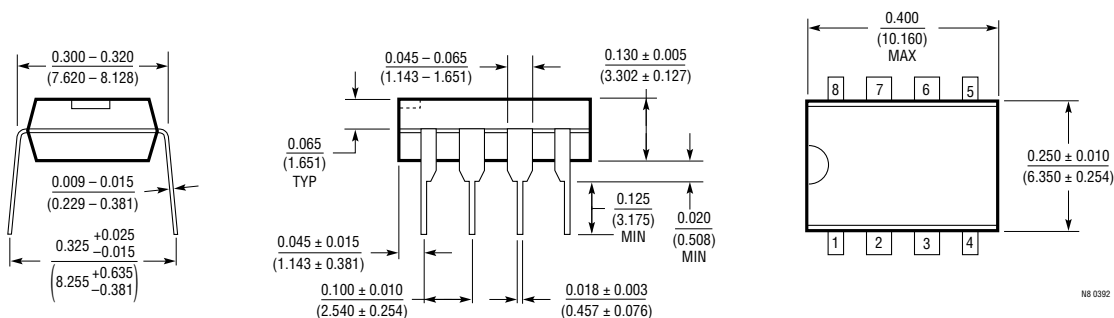
5V to -5V Converter



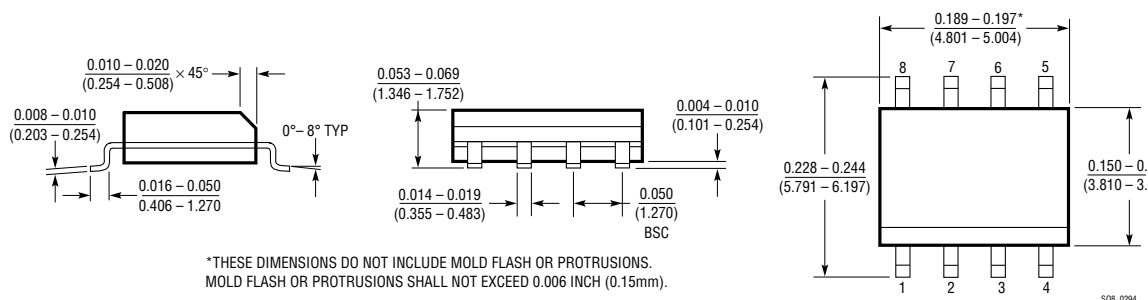
PACKAGE DESCRIPTION

Dimensions in inches (millimeters) unless otherwise noted.

N8 Package 8-Lead Plastic DIP



S8 Package 8-Lead Plastic SOIC



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