Table 2. Absolute Maximum Ratings

Symbol	Pin	Parameter	Value	Unit	
I _{Vcc}	8	$I_{q} + I_{Z}; (I_{GD} = 0)$	30	mA	
I _{GD}	7	Output Totem Pole Peak Current (2µs)	±700	mA	
INV, COMP MULT	1, 2, 3	Analog Inputs & Outputs	-0.3 to 7	V	
CS	4	Current Sense Input	-0.3 to 7	V	
ZCD	5	Zero Current Detector	50 (source) -10 (sink)	mA mA	
P _{tot}		Power Dissipation @T _{amb} = 50 °C	(DIP-8) (SO-8)	1 0.65	W W
Tj		Junction Temperature Operating Range	-40 to 150	°C	
T _{stg}		Storage Temperature	-55 to 150	°C	

Figure 3. Pin Connection (Top view)

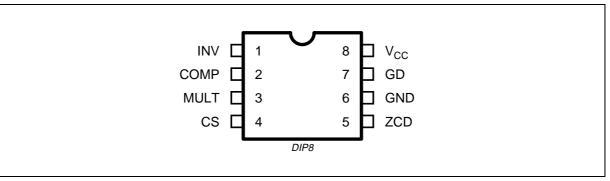


Table 3. Thermal Data

Symbol	Parameter	SO 8	MINIDIP	Unit
R _{th j-amb}	Thermal Resistance Junction to ambient	150	100	°C/W

Table 4. Pin Description

Ν.	Name	Function
1	INV	Inverting input of the error amplifier. A resistive divider is connected between the output regulated voltage and this point, to provide voltage feedback.
2	COMP	Output of error amplifier. A feedback compensation network is placed between this pin and the INV pin.
3	MULT	Input of the multiplier stage. A resistive divider connects to this pin the rectified mains. A voltage signal, proportional to the rectified mains, appears on this pin.
4	CS	Input to the comparator of the control loop. The current is sensed by a resistor and the resulting voltage is applied to this pin.
5	ZCD	Zero current detection input. If it is connected to GND, the device is disabled.
6	GND	Current return for driver and control circuits.
7	GD	Gate driver output. A push pull output stage is able to drive the Power MOS with peak current of 400mA (source and sink).
8	V _{CC}	Supply voltage of driver and control circuits.

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(1) Parameter guaranteed by design, not tested in production.

Table 5. Electrical Characteristics

(V_{CC} = 14.5V; T_{amb} = -25°C to 125°C;unless otherwise specified)

Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
SUPPLY	VOLTA	GE SECTION					
V _{CC}	8	Operating Range	after turn-on	11		18	V
V _{CC ON}	8	Turn-on Threshold		11	12	13	V
V _{CC OFF}	8	Turn-off Threshold		8.7	9.5	10.3	V
Hys	8	Hysteresis		2.2	2.5	2.8	V
SUPPLY	CURR	ENT SECTION	•				
ISTART-U	8	Start-up Current	before turn-on (V _{CC} =11V)	20	50	90	μΑ
۱ _q	8	Quiescent Current			2.6	4	mA
Icc	8	Operating Supply Current	C _L = 1nF @ 70KHz		4	5.5	mA
			in OVP condition $V_{pin1} = 2.7V$		1.4	2.1	mA
lq	8	Quiescent Current	$V_{PIN5} \leq 150 \text{mV}, V_{CC} > V_{CC off}$		1.4	2.1	mA
	8		$V_{PIN5} \le 150 mV$, $V_{CC} < V_{CC off}$	20	50	90	μΑ
VZ	8	Zener Voltage	I _{CC} = 25mA	18	20	22	V
ERROR A	MPLI	FIER SECTION	•				
VINV	1	Voltage Feedback Input	$T_{amb} = 25^{\circ}C$	2.465	2.5	2.535	V
		Threshold	12V < V _{CC} < 18V	2.44		2.56	V
		Line Regulation	V _{CC} = 12 to 18V		2	5	mV
I _{INV}	1	Input Bias Current			-0.1	-1	μA
Gv		Voltage Gain	Open loop	60	80		dB
GB		Gain Bandwidth			1		MHz
I _{COMP}	2	Source Current	$V_{COMP} = 4V, V_{INV} = 2.4V$	-2	-4	-8	mA
		Sink Current	$V_{COMP} = 4V, V_{INV} = 2.6V$	2.5	4.5		mA
VCOMP	2	Upper Clamp Voltage	I _{SOURCE} = 0.5mA		5.8		V
		Lower Clamp Voltage	I _{Sink} = 0.5mA		2.25		V
MULTIPL	IER SE	ECTION	-				
V _{MULT}	3	Linear Operating Voltage		0 to 3	0 to 3.5		V
ΔV_{CS}		Output Max. Slope	V _{MULT} = from 0V to 0.5V	1.65	1.9		
ΔV_{mult}			V _{COMP} = Upper Clamp Voltage				
K		Gain	V _{MULT} = 1V V _{COMP} = 4V	0.45	0.6	0.75	1/V
		SE COMPARATOR		0.40	0.0	0.75	17 V
V _{CS}	4	Current Sense Reference Clamp	$V_{MULT} = 2.5V$ $V_{COMP} = Upper Clamp Voltage$	1.6	1.7	1.8	V
I _{CS}	4	Input Bias Current	$V_{OS} = 0$		-0.05	-1	μA
t _{d (H-L)}	4	Delay to Output		1	200	450	ns
(() L)	4	Current Sense Offset			0	15	mV
ZERO CU		IT DETECTOR	1		-	-	
V _{ZCD}	5	Input Threshold Voltage Rising Edge	(1)		2.1		V
		Hysteresis	(1)	0.3	0.5	0.7	V
V _{ZCD}	5	Upper Clamp Voltage	$I_{ZCD} = 20\mu A$	4.5	5.1	5.9	V
	1	Upper Clamp Voltage	$I_{ZCD} = 3mA$		1		

Table 5. Electrical Characteristics (continued)

(V_{CC} = 14.5V; T_{amb} = -25°C to 125°C;unless otherwise specified)

Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
V _{ZCD}	5	Lower Clamp Voltage	I _{ZCD} = -3mA	0.3	0.65	1	V
I _{ZCD}	5	Sink Bias Current	$1V \le V_{ZCD} \le 4.5V$		2		μΑ
I _{ZCD}	5	Source Current Capability		-3		-10	mA
I _{ZCD}	5	Sink Current Capability		3		10	mA
V _{DIS}	5	Disable threshold		150	200	250	mV
I _{ZCD}	5	Restart Current After Disable	V _{ZCD} < V _{dis} ; V _{CC} > V _{CCOFF}	-100	-200	-300	μA
OUTPUT	SECT	ION					
V _{GD}	7	Dropout Voltage	I _{GDsource} = 200mA		1.2	2	V
			I _{GDsource} = 20mA		0.7	1	V
			I _{GDsink} = 200mA			1.5	V
			I _{GDsink} = 20mA			0.3	V
tr	7	Output Voltage Rise Time	$C_L = 1nF$		40	100	ns
t _f	7	Output Voltage Fall Time	$C_L = 1nF$		40	100	ns
I _{GD off}	7	IGD Sink Current	V _{CC} =3.5V V _{GD} = 1V	5	10	-	mA
OUTPUT	OVER	VOLTAGE SECTION					
I _{OVP}	2	OVP Triggering Current		35	40	45	μA
		Static OVP Threshold		2.1	2.25	2.4	V
RESTAR	TIME	R					
t _{START}		Start Timer		70	150	400	μs

3 OVER VOLTAGE PROTECTION OVP

The output voltage is expected to be kept by the operation of the PFC circuit close to its nominal value. This is set by the ratio of the two external resistors R1 and R2 (see fig. 5), taking into consideration that the non inverting input of the error amplifier is biased inside the L6561 at 2.5V.

In steady state conditions, the current through R1 and R2 is:

$$I_{R1sc} = \frac{V_{out} - 2.5}{R1} = I_{R2} = \frac{2.5V}{R2}$$

and, if the external compensation network is made only with a capacitor C_{comp} , the current through C_{comp} equals zero. When the output voltage increases abruptly the current through R1 becomes:

$$I_{R1} = \frac{V_{outsc} + \Delta V_{out} - 2.5}{R1} = I_{R1sc} + \Delta I_{R1}$$

Since the current through R2 does not change, ΔI_{R1} must flow through the capacitor C_{comp} and enter the error amplifier.

This current is monitored inside the L6561 and when reaches about 37μ A the output voltage of the multiplier is forced to decrease, thus reducing the energy drawn from the mains. If the current exceeds 40μ A, the OVP protection is triggered (Dynamic OVP), and the external power transistor is switched off until the current falls approximately below 10μ A.

However, if the overvoltage persists, an internal comparator (Static OVP) confirms the OVP condition keeping the external power switch turned off (see fig. 4). Finally, the overvoltage that triggers the OVP function is:

$$\Delta V_{out} = R_1 \cdot 40 \mu A.$$

Typical values for R1, R2 and C are shown in the application circuits. The overvoltage can be set indepen-

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dently from the average output voltage. The precision in setting the overvoltage threshold is 7% of the overvoltage value (for instance $\Delta V = 60V \pm 4.2V$).

3.1 Disable function

The zero current detector (ZCD) pin can be used for device disabling as well. By grounding the ZCD voltage the device is disabled reducing the supply current consumption at 1.4mA typical (@ 14.5V supply voltage).

Releasing the ZCD pin the internal start-up timer will restart the device.

Figure 4.

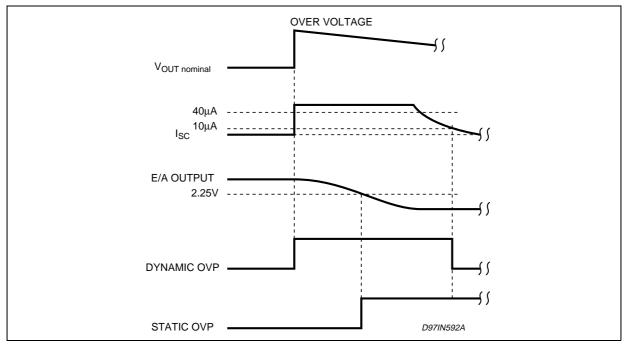
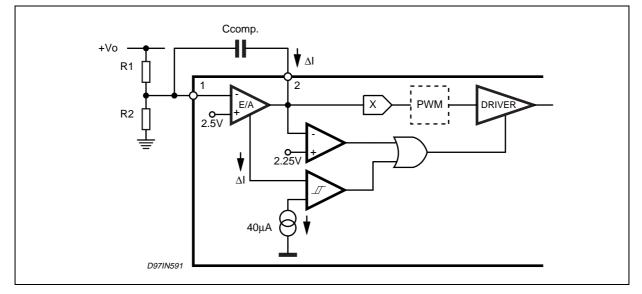
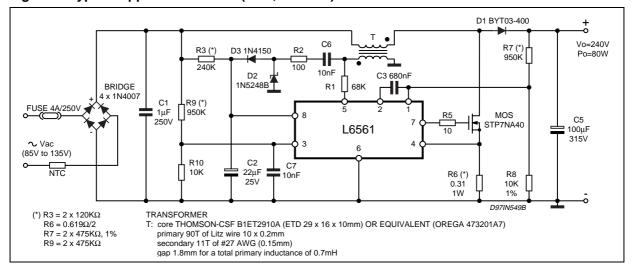


Figure 5. Overvoltage Protection Circuit



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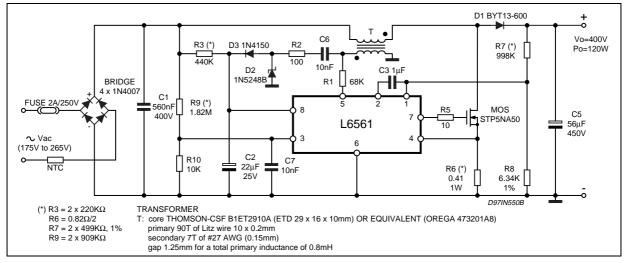
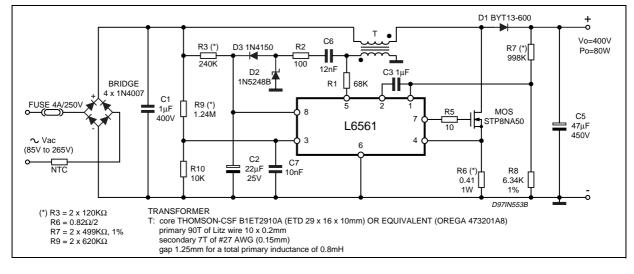


Figure 8. Typical Application Circuit (80W, Wide-range Mains)



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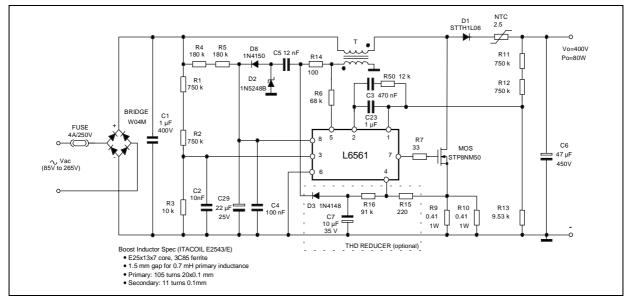


Figure 9. Demo Board (EVAL6561-80) Electrical Schematic

Figure 10. EVAL6561-80: PCB and Component Layout (Top view, real size 57x108mm)

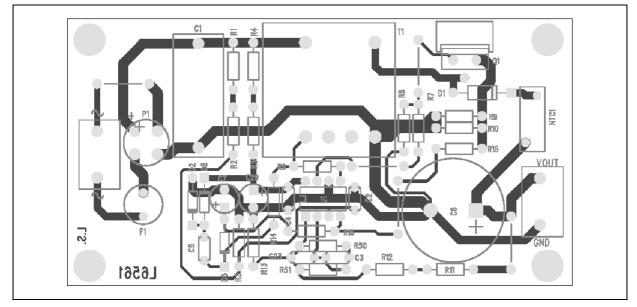
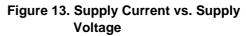


Table 6. EVAL6561-80	: Evaluation Results.
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V _{in} (Vac)		V _o (Vdc)	4)/a ()/da)		m (9/)	w/o THD	reducer	with THE) reducer
	Pin (W)	v ₀ (vuc)	∆Vo (Vdc)	Po (W)	'ο (W) η (%)	PF	THD (%)	PF	THD (%)
85	87.2	400.1	14	80.7	92.8	0.999	3.7	0.999	2.9
110	85.2	400.1	14	80.7	94.7	0.996	5.0	0.996	3.2
135	84.2	400.1	14	80.7	95.8	0.989	6.2	0.989	3.7
175	83.5	400.1	14	80.7	96.6	0.976	8.3	0.976	4.3
220	83.1	400.1	14	80.7	97.1	0.940	10.7	0.941	5.6
265	82.9	400.1	14	80.7	97.3	0.890	13.7	0.893	8.1

D94IN047A 10VP (μA) 41 40 39 38 -50 -25 0 25 50 75 100 125 T (°C)

Figure 11. OVP Current Threshold vs. Temperature



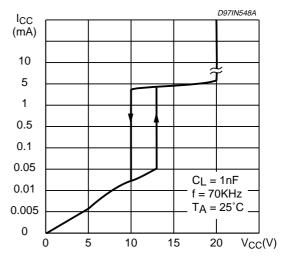


Figure 12. Undervoltage Lockout Threshold vs. Temperature

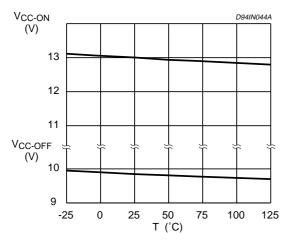
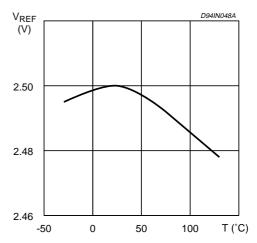


Figure 14. Voltage Feedback Input Threshold vs. Temperature



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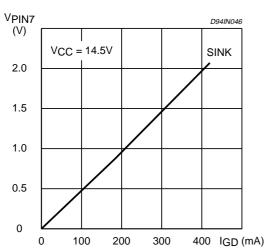


Figure 15. Output Saturation Voltage vs. Sink Current



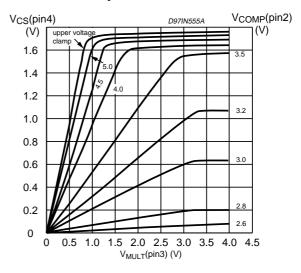
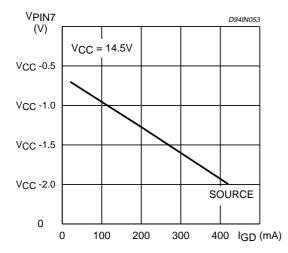
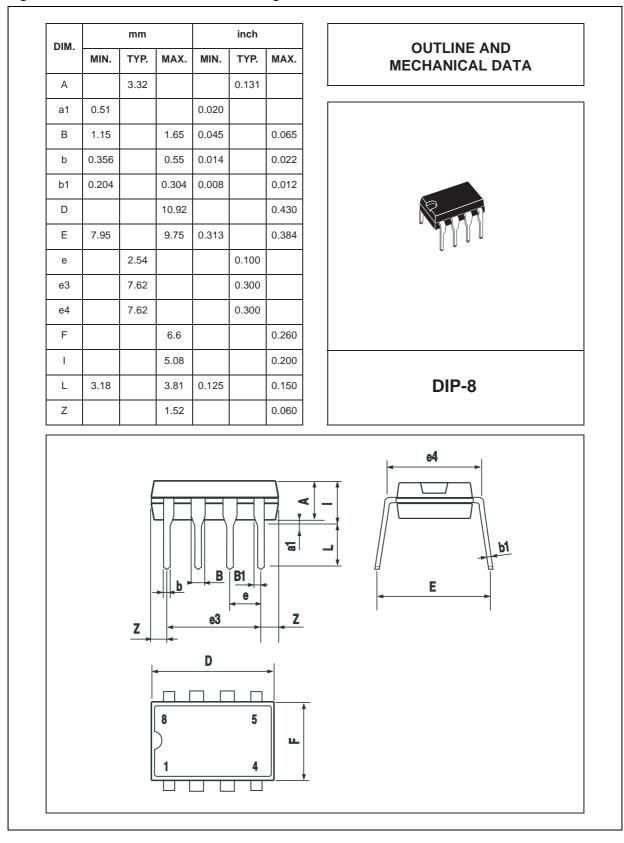


Figure 16. Output Saturation Voltage vs. Source Current



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Figure 18. DIP-8 Mechanical Data & Package Dimensions

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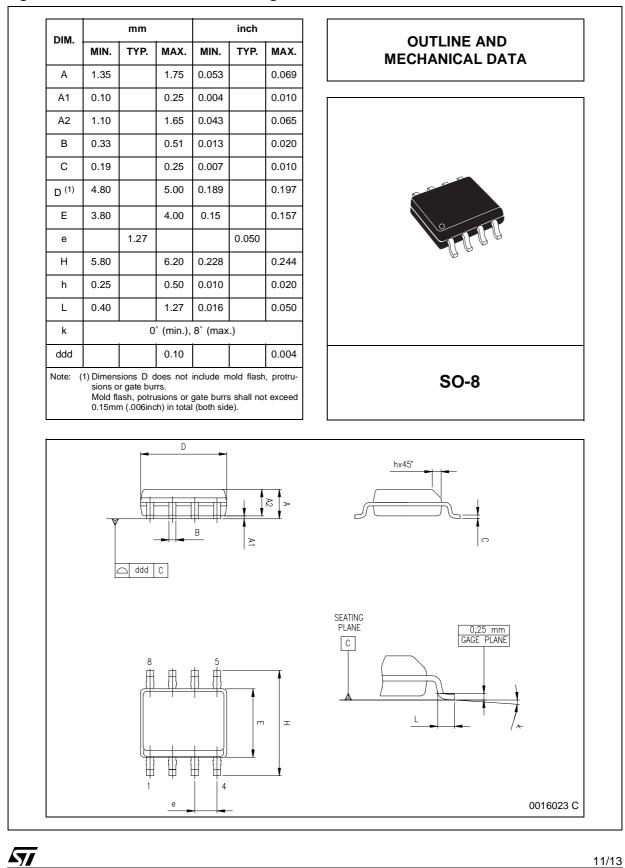


Figure 19. SO-8 Mechanical Data & Package Dimensions

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Table 7. Revision History

Date	Revision	Description of Changes
January 2004	15	First Issue
June 2004	16	Modified the Style-look in compliance with the "Corporate Technical Publications Design Guide". Changed input of the power amplifier connected to Multiplier (Fig. 2).

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